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Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

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Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China



Dual-Cell Li-Ion/Li-Polymer Battery Charge Management Controller with Input Overvoltage Protection

Features

- Complete Linear Charge Management Controller:
 - Integrated Input Overvoltage Protection
 - Integrated Pass Transistor
 - Integrated Current Sense
 - Integrated Reverse Discharge Protection
- Constant Current/Constant Voltage Operation with Thermal Regulation
- 4.15V Undervoltage Lockout (UVLO)
- 13V Input Overvoltage Protection
- High Accuracy Preset Voltage Regulation through Full Temperature Range (-5°C to +55°C ±0.6%)
- Battery Charge Voltage Options:
 - 8.20V, 8.40V, 8.7V or 8.8V
- Resistor Programmable Fast Charge Current:
 - 130 mA-1100 mA
- Preconditioning of Deeply Depleted Cells:
 - Available Options: 10% or Disable
- Integrated Precondition Timer:
 - 32 Minutes or Disable
- Automatic End-of-Charge Control:
 - Selectable Minimum Current Ratio: 5%, 7.5%, 10% or 20%
 - Elapse Safety Timer: 4 hr, 6 hr, 8 hr or Disable
- Automatic Recharge:
 - Available Options: 95% or Disable
- Factory Preset Charge Status Output:
 - On/Off or Flashing
- Soft Start
- Temperature Range: -40°C to +85°C
- Packaging: DFN-10 (3 mm x 3 mm)

Applications

- Digital Camcorders
- Portable Media Players
- Ultra Mobile Personal Computers
- Netbook Computers
- Handheld Devices
- Walkie-Talkie
- Low-Cost 2-Cell Li-Ion/Li-Poly Chargers/Cradles

Description

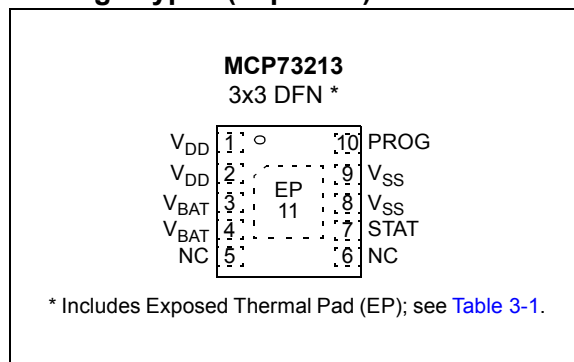
The MCP73213 is a highly integrated Li-Ion battery charge management controller for use in space-limited and cost-sensitive applications. The MCP73213 provides specific charge algorithms for dual-cell Li-Ion/Li-Polymer batteries to achieve optimal capacity and safety in the shortest charging time possible. Along with its small physical size, the low number of external components makes the MCP73213 ideally suitable for portable applications. The absolute maximum voltage, up to 18V, allows the use of MCP73213 in harsh environments, such as low-cost wall wart or voltage spikes from plug/unplug.

The MCP73213 employs a constant current/constant voltage charge algorithm. The various charging voltage regulations provide design engineers flexibility to use in different applications. The fast charge, constant current value is set with one external resistor from 130 mA to 1100 mA. The MCP73213 limits the charge current based on die temperature during high-power or high-ambient conditions. This thermal regulation optimizes the charge cycle time while maintaining device reliability.

The PROG pin of the MCP73213 also serves as enable pin. When high impedance is applied, the MCP73213 will be in Standby mode.

The MCP73213 is fully specified over the ambient temperature range of -40°C to +85°C. The MCP73213 is available in a 10-lead DFN package.

Package Types (Top View)



MCP73213

Typical Application

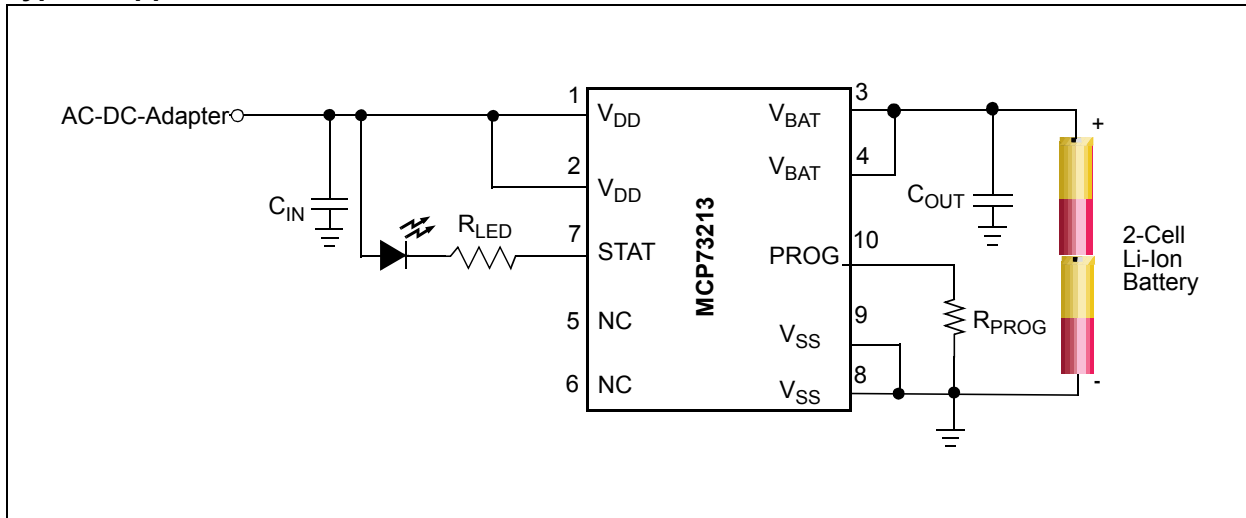


TABLE 1: AVAILABLE FACTORY PRESET OPTIONS

Charge Voltage	OVP	Preconditioning Charge Current	Preconditioning Threshold	Precondition Timer	Elapse Timer	End-of-Charge Control	Automatic Recharge	Output Status
8.2V	13V	Disable/10%	66.5%/71.5%	Disable/ 32 Minimum	Disable/4 hr/ 6 hr/8 hr	5%/7.5%/ 10%/20%	No/ Yes	Type 1/ Type 2
8.4V	13V	Disable/10%	66.5%/71.5%	Disable/ 32 Minimum	Disable/4 hr/ 6 hr/8 hr	5%/7.5%/ 10%/20%	No/ Yes	Type 1/ Type 2
8.7V	13V	Disable/10%	66.5%/71.5%	Disable/ 32 Minimum	Disable/4 hr/ 6 hr/8 hr	5%/7.5%/ 10%/20%	No/ Yes	Type 1/ Type 2
8.8V	13V	Disable/10%	66.5%/71.5%	Disable/ 32 Minimum	Disable/4 hr/ 6 hr/8 hr	5%/7.5%/ 10%/20%	No/ Yes	Type 1/ Type 2

- Note 1:** I_{REG} : Regulated fast charge current.
Note 2: V_{REG} : Regulated charge voltage.
Note 3: I_{PREG}/I_{REG} : Preconditioning charge current; ratio of regulated fast charge current.
Note 4: I_{TERM}/I_{REG} : End-of-Charge control; ratio of regulated fast charge current.
Note 5: V_{RTH}/V_{REG} : Recharge threshold; ratio of regulated battery voltage.
Note 6: V_{PTH}/V_{REG} : Preconditioning threshold voltage.
Note 7: Type 1: On/Off; Type 2: Flashing. Please refer to [Table 5-2](#).

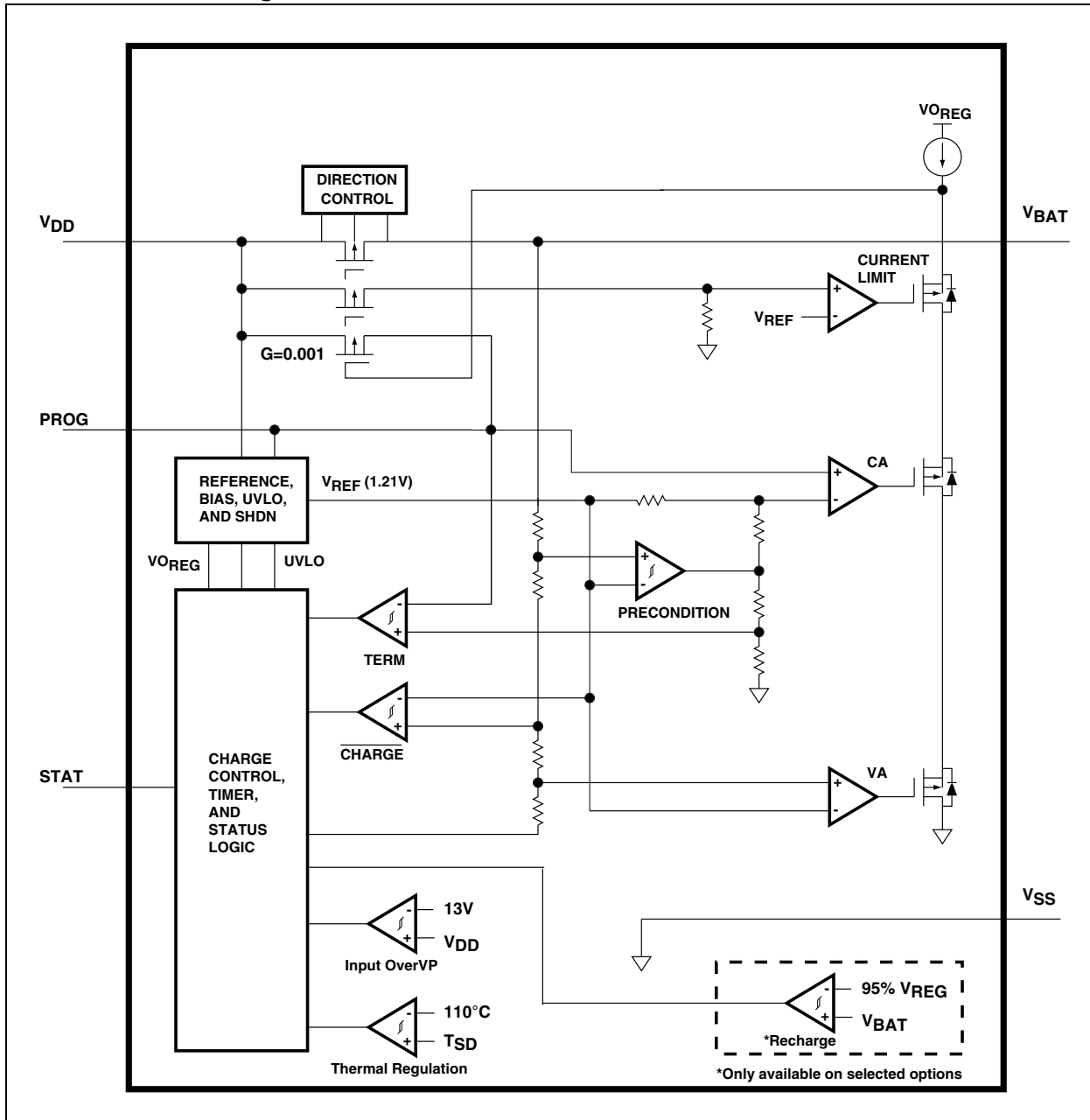
TABLE 2: STANDARD SAMPLE OPTIONS

Part Number	V_{REG}	OVP	I_{PREG}/I_{REG}	Precharge Timer	Elapse Timer	I_{TERM}/I_{REG}	V_{RTH}/V_{REG}	V_{PTH}/V_{REG}	Output Status
MCP73213-B6S/MF	8.20V	13V	10%	32 Minimum	6 hr	10%	95%	71.5%	Type 1
MCP73213-A6S/MF	8.40V	13V	10%	32 Minimum	6 hr	10%	95%	71.5%	Type 1

- Note 1:** Customers should contact their distributor, representatives or field application engineer (FAE) for support and sample. Local sales offices are also available to help customers. A listing of sales offices and locations is included at the back of this document. Technical support is available through the web site at: <http://www.microchip.com/support>

MCP73213

Functional Block Diagram



1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings†

V_{DD}	18.0V
V_{PROG}	6.0V
All Inputs and Outputs w.r.t. V_{SS}	-0.3 to ($V_{DD}+0.3$)V
Maximum Junction Temperature, T_J . Internally Limited	
Storage Temperature	-65°C to +150°C
ESD Protection on All Pins	≥ 4 kV HBM
ESD Protection on All Pins	≥ 300 V MM

† **Notice:** Stresses above those listed under “Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

DC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG}(\text{Typical}) + 0.3V]$ to 12V, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$. Typical values are at $+25^\circ\text{C}$, $V_{DD} = [V_{REG}(\text{Typical}) + 1.0V]$						
Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Supply Input						
Input Voltage Range	V_{DD}	4	—	16	V	
Operating Supply Voltage	V_{DD}	4.2	—	13	V	
Supply Current	I_{SS}	—	4	5.5	μA	Shutdown ($V_{DD} \leq V_{BAT} - 150$ mV)
		—	700	1500	μA	Charging
		—	50	125	μA	Standby (PROG Floating)
		—	50	150	μA	Charge Complete; No Battery; $V_{DD} < V_{STOP}$
Battery Discharge Current						
Output Reverse Leakage Current	$I_{DISCHARGE}$	—	0.5	2	μA	Standby (PROG Floating)
		—	0.5	2	μA	Shutdown ($V_{DD} \leq V_{BAT}$ or $V_{DD} < V_{STOP}$)
		—	10	17	μA	Charge Complete; V_{DD} is present
Undervoltage Lockout						
UVLO Start Threshold	V_{START}	4.10	4.15	4.25	V	
UVLO Stop Threshold	V_{STOP}	4.00	4.05	4.10	V	
UVLO Hysteresis	V_{HYS}	—	100	—	mV	
Overvoltage Protection						
OVP Start Threshold	V_{OVP}	12.8	13	13.2	V	
OVP Hysteresis	V_{OVPHYS}	—	150	—	mV	
Voltage Regulation (Constant Voltage Mode)						
Regulated Output Voltage Options	V_{REG}	8.15	8.20	8.25	V	$T_A = -5^\circ\text{C}$ to $+55^\circ\text{C}$
		8.35	8.40	8.45	V	$V_{DD} = [V_{REG}(\text{Typical})+1V]$
		8.65	8.70	8.75	V	$I_{OUT} = 50$ mA
		8.75	8.80	8.85	V	
Output Voltage Tolerance	V_{RTOL}	-0.6	—	0.6	%	
Line Regulation	$ \Delta V_{BAT}/V_{BAT} /\Delta V_{DD}$	—	0.05	0.20	%/V	$V_{DD} = [V_{REG}(\text{Typical})+1V]$ to 12V $I_{OUT} = 50$ mA
Load Regulation	$ \Delta V_{BAT}/V_{BAT} $	—	0.05	0.20	%	$I_{OUT} = 50$ mA - 150 mA $V_{DD} = [V_{REG}(\text{Typical})+1V]$

Note 1: Not production tested. Ensured by design.

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DC CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG}(\text{Typical}) + 0.3V]$ to 12V, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$. Typical values are at $+25^\circ\text{C}$, $V_{DD} = [V_{REG}(\text{Typical}) + 1.0V]$						
Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Supply Ripple Attenuation	PSRR	—	-46	—	dB	$I_{OUT} = 20\text{ mA}$, 10 Hz to 1 kHz
		—	-30	—	dB	$I_{OUT} = 20\text{ mA}$, 10 Hz to 10 kHz
Battery Short Protection						
BSP Start Threshold	V_{SHORT}	—	3.4	—	V	
BSP Hysteresis	V_{BSPHYS}	—	150	—	mV	
BSP Regulation Current	I_{SHORT}	—	25	—	mA	
Current Regulation (Fast Charge, Constant-Current Mode)						
Fast Charge Current Regulation	I_{REG}	130	—	1100	mA	$T_A = -5^\circ\text{C}$ to $+55^\circ\text{C}$
		117	130	143	mA	PROG = 10 k Ω
		900	1000	1100	mA	PROG = 1.1 k Ω
Preconditioning Current Regulation (Trickle Charge Constant-Current Mode)						
Precondition Current Ratio	I_{PREG}/I_{REG}	—	10	—	%	PROG = 1 k Ω to 10 k Ω $T_A = -5^\circ\text{C}$ to $+55^\circ\text{C}$
		—	100	—	%	No Preconditioning
Precondition Voltage Threshold Ratio	V_{PTH}/V_{REG}	64	66.5	69	%	V_{BAT} Low-to-High
		69	71.5	74	%	V_{BAT} Low-to-High
Precondition Hysteresis	V_{PHYS}	—	100	—	mV	V_{BAT} High-to-Low (Note 1)
Charge Termination						
Charge Termination Current Ratio	I_{TERM}/I_{REG}	3.7	5	6.3	%	PROG = 1 k Ω to 10 k Ω $T_A = -5^\circ\text{C}$ to $+55^\circ\text{C}$
		5.6	7.5	9.4	—	
		7.5	10	12.5	—	
		15	20	25	—	
Automatic Recharge						
Recharge Voltage Threshold Ratio	V_{RTH}/V_{REG}	93	95.0	97	%	V_{BAT} High-to-Low No Automatic Recharge
		—	0	—	%	
Pass Transistor ON-Resistance						
ON-Resistance	R_{DSON}	—	350	—	m Ω	$V_{DD} = 4.5V$, $T_J = 105^\circ\text{C}$ (Note 1)
Status Indicator - STAT						
Sink Current	I_{SINK}	—	20	35	mA	
Low Output Voltage	V_{OL}	—	0.2	0.5	V	$I_{SINK} = 4\text{ mA}$
Input Leakage Current	I_{LK}	—	0.001	1	μA	High Impedance, V_{DD} on pin
PROG Input						
Charge Impedance Range	R_{PROG}	1	—	22	k Ω	
Shutdown Impedance	R_{PROG}	—	200	—	k Ω	Impedance for Shutdown
Automatic Power-Down						
Automatic Power-Down Entry Threshold	V_{PENTRY}	$V_{BAT} + 10\text{ mV}$	$V_{BAT} + 50\text{ mV}$	—	V	V_{DD} Falling
Automatic Power-Down Exit Threshold	V_{PEXIT}	—	$V_{BAT} + 150\text{ mV}$	$V_{BAT} + 250\text{ mV}$	V	V_{DD} Rising

Note 1: Not production tested. Ensured by design.

DC CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG}(\text{Typical}) + 0.3V]$ to 12V, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$. Typical values are at $+25^\circ\text{C}$, $V_{DD} = [V_{REG}(\text{Typical}) + 1.0V]$

Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Thermal Shutdown						
Die Temperature	T_{SD}	—	150	—	$^\circ\text{C}$	
Die Temperature Hysteresis	T_{SDHYS}	—	10	—	$^\circ\text{C}$	

Note 1: Not production tested. Ensured by design.

AC CHARACTERISTICS

Electrical Specifications: Unless otherwise specified, all limits apply for $V_{DD} = [V_{REG}(\text{Typical}) + 0.3V]$ to 12V, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$. Typical values are at $+25^\circ\text{C}$, $V_{DD} = [V_{REG}(\text{Typical}) + 1.0V]$

Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Elapsed Timer						
Elapsed Timer Period	$t_{EL-APSED}$	—	0	—	Hours	Timer Disabled
		3.6	4.0	4.4	Hours	
		5.4	6.0	6.6	Hours	
		7.2	8.0	8.8	Hours	
Preconditioning Timer						
Preconditioning Timer Period	t_{PRECHG}	—	0	—	Hours	Disabled Timer
		0.4	0.5	0.6	Hours	
Status Indicator						
Status Output Turn-Off	t_{OFF}	—	—	500	μs	$I_{SINK} = 1 \text{ mA to } 0 \text{ mA}$ (Note 1)
Status Output Turn-On	t_{ON}	—	—	500	—	$I_{SINK} = 0 \text{ mA to } 1 \text{ mA}$ (Note 1)

Note 1: Not production tested. Ensured by design.

TEMPERATURE SPECIFICATIONS

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{DD} = [V_{REG}(\text{Typical}) + 0.3V]$ to 6V. Typical values are at $+25^\circ\text{C}$, $V_{DD} = [V_{REG}(\text{Typical}) + 1.0V]$

Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Temperature Ranges						
Specified Temperature Range	T_A	-40	—	+85	$^\circ\text{C}$	
Operating Temperature Range	T_J	-40	—	+125	$^\circ\text{C}$	
Storage Temperature Range	T_A	-65	—	+150	$^\circ\text{C}$	
Thermal Package Resistances						
Thermal Resistance, DFN-10LD (3x3)	θ_{JA}	—	62	—	$^\circ\text{C/W}$	4-Layer JC51-7 Standard Board, Natural Convection
	θ_{JC}	—	20.5	—	$^\circ\text{C/W}$	

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NOTES:

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Note: Unless otherwise indicated, $V_{DD} = [V_{REG}(Typical) + 1V]$, $I_{OUT} = 50\text{ mA}$ and $T_A = +25^\circ\text{C}$, Constant Voltage mode.

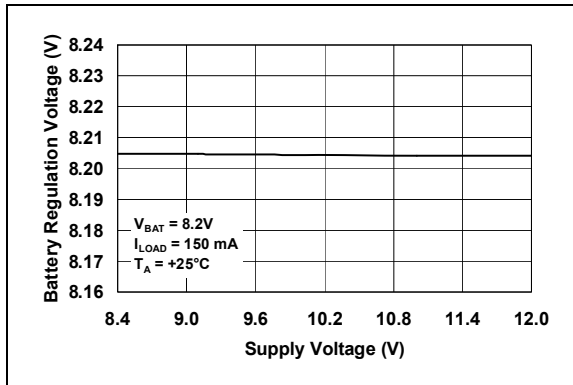


FIGURE 2-1: Battery Regulation Voltage (V_{BAT}) vs. Supply Voltage (V_{DD}).

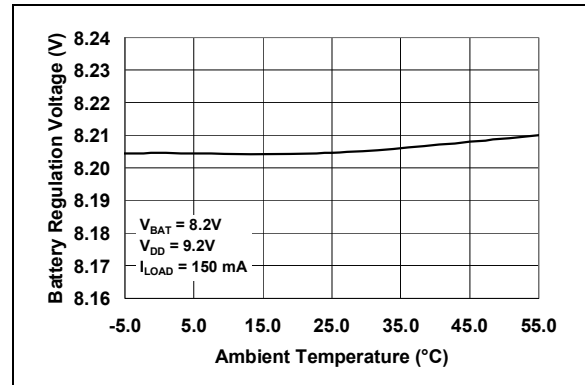


FIGURE 2-4: Battery Regulation Voltage (V_{BAT}) vs. Ambient Temperature (T_A).

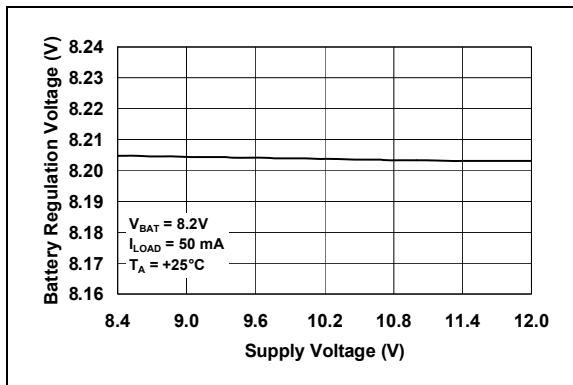


FIGURE 2-2: Battery Regulation Voltage (V_{BAT}) vs. Supply Voltage (V_{DD}).

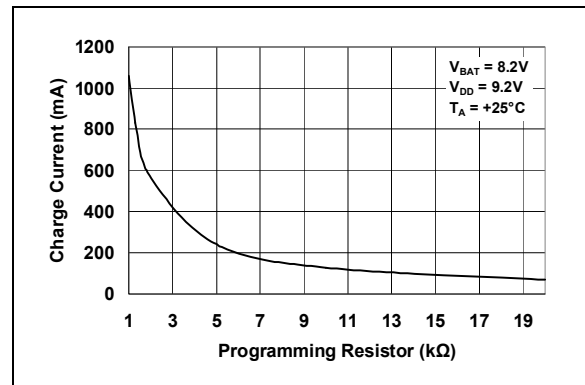


FIGURE 2-5: Charge Current (I_{OUT}) vs. Programming Resistor (R_{PROG}).

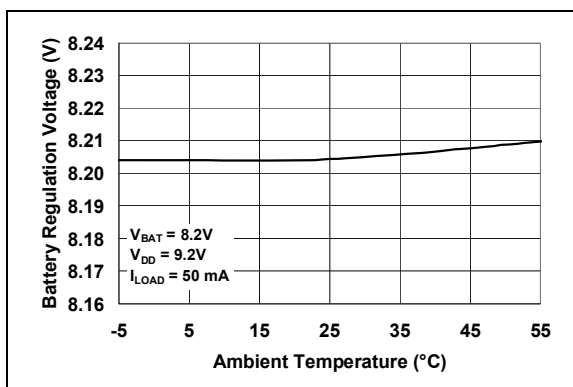


FIGURE 2-3: Battery Regulation Voltage (V_{BAT}) vs. Ambient Temperature (T_A).

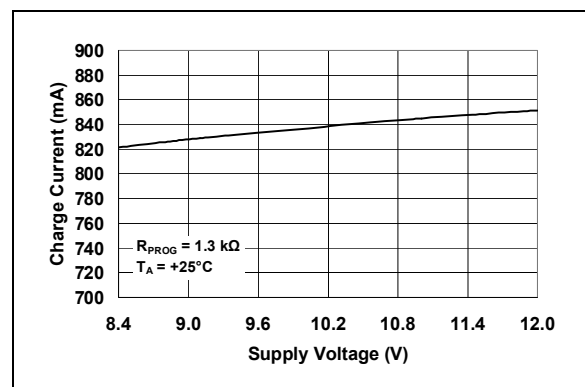


FIGURE 2-6: Charge Current (I_{OUT}) vs. Supply Voltage (V_{DD}).

MCP73213

TYPICAL PERFORMANCE CURVES (CONTINUED)

Note: Unless otherwise indicated, $V_{DD} = [V_{REG}(Typical) + 1V]$, $I_{OUT} = 10\text{ mA}$ and $T_A = +25^\circ\text{C}$, Constant-voltage mode.

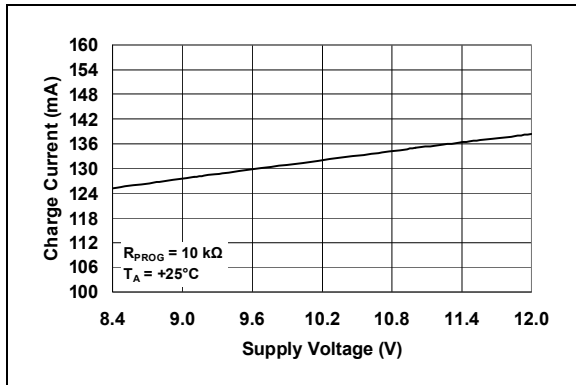


FIGURE 2-7: Charge Current (I_{OUT}) vs. Supply Voltage (V_{DD}).

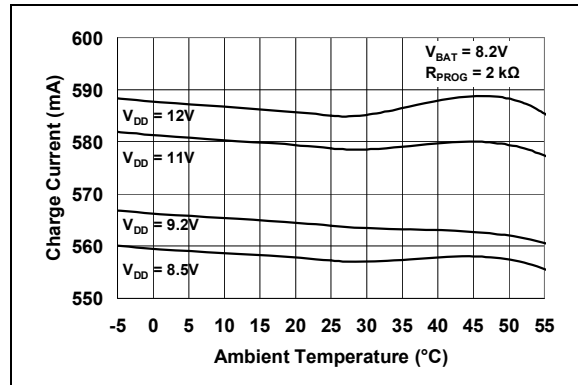


FIGURE 2-10: Charge Current (I_{OUT}) vs. Ambient Temperature (T_A).

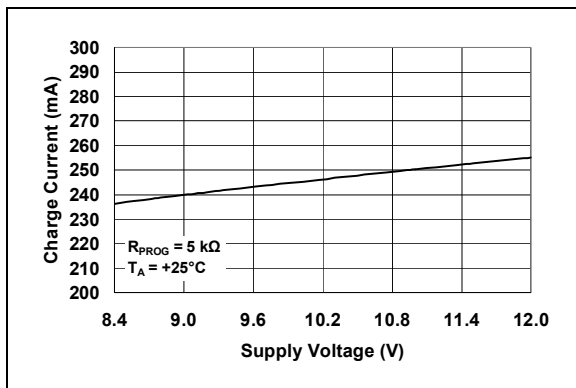


FIGURE 2-8: Charge Current (I_{OUT}) vs. Supply Voltage (V_{DD}).

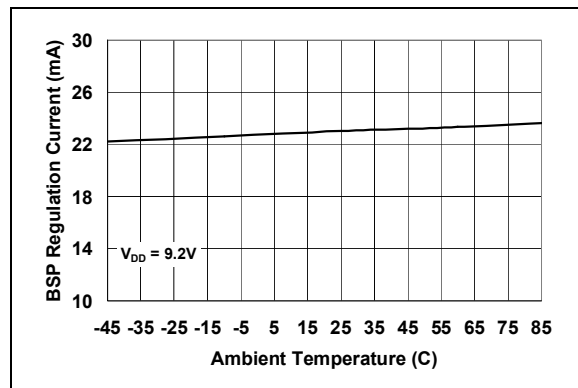


FIGURE 2-11: Battery Short Protection Regulation Current (I_{SHORT}) vs. Ambient Temperature (T_A).

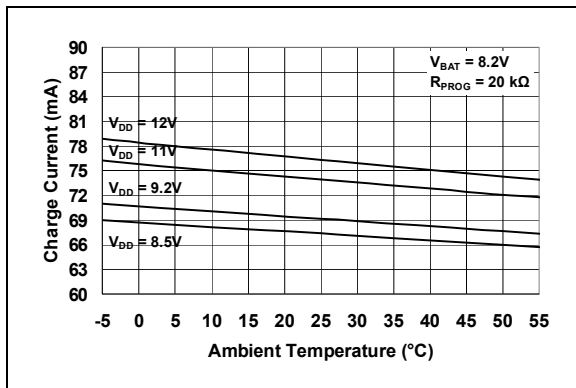


FIGURE 2-9: Charge Current (I_{OUT}) vs. Ambient Temperature (T_A).

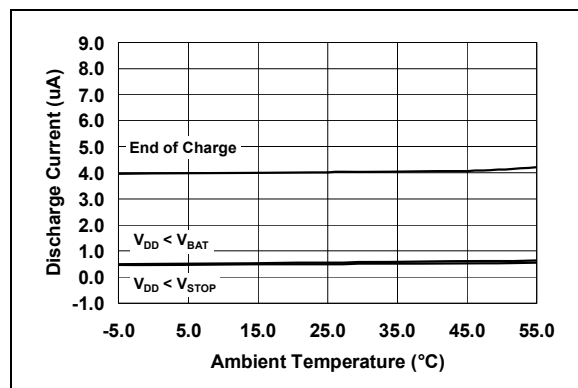


FIGURE 2-12: Output Leakage Current ($I_{DISCHARGE}$) vs. Ambient Temperature (T_A).

TYPICAL PERFORMANCE CURVES (CONTINUED)

Note: Unless otherwise indicated, $V_{DD} = [V_{REG}(Typical) + 1V]$, $I_{OUT} = 10\text{ mA}$ and $T_A = +25^\circ\text{C}$, Constant-voltage mode.

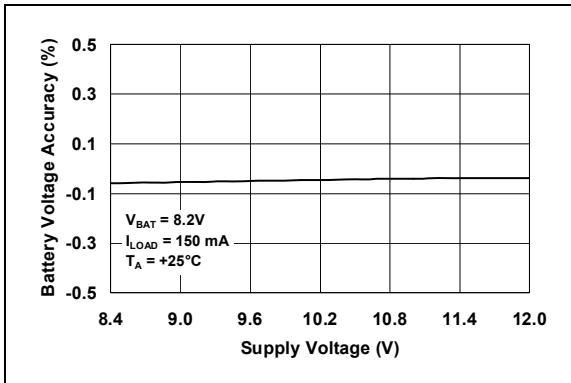


FIGURE 2-13: Battery Voltage Accuracy (V_{RTOL}) vs. Supply Voltage (V_{DD}).

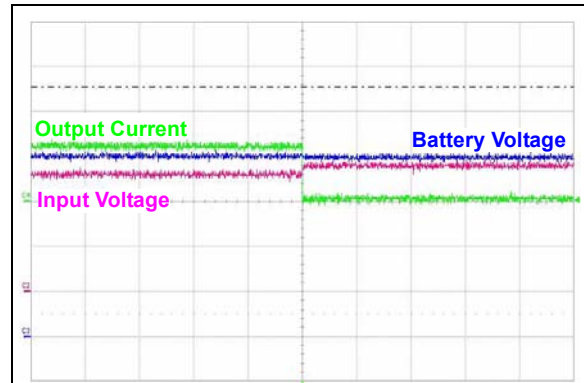


FIGURE 2-16: Input Overvoltage Protection.

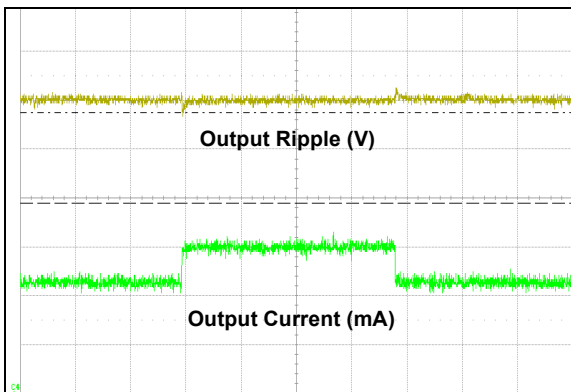


FIGURE 2-14: Load Transient Response ($I_{LOAD} = 50\text{ mA/Div}$, Output: 100 mV/Div , Time: $100\text{ }\mu\text{s/Div}$).

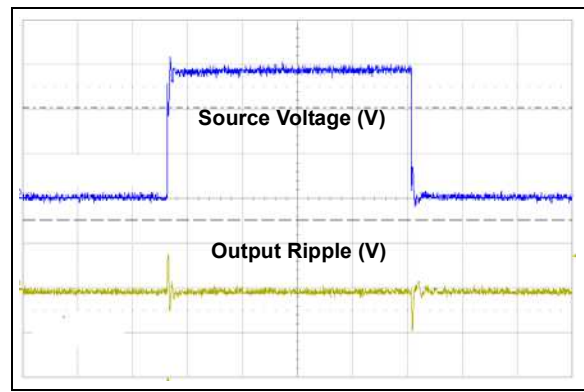


FIGURE 2-17: Line Transient Response ($I_{LOAD} = 10\text{ mA}$) ($100\text{ }\mu\text{s/Div}$).

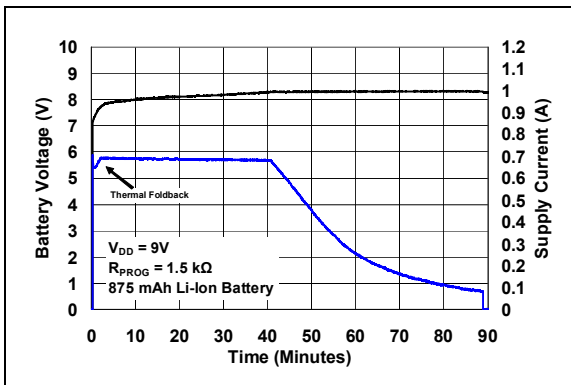


FIGURE 2-15: Complete Charge Cycle.

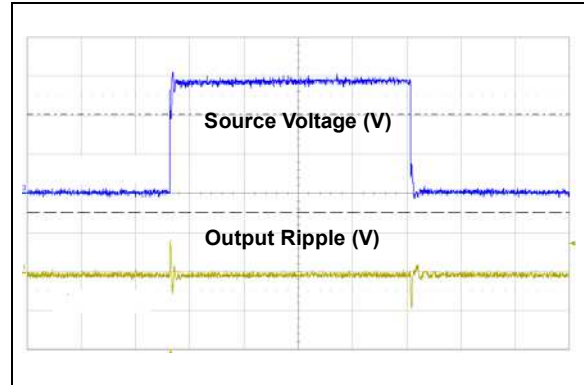


FIGURE 2-18: Line Transient Response ($I_{LOAD} = 100\text{ mA}$) ($100\text{ }\mu\text{s/Div}$).

MCP73213

NOTES:

3.0 PIN DESCRIPTION

The descriptions of the pins are listed in [Table 3-1](#).

TABLE 3-1: PIN FUNCTION TABLE

MCP73213	Symbol	I/O	Description
DFN-10			
1, 2	V _{DD}	I	Battery Management Input Supply Pin
3, 4	V _{BAT}	I/O	Battery Charge Control Output Pin
5, 6	NC	—	No Connection Pin
7	STAT	O	Battery Charge Status Output Pin
8, 9	V _{SS}	—	Battery Management 0V Reference Pin
10	PROG	I/O	Battery Charge Current Regulation Program and Charge Control Enable Pin
11	EP	—	Exposed Pad Pin

3.1 Battery Management Input Supply (V_{DD})

A supply voltage of [V_{REG} (Typical) + 0.3V] to 13.0V is recommended. Bypass to V_{SS} with a minimum of 1 μF. The V_{DD} pin is rated 18V absolute maximum to prevent a sudden rise in input voltage from spikes or low-cost AC-DC wall adapters from causing an over-voltage condition and damaging the device.

3.2 Battery Charge Control Output (V_{BAT})

Connect to the positive terminal of the battery. Bypass to V_{SS} with a minimum of 1 μF to ensure loop stability when the battery is disconnected.

3.3 No Connection (NC)

No connection.

3.4 Status Output (STAT)

STAT is an open-drain logic output for connection to an LED for charge status indication in stand-alone applications. Alternatively, a pull-up resistor can be applied for interfacing to a host microcontroller. Refer to [Table 5-2](#) for a summary of the status output during a charge cycle.

3.5 Battery Management 0V Reference (V_{SS})

Connect to the negative terminal of the battery and input supply.

3.6 Current Regulation Set (PROG)

The fast charge current is set by placing a resistor from PROG to V_{SS} during constant current (CC) mode. PROG pin also serves as charge control enable. When a typical 200 kΩ impedance is applied to the PROG pin, the MCP73213 will go into standby mode until the high impedance is removed. Refer to [Section 5.5 “Constant-Current Mode - Fast Charge”](#) for details.

3.7 Exposed Pad (EP)

Connect the Exposed Thermal Pad (EP) to the exposed copper area on the Printed Circuit Board (PCB) for thermal enhancement. Additional vias in the copper area under the MCP73213 device can improve heat dissipation performance and simplify the assembly process.

MCP73213

NOTES:

4.0 DEVICE OVERVIEW

The MCP73213 are simple, but fully integrated linear charge management controllers. Figure 4-1 depicts the operational flow algorithm.

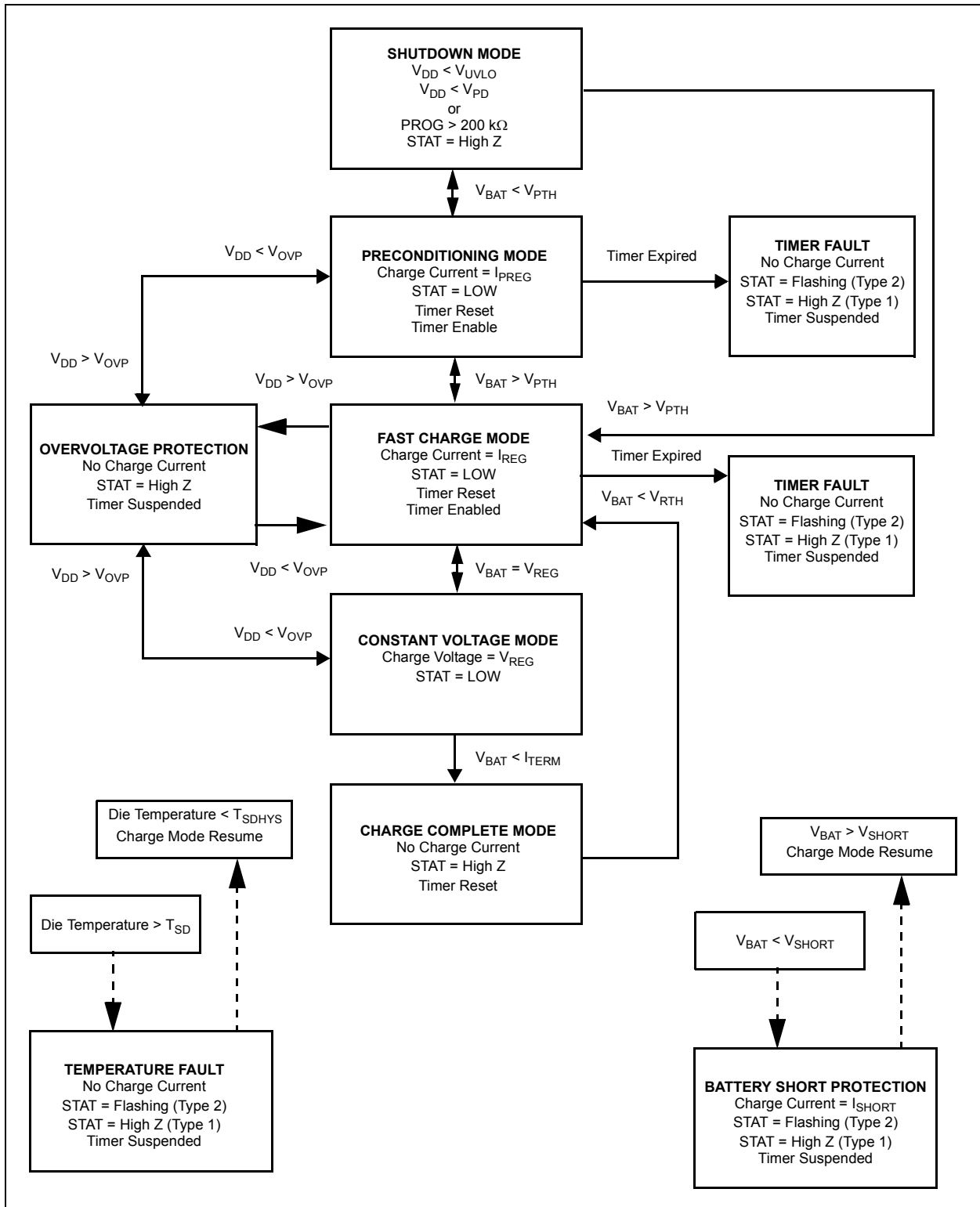


FIGURE 4-1: The MCP73213 Flow Chart.

MCP73213

NOTES:

5.0 DETAILED DESCRIPTION

5.1 Undervoltage Lockout (UVLO)

An internal undervoltage lockout (UVLO) circuit monitors the input voltage and keeps the charger in shutdown mode until the input supply rises above the UVLO threshold. In the event a battery is present when the input power is applied, the input supply must rise approximately 150 mV above the battery voltage before the MCP73213 device becomes operational.

The UVLO circuit places the device in shutdown mode if the input supply falls to approximately 150 mV above the battery voltage. The UVLO circuit is always active. Any time the input supply is below the UVLO threshold or approximately 150 mV of the voltage at the V_{BAT} pin, the MCP73213 device is placed in a shutdown mode.

5.2 Overvoltage Protection (OVP)

An internal overvoltage protection (OVP) circuit monitors the input voltage and keeps the charger in shutdown mode when the input supply rises above the typical 13V OVP threshold. The OVP hysteresis is approximately 150 mV for the MCP73213 device.

The MCP73213 device is operational between UVLO and OVP thresholds. The OVP circuit is also recognized as overvoltage lockout (OVLO).

5.3 Charge Qualification

When the input power is applied, the input supply must rise 150 mV above the battery voltage before the MCP73213 becomes operational.

The automatic power-down circuit places the device in a shutdown mode if the input supply falls to within +50 mV of the battery voltage.

The automatic circuit is always active. At any time the input supply is within +50 mV of the voltage at the V_{BAT} pin, the MCP73213 is placed in a shutdown mode.

For a charge cycle to begin, the automatic power-down conditions must be met and the charge enable input must be above the input high threshold.

5.3.1 BATTERY MANAGEMENT INPUT SUPPLY (V_{DD})

The V_{DD} input is the input supply to the MCP73213. The MCP73213 automatically enters a power-down mode if the voltage on the V_{DD} input falls to within +50 mV of the battery voltage. This feature prevents draining the battery pack when the V_{DD} supply is not present.

5.3.2 BATTERY CHARGE CONTROL OUTPUT (V_{BAT})

The battery charge control output is the drain terminal of an internal P-channel MOSFET. The MCP73213 provides constant current and voltage regulation to the battery pack by controlling this MOSFET in the linear region. The battery charge control output should be connected to the positive terminal of the battery pack.

5.3.3 BATTERY DETECTION

The MCP73213 detects the battery presence with charging of the output capacitor. The charge flow will initiate when the voltage on V_{BAT} is pulled below the $V_{RECHARGE}$ threshold. Refer to [Section 1.0 “Electrical Characteristics”](#) for $V_{RECHARGE}$ values. The value will be the same for nonrechargeable devices.

When $V_{BAT} > V_{REG} + \text{Hysteresis}$, the charge will be suspended (or not started, depending on the condition) to prevent overcharging.

5.4 Preconditioning

If the voltage at the V_{BAT} pin is less than the preconditioning threshold, the MCP73213 device enters a preconditioning mode. The preconditioning threshold is factory set. Refer to [Section 1.0 “Electrical Characteristics”](#) for preconditioning threshold options.

In this mode, the MCP73213 device supplies 10% of the fast charge current (established with the value of the resistor connected to the PROG pin) to the battery.

When the voltage at the V_{BAT} pin rises above the preconditioning threshold, the MCP73213 device enters the Constant Current (Fast Charge) mode.

Note: The MCP73213 device also offers options with no preconditioning.

5.4.1 TIMER EXPIRED DURING PRECONDITIONING MODE

If the internal timer expires before the voltage threshold is reached for Fast Charge mode, a timer fault is indicated and the charge cycle terminates. The MCP73213 device remains in this condition until the battery is removed or input power is cycled. If the battery is removed, the MCP73213 device enters the Standby mode, where it remains until a battery is reinserted.

Note: The typical preconditioning timer for MCP73213 is 32 minutes. The MCP73213 also offers options with no preconditioning timer.

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5.5 Constant-Current Mode - Fast Charge

During Constant-Current mode, the programmed charge current is supplied to the battery or load.

The charge current is established using a single resistor from PROG to V_{SS}. The program resistor and the charge current are calculated using the following equation:

EQUATION 5-1:

$$I_{REG} = 1104 \times R_{PROG}^{-0.93}$$

Where:

R_{PROG} = kilohm (kΩ)
 I_{REG} = milliampere (mA)

EQUATION 5-2:

$$R_{PROG} = 10^{\left(\log\left(\frac{I_{REG}}{1104}\right)\right)/(-0.93)}$$

Where:

R_{PROG} = kilohm (kΩ)
 I_{REG} = milliampere (mA)

Table 5-1 provides commonly seen E96 (1%) and E24 (5%) resistors for various charge current to reduce design time.

TABLE 5-1: RESISTOR LOOKUP TABLE

Charge Current (mA)	Recommended E96 Resistor (Ω)	Recommended E24 Resistor (Ω)
130	10k	10k
150	8.45k	8.20k
200	6.20k	6.20k
250	4.99k	5.10k
300	4.02k	3.90k
350	3.40k	3.30k
400	3.00k	3.00k
450	2.61k	2.70k
500	2.32k	2.37k
550	2.10k	2.20k
600	1.91k	2.00k
650	1.78k	1.80k
700	1.62k	1.60k
750	1.50k	1.50k
800	1.40k	1.50k
850	1.33k	1.30k
900	1.24k	1.20k
950	1.18k	1.20k
1000	1.10k	1.10k
1100	1.00k	1.00k

Constant-Current mode is maintained until the voltage at the V_{BAT} pin reaches the regulation voltage, V_{REG}. When Constant Current mode is invoked, the internal timer is reset.

5.5.1 TIMER EXPIRED DURING CONSTANT-CURRENT - FAST CHARGE MODE

If the internal timer expires before the recharge voltage threshold is reached, a timer fault is indicated and the charge cycle terminates. The MCP73213 device remains in this condition until the battery is removed. If the battery is removed or input power is cycled, the MCP73213 device enters the Standby mode where it remains until a battery is reinserted.

5.6 Constant-Voltage Mode

When the voltage at the V_{BAT} pin reaches the regulation voltage, V_{REG}, constant voltage regulation begins. The regulation voltage is factory set to 8.2V, 8.4V, 8.7V or 8.8V with a tolerance of ± 0.5%.

5.7 Charge Termination

The charge cycle is terminated when, during Constant-Voltage mode, the average charge current diminishes below a threshold established with the value of 5%, 7.5%, 10% or 20% of fast charge current or the internal timer expires. A 1 ms filter time on the termination comparator ensures that transient load conditions do not result in premature charge cycle termination. The timer period is factory set and can be disabled. Refer to Section 1.0 “Electrical Characteristics” for timer period options.

5.8 Automatic Recharge

The MCP73213 device continuously monitors the voltage at the V_{BAT} pin in the charge complete mode. If the voltage drops below the recharge threshold, another charge cycle begins and current is once again supplied to the battery or load. The recharge threshold is factory set. Refer to Section 1.0 “Electrical Characteristics” for recharge threshold options.

Note: The MCP73213 also offers options with no automatic recharge.

For the MCP73213 device with no recharge option, the MCP73213 will go into Standby mode when the termination condition is met. The charge will not restart until the following conditions have been met:

- Battery is removed from the system and inserted again
- V_{DD} is removed and plugged in again
- R_{PROG} is disconnected (or high-impedance) and reconnected

5.9 Thermal Regulation

The MCP73213 shall limit the charge current based on the die temperature. The thermal regulation optimizes the charge cycle time while maintaining device reliability. Figure 5-1 depicts the thermal regulation for the MCP73213 device. Refer to Section 1.0 “Electrical Characteristics” for thermal package resistances and Section 6.1.1.2 “Thermal Considerations” for calculating power dissipation.

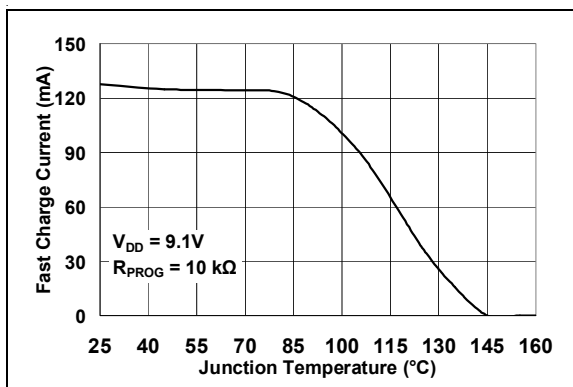


FIGURE 5-1: Thermal Regulation.

5.10 Thermal Shutdown

The MCP73213 suspends charge if the die temperature exceeds +150°C. Charging will resume when the die temperature has cooled by approximately 10°C. The thermal shutdown is a secondary safety feature in the event that there is a failure within the thermal regulation circuitry.

5.11 Status Indicator

The charge status outputs are open-drain outputs with two different states: Low (L), and High-Impedance (High Z). The charge status outputs can be used to illuminate LEDs. Optionally, the charge status outputs can be used as an interface to a host microcontroller. Table 5-2 summarizes the state of the status outputs during a charge cycle.

TABLE 5-2: STATUS OUTPUTS

CHARGE CYCLE STATE	STAT
Shutdown	High Z
Standby	High Z
Preconditioning	L
Constant Current Fast Charge	L
Constant Voltage	L
Charge Complete - Standby	High Z
Temperature Fault	1.6 second 50% D.C. Flashing (Type 2) High Z (Type 1)
Timer Fault	1.6 second 50% D.C. Flashing (Type 2) High Z (Type 1)
Preconditioning Timer Fault	1.6 second 50% D.C. Flashing (Type 2) High Z (Type 1)

5.12 Battery Short Protection

Once a single-cell Li-Ion battery is detected, an internal battery short protection (BSP) circuit starts monitoring the battery voltage. When V_{BAT} falls below a typical 1.7V battery short protection threshold voltage, the charging behavior is postponed. A typical 25 mA detection current is supplied for recovering from the battery short condition.

Preconditioning mode resumes when V_{BAT} raises above the battery short protection threshold. The battery voltage must rise approximately 150 mV above the battery short protection voltage before the MCP73213 device becomes operational.

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6.0 APPLICATIONS

The MCP73213 is designed to operate in conjunction with a host microcontroller or in stand-alone applications. The MCP73213 provides the preferred charge algorithm for dual Lithium-Ion or Lithium-Polymer cells: Constant Current followed by Constant Voltage. [Figure 6-1](#) depicts a typical stand-alone application circuit, while [Figure 6-2](#) depicts the accompanying charge profile.

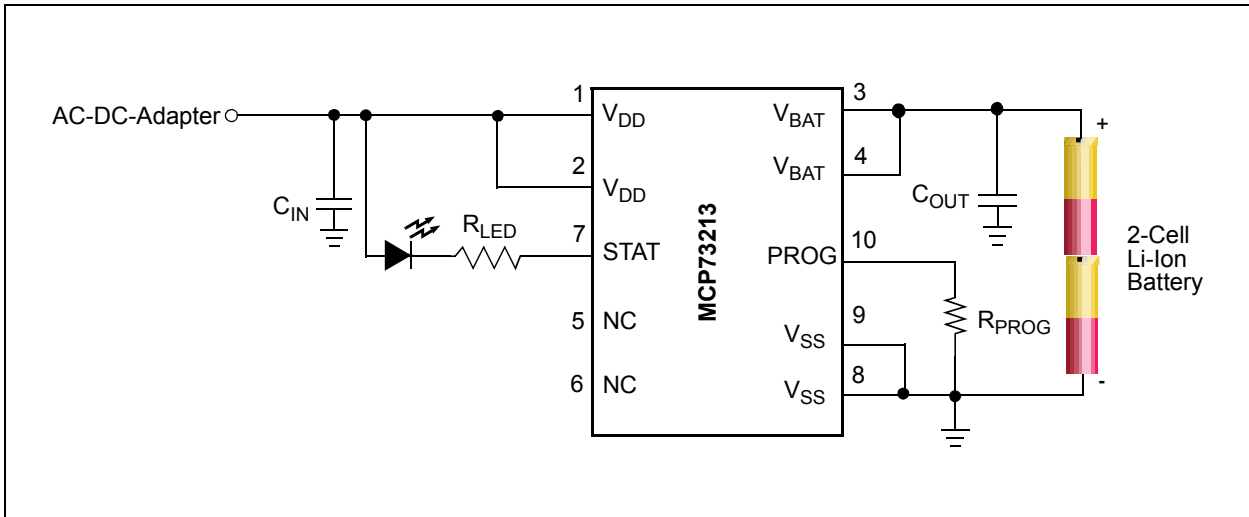


FIGURE 6-1: Typical Application Circuit.

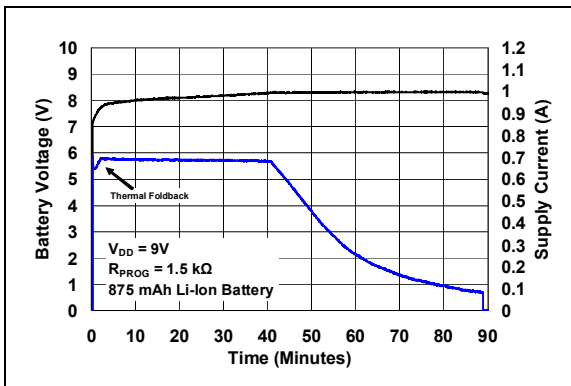


FIGURE 6-2: Typical Charge Profile (875 mAh Li-Ion Battery).

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6.1 Application Circuit Design

Due to the low efficiency of linear charging, the most important factors are thermal design and cost, which are a direct function of the input voltage, output current and thermal impedance between the battery charger and the ambient cooling air. The worst-case situation is when the device has transitioned from the Preconditioning mode to the Constant Current mode. In this situation, the battery charger has to dissipate the maximum power. A trade-off must be made between the charge current, cost and thermal requirements of the charger.

6.1.1 COMPONENT SELECTION

Selection of the external components in [Figure 6-1](#) is crucial to the integrity and reliability of the charging system. The following discussion is intended as a guide for the component selection process.

6.1.1.1 Charge Current

The preferred fast charge current for Li-Ion/Li-Poly cells is below the 1C rate, with an absolute maximum current at the 2C rate. **The recommended fast charge current should be obtained from the battery manufacturer.** For example, a 500 mAh battery pack with 0.7C preferred fast charge current has a charge current of 350 mA. Charging at this rate provides the shortest charge cycle times without degradation to the battery pack performance or life.

Note: Please consult with your battery supplier or refer to the battery data sheet for the preferred charge rate.

6.1.1.2 Thermal Considerations

The worst-case power dissipation in the battery charger occurs when the input voltage is at the maximum and the device has transitioned from Preconditioning mode to Constant-Current mode. In this case, the power dissipation is:

EQUATION 6-1:

$$PowerDissipation = (V_{DDMAX} - V_{PTHMIN}) \times I_{REGMAX}$$

Where:

V_{DDMAX} = the maximum input voltage
 I_{REGMAX} = the maximum fast charge current
 V_{PTHMIN} = the minimum transition threshold voltage

Power dissipation with a 9V, $\pm 10\%$ input voltage source, 400 mA $\pm 10\%$ and preconditioning threshold voltage at 6V is:

EQUATION 6-2:

$$Power\ dissipation = (9.9V - 6.0V) \times 440\ mA = 1.58W$$

This power dissipation with the battery charger in the DFN-10 package will result approximately 98°C above room temperature.

6.1.1.3 External Capacitors

The MCP73213 is stable with or without a battery load. In order to maintain good AC stability in Constant-Voltage mode, a minimum capacitance of 1 μ F is recommended to bypass the V_{BAT} pin to V_{SS} . This capacitance provides compensation when there is no battery load. In addition, the battery and interconnections appear inductive at high frequencies. These elements are in the control feedback loop during Constant Voltage mode. Therefore, bypass capacitance may be necessary to compensate for the inductive nature of the battery pack.

For typical applications, it is recommended to apply a minimum of 16V rated 1 μ F to the output capacitor and a minimum of 25V rated 1 μ F to the input capacitor.

TABLE 6-1: MLCC CAPACITOR EXAMPLE

MLCC Capacitors	Temperature Range	Tolerance
X7R	-55°C to +125°C	$\pm 15\%$
X5R	-55°C to +85°C	$\pm 15\%$

Virtually any good quality output filter capacitor can be used independent of the capacitor's minimum Effective Series Resistance (ESR) value. The actual value of the capacitor (and its associated ESR) depends on the output load current. A 1 μ F ceramic, tantalum or aluminum electrolytic capacitor at the output is usually sufficient to ensure stability.

6.1.1.4 Reverse-Blocking Protection

The MCP73213 provides protection from a faulted or shorted input. Without the protection, a faulted or shorted input would discharge the battery pack through the body diode of the internal pass transistor.

6.2 PCB Layout Issues

For optimum voltage regulation, place the battery pack as close as possible to the device's V_{BAT} and V_{SS} pins to minimize voltage drops along the high-current-carrying PCB traces.

If the PCB layout is used as a heatsink, adding multiple vias in the heatsink pad can help conduct more heat to the backplane of the PCB, thus reducing the junction temperature. Figures 6-4 and 6-5 depict a typical layout with PCB heatsinking.

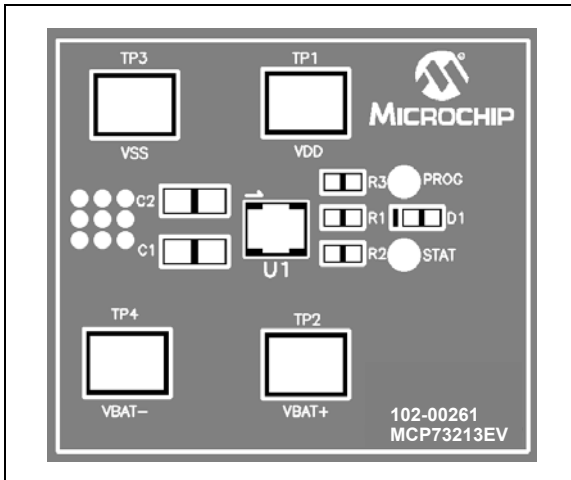


FIGURE 6-3: Typical Layout (Top).

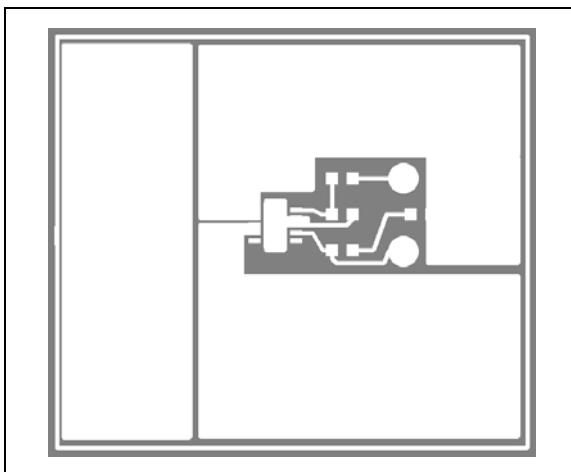


FIGURE 6-4: Typical Layout (Top Metal).

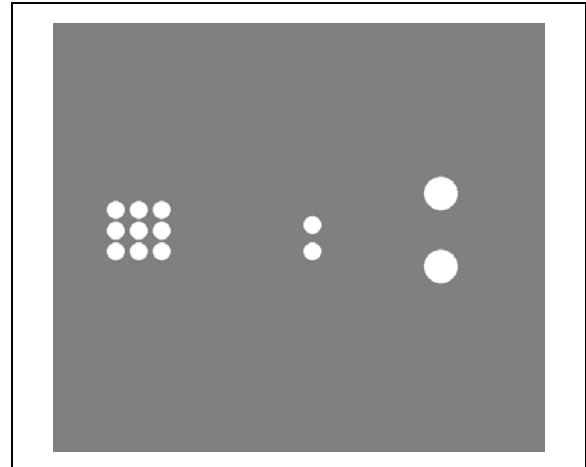


FIGURE 6-5: Typical Layout (Bottom).

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NOTES:

7.0 PACKAGING INFORMATION

7.1 Package Marking Information

10-Lead DFN (3x3)

XXXX
YYWW
NNN

Standard *	
Part Number	Code
MCP73213-A6SI/MF	Z3HI
MCP73213T-A6SI/MF	Z3HI
MCP73213-B6SI/MF	Y3HI
MCP73213T-B6SI/MF	Y3HI

Example:

Z3HI
1443
256

Legend: XX...X Customer-specific information
 Y Year code (last digit of calendar year)
 YY Year code (last 2 digits of calendar year)
 WW Week code (week of January 1 is week '01')
 NNN Alphanumeric traceability code
 Ⓔ3 Pb-free JEDEC® designator for Matte Tin (Sn)
 * This package is Pb-free. The Pb-free JEDEC designator (Ⓔ3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.