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## Product Preview

# 300 mA CMOS Low Dropout Regulator

The NCV8114 is 300 mA LDO that provides the engineer with a very stable, accurate voltage with low noise suitable for space constrained, noise sensitive applications. In order to optimize performance for battery operated portable applications, the NCV8114 employs the dynamic quiescent current adjustment for very low  $I_Q$  consumption at no–load.

#### **Features**

- Operating Input Voltage Range: 1.7 V to 5.5 V
- Available in Fixed Voltage Options: 0.9 V to 3.6 V Contact Factory for Other Voltage Options
- Very Low Quiescent Current of Typ. 50 μA
- Standby Current Consumption: Typ. 0.1 μA
- Low Dropout: 135 mV Typical at 300 mA
- ±1% Accuracy at Room Temperature
- High Power Supply Ripple Rejection: 75 dB at 1 kHz
- Thermal Shutdown and Current Limit Protections
- Stable with a 1 μF Ceramic Output Capacitor
- Available in TSOP Package
- NCV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC-Q100 Qualified and PPAP Capable
- These Devices are Pb–Free, Halogen Free/BFR Free and are RoHS Compliant

#### **Typical Applicaitons**

- Parking Camera Modules
- Wireless Handsets, Wireless LAN, Bluetooth<sup>®</sup>, Zigbee<sup>®</sup>
- Automotive Infotainment Systems
- Other Battery Powered Applications

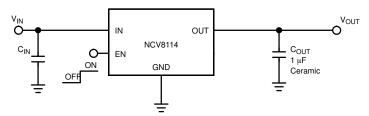


Figure 1. Typical Application Schematic

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### DIAGRAM 5☐ ☐



TSOP-5 SN SUFFIX CASE 483



**MARKING** 

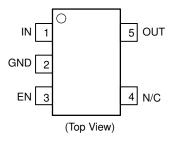
XXX = Device Code

M = Date Code\*

= Pb-Free Package

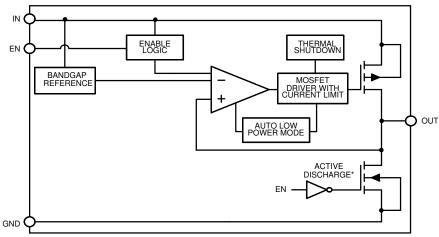
(Note: Microdot may be in either location)
\*Date Code orientation and/or position may vary depending upon manufacturing location.

#### **PIN CONNECTIONS**



#### ORDERING INFORMATION

See detailed ordering, marking and shipping information on page 11 of this data sheet.



\*Active output discharge function is present only in NCV8114ASNyyyTCG devices. yyy denotes the particular V<sub>OUT</sub> option.

Figure 2. Simplified Schematic Block Diagram

#### PIN FUNCTION DESCRIPTION

Pin No.	Pin Name	Description		
5	OUT	Regulated output voltage pin. A small ceramic capacitor with minimum value of 1 $\mu$ F is needed from this pin to ground to assure stability.		
2	GND	Power supply ground.		
3	EN	Driving EN over 0.9 V turns on the regulator. Driving EN below 0.4 V puts the regulator into shutdown mode.		
1	IN	Input pin. A small capacitor is needed from this pin to ground to assure stability.		
4	N/C	Not connected. This pin can be tied to ground to improve thermal dissipation.		

#### **ABSOLUTE MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Input Voltage (Note 1)	V <sub>IN</sub>	−0.3 V to 6 V	V
Output Voltage	Vout	-0.3 V to VIN + 0.3 V or 6 V	V
Enable Input	VEN	-0.3 V to VIN + 0.3 V or 6 V	V
Output Short Circuit Duration	tsc	∞	s
Maximum Junction Temperature	$T_{J(MAX)}$	150	°C
Operating Ambient Temperature	T <sub>A</sub>	-40 to 125	°C
Storage Temperature	T <sub>STG</sub>	-55 to 150	°C
ESD Capability, Human Body Model (Note 2)	ESD <sub>HBM</sub>	2000	V
ESD Capability, Machine Model (Note 2)	ESD <sub>MM</sub>	200	V

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

- 1. Refer to ELECTRICAL CHĂRACTERISTICS and APPLICATION INFORMATION for Safe Operating Area.
- 2. This device series incorporates ESD protection and is tested by the following methods:
  - ESD Human Body Model tested per EIA/JESD22-A114, ESD Machine Model tested per EIA/JESD22-A115,

  - Latchup Current Maximum Rating tested per JEDEC standard: JESD78.

#### THERMAL CHARACTERISTICS (Note 3)

Rating	Symbol	Value	Unit
Thermal Characteristics, TSOP-5 Thermal Resistance, Junction-to-Air	$R_{ heta JA}$	259.9	°C/W

<sup>3.</sup> Single component mounted on 1 oz, FR 4 PCB with 645 mm<sup>2</sup> Cu area.

#### RECOMMENDED OPERATING CONDITIONS

Rating	Symbol	Min	Тур	Max	Unit
Input Voltage	$V_{IN}$	1.7		5.5	V
Junction Temperature		-40		+125	°C

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

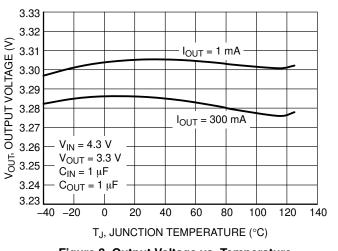
**ELECTRICAL CHARACTERISTICS**  $-40^{\circ}\text{C} \le T_J \le 125^{\circ}\text{C}$ ;  $V_{IN} = V_{OUT(NOM)} + 1$  V for  $V_{OUT}$  options greater than 1.5 V. Otherwise  $V_{IN} = 2.5$  V, whichever is greater;  $I_{OUT} = 1$  mA,  $C_{IN} = C_{OUT} = 1$   $\mu\text{F}$ , unless otherwise noted.  $V_{EN} = 0.9$  V. Typical values are at  $T_J = +25^{\circ}\text{C}$ . Min./Max. are for  $T_J = -40^{\circ}\text{C}$  and  $T_J = +125^{\circ}\text{C}$  respectively (Note 4).

Parameter	Test Condition	Test Conditions		Min	Тур	Max	Unit
Operating Input Voltage			V <sub>IN</sub>	1.7		5.5	V
		V <sub>OUT</sub> ≤ 2.0 V	V <sub>OUT</sub>	-40		+50	mV
Output Voltage Accuracy	$-40$ °C $\leq$ T <sub>J</sub> $\leq$ 125°C	V <sub>OUT</sub> > 2.0 V		-2		+3	%
Line Regulation	Vout + 0.5 V ≤ Vin ≤ 5.5	V (V <sub>IN</sub> ≥ 1.7 V)	Reg <sub>LINE</sub>		0.01	0.1	%/V
Load Regulation	IOUT = 1 mA to 30	IOUT = 1 mA to 300 mA			28	45	mV
Load Transient	$I_{OUT}$ = 1 mA to 300 mA or in 1 $\mu$ s, $C_{OUT}$ =		Tran <sub>LOAD</sub>		-50/ +30		mV
		V <sub>OUT</sub> = 1.5 V	V <sub>DO</sub>		380	500	mV
		V <sub>OUT</sub> = 1.85 V			260	370	
December 1 Vallage (Nata 5)	000 4	V <sub>OUT</sub> = 2.8 V			170	270	
Dropout Voltage (Note 5)	I <sub>OUT</sub> = 300 mA	V <sub>OUT</sub> = 3.0 V			160	260	
		V <sub>OUT</sub> = 3.1 V			155	250	
		V <sub>OUT</sub> = 3.3 V			150	240	
Output Current Limit	V <sub>OUT</sub> = 90% V <sub>OUT(nom)</sub>		I <sub>CL</sub>	300	600		mA
Ground Current	ent Iout = 0 mA		ΙQ		50	95	μΑ
Shutdown Current $VEN \le 0.4 \text{ V}, VIN = 5.5 \text{ V}$		= 5.5 V	I <sub>DIS</sub>		0.01	1	μΑ
EN Pin Threshold Voltage High Threshold Low Threshold	V <sub>EN</sub> Voltage increasing V <sub>EN</sub> Voltage decreasing		V <sub>EN_HI</sub> V <sub>EN_LO</sub>	0.9		0.4	V
EN Pin Input Current	VEN = 5.5 V		I <sub>EN</sub>		0.3	1.0	μΑ
Power Supply Rejection Ratio	V <sub>IN</sub> = 4.3 V, V <sub>OUT</sub> = 3.3 V I <sub>OUT</sub> = 10 mA	f = 1 kHz	PSRR		75		dB
Output Noise Voltage	tput Noise Voltage $V_{\text{IN}} = 2.5 \text{ V}, V_{\text{OUT}} = 1.8 \text{ V}, I_{\text{OUT}} = 150 \text{ mA}$ $f = 10 \text{ Hz to } 100 \text{ kHz}$		V <sub>N</sub>		70		$\mu V_{rms}$
Thermal Shutdown Temperature	Temperature increasing fr	Temperature increasing from T <sub>J</sub> = +25°C			160		°C
Thermal Shutdown Hysteresis	rmal Shutdown Hysteresis Temperature falling from T <sub>SD</sub>		T <sub>SDH</sub>		20		°C
Active Output Discharge Resistance	VEN < 0.4 V, Versio	VEN < 0.4 V, Version A only			100		Ω

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

<sup>4.</sup> Performance guaranteed over the indicated operating temperature range by design and/or characterization. Production tested at T<sub>J</sub> = T<sub>A</sub> = 25°C. Low duty cycle pulse techniques are used during testing to maintain the junction temperature as close to ambient as possible.

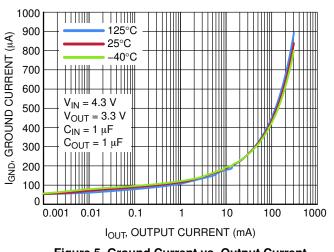
<sup>5.</sup> Characterized when Vout falls 100 mV below the regulated voltage at Vin = Vout(NOM) + 1 V.



100  $V_{OUT} = 3.3 V$ 90 QUIESCENT CURRENT (µA)  $I_{OUT} = 0 \text{ mA}$ 80  $C_{IN}=1\;\mu\text{F}$ 70  $C_{OUT} = 1 \mu F$  $T_{.1} = -40^{\circ}C$ 60  $T_J = 125^{\circ}C$ 50 40  $T_J = 25^{\circ}C$ 30 20 ĝ 10 0 2 3 4 0 5 V<sub>IN</sub>, INPUT VOLTAGE (V)

Figure 3. Output Voltage vs. Temperature - $V_{OUT} = 3.3 V$ 

Figure 4. Quiescent Current vs. Input Voltage



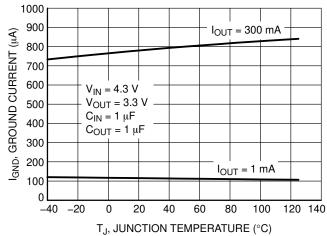
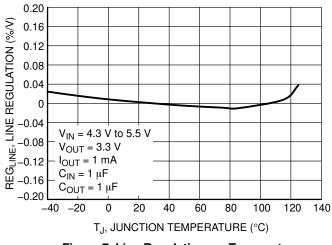


Figure 5. Ground Current vs. Output Current

Figure 6. Ground Current vs. Temperature



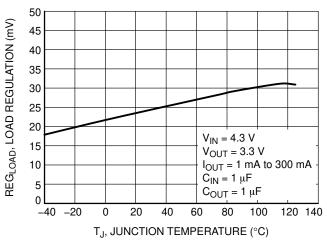


Figure 7. Line Regulation vs. Temperature

Figure 8. Load Regulation vs. Temperature

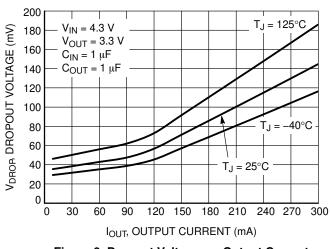


Figure 9. Dropout Voltage vs. Output Current

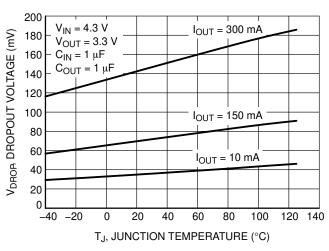


Figure 10. Dropout Voltage vs. Temperature

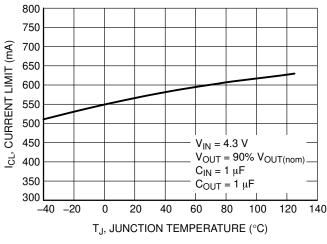


Figure 11. Current Limit vs. Temperature

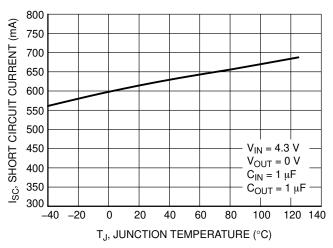


Figure 12. Short Circuit Current vs. Temperature

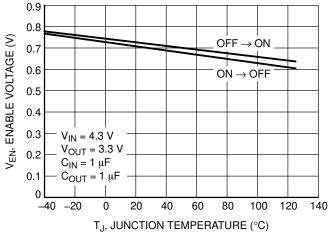


Figure 13. Enable Voltage Threshold vs.
Temperature

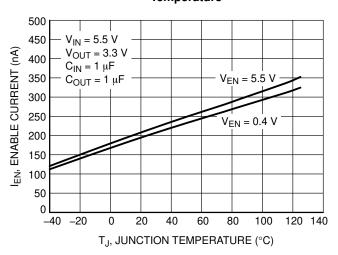
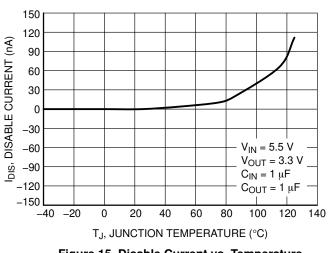


Figure 14. Current to Enable Pin vs. Temperature

#### TYPICAL CHARACTERISTICS

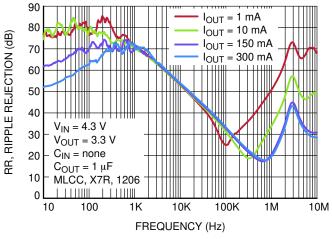
150



135 R<sub>DIS</sub>, DISCHARGE RESISTANCE 120 105 90 75 60  $V_{IN} = 4.3 V$ 45  $V_{EN} = 0 V$ 30  $C_{IN}=1\ \mu F$ 15  $C_{OUT} = 1 \mu F$ -20 20 40 80 120 140 -40 60 100 T<sub>J</sub>, JUNCTION TEMPERATURE (°C)

Figure 15. Disable Current vs. Temperature

Figure 16. Discharge Resistance vs. Temperature



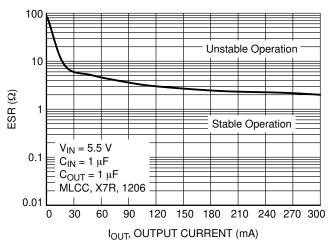
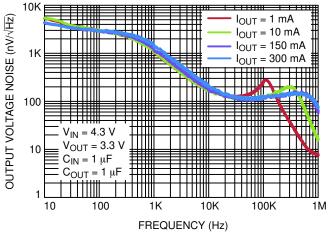


Figure 17. Power Supply Rejection Ratio –  $C_{OUT} = 1 \mu F$ 

Figure 18. Output Capacitor ESR vs. Output Current



	RMS Output Noise (μV)			
lout	10 Hz – 100 kHz	100 Hz – 100 kHz		
1 mA	90.25	83.61		
10 mA	84.55	77.23		
150 mA	86.57	80.86		
300 mA	95.36	90.17		

Figure 19. Output Voltage Noise Spectral Density –  $C_{OUT}$  = 1  $\mu F$ 

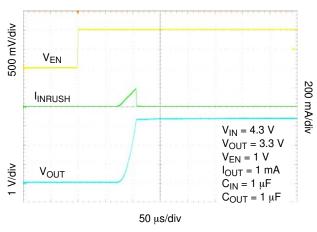


Figure 20. Enable Turn-on Response –  $C_{OUT}$  = 1  $\mu$ F,  $I_{OUT}$  = 1 mA

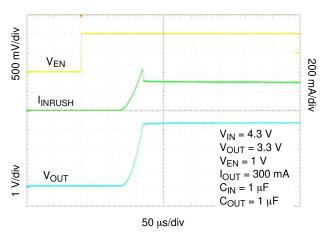


Figure 21. Enable Turn-on Response –  $C_{OUT}$  = 1  $\mu$ F,  $I_{OUT}$  = 300 mA

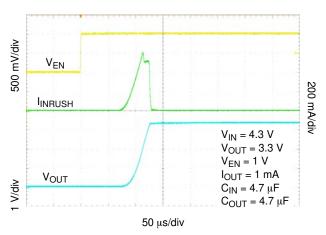


Figure 22. Enable Turn-on Response –  $C_{OUT}$  = 4.7  $\mu$ F,  $I_{OUT}$  = 1 mA

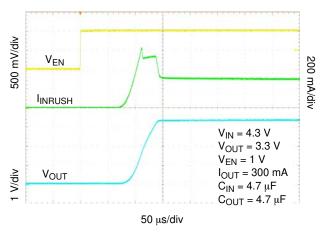


Figure 23. Enable Turn-on Response –  $C_{OUT}$  = 4.7  $\mu$ F,  $I_{OUT}$  = 300 mA

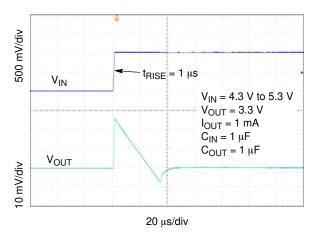


Figure 24. Line Transient Response – Rising Edge, I<sub>OUT</sub> = 1 mA

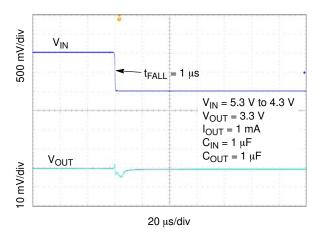


Figure 25. Line Transient Response – Falling Edge, I<sub>OUT</sub> = 1 mA

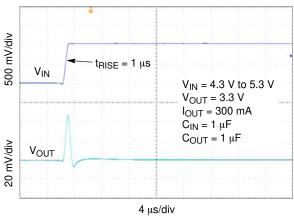


Figure 26. Line Transient Response – Rising Edge, I<sub>OUT</sub> = 300 mA

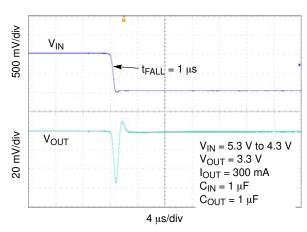


Figure 27. Line Transient Response – Falling Edge, I<sub>OUT</sub> = 300 mA

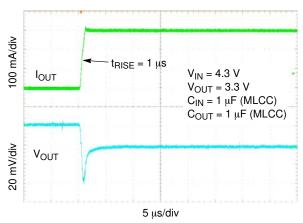


Figure 28. Load Transient Response – Rising Edge, V<sub>OUT</sub> = 3.3 V, I<sub>OUT</sub> = 1 mA to 300 mA

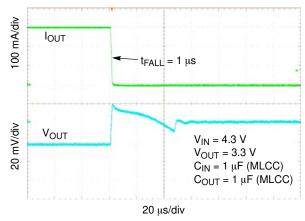


Figure 29. Load Transient Response – Falling Edge, V<sub>OUT</sub> = 3.3 V, I<sub>OUT</sub> = 1 mA to 300 mA

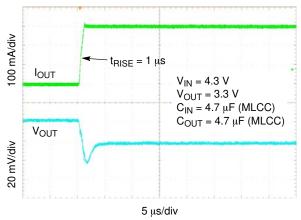


Figure 30. Load Transient Response – Rising Edge, V<sub>OUT</sub> = 3.3 V, I<sub>OUT</sub> = 1 mA to 300 mA

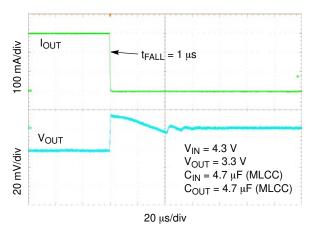


Figure 31. Load Transient Response – Falling Edge, V<sub>OUT</sub> = 3.3 V, I<sub>OUT</sub> = 1 mA to 300 mA

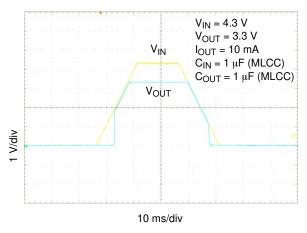


Figure 32. Turn-on/off - Slow Rising V<sub>IN</sub>

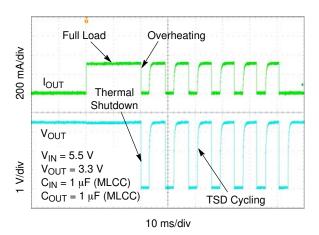


Figure 33. Short Circuit and Thermal Shutdown

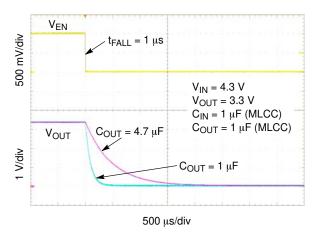
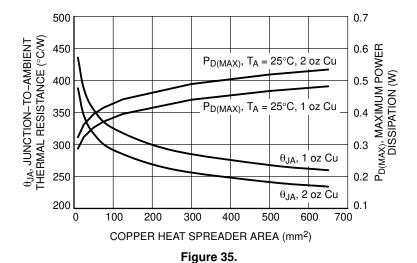


Figure 34. Enable Turn-off



#### **APPLICATIONS INFORMATION**

#### General

The NCV8114 is a high performance 300 mA Low Dropout Linear Regulator. This device delivers very high PSRR (over 75 dB at 1 kHz) and excellent dynamic performance as load/line transients. In connection with very low quiescent current this device is very suitable for various battery powered applications such as tablets, cellular phones, wireless and many others. The device is fully protected in case of output overload, output short circuit condition and overheating, assuring a very robust design.

#### Input Capacitor Selection (CIN)

It is recommended to connect at least a 1  $\mu$ F Ceramic X5R or X7R capacitor as close as possible to the IN pin of the device. This capacitor will provide a low impedance path for unwanted AC signals or noise modulated onto constant input voltage. There is no requirement for the min. /max. ESR of the input capacitor but it is recommended to use ceramic capacitors for their low ESR and ESL. A good input capacitor will limit the influence of input trace inductance and source resistance during sudden load current changes. Larger input capacitor may be necessary if fast and large load transients are encountered in the application.

#### Output Decoupling (COUT)

The NCV8114 requires an output capacitor connected as close as possible to the output pin of the regulator. The recommended capacitor value is 1  $\mu F$  and X7R or X5R dielectric due to its low capacitance variations over the specified temperature range. The NCV8114 is designed to remain stable with minimum effective capacitance of 0.22 $\mu F$  to account for changes with temperature, DC bias and package size. Especially for small package size capacitors such as 0402 the effective capacitance drops rapidly with the applied DC bias.

There is no requirement for the minimum value of Equivalent Series Resistance (ESR) for the  $C_{OUT}$  but the maximum value of ESR should be less than 2  $\Omega$ . Larger output capacitors and lower ESR could improve the load transient response or high frequency PSRR. It is not recommended to use tantalum capacitors on the output due to their large ESR. The equivalent series resistance of tantalum capacitors is also strongly dependent on the temperature, increasing at low temperature.

#### **Enable Operation**

The NCV8114 uses the EN pin to enable/disable its device and to deactivate/activate the active discharge function.

If the EN pin voltage is <0.4 V the device is guaranteed to be disabled. The pass transistor is turned—off so that there is virtually no current flow between the IN and OUT. The active discharge transistor is active so that the output voltage  $V_{OUT}$  is pulled to GND through a 100  $\Omega$  resistor. In the

disable state the device consumes as low as typ. 10 nA from the  $V_{\rm IN}$ .

If the EN pin voltage >0.9 V the device is guaranteed to be enabled. The NCV8114 regulates the output voltage and the active discharge transistor is turned–off.

The EN pin has internal pull-down current source with typ. value of 300 nA which assures that the device is turned-off when the EN pin is not connected. In the case where the EN function isn't required the EN should be tied directly to IN.

#### **Output Current Limit**

Output Current is internally limited within the IC to a typical 600 mA. The NCV8114 will source this amount of current measured with a voltage drops on the 90% of the nominal  $V_{OUT}$ . If the Output Voltage is directly shorted to ground ( $V_{OUT} = 0$  V), the short circuit protection will limit the output current to 630 mA (typ). The current limit and short circuit protection will work properly over whole temperature range and also input voltage range. There is no limitation for the short circuit duration.

#### **Thermal Shutdown**

When the die temperature exceeds the Thermal Shutdown threshold ( $T_{SD}$  – 160°C typical), Thermal Shutdown event is detected and the device is disabled. The IC will remain in this state until the die temperature decreases below the Thermal Shutdown Reset threshold ( $T_{SDU}$  – 140°C typical). Once the IC temperature falls below the 140°C the LDO is enabled again. The thermal shutdown feature provides the protection from a catastrophic device failure due to accidental overheating. This protection is not intended to be used as a substitute for proper heat sinking.

#### **Power Dissipation**

As power dissipated in the NCV8114 increases, it might become necessary to provide some thermal relief. The maximum power dissipation supported by the device is dependent upon board design and layout. Mounting pad configuration on the PCB, the board material, and the ambient temperature affect the rate of junction temperature rise for the part. For reliable operation, junction temperature should be limited to +125°C.

The maximum power dissipation the NCV8114 can handle is given by:

$$P_{D(MAX)} = \frac{\left[125^{\circ}C - T_{A}\right]}{\theta_{JA}}$$
 (eq. 1)

The power dissipated by the NCV8114 for given application conditions can be calculated from the following equations:

$$P_D \approx V_{IN} \! \! \left( I_{GND} \! \! @ I_{OUT} \right) + I_{OUT} \! \! \left( V_{IN} - V_{OUT} \right) \hspace{0.5cm} \text{(eq. 2)} \label{eq:PD}$$

#### **Reverse Current**

The PMOS pass transistor has an inherent body diode which will be forward biased in the case that  $V_{OUT} > V_{IN}$ . Due to this fact in cases, where the extended reverse current condition can be anticipated the device may require additional external protection.

#### **Power Supply Rejection Ratio**

The NCV8114 features very good Power Supply Rejection ratio. If desired the PSRR at higher frequencies in the range 100 kHz - 10 MHz can be tuned by the selection of  $C_{OUT}$  capacitor and proper PCB layout.

#### Turn-On Time

The turn–on time is defined as the time period from EN assertion to the point in which  $V_{OUT}$  will reach 98% of its

nominal value. This time is dependent on various application conditions such as  $V_{OUT(NOM)}$ ,  $C_{OUT}$  and  $T_A$ . For example typical value for  $V_{OUT}$  = 1.2 V,  $C_{OUT}$  = 1  $\mu$ F,  $I_{OUT}$  = 1 mA and  $T_A$  = 25°C is 90  $\mu$ s.

#### **PCB Layout Recommendations**

To obtain good transient performance and good regulation characteristics place  $C_{\rm IN}$  and  $C_{\rm OUT}$  capacitors close to the device pins and make the PCB traces wide. In order to minimize the solution size, use 0402 capacitors. Larger copper area connected to the pins will also improve the device thermal resistance. The actual power dissipation can be calculated from the equation above (Equation 2). Expose pad should be tied the shortest path to the GND pin.

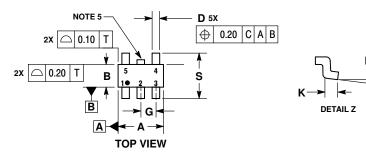
#### **ORDERING INFORMATION**

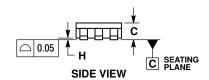
Device	Voltage Option	Marking	Option	Package	Shipping <sup>†</sup>
NCV8114ASN120T1G	1.2 V	DEC			
NCV8114ASN150T1G	1.5 V	DED			
NCV8114ASN180T1G	1.8 V	DEE	With output active		
NCV8114ASN280T1G	2.8 V	DEF	discharge function		
NCV8114ASN300T1G	3.0 V	DEG			
NCV8114ASN330T1G	3.3 V	DEA		TSOP-5	3000 / Tape & Reel (Contact sales office for availability)
NCV8114BSN120T1G	1.2 V	DFC		(Pb-Free)	
NCV8114BSN150T1G	1.5 V	DFD			avallability)
NCV8114BSN180T1G	1.8 V	DFE	Without output active		
NCV8114BSN280T1G	2.8 V	DFF	discharge function		
NCV8114BSN300T1G	3.0 V	DFG			
NCV8114BSN330T1G	3.3 V	DFA			

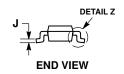
<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

#### PACKAGE DIMENSIONS

#### TSOP-5 CASE 483-02 **ISSUE L**





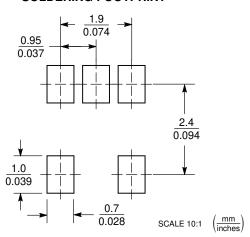


- NOTES:
  1. DIMENSIONING AND TOLERANCING PER ASME
- DIMENSIONING AND TOLERANCING FER ASME Y14.5M, 1994. CONTROLLING DIMENSION: MILLIMETERS. MAXIMUM LEAD THICKNESS INCLUDES LEAD FINISH THICKNESS. MINIMUM LEAD THICKNESS IS THE
- I HICKNESS. MINIMOM LEAD I HICKNESS IS THE MINIMOM THICKNESS OF BASE MATERIAL.

  DIMENSIONS A AND B DO NOT INCLUDE MOLD FLASH, PROTRUSIONS, OR GATE BURRS. MOLD FLASH, PROTRUSIONS, OR GATE BURRS SHALL NOT EXCEED 0.15 PER SIDE. DIMENSION A.
- OPTIONAL CONSTRUCTION: AN ADDITIONAL TRIMMED LEAD IS ALLOWED IN THIS LOCATION. TRIMMED LEAD NOT TO EXTEND MORE THAN 0.2 FROM BODY.

	MILLIMETERS			
DIM	MIN	MAX		
Α	3.00	BSC		
В	1.50 BSC			
С	0.90	1.10		
D	0.25	0.50		
G	0.95 BSC			
Н	0.01 0.10			
J	0.10	0.26		
K	0.20	0.60		
M	0° 10°			
S	2.50 3.00			

#### **SOLDERING FOOTPRINT\***



\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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