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DUAL VIDEO 6dB AMPLIFIER WITH 75Ω DRIVER

GENERAL DESCRIPTION

NJM2267 is a dual video 6dB amplifier with 75 Ω drivers for S-VHS VCRs, HI-BAND VCRs, etc..Each channel has clamp function that fixes DC level of video signal and 75 Ω drivers to be connected to TV monitors directly. Further more it has sag corrective circuits that prevent the generation of sag with smaller capacitance than ever.

Its operating supply voltage is 4.85 to 9V and bandwidth is 7MHz.

FEATURES

- Wide Operating Voltage (4.85V to 9.0V)
- Dual Channel
- Internal Clamp Function
- Internal Driver Circuit For 75Ω Load
- SAG Corrective Function
- Wide Frequency Range (7MHz)
- Low Operating Current 14.0mA (Dual)
- Package Outline DIP8, DMP8, SSOP8
- Bipolar Technology

APPLICATIONS

•VCR, Video Camera, TV, Video Disc Player.

BLOCK DIAGRAM

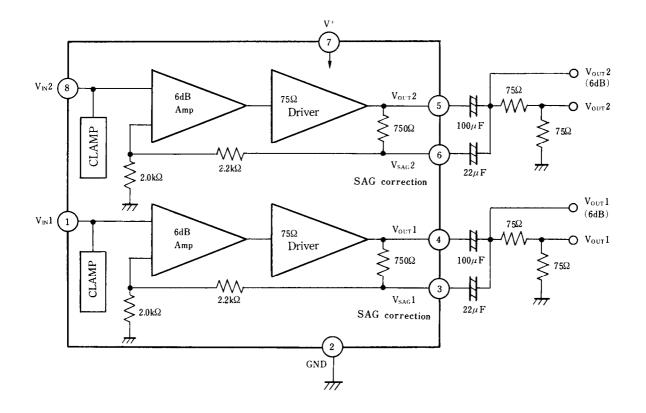




NJM2267D



NJM2267V



■ ABSOLUTE MAXIMUM RATINGS (Ta=25°C)						
PARAMETER	SYMBOL	RATINGS	UNIT			
Supply Voltage	V*	10	V			
Power Dissipation	P _D	(DIP8) 500 (DMP8) 300 (SSOP8) 250	mW mW mW			
Operating Temperature Range	T _{opr}	-40 to +85	°C			
Storage Temperature Range	T _{stg}	-40 to +125	°C			

■ ELECTRICAL CHARACTERISTICS

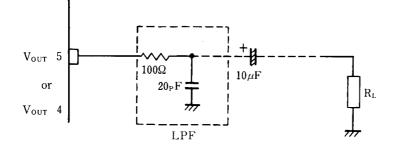
(V⁺=5V, Ta=25±2°C)

PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Operating Current	I _{CC}	No Signal	-	14.0	18.2	mA
Voltage Gain	Gv	$V_{IN}=1MHz$, $1V_{P-P}Sinewave$	5.7	6.2	6.7	dB
Frequency Characteristics	G _f	V_{IN} =1 V_{P-P} , Sinewave, 7MHz / 1MHz	-	-	±1.0	dB
Differential Gain	DG	$V_{IN}=1V_{P-P}$, Staircase	-	1.0	3.0	%
Differential Phase	DP	$V_{IN}=1V_{P-P}$, Staircase	-	1.0	3.0	deg
Crosstalk	СТ	$V_{IN}=4.43MHz$, $1V_{P.P}$, Sinewave	-	-70	-	dB
Gain Offset	G _{CH}	$V_{IN}=1MHz, 1V_{P-P}, G_{CH}=V_{OUT1}-V_{OUT2}$	-	-	±0.5	dB
Input Clamp Voltage	V _{CL}		1.79	1.91	2.03	v
SAG Terminal Gain	Gsag		35	45	-	dB

APPLICATION

Oscillation Prevention

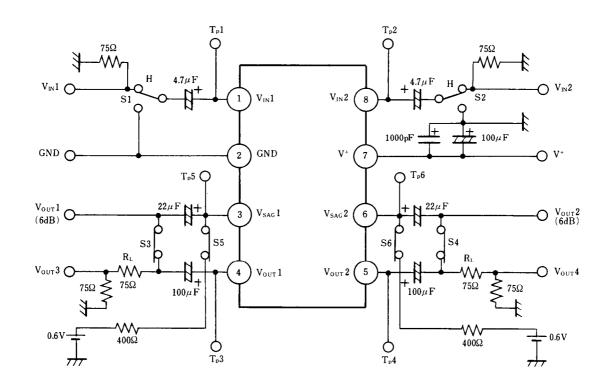
It is much effective to insert LPF (Cutoff Frequency 70MHz) under light loading conditions (RL » 1kΩ)



NJM2267

	NAL FUNC			(V ⁺ =5.0V, Ta=25°
PIN No.	PIN NAME	SYMBOL	EQUIVALENT CIRCUIT	FUNCTIONS
1	Input Clamp Terminal	V _{IN1}	V ⁴ (1) 300 300 µ A	Input terminal of 1V _{P-P} composite signal or Y signal. Clamp level is 1.9V
2	GND	GND	•	Ground
3	SAG correction	V _{SAG1}	V·	SAG caused by a coupling capacitor of the output can be prevented by connecting this terminal with the output termina through an external capacitor.(see block diagram) When SAG correcting function is not necessary, this terminal must be connected with pin "4" directly.
4	Video Output1	Vouti	V ⁺ 2.2k 750 4	Output terminal that can drive 75Ω line.
5	Video Output2	V _{OUT2}	V- 2.2k 750 5	Output terminal that can drive 75Ω line.
6	SAG correction	V _{SAG2}	V- 2.2k 6 750 3mA	SAG caused by a coupling capacitor of the output can be prevented by connecting this terminal with the output termina through an external capacitor.(see block diagram) When SAG correcting function is not necessary, this terminal must be connected with pin "5" directly.
7	V ⁺	V ⁺	///	Cumple Vallage
7 8	Input Clamp Terminal	V V _{IN2}	V- 8 300 300/(A	Supply Voltage Input terminal of 1V _{P-P} composite signal or Y signal. Clamp level is 1.9V

TEST CIRCUIT



TEST METHODES

PARAMETER	SYMBOL	SWITCH CONDITIONS				TIONS	6	CONDITIONS
		S1	S2	S3	S4	S5	S6	CONDITIONO
Supply Current	I _{CC}	Н	Н					7PIN Sink Current
Voltage Gain	Gv	Н	Н	ON	ON			$V_{OUT1} / V_{IN}, V_{OUT2} / V_{IN2}$ at $V_{IN1}(V_{IN2})=1MHz$, $1V_{P:P}$, Sinewave
Frequency Characteristic	G _f	Н	Н	ON	ON			G_{V1M} ; Voltage Gain at V_{IN1} (V_{IN2})=1MHz, $1V_{P\cdot P}$ G_{V10M} ; Voltage Gain at V_{IN1} (V_{IN2})=7MHz, $1V_{P\cdot P}$ G_f = G_{V10M} - G_{V1M}
Differential Gain	DG	Н	Н	ON	ON			Measuring VOUT3 at VIN1=Staircase Signal
Differential Phase	DP	Н	Н	ON	ON			Measuring V _{OUT3} at V _{IN1} =Staircase Signal
Crosstalk	СТ	н	L	ON	ON			V_{OUT2} / V_{OUT1} at V_{IN1} =4.43MHz, 1 V_{P-P} , Sinewave $V_{OUT1} / VIN2$ at V_{IN12} =4.43MHz, 1 V_{P-P} , Sinewave
Gain Offset	G _{CH}	Н	н	ON	ON			$G_{V1} = V_{OUT1} / V_{IN1}, G_{V2} = V_{OUT2} / V_{IN2}$ $G_{CH} = G_{V1} - G_{V2}$
Input Clamp Voltage	V _{CL}	Н	Н					Measuring at TP1 (TP2)
SAG Terminal Gain	G _{SAG}	H H	H H			ON	ON	TP3 (TP4) Voltage; Vo1A (Vo2A), TP5 (TP6) voltage; Vso1A (Vso2A) TP3 (TP4) Voltage; Vo1B(Vo2B), TP5 (TP6) voltage; Vso1B (Vso2B) G _{SAG} =20log{ (Vo1B-Vo1A) / (Vso1A-Vso1B) } G _{SAG} =20log{ (Vo2B-Vo2A) / (Vso2A-Vso2B) }

Clamp circuit

1. Operation of Sync-tip-clamp

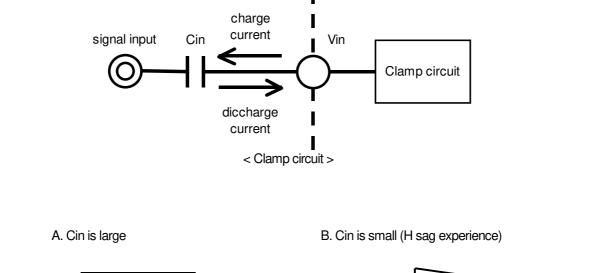
Input circuit will be explained. Sync-tip clamp circuit (below the clamp circuit) operates to keep a sync tip of the minimum potential of the video signal. Clamp circuit is a circuit of the capacitor charging and discharging of the external input Cin. It is charged to the capacitor to the external input Cin at sync tip of the video signal. Therefore, the potential of the sync tip is fixed.

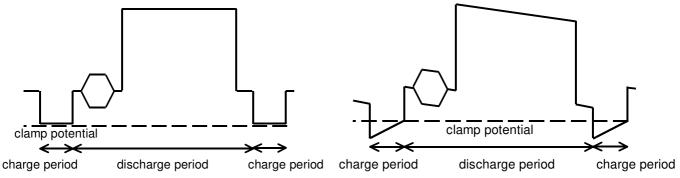
And it is discharged charge by capacitor Cin at period other than the video signal sync tip. This is due to a small discharge current to the IC.

In this way, this clamp circuit is fixed sync tip of video signal to a constant potential from charging of Cin and discharging of Cin at every one horizontal period of the video signal.

The minute current be discharged an electrical charge from the input capacitor at the period other than the sync tip of video signals. Decrease of voltage on discharge is dependent on the size of the input capacitor Cin.

If you decrease the value of the input capacitor, will cause distortion, called the H sag. Therefore, the input capacitor recommend on more than 0.1 uF.





< Waveform of input terminal >

2. Input impedance

The input impedance of the clamp circuit is different at the capacitor discharge period and the charge period.

The input impedance of the charging period is a few k Ω . On the other hand, the input impedance of the discharge period is several M Ω . Because is a small discharge-current through to the IC.

Thus the input impedance will vary depending on the operating state of the clamp circuit.

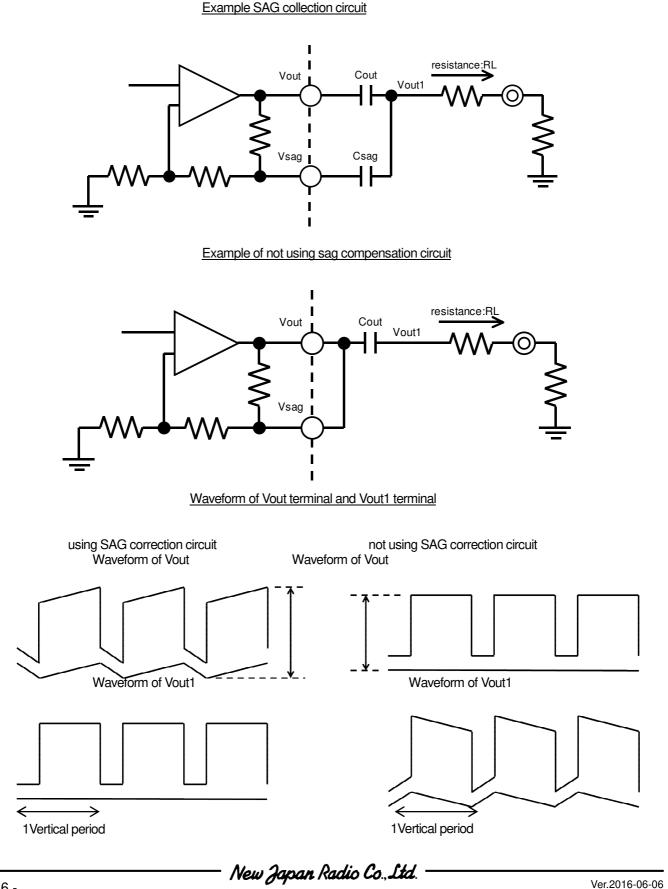
3. Impedance of signal source

Source impedance to the input terminal, please lower than 200Ω . A high source impedance, the signal may be distorted. If so, please to connect a buffer for impedance conversion.

♦ SAG correction circuit

SAG correction circuit is a circuit to correct for low-frequency attenuation by high-pass filter consisting of the output coupling capacitance and load resistance. Low-frequency attenuation raises the sag in the vertical period of the video signal.

Capacitor for Vsag (Csag) is connected to the negative feedback of the amplifier. This Csag increase the low frequency gain to correct for the attenuation of low frequency gain.

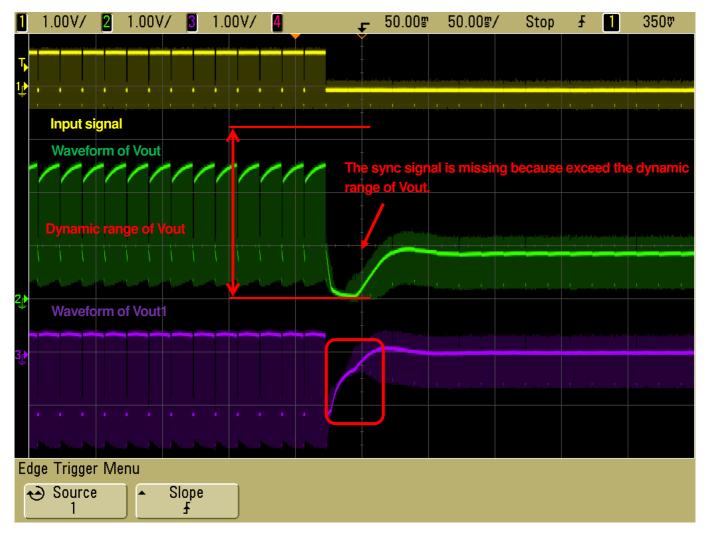


SAG correction circuit generates a low frequency component signal amplified to Vout terminal.

Changes of the luminance signal will be low-frequency components, if you want to output a large signal luminance changes. Therefore, generate correction signal of change of a luminance signal to Vout pin.

At this time, signal is over the dynamic range of Vout pin. This may cause a lack of sync signal, and waveform distortion. Please see diagram below (green waveform), if you want to output large changes of a signal luminance, such as 100% white

video signal and black signal. Thus, output signal exceed dynamic range of Vout pin and may be the signal lack.



< Countermeasure for waveform distortion >

1. Please using small value the Sag compensation capacitor (CSAG).

It can ensure the dynamic range by using small value the capacitor (CSAG). It because of low-frequency variation of Vout pin is smaller. However, the output (COUT) must be use large capacitor for this reason sag characteristics become exacerbated.

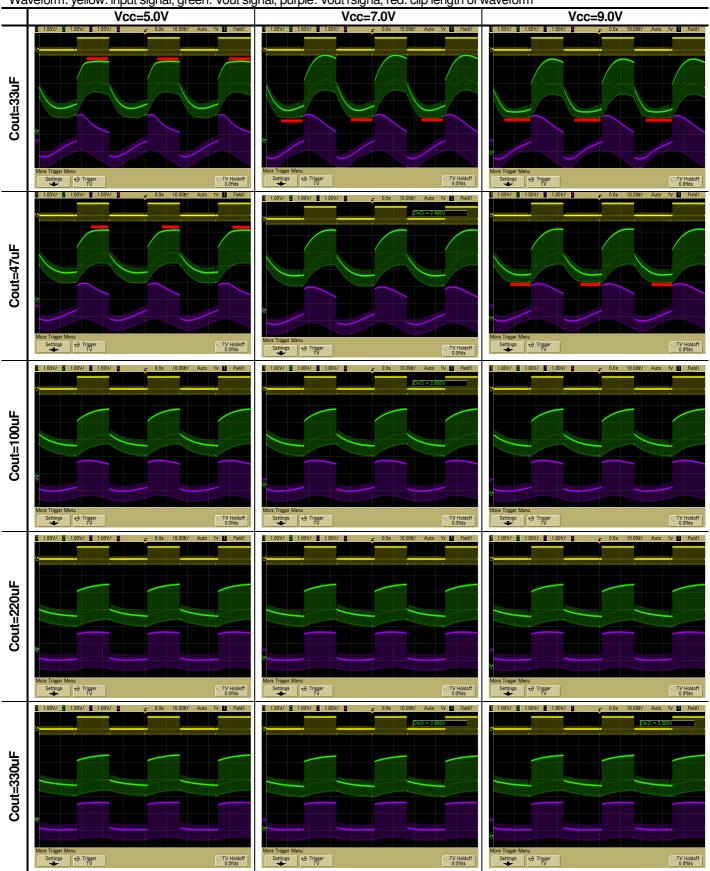
Please do not use the sag correction circuit.
Signal can output within dynamic range for reason it does not change the DC level of the output terminal.
However, the output (COUT) must be use large capacitor for this reason sag characteristics become exacerbated.

3. To increase the power supply voltage

Internal reference voltage is dividing the power supply voltage and GND.

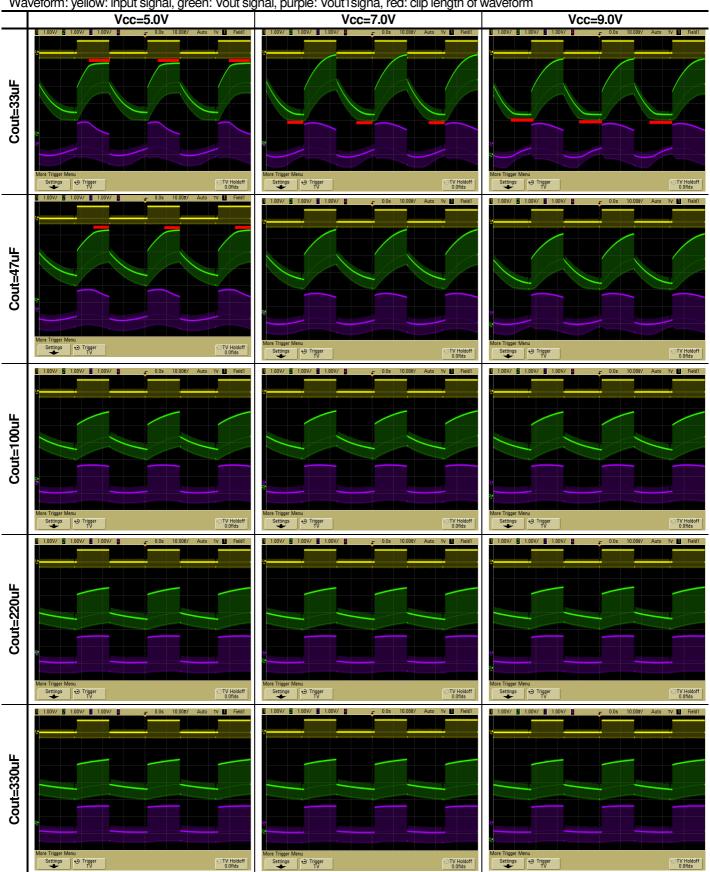
Therefore, it can ensure the dynamic range by increasing the power supply voltage.

Csag=10uF, Input signal: bounce signal (IRE0%, IRE100%, 30Hz), resistance=150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signa, red: clip length of waveform



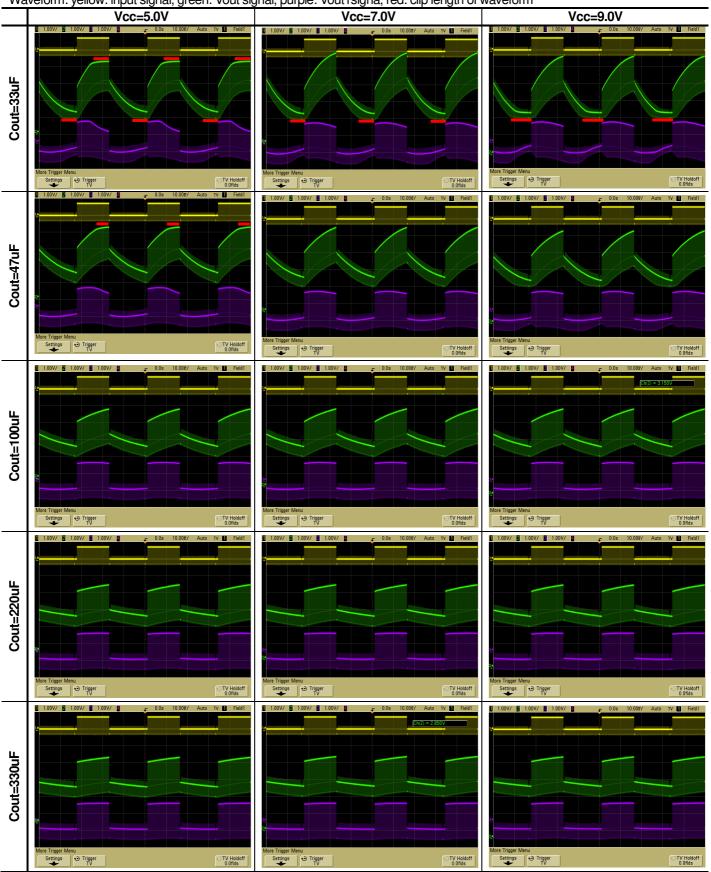
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< Using SAG correction circuit > Csag=22uF, Input signal: bounce signal (IRE0%, IRE100%, 30Hz), resistance=150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signa, red: clip length of waveform



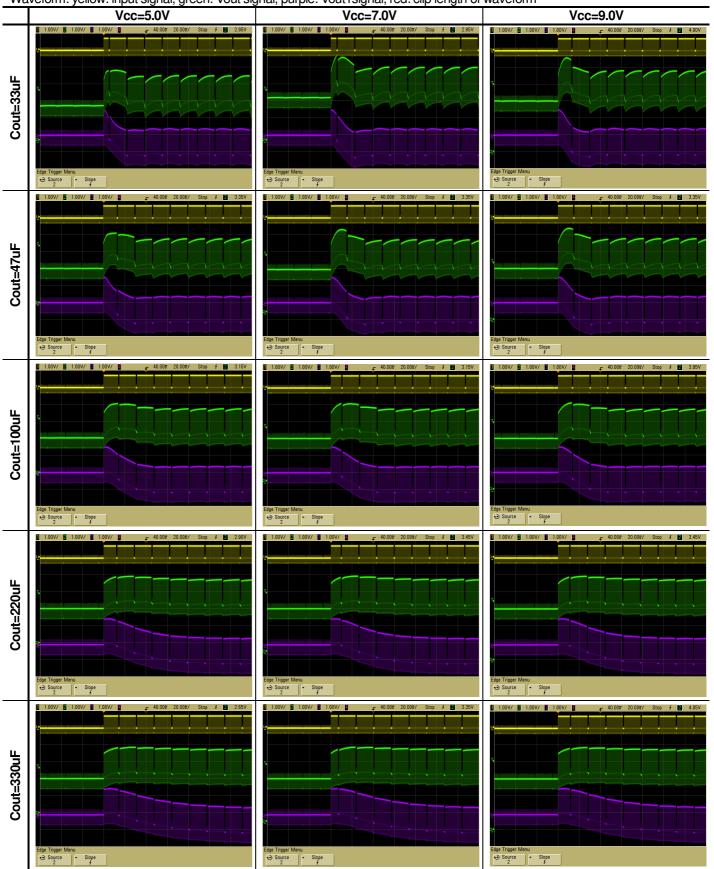
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Csag=33uF, Input signal: bounce signal (IRE0%, IRE100%, 30Hz), resistance=150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signa, red: clip length of waveform



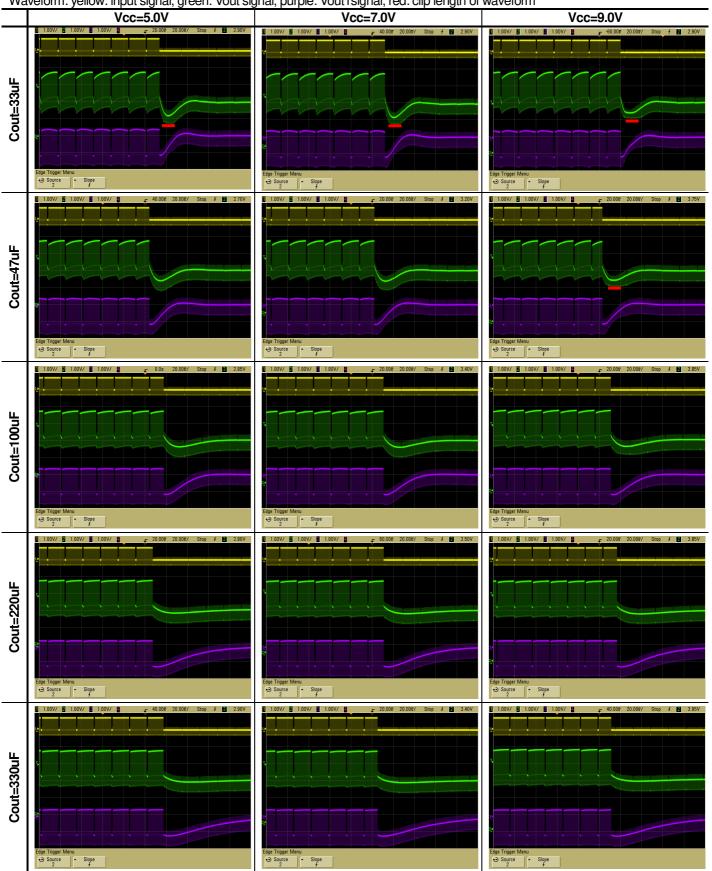
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Csag=10uF, Input signal: Black to White100%, resistance150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal, red: clip length of waveform



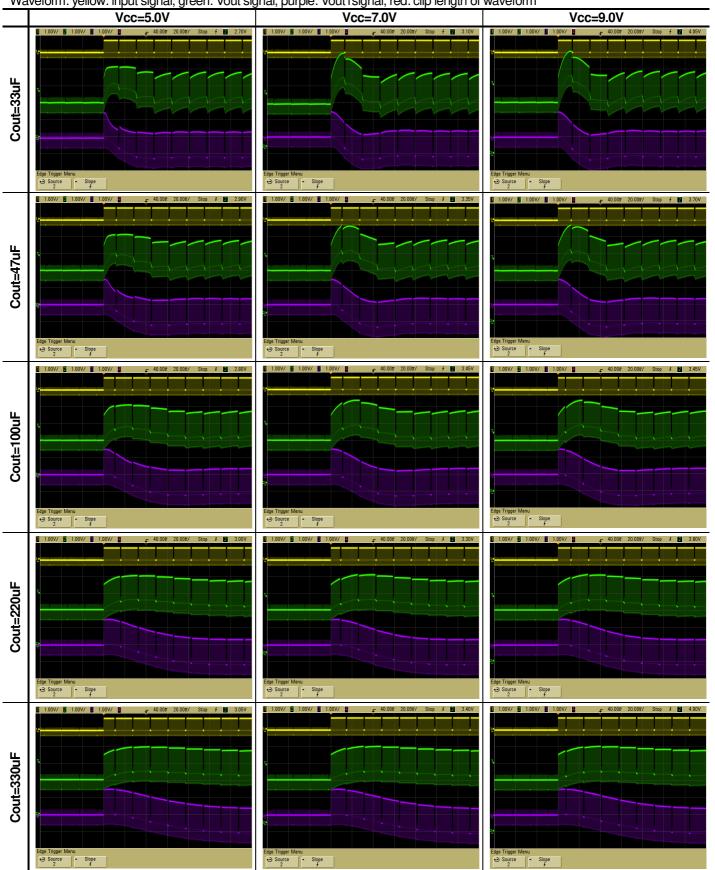
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Csag=10uF, Input signal: White100% to Black, resistance150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal, red: clip length of waveform



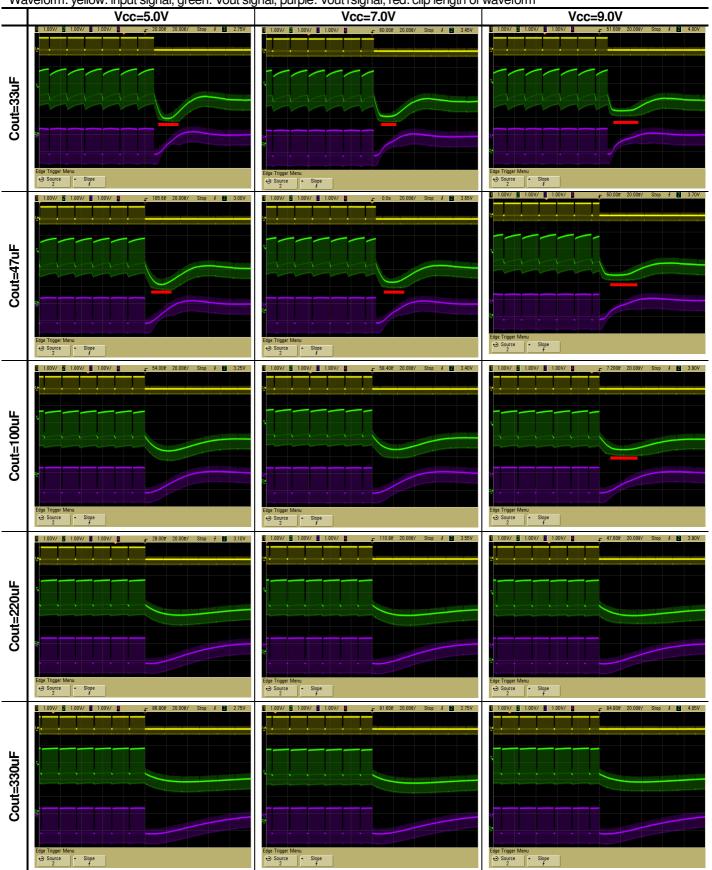
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Csag=22uF, Input signal: Black to White100%, resistance150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal, red: clip length of waveform



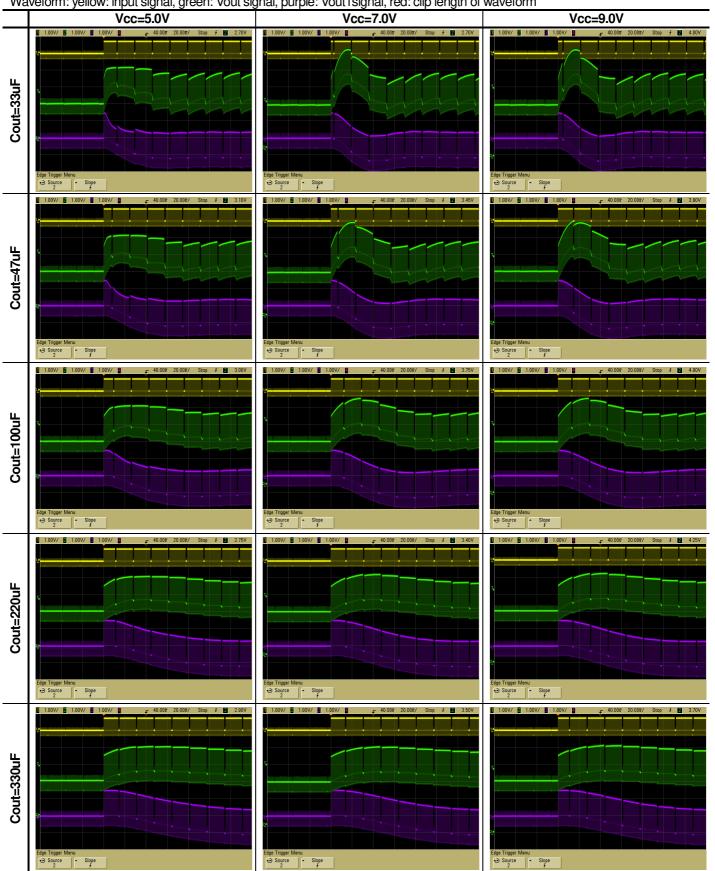
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Csag=22uF, Input signal: White100% to Black, resistance150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal, red: clip length of waveform



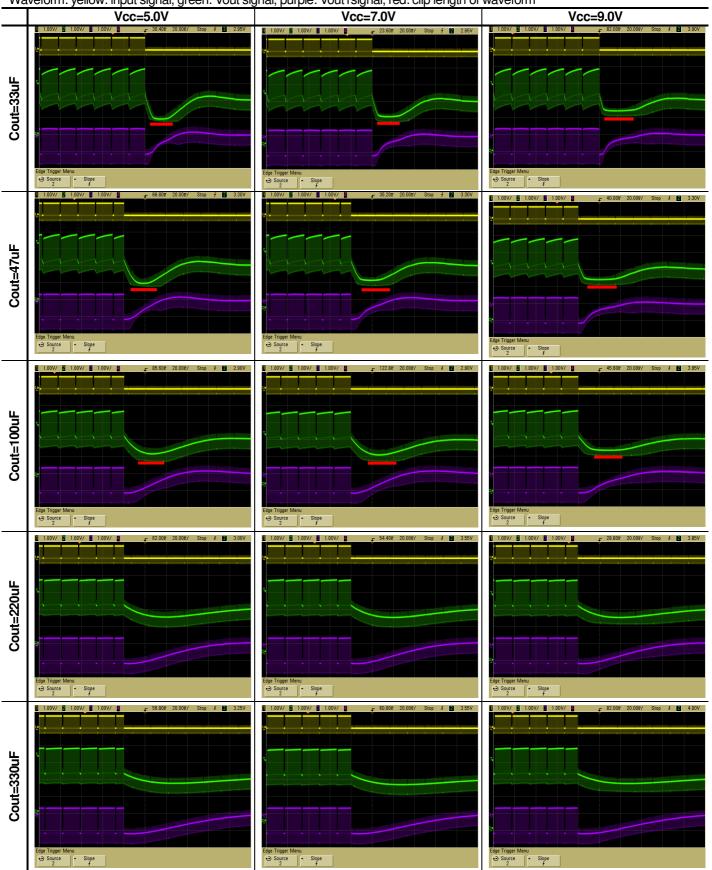
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Csag=33uF, Input signal: Black to White100%, resistance150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal, red: clip length of waveform



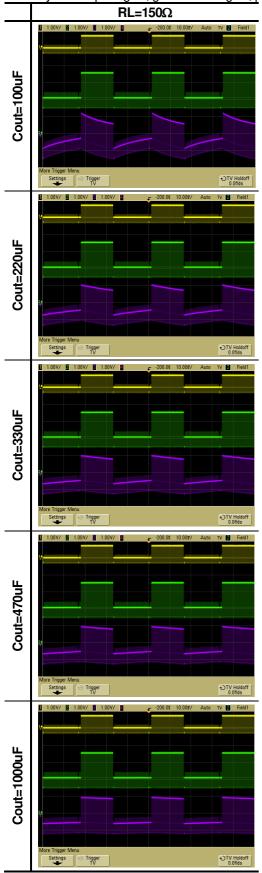
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Csag=33uF, Input signal: White100% to Black, resistance150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal, red: clip length of waveform

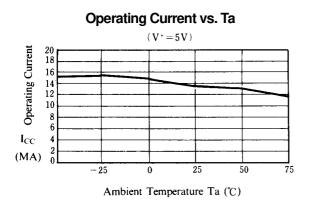


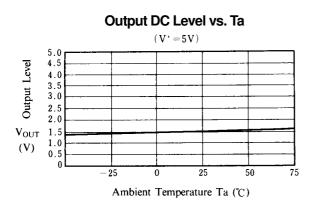
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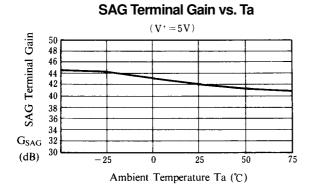
< Not using SAG correction circuit > Vcc=5V, Input signal: bounce signal (IRE0%, IRE100%, 30Hz), resistance=150 Ω Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal

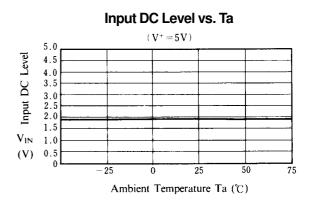


TYPICAL CHARACTERISTICS

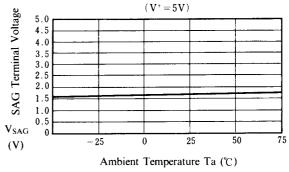


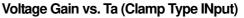


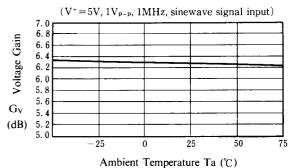




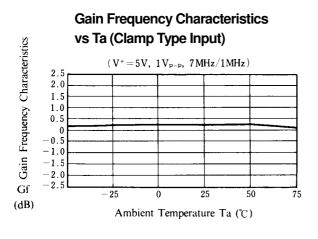


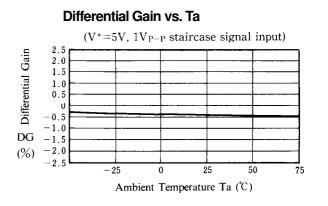






■ TYPICAL CHARACTERISTICS

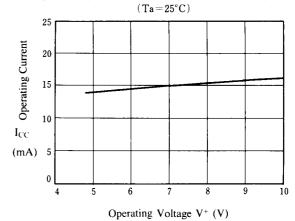


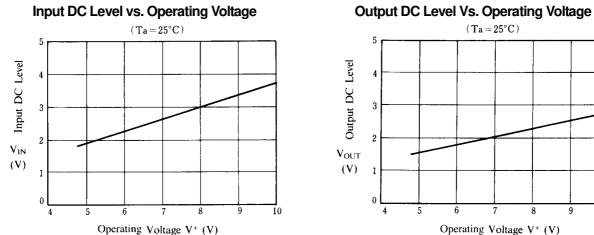


Differential Phase vs. Ta (V⁺=5V, 1V_{P-P} staircase signal input) 2.5 2.0 1.5 1.0 Differential Phase 0.5 0 -0.5 -1.0 DP - 1.5 -2.0 $(deg) = \frac{1}{1.5}$ - 25 25 50 0 75

Ambient Temperature Ta (°C)

Operating Current vs. Operating Voltage





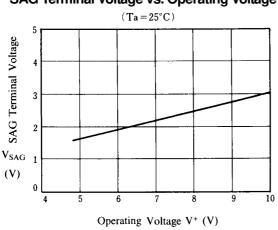
Operating Voltage V+ (V)

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10

9

■ TYPICAL CHARACTERISTICS



SAG Terminal Voltage vs. Operating Voltage

SAG Terminal Gain 50 40 30 GSAG (dB) 20 5 4 6 7 8 9 Operating Voltage V⁺ (V)

SAG Terminal Gain vs. Operating Voltage

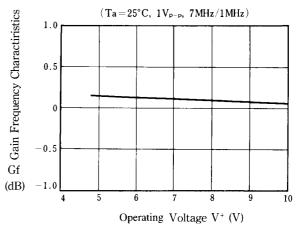
60

 $(Ta = 25^{\circ}C)$

Voltage Gain vs. Operating Voltage (Ta=25°C, 1V_{P-P}, 1MHz sinewave signal input) 7 Voltage Gain 6.5 6 Gv 5.5 (dB) 5 5 4 6 7 8 9 10 Operating Voltage V⁺ (V)

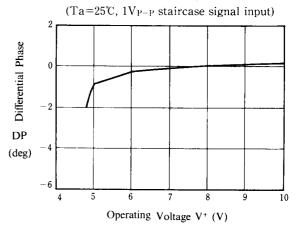
Gain Frequency Characteristics vs. Operating Voltage

10



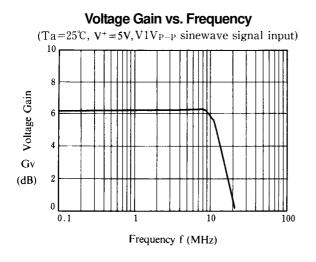
Differential Gain vs. Operating Voltage (Ta=25°C, 1VP.P staircase signal input) 2 Differential Gain 0 - 2 DG (%) 5 4 6 7 8 9 10 Operating Voltage V⁺ (V)

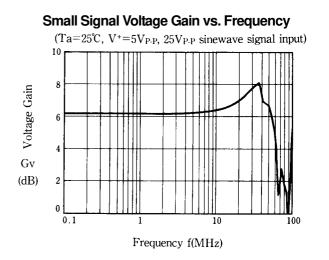
Differential Phase vs. Operating Voltage



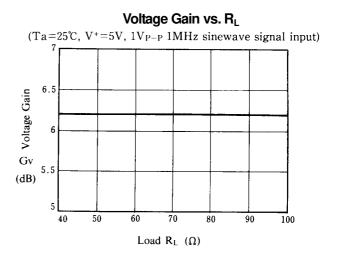
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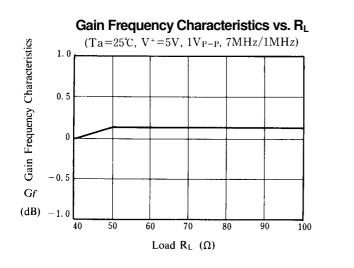
TYPICAL CHARACTERISTICS

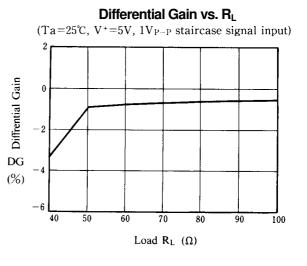




Cross Talk vs. Frequency (Ta=25°C, V⁺=5V, 1V_{P-P} sinewave signal input) y = -20 y = -20 CT = -60(dB) -100 0 -100 0 0 -100 0 0 0 1 10 100100

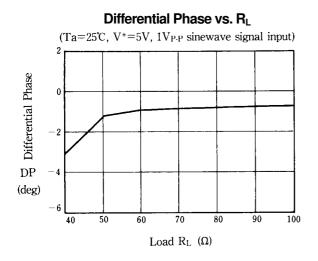


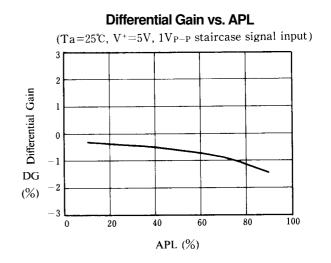




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TYPICAL CHARACTERISTICS



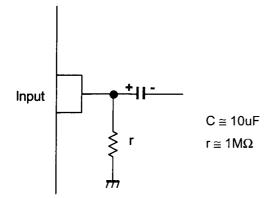


Differential Phase vs. APL (Ta=25°C, V⁺=5V, $1V_{P-P}$, staircase signal input) 3 2 Differential Phase 1 0 -1 (deg) - 2 -3 ∟ 0 80 100 20 40 60 APL (%)

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APPLICATION

This IC requires $1M\Omega$ resistance between INPUT and GND pin for clamp type input since the minute current causes an unstable pin voltage.



[CAUTION]

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