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# Triple 1.4 GHz Current Feedback Op Amp with Enable Feature 

NCS2535 is a triple 1.4 GHz current feedback monolithic operational amplifier featuring high slew rate and low differential gain and phase error. The current feedback architecture allows for a superior bandwidth and low power consumption. This device features an enable pin.

## Features

- -3.0 dB Small Signal $\mathrm{BW}\left(\mathrm{A}_{\mathrm{V}}=+2.0, \mathrm{~V}_{\mathrm{O}}=0.5 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}\right)$ 1.4 GHz Typ
- Slew Rate $2500 \mathrm{~V} / \mu \mathrm{s}$
- Supply Current 12 mA per Amplifier
- Input Referred Voltage Noise $5.0 \mathrm{nV} / \sqrt{\mathrm{Hz}}$
- THD -69 dB (f = 5.0 MHz, $\left.\mathrm{V}_{\mathrm{O}}=2.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}\right)$
- Output Current 120 mA
- Enable Pin Available
- This is a $\mathrm{Pb}-$ Free Device


## Applications

- High Resolution Video
- Line Driver
- High-Speed Instrumentation
- Wide Dynamic Range IF Amp


Figure 1. Frequency Response: Gain (dB) vs. Frequency $A v=+2.0$


ON Semiconductor ${ }^{\circledR}$
http://onsemi.com


NCS2535 = Specific Device Code
A = Assembly Location
L = Wafer Lot
Y = Year
W = Work Week

- = Pb-Free Package
(Note: Microdot may be in either location)


ORDERING INFORMATION

| Device | Package | Shipping $^{\dagger}$ |
| :--- | :---: | :---: |
| NCS2535DTG | TSSOP-16 | 96 Units/Rail |
| NCS2535DTR2G | TSSOP-16 | 2500 Tape \& Reel |

$\dagger$ For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

[^0]PIN FUNCTION DESCRIPTION

| Pin | Symbol | Function | Equivalent Circuit |
| :---: | :---: | :---: | :---: |
| 9,12, 15 | OUTx | Output |  |
| 3, 6 | $\mathrm{V}_{\text {EE }}$ | Negative Power Supply |  |
| 2, 5, 8 | +INx | Non-inverted Input |  |
| 1, 4, 7 | -INx | Inverted Input | See Above |
| 11, 14 | $\mathrm{V}_{\mathrm{CC}}$ | Positive Power Supply |  |
| 10, 13, 16 | EN | Enable |  |

ENABLE PIN TRUTH TABLE

|  | High | Low* |
| :--- | :---: | :---: |
| Enable | Disabled | Enabled |

*Default open state


Figure 2. Simplified Device Schematic

## ATTRIBUTES

| Characteristics | Value |
| :---: | :---: |
| ESD |  |
| Human Body Model | 2.0 kV (Note 1) |
| Machine Model | 200 V |
| Charged Device Model | 1.0 kV |
| Moisture Sensitivity (Note 2) | Level 1 |
| Flammability Rating Oxygen Index: 28 to 34 | UL $94 \mathrm{~V}-0 @ 0.125 \mathrm{in}$ |

1. 0.8 kV between the input pairs +IN and -IN pins only. All other pins are 2.0 kV .
2. For additional information, see Application Note AND8003/D.

MAXIMUM RATINGS

| Parameter | Symbol | Rating | Unit |
| :---: | :---: | :---: | :---: |
| Power Supply Voltage | $\mathrm{V}_{\mathrm{S}}$ | 11 | Vdc |
| Input Voltage Range | $V_{1}$ | $\leq \mathrm{V}_{\mathrm{S}}$ | Vdc |
| Input Differential Voltage Range | $\mathrm{V}_{\text {ID }}$ | $\leq \mathrm{V}_{\mathrm{S}}$ | Vdc |
| Output Current | $\mathrm{I}_{0}$ | 120 | mA |
| Maximum Junction Temperature (Note 3) | $\mathrm{T}_{J}$ | 150 | ${ }^{\circ} \mathrm{C}$ |
| Operating Ambient Temperature | $\mathrm{T}_{\mathrm{A}}$ | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $\mathrm{T}_{\text {stg }}$ | -60 to +150 | ${ }^{\circ} \mathrm{C}$ |
| Power Dissipation | $\mathrm{P}_{\mathrm{D}}$ | (See Graph) | mW |
| Thermal Resistance, Junction-to-Air | $\mathrm{R}_{\theta \mathrm{JA}}$ | 156 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.
3. Power dissipation must be considered to ensure maximum junction temperature $\left(T_{\mathrm{J}}\right)$ is not exceeded.

## MAXIMUM POWER DISSIPATION

The maximum power that can be safely dissipated is limited by the associated rise in junction temperature. For the plastic packages, the maximum safe junction temperature is $150^{\circ} \mathrm{C}$. If the maximum is exceeded momentarily, proper circuit operation will be restored as soon as the die temperature is reduced. Leaving the device in the "overheated" condition for an extended period can result in device damage. To ensure proper operation, it is important to observe the derating curves.


Figure 3. Power Dissipation vs. Temperature

NCS2535

AC ELECTRICAL CHARACTERISTICS $\left(\mathrm{V}_{\mathrm{CC}}=+5.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{EE}}=-5.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}\right.$ to $+85^{\circ} \mathrm{C}, \mathrm{R}_{\mathrm{L}}=150 \Omega$ to $\mathrm{GND}, \mathrm{R}_{\mathrm{F}}=330 \Omega$, $A_{V}=+2.0$, Enable is left open, unless otherwise specified).

| Symbol | Characteristic | Conditions | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |

## FREQUENCY DOMAIN PERFORMANCE

$\left.\begin{array}{|c|c|c|c|c|c|c|}\hline \text { BW } & \begin{array}{c}\text { Bandwidth } \\ 3.0 \mathrm{~dB} \text { Small Signal } \\ 3.0 \mathrm{~dB} \text { Large Signal }\end{array} & \begin{array}{c}\mathrm{A}_{V}=+2.0, \mathrm{~V}_{\mathrm{O}}=0.5 \mathrm{~V}_{\mathrm{p}-\mathrm{p}} \\ \mathrm{A}_{\mathrm{V}}=+2.0, \mathrm{~V}_{\mathrm{O}}=2.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}\end{array} & & 1400 \\ 650\end{array}\right)$

TIME DOMAIN RESPONSE

| SR | Slew Rate | $A_{V}=+2.0, V_{\text {step }}=2.0 \mathrm{~V}$ |  | 2500 |  | $\mathrm{~V} / \mathrm{us}$ |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{s}}$ | Settling Time <br> $0.1 \%$ | $\mathrm{~A}_{\mathrm{V}}=+2.0, \mathrm{~V}_{\text {step }}=2.0 \mathrm{~V}$ |  | 13 |  | ns |
| $\mathrm{t}_{\mathrm{r}} \mathrm{t}_{\mathrm{f}}$ | Rise and Fall Time | $(10 \%-90 \%) \mathrm{A}_{\mathrm{V}}=+2.0, \mathrm{~V}_{\text {step }}=2.0 \mathrm{~V}$ |  | 1.5 |  | ns |
| $\mathrm{t}_{\mathrm{ON}}$ | Turn-on Time |  |  | 55 |  | ns |
| $\mathrm{t}_{\mathrm{OFF}}$ | Turn-off Time |  |  | 55 | ns |  |

HARMONIC/NOISE PERFORMANCE

| THD | Total Harmonic Distortion | $\mathrm{f}=5.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=2.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | -69 | dB |
| :---: | :--- | :---: | :---: | :---: | :---: |
| HD 2 | 2nd Harmonic Distortion | $\mathrm{f}=5.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=2.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | -73 |  |
| HD 3 | 3rd Harmonic Distortion | $\mathrm{f}=5.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=2.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | -73 | dBc |
| IP3 | Third-Order Intercept | $\mathrm{f}=10 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=1.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | 34 | dBc |
| SFDR | Spurious-Free Dynamic <br> Range | $\mathrm{f}=5.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=2.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | 73 | dBm |
| $\mathrm{e}_{\mathrm{N}}$ | Input Referred Voltage Noise | $\mathrm{f}=1.0 \mathrm{MHz}$ | dBc |  |  |
| $\mathrm{i}_{\mathrm{N}}$ | Input Referred Current Noise | $\mathrm{f}=1.0 \mathrm{MHz}$, Inverting <br> $\mathrm{f}=1.0 \mathrm{MHz}$, Non-Inverting |  | 5.0 |  |

NCS2535

DC ELECTRICAL CHARACTERISTICS $\left(\mathrm{V}_{\mathrm{CC}}=+5.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{EE}}=-5.0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}\right.$ to $+85^{\circ} \mathrm{C}, \mathrm{R}_{\mathrm{L}}=150 \Omega$ to $\mathrm{GND}, \mathrm{R}_{\mathrm{F}}=330 \Omega$,
$A_{V}=+2.0$, Enable is left open, unless otherwise specified).

| Symbol | Characteristic | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |

DC PERFORMANCE

| $\mathrm{V}_{\mathrm{IO}}$ | Input Offset Voltage |  | -10 | 0 | 10 | mV |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\Delta \mathrm{V}_{\mathrm{IO}} / \Delta \mathrm{T}$ | Input Offset Voltage <br> Temperature Coefficient |  | 6.0 |  | $\mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ |  |
| $\mathrm{I}_{\mathrm{IB}}$ | Input Bias Current | +Input (Non-Inverting), $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ <br> -Input (Inverting), $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}($ Note 4) |  | $\pm 3.0$ <br> $\pm 6.0$ | $\pm 35$ <br> $\pm 35$ | $\mu \mathrm{~A}$ |
| $\Delta \mathrm{l}_{\mathrm{IB}} / \Delta \mathrm{T}$ | Input Bias Current <br> Temperature Coefficient | +Input (Non-Inverting), $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ <br> -Input (Inverting), $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ |  | +40 <br> -10 | $\mathrm{nA} /{ }^{\circ} \mathrm{C}$ |  |
| $\mathrm{V}_{\mathrm{IH}}$ | Input High Voltage (Enable) <br> (Note 4) |  | +3.0 |  | V |  |
| $\mathrm{~V}_{\mathrm{IL}}$ | Input Low Voltage (Enable) <br> (Note 4) |  |  | +1.0 | V |  |

INPUT CHARACTERISTICS
$\left.\begin{array}{|c|c|c|c|c|c|}\hline \mathrm{V}_{\mathrm{CM}} & \begin{array}{c}\text { Input Common Mode Voltage } \\ \text { Range (Note 4) }\end{array} & \pm 3.0 & \pm 4.0 & \mathrm{~V} \\ \hline \text { CMRR } & \begin{array}{l}\text { Common Mode Rejection } \\ \text { Ratio }\end{array} & \text { (See Graph) } & 40 & 50 & \\ \hline \mathrm{R}_{\mathrm{IN}} & \text { Input Resistance } & \begin{array}{c}\text { +Input (Non-Inverting) } \\ \text {-Input (Inverting) }\end{array} & & \begin{array}{c}150 \\ 70\end{array} & \mathrm{~dB} \\ \hline \mathrm{C}_{\mathrm{IN}} & \begin{array}{l}\text { Differential Input } \\ \text { Capacitance }\end{array} & & 1.0 & \mathrm{k} \Omega \\ \Omega\end{array}\right]$

OUTPUT CHARACTERISTICS

| $R_{\text {OUT }}$ | Output Resistance | Closed Loop <br> Open Loop |  | 0.1 <br> 13 | $\Omega$ |  |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{~V}_{\mathrm{O}}$ | Output Voltage Range |  | $\pm 3.0$ | $\pm 4.0$ |  | V |
| $\mathrm{I}_{\mathrm{O}}$ | Output Current |  | $\pm 80$ | $\pm 120$ |  | mA |

POWER SUPPLY

| $\mathrm{V}_{\mathrm{S}}$ | Operating Voltage Supply |  | 10 |  | V |  |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{I}_{\mathrm{S}, \mathrm{ON}}$ | Power Supply Current - <br> Enabled per amplifier <br> (Note 4) | $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ | 6.0 | 12 | 18 | mA |
| $\mathrm{I}_{\mathrm{S}, \mathrm{OFF}}$ | Power Supply Current - <br> Disabled per amplifier | $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ | 0.1 | 0.3 | mA |  |
|  | Crosstalk | Channel to Channel, $\mathrm{f}=5.0 \mathrm{MHz}$ |  | 60 |  | dB |
| PSRR | Power Supply Rejection <br> Ratio | (See Graph) | 40 | 55 |  | dB |

4. Guaranteed by design and/or characterization.

NCS2535

AC ELECTRICAL CHARACTERISTICS $\left(\mathrm{V}_{\mathrm{CC}}=+2.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{EE}}=-2.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}\right.$ to $+85^{\circ} \mathrm{C}, \mathrm{R}_{\mathrm{L}}=150 \Omega$ to $\mathrm{GND}, \mathrm{R}_{\mathrm{F}}=330 \Omega$, $A_{V}=+2.0$, Enable is left open, unless otherwise specified).

| Symbol | Characteristic | Conditions | Min | Typ | Max | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |

## FREQUENCY DOMAIN PERFORMANCE

| BW | Bandwidth 3.0 dB Small Signal 3.0 dB Large Signal | $\begin{aligned} & A_{V}=+2.0, V_{O}=0.5 \mathrm{~V}_{\mathrm{p}-\mathrm{p}} \\ & \mathrm{~A}_{\mathrm{V}}=+2.0, \mathrm{~V}_{\mathrm{O}}=1.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}} \end{aligned}$ | $\begin{aligned} & 800 \\ & 450 \end{aligned}$ | MHz |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{GF}_{0.1 \mathrm{~dB}}$ | 0.1 dB Gain Flatness Bandwidth | $A_{V}=+2.0$ | 100 | MHz |
| dG | Differential Gain | $\mathrm{A}_{V}=+2.0, \mathrm{R}_{\mathrm{L}}=150 \Omega, \mathrm{f}=3.58 \mathrm{MHz}$ | 0.02 | \% |
| dP | Differential Phase | $\mathrm{A}_{V}=+2.0, \mathrm{R}_{\mathrm{L}}=150 \Omega, \mathrm{f}=3.58 \mathrm{MHz}$ | 0.02 | 。 |

TIME DOMAIN RESPONSE

| SR | Slew Rate | $\mathrm{A}_{\mathrm{V}}=+2.0, \mathrm{~V}_{\text {step }}=1.0 \mathrm{~V}$ |  | 1500 |  | $\mathrm{~V} / \mathrm{us}$ |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{s}}$ | Settling Time <br> $0.1 \%$ | $\mathrm{~A}_{\mathrm{V}}=+2.0, \mathrm{~V}_{\text {step }}=1.0 \mathrm{~V}$ |  | 10 |  | ns |
| $\mathrm{t}_{\mathrm{r}} \mathrm{t}_{\mathrm{f}}$ | Rise and Fall Time | $(10 \%-90 \%) \mathrm{A}_{\mathrm{V}}=+2.0, \mathrm{~V}_{\text {step }}=1.0 \mathrm{~V}$ |  | 1.2 |  | ns |
| $\mathrm{t}_{\mathrm{ON}}$ | Turn-on Time |  |  | 55 |  | ns |
| $\mathrm{t}_{\mathrm{OFF}}$ | Turn-off Time |  |  | 55 |  | ns |

HARMONIC/NOISE PERFORMANCE

| THD | Total Harmonic Distortion | $\mathrm{f}=5.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=1.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | -58 | dB |
| :---: | :--- | :---: | :---: | :---: | :---: |
| HD 2 | 2nd Harmonic Distortion | $\mathrm{f}=5.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=1.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | -61 | dBc |
| HD 3 | 3rd Harmonic Distortion | $\mathrm{f}=5.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=1.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | -64 |  |
| IP3 | Third-Order Intercept | $\mathrm{f}=10 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=0.5 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | 28 | dBc |
| SFDR | Spurious-Free Dynamic <br> Range | $\mathrm{f}=5.0 \mathrm{MHz}, \mathrm{V}_{\mathrm{O}}=1.0 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}$ |  | 61 | dBm |
| $\mathrm{e}_{\mathrm{N}}$ | Input Referred Voltage Noise | $\mathrm{f}=1.0 \mathrm{MHz}$ | dBc |  |  |
| $\mathrm{i}_{\mathrm{N}}$ | Input Referred Current Noise | $\mathrm{f}=1.0 \mathrm{MHz}$, Inverting <br> $\mathrm{f}=1.0 \mathrm{MHz}$, Non-Inverting |  | 5.0 |  |

DC ELECTRICAL CHARACTERISTICS ( $\mathrm{V}_{\mathrm{CC}}=+2.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{EE}}=-2.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}, \mathrm{R}_{\mathrm{L}}=150 \Omega$ to $\mathrm{GND}, \mathrm{R}_{\mathrm{F}}=330 \Omega$,
$A_{V}=+2.0$, Enable is left open, unless otherwise specified).

| Symbol | Characteristic | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |

DC PERFORMANCE

| $\mathrm{V}_{\mathrm{IO}}$ | Input Offset Voltage |  | -10 | 0 | +10 | mV |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\Delta \mathrm{V}_{\mathrm{IO}} / \Delta \mathrm{T}$ | Input Offset Voltage <br> Temperature Coefficient |  | 6.0 |  | $\mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ |  |
| $\mathrm{I}_{\mathrm{IB}}$ | Input Bias Current | +Input (Non-Inverting), $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ <br> -Input (Inverting), $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}($ (Note 5$)$ |  | $\pm 3.0$ <br> $\pm 6.0$ | $\pm 35$ <br> $\pm 35$ | $\mu \mathrm{~A}$ |
| $\Delta \mathrm{I}_{\mathrm{IB}} / \Delta \mathrm{T}$ | Input Bias Current Temperature <br> Coefficient | +Input (Non-Inverting), $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ <br> -Input (Inverting), $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ |  | +40 <br> -10 |  | $\mathrm{nA} /{ }^{\circ} \mathrm{C}$ |
| $\mathrm{V}_{\mathrm{IH}}$ | Input High Voltage (Enable) <br> (Note 5) |  | +1.5 |  |  | V |
| $\mathrm{~V}_{\mathrm{IL}}$ | Input Low Voltage (Enable) <br> (Note 5) |  |  | +0.5 | V |  |

INPUT CHARACTERISTICS

| $\mathrm{V}_{\mathrm{CM}}$ | Input Common Mode Voltage <br> Range (Note 5) | $\pm 1.1$ | $\pm 1.5$ | V |  |  |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| CMRR | Common Mode Rejection Ratio | (See Graph) | 40 | 50 |  | dB |
| $\mathrm{R}_{\mathrm{IN}}$ | Input Resistance | +Input (Non-Inverting) <br> -Input (Inverting) |  | 150 |  | $\mathrm{k} \Omega$ |
| $\mathrm{C}_{\mathrm{IN}}$ | Differential Input Capacitance |  |  | 1.0 | pF |  |

OUTPUT CHARACTERISTICS

| $R_{\text {OUT }}$ | Output Resistance | Closed Loop <br> Open Loop |  | 0.1 <br> 13 |  | $\Omega$ |
| :---: | :--- | :--- | :--- | :--- | :---: | :---: |
| $\mathrm{~V}_{\mathrm{O}}$ | Output Voltage Range |  | $\pm 1.1$ | $\pm 1.5$ |  | V |
| $\mathrm{I}_{\mathrm{O}}$ | Output Current |  | $\pm 80$ | $\pm 120$ |  | mA |

POWER SUPPLY

| $\mathrm{V}_{\mathrm{S}}$ | Operating Voltage Supply |  | 5.0 |  | V |  |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{I}_{\mathrm{S}, \mathrm{ON}}$ | Power Supply Current - Enabled <br> per amplifier (Note 5) | $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ | 6.0 | 11 | 18 | mA |
| $\mathrm{I}_{\mathrm{S}, \mathrm{OFF}}$ | Power Supply Current - <br> Disabled per amplifier | $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ | 0.09 | 0.3 | mA |  |
|  | Crosstalk | Channel to Channel, $\mathrm{f}=5.0 \mathrm{MHz}$ |  | 60 |  | dB |
| PSRR | Power Supply Rejection Ratio | (See Graph) | 40 | 55 |  | dB |

5. Guaranteed by design and/or characterization.


Figure 4. Typical Test Setup
$\left(A_{V}=+2.0, R_{F}=330 \Omega, R_{L}=150 \Omega\right)$


Figure 5. Frequency Response:
Gain (dB) vs. Frequency

$$
A_{V}=+2.0
$$



Figure 7. Large Signal Frequency Response Gain (dB) vs. Frequency


Figure 9. Small Signal Step Response Vertical: $\mathbf{5 0 0} \mathbf{~ m V / d i v}$ Horizontal: 10 ns/div


Figure 6. Frequency Response:
Gain (dB) vs. Frequency $A_{V}=+1.0$


Figure 8. Small Signal Frequency Response Gain (dB) vs. Frequency


Figure 10. Large Signal Step Response Vertical: $2 \mathrm{~V} / \mathrm{div}$
Horizontal: $10 \mathrm{~ns} / \mathrm{div}$


Figure 11. THD, HD2, HD3 vs. Frequency


Figure 13. Input Referred Voltage Noise vs. Frequency


Figure 15. PSRR vs. Frequency


Figure 12. THD, HD2, HD3 vs. Frequency


Figure 14. CMRR vs. Frequency


Figure 16. Differential Gain


Figure 17. Differential Phase


Figure 19. Supply Current per Amplifier vs. Temperature (Disabled)


Figure 18. Supply Current per Amplifier vs. Power Supply (Enabled)


Figure 20. Output Voltage Swing vs. Supply Voltage


Figure 22. Closed Loop Output Resistance vs. Frequency


Figure 23. Frequency Response vs. Capacitive Load


Figure 24. Turn ON Time Delay
Vertical: (EN) $500 \mathrm{mV} / \mathrm{div}$ (OUT) $1 \mathrm{~V} / \mathrm{div}$ Horizontal: 40ns/div


Figure 26. Crosstalk (dBc) vs. Frequency (Crosstalk measured on Channel 2 with input signal on Channel 1 and 3)


Figure 25. Turn OFF Time Delay Vertical: (EN) $500 \mathrm{mV} / \mathrm{div}$ (OUT) 1V/div Horizontal: 40ns/div


Figure 27. Channel Matching vs. Frequency

## General Design Considerations

The current feedback amplifier is optimized for use in high performance video and data acquisition systems. For current feedback architecture, its closed-loop bandwidth depends on the value of the feedback resistor. The closed-loop bandwidth is not a strong function of gain, as is for a voltage feedback amplifier, as shown in Figure 28.


Figure 28. Frequency Response vs. $\mathbf{R}_{\mathbf{F}}$
The -3.0 dB bandwidth is, to some extent, dependent on the power supply voltages. By using lower power supplies, the bandwidth is reduced, because the internal capacitance increases. Smaller values of feedback resistor can be used at lower supply voltages, to compensate for this affect.

## Feedback and Gain Resistor Selection for Optimum Frequency Response

A current feedback operational amplifier's key advantage is the ability to maintain optimum frequency response independent of gain by using appropriate values for the feedback resistor. To obtain a very flat gain response, the feedback resistor tolerance should be considered as well. Resistor tolerance of $1 \%$ should be used for optimum flatness. Normally, lowering RF resistor from its recommended value will peak the frequency response and extend the bandwidth while increasing the value of RF resistor will cause the frequency response to roll off faster. Reducing the value of RF resistor too far below its recommended value will cause overshoot, ringing, and eventually oscillation.

Since each application is slightly different, it is worth some experimentation to find the optimal RF for a given circuit. A value of the feedback resistor that produces $\sim 0.1 \mathrm{~dB}$ of peaking is the best compromise between stability and maximal bandwidth. It is not recommended to
use a current feedback amplifier with the output shorted directly to the inverting input.

## Printed Circuit Board Layout Techniques

Proper high speed PCB design rules should be used for all wideband amplifiers as the PCB parasitics can affect the overall performance. Most important are stray capacitances at the output and inverting input nodes as it can effect peaking and bandwidth. A space ( $3 / 16^{\prime \prime}$ is plenty) should be left around the signal lines to minimize coupling. Also, signal lines connecting the feedback and gain resistors should be short enough so that their associated inductance does not cause high frequency gain errors. Line lengths less than $1 / 4^{\prime \prime}$ are recommended.

## Video Performance

This device designed to provide good performance with NTSC, PAL, and HDTV video signals. Best performance is obtained with back terminated loads as performance is degraded as the load is increased. The back termination reduces reflections from the transmission line and effectively masks transmission line and other parasitic capacitances from the amplifier output stage.

## ESD Protection

All device pins have limited ESD protection using internal diodes to power supplies as specified in the attributes table (see Figure 29). These diodes provide moderate protection to input overdrive voltages above the supplies. The ESD diodes can support high input currents with current limiting series resistors. Keep these resistor values as low as possible since high values degrade both noise performance and frequency response. Under closed-loop operation, the ESD diodes have no effect on circuit performance. However, under certain conditions the ESD diodes will be evident. If the device is driven into a slewing condition, the ESD diodes will clamp large differential voltages until the feedback loop restores closed-loop operation. Also, if the device is powered down and a large input signal is applied, the ESD diodes will conduct.
NOTE: Human Body Model for +IN and -IN pins are rated at 0.8 kV while all other pins are rated at 2.0 kV .


Figure 29. Internal ESD Protection

## PACKAGE DIMENSIONS

TSSOP-16
CASE 948F-01
ISSUE A


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