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Low G Micromachined Accelerometer

The MMA series of silicon capacitive, micromachined accelerometers features signal conditioning, a 2-pole low pass filter and temperature compensation. Zero-g offset full scale span and filter cut-off are factory set and require no external devices. A full system self-test capability verifies system functionality.

Features

- Integral Signal Conditioning
- Linear Output
- · 2nd Order Bessel Filter
- Calibrated Self-test
- EPROM Parity Check Status
- Transducer Hermetically Sealed at Wafer Level for Superior Reliability
- · Robust Design, High Shock Survivability

Typical Applications

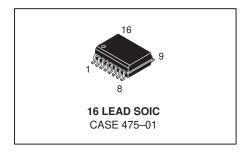
- Vibration Monitoring and Recording
- Appliance Control
- Mechanical Bearing Monitoring
- · Computer Hard Drive Protection
- Computer Mouse and Joysticks
- Virtual Reality Input Devices
- Sports Diagnostic Devices and Systems

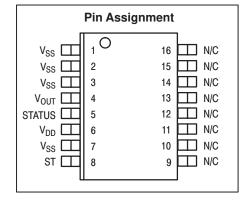
ORDERING INFORMATION

Device	Temperature Range	Case No.	Package	
MMA1250D	-40 to +105°C	Case 475-01	SOIC-16	

MMA1250D

MMA1250D: Z AXIS SENSITIVITY
MICROMACHINED
ACCELEROMETER
±5g





SIMPLIFIED ACCELEROMETER FUNCTIONAL BLOCK DIAGRAM

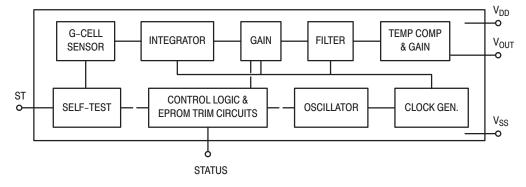


Figure 1. Simplified Accelerometer Functional Block Diagram

REV 0



MAXIMUM RATINGS (Maximum ratings are the limits to which the device can be exposed without causing permanent damage.)

Rating	Symbol	Value	Unit	
Powered Acceleration (all axes)	9 _{pd}	1500	g	
Unpowered Acceleration (all axes)	g _{upd}	2000	g	
Supply Voltage	V _{DD}	-0.3 to +7.0	V	
Drop Test ⁽¹⁾	H _{drop}	1.2	m	
Storage Temperature Range	T _{stg}	-40 to +125	°C	

NOTES:

1. Dropped onto concrete surface from any axis.

ELECTRO STATIC DISCHARGE (ESD)

WARNING: This device is sensitive to electrostatic discharge.

Although the Motorola accelerometers contain internal 2kV ESD protection circuitry, extra precaution must be taken by the user to protect the chip from ESD. A charge of over

2000 volts can accumulate on the human body or associated test equipment. A charge of this magnitude can alter the performance or cause failure of the chip. When handling the accelerometer, proper ESD precautions should be followed to avoid exposing the device to discharges which may be detrimental to its performance.

OPERATING CHARACTERISTICS

(Unless otherwise noted: -40° C \leq T_A \leq +105 $^{\circ}$ C, 4.75 \leq V_{DD} \leq 5.25, Acceleration = 0g, Loaded output⁽¹⁾)

Characteristic	Symbol	Min	Тур	Max	Unit
Operating Range ⁽²⁾ Supply Voltage ⁽³⁾ Supply Current Operating Temperature Range Acceleration Range	V _{DD} I _{DD} T _A 9FS	4.75 1.1 - 40 	5.00 2.1 — 5	5.25 3.0 +105 —	V mA °C g
Output Signal Zero g ($T_A = 25^{\circ}C$, $V_{DD} = 5.0 \text{ V}$) ⁽⁴⁾ Zero g ($V_{DD} = 5.0 \text{ V}$) Sensitivity ($T_A = 25^{\circ}C$, $V_{DD} = 5.0 \text{ V}$) ⁽⁵⁾ Sensitivity ($V_{DD} = 5.0 \text{ V}$) Bandwidth Response Nonlinearity	Voff Voff S S f_3dB NLOUT	2.25 2.0 380 370 42.5 -1.0	2.5 2.5 400 400 50	2.75 3.0 420 430.1 57.5 +1.0	V V mV/g mV/g Hz % FSO
Noise RMS (0.1 Hz – 1.0 kHz) Spectral Density (RMS, 0.1 Hz – 1.0 kHz) ⁽⁶⁾	n _{RMS}		2.0 700	4.0 —	mVrms μg/√ Hz
Self-Test Output Response (V _{DD} = 5.0 V) Input Low Input High Input Loading ⁽⁷⁾ Response Time ⁽⁸⁾	ΔV _{ST} V _{IL} V _{IH} I _{IN} t _{ST}	1.0 V _{SS} 0.7 V _{DD} -50	1.25 — — —125 10	1.5 0.3 V _{DD} V _{DD} - 300 25	V V V μA ms
Status ⁽¹²⁾ (13) Output Low ($I_{load} = 100 \mu A$) Output High ($I_{load} = -100 \mu A$)	V _{OL} V _{OH}	 V _{DD} -0.8	_ _	0.4	V V
Output Stage Performance Electrical Saturation Recovery Time ⁽⁹⁾ Full Scale Output Range (I _{OUT} = -200 μA) Capacitive Load Drive ⁽¹⁰⁾ Output Impedance	t _{DELAY} V _{FSO} C _L Z _O	 V _{SS} +0.25 	— — — 50	2.0 V _{DD} – 0.25 100	ms V pF Ω
Mechanical Characteristics Transverse Sensitivity ⁽¹¹⁾	V _{XZ,YZ}	_	_	5.0	% FSO

NOTES:

- 1. For a loaded output the measurements are observed after an RC filter consisting of a 1 kΩ resistor and a 0.1 μF capacitor to ground.
- 2. These limits define the range of operation for which the part will meet specification.
- 3. Within the supply range of 4.75 and 5.25 volts, the device operates as a fully calibrated linear accelerometer. Beyond these supply limits the device may operate as a linear device but is not guaranteed to be in calibration.
- 4. The device can measure both + and acceleration. With no input acceleration the output is at midsupply. For positive acceleration the output will increase above V_{DD}/2 and for negative acceleration the output will decrease below V_{DD}/2.
- 5. Sensitivity limits apply to 0 Hz acceleration.
- 6. At clock frequency \approx 35 kHz.
- 7. The digital input pin has an internal pull-down current source to prevent inadvertent self test initiation due to external board level leakages.
- 8. Time for the output to reach 90% of its final value after a self-test is initiated.
- $9. \ \,$ Time for amplifiers to recover after an acceleration signal causing them to saturate.
- 10. Preserves phase margin (60°) to guarantee output amplifier stability.
- 11. A measure of the device's ability to reject an acceleration applied 90° from the true axis of sensitivity.
- 12. The Status pin output is not valid following power—up until at least one rising edge has been applied to the self—test pin. The Status pin is high whenever the self—test input is high.
- 13. The Status pin output latches high if the EPROM parity changes to odd. The Status pin can be reset by a rising edge on self-test, unless a fault condition continues to exist.

PRINCIPLE OF OPERATION

The Motorola accelerometer is a surface—micromachined integrated—circuit accelerometer.

The device consists of a surface micromachined capacitive sensing cell (g–cell) and a CMOS signal conditioning ASIC contained in a single integrated circuit package. The sensing element is sealed hermetically at the wafer level using a bulk micromachined "cap" wafer.

The g–cell is a mechanical structure formed from semiconductor materials (polysilicon) using semiconductor processes (masking and etching). It can be modeled as two stationary plates with a moveable plate in–between. The center plate can be deflected from its rest position by subjecting the system to an acceleration (Figure 2).

When the center plate deflects, the distance from it to one fixed plate will increase by the same amount that the distance to the other plate decreases. The change in distance is a measure of acceleration.

The g-cell plates form two back-to-back capacitors (Figure 3). As the center plate moves with acceleration, the distance between the plates changes and each capacitor's value will change, (C = A ϵ /D). Where A is the area of the plate, ϵ is the dielectric constant, and D is the distance between the plates.

The CMOS ASIC uses switched capacitor techniques to measure the g-cell capacitors and extract the acceleration data from the difference between the two capacitors. The ASIC also signal conditions and filters (switched capacitor) the signal, providing a high level output voltage that is ratiometric and proportional to acceleration.

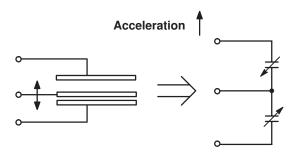


Figure 2. Transducer Physical Model

Figure 3. Equivalent Circuit Model

SPECIAL FEATURES

Filtering

The Motorola accelerometers contain an onboard 2-pole switched capacitor filter. A Bessel implementation is used because it provides a maximally flat delay response (linear phase) thus preserving pulse shape integrity. Because the filter is realized using switched capacitor techniques, there is no requirement for external passive components (resistors and capacitors) to set the cut-off frequency.

Self-Test

The sensor provides a self–test feature that allows the verification of the mechanical and electrical integrity of the accelerometer at any time before or after installation. This feature is critical in applications such as automotive airbag systems where system integrity must be ensured over the life of the vehicle. A fourth "plate" is used in the g–cell as a self–test plate. When the user applies a logic high input to the self–test pin, a calibrated potential is applied across the self–test plate and the moveable plate. The resulting electrostatic force (Fe = $^{1}/_{2}$ AV $^{2}/_{0}$ d²) causes the center plate to deflect. The resultant deflection is measured by the accelerometer's control ASIC and a proportional output voltage results. This procedure assures that both the mechanical (g–cell) and electronic sections of the accelerometer are functioning.

Status

Motorola accelerometers include fault detection circuitry and a fault latch. The Status pin is an output from the fault latch, OR'd with self-test, and is set high whenever the following event occurs:

Parity of the EPROM bits becomes odd in number.

The fault latch can be reset by a rising edge on the selftest input pin, unless one (or more) of the fault conditions continues to exist.

BASIC CONNECTIONS

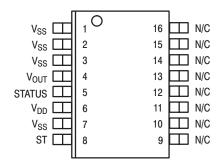


Figure 4. Pinout Description

Pin No.	Pin Name	Description
1 thru 3	V _{SS}	Redundant connections to the internal V _{SS} and may be left unconnected.
4	V _{OUT}	Output voltage of the accelerometer.
5	STATUS	Logic output pin used to indicate fault.
6	V_{DD}	The power supply input.
7	V_{SS}	The power supply ground.
8	ST	Logic input pin used to initiate self–test.
9 thru 13	Trim pins	Used for factory trim. Leave unconnected.
14 thru 16	_	No internal connection. Leave unconnected.

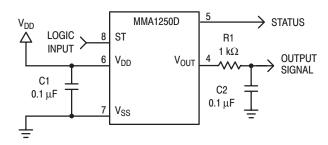


Figure 5. SOIC Accelerometer with Recommended Connection Diagram

PCB Layout

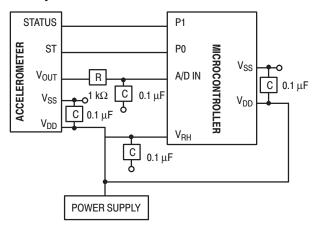
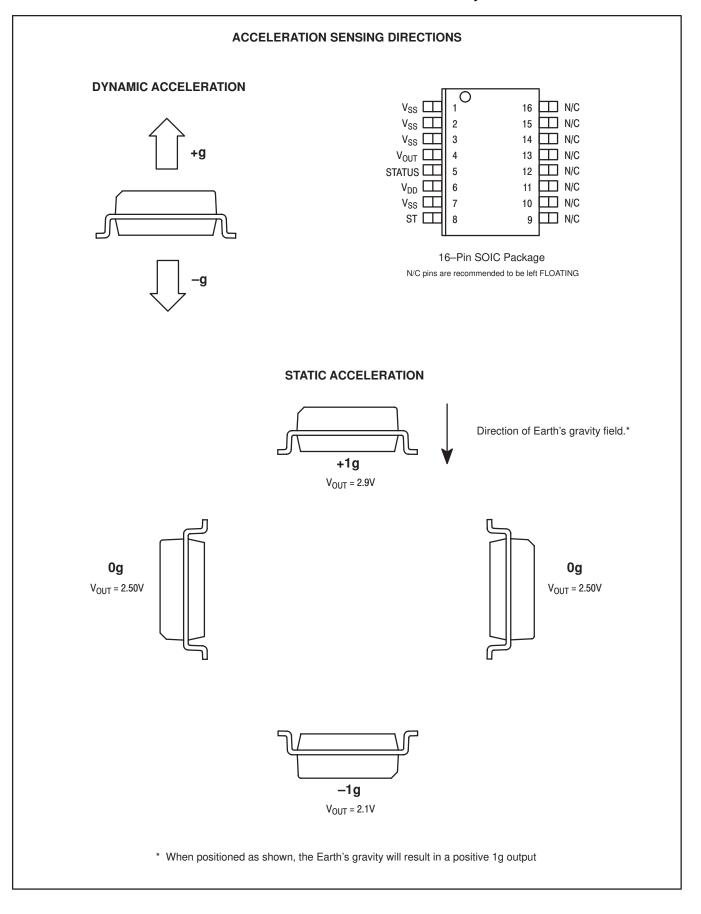


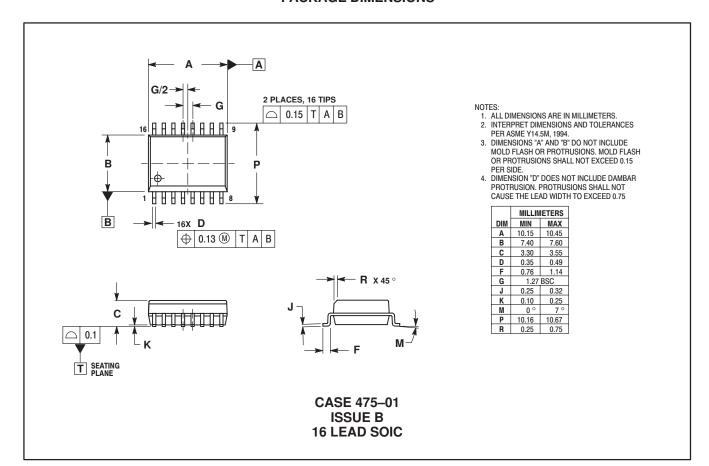
Figure 6. Recommended PCB Layout for Interfacing Accelerometer to Microcontroller

NOTES:

- Use a 0.1 μF capacitor on V_{DD} to decouple the power source.
- Physical coupling distance of the accelerometer to the microcontroller should be minimal.
- Place a ground plane beneath the accelerometer to reduce noise, the ground plane should be attached to all internal V_{SS} terminals shown in Figure 4.
- Use an RC filter of 1 k Ω and 0.1 μ F on the output of the accelerometer to minimize clock noise (from the switched capacitor filter circuit).
- PCB layout of power and ground should not couple power supply noise.
- Accelerometer and microcontroller should not be a high current path.
- A/D sampling rate and any external power supply switching frequency should be selected such that they do not interfere with the internal accelerometer sampling frequency. This will prevent aliasing errors.



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