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FILTERLESS CLASS-D MONO AUDIO AMPLIFIER

Description

The PAM8301 is a 1.5W Class-D mono audio amplifier. Its low THD+N feature offers high quality sound reproduction. The new filterless architecture allows the device to drive speaker directly instead of using low-pass output filters, therefore saving system cost and PCB area.

With the same number of external components, the efficiency of the PAM8301 is much better than that of Class-AB cousins. It can optimize battery life thus is ideal for portable applications.

The PAM8301 is available in TSOT26 package.

Pin Assignments

Top View
TSOT26

FPXYW
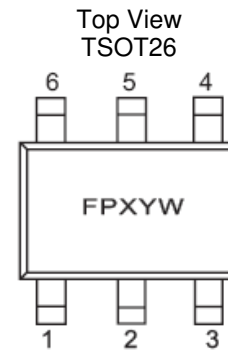
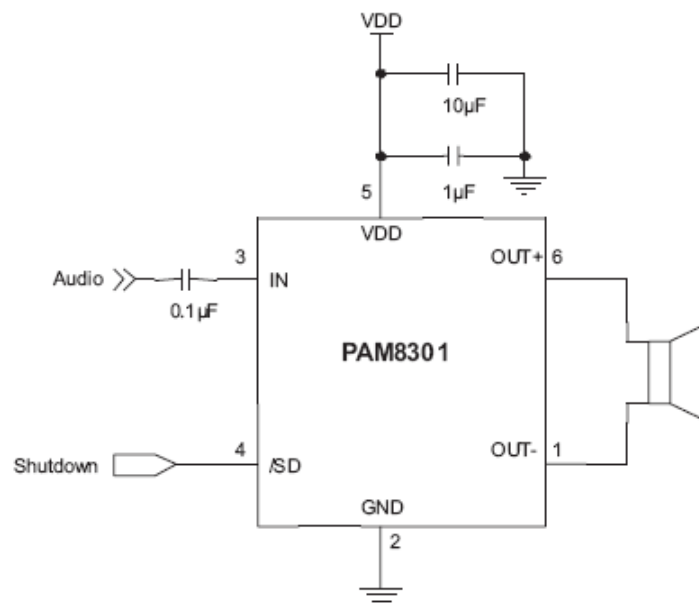## Features

- 1.5W Output at 10% THD with a 8Ω Load and 5V Power Supply
- Filterless, Low Quiescent Current and Low EMI
- High Efficiency up to 88%
- Superior Low Noise
- Short Circuit Protection
- Thermal Shutdown
- Few External Components to Save Space and Cost
- Tiny TSOT26 Package
- Pb-Free Package

Applications

- PMP/MP4
- GPS
- Portable Speakers
- Walkie Talkie
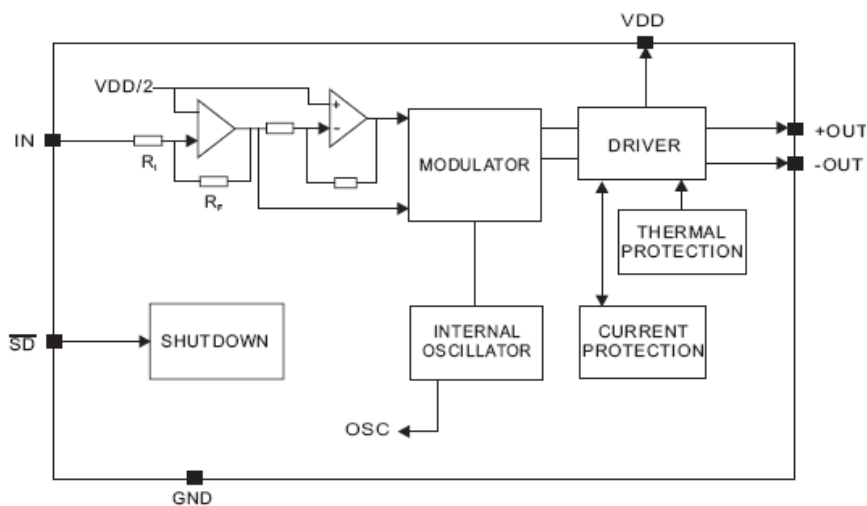- Handsfree Phones/Speaker Phones
- Cellular Phones

Typical Applications Circuit

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Pin Descriptions

Pin Number	Pin Name	Function
1	OUT-	Negative Output
2	GND	Ground
3	IN	Input
4	$\overline{SD}$	Shutdown, Active Low
5	VDD	Power Supply
6	OUT+	Positive Output

Functional Block Diagram

Absolute Maximum Ratings (@T$_A$ = +25°C, unless otherwise specified.)

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability. All voltages are with respect to ground.

Parameter	Rating	Unit
Supply Voltage at No Input Signal	6.0	V
Input Voltage Range	-0.3 to V$_{DD}$ +0.3	
Maximum Junction Temperature	150	°C
Storage Temperature	-65 to +150	
Soldering Temperature	300, 5sec	

Recommended Operating Conditions (@T$_A$ = +25°C, unless otherwise specified.)

Parameter	Rating	Unit
Supply Voltage Range	2.5 to 5.5	V
Operation Temperature Range	-40 to +85	°C
Junction Temperature Range	-40 to +125	°C

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Thermal Information

Parameter	Package	Symbol	Max	Unit
Thermal Resistance (Junction to Ambient)	TSOT26	θ_{JA}	250	°C/W
Thermal Resistance (Junction to Case)	TSOT26	θ_{JC}	130	

Electrical Characteristics (@T_A = +25°C, V_{DD} = 5V, Gain = 24dB, R_L = 8Ω, unless otherwise specified.)

Symbol	Parameter	Test Conditions		Min	Typ	Max	Units
V_{DD}	Supply Voltage Range			2.5		5.5	V
I_Q	Quiescent Current	No Load			4	8	mA
I_{SHDN}	Shutdown Current	V_{SHDN} = 0V				1	µA
V_{SH}	$\overline{SHDN}$ Input High			1.2			V
V_{SL}	SHDN Input Low					0.4	
$R_{DS(ON)}$	Drain-Source On-State Resistance	I_{DS} = 100mA	P MOSFET		0.45		Ω
			N MOSFET		0.20		
P_O	Output Power	f = 1kHz	THD+N = 1%		1.2		W
			THD+N = 10%		1.5		
THD+N	Total Harmonic Distortion Plus Noise	R_L = 8Ω, P_O = 200mW			0.2		%
		R_L = 8Ω, P_O = 0.5W			0.3		
PSRR	AC Power Supply Ripple Rejection	No Inputs, f = 1kHz, V_{PP} = 200mV		45	50		dB
G_V	Gain				24		dB
V_N	Output Noise	No A-Weighting			180		µV
		A-Weighting			120		
f_{OSC}	Oscillator Frequency			200	250	300	kHz
η	Peak Efficiency	f = 1kHz		85	88		%
SNR	Signal to Noise Ratio	f = 20 to 20kHz			78		dB
OTP	Over Temperature Protection				135		°C
OTH	Over Temperature Hysterisis				30		°C

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Typical Performance Characteristics (@T$_A$ = +25°C, unless otherwise specified.)

1. Frequency VS Supply Voltage

2. Frequency VS Temperature

3. Efficiency VS Output Power

4. Efficiency VS Output Power

5. THD+N VS Output Power

6. THD+N VS Output Power

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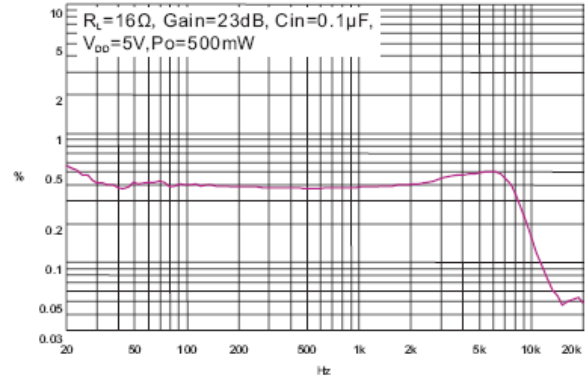
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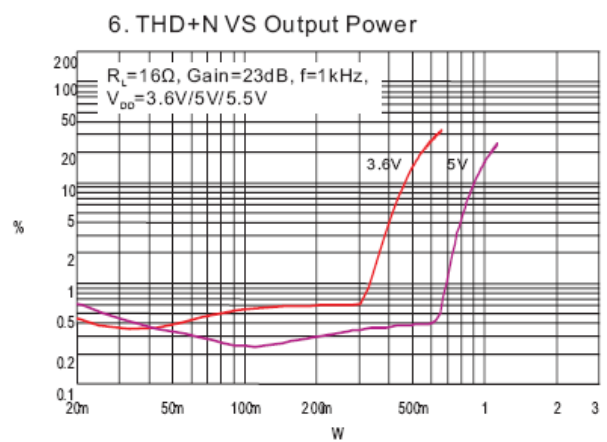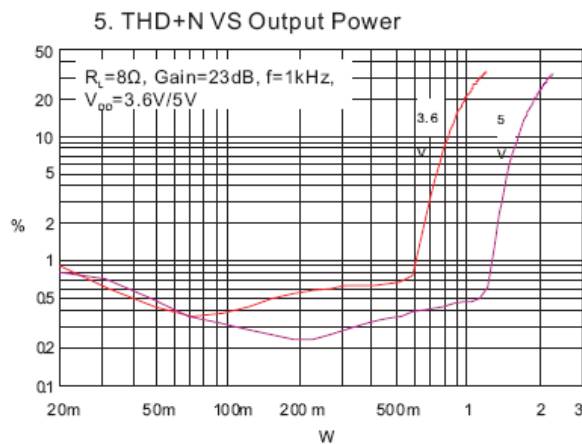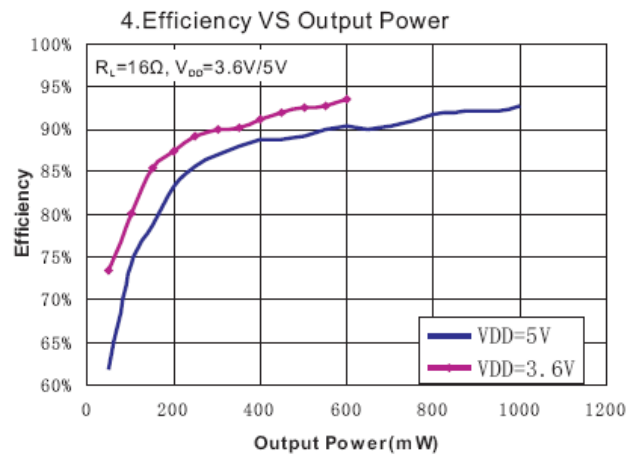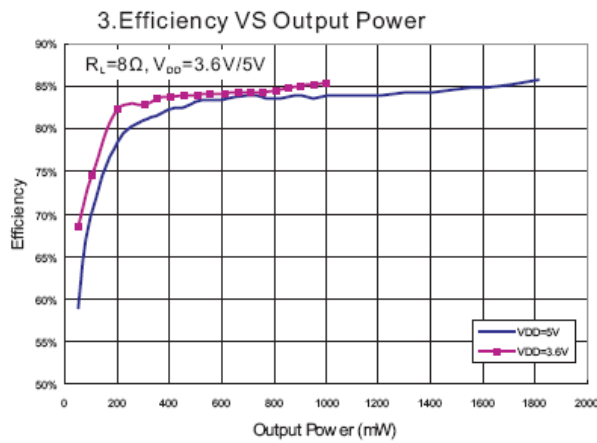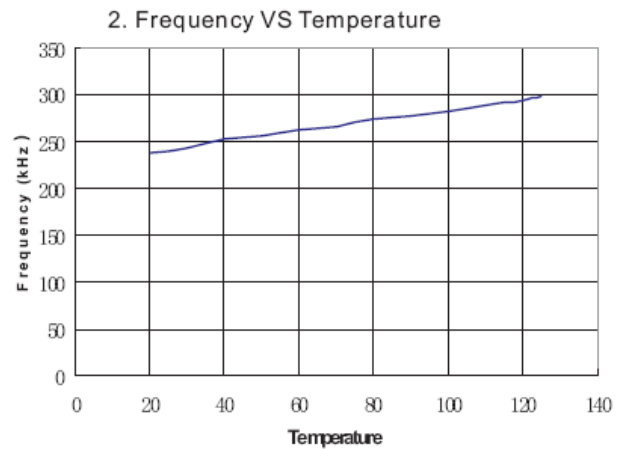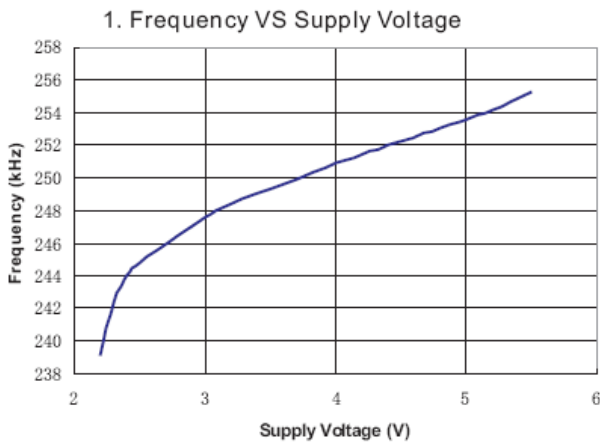
Typical Performance Characteristics (cont.) (@T$_A$ = +25°C, unless otherwise specified.)

7. THD+N VS Frequency
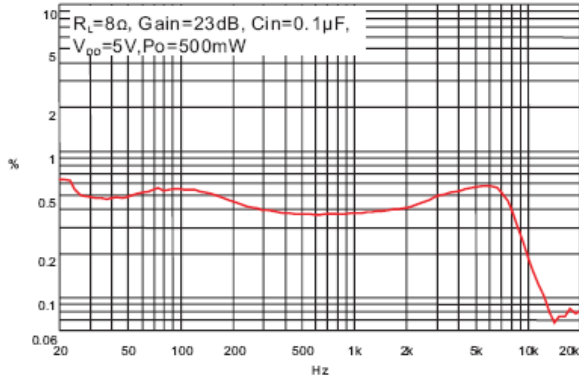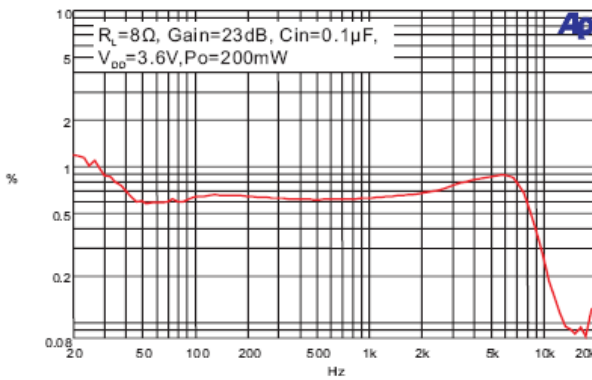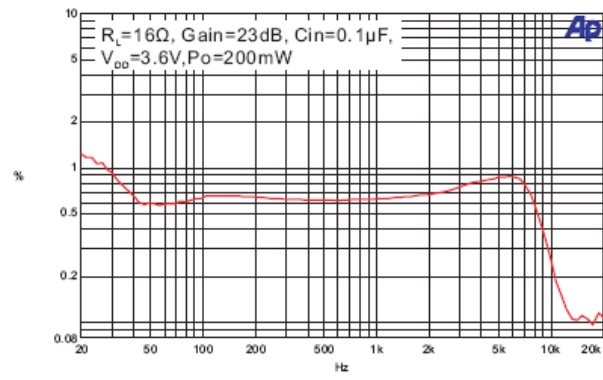
R$_L$=8Ω, Gain=23dB, Cin=0.1μF, V$_{DD}$=5V,Po=500mW

8. THD+N VS Frequency

R$_L$=16Ω, Gain=23dB, Cin=0.1μF, V$_{DD}$=5V,Po=500mW

9. THD+N VS Frequency

R$_L$=8Ω, Gain=23dB, Cin=0.1μF, V$_{DD}$=3.6V,Po=200mW

10. THD+N VS Frequency

R$_L$=16Ω, Gain=23dB, Cin=0.1μF, V$_{DD}$=3.6V,Po=200mW

11. Frequency Response

R$_L$=16Ω, Cin=0.1μF,V$_{DD}$=5V

12. Noise Floor

Inputs floating, R$_L$=8Ω, Cin=0.1μF,V$_{DD}$=5V

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Typical Performance Characteristics (cont.) (@T$_A$ = +25°C, unless otherwise specified.)

13. PSRR

R$_L$=8Ω, Gain=23dB,
V$_{DD}$=5V, Input floating

14. EMI vs Frequency

FCC Class B

R$_L$=8Ω, Gain=23dB,
V$_{DD}$=5V, Po=500mW
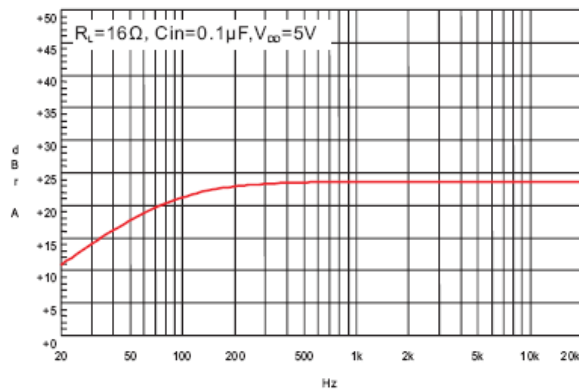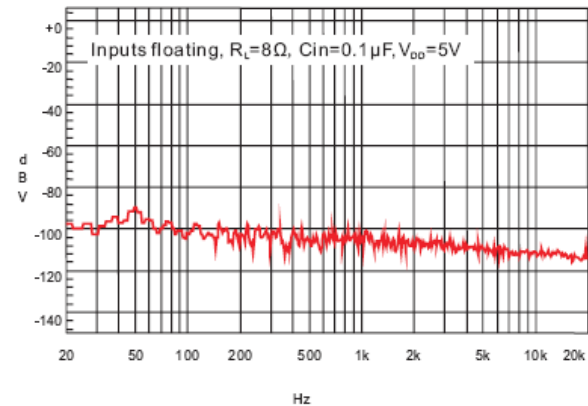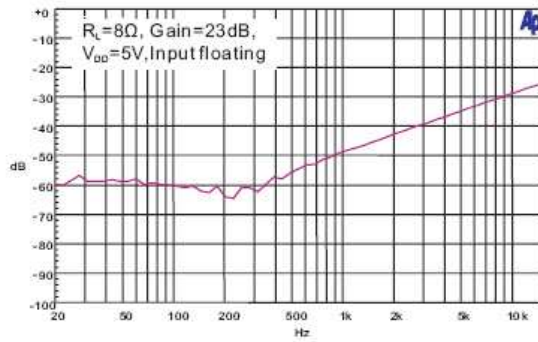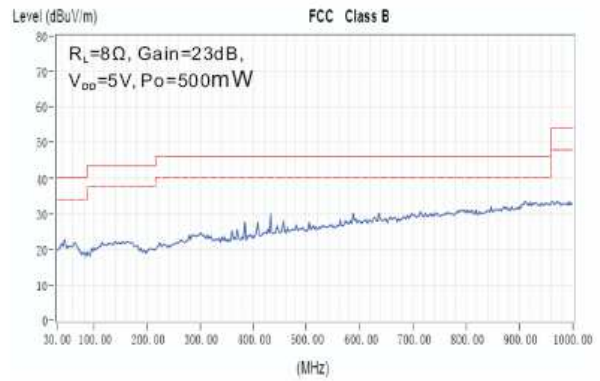
Application Information

Test Setup for Performance Testing
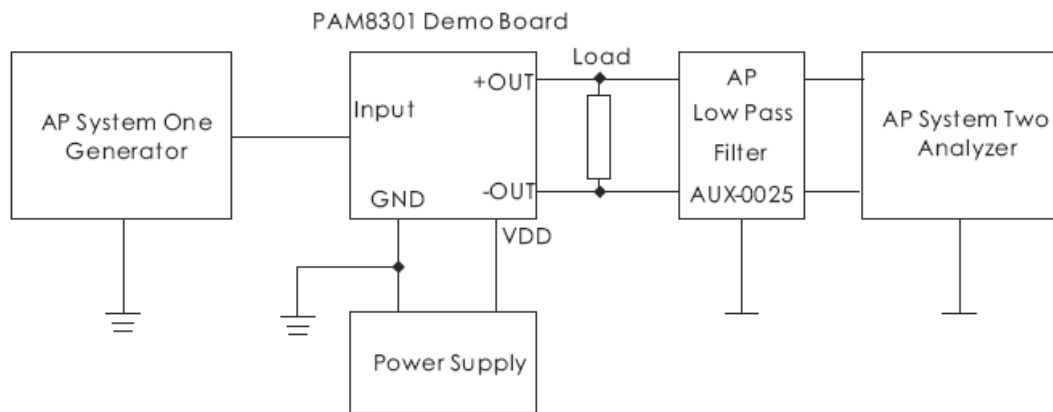
Notes: 1. The AP AUX-0025 low pass filter is necessary for every class-D amplifier measurement with AP analyzer.
         2. Two 22µH inductors are used in series with load resistor to emulate the small speaker for efficiency measurement.

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Application Information (cont.)

Maximum Gain

As shown in block diagram (Page 2), the PAM8301 has two internal amplifier stages. The first stage's gain is externally configurable, while the second stage's is internally fixed. The closed-loop gain of the first stage is set by selecting the ratio of RF to RI while the second stage's gain is fixed at 2x.The output of amplifier one serves as the input to amplifier two, thus the two amplifiers produce signals identical in magnitude, but different in phase by +180°. Consequently, the differential gain for the IC is

$$A_{VD} = 20*log\ [2*(R_F/R_I)]$$

The PAM8301 sets maximum R_F = 80kΩ, minimum R_I = 10kΩ, so the maximum closed-gain is 24dB.

Input Capacitors (C$_I$)

In typical application, an input capacitor, CI, is required to allow the amplifier to bias input signals to a proper DC level for optimum operation. In this case, CI and the minimum input impedance RI (10k internal) form a high pass filter with a corner frequency determined by the following equation:

$$f_C = \frac{1}{2\Pi R_I\ C_I}$$

It is important to choose the value of CI as it directly affects low frequency performance of the circuit, for example, when an application requires a flat bass response as low as 100Hz. Equation is reconfigured as follows:

$$C_I = \frac{1}{2\Pi R_I\ f_I}$$

As the input resistance is variable, for the C_I value of 0.16µF, one should actually choose the C_I within the range of 0.1µF to 0.22µF. A further consideration for this capacitor is the leakage path from the input source through the input network (R_I, R_F, C_I) to the load. This leakage current creates a DC offset voltage at the input to the amplifier that reduces useful headroom, especially in high gain application. For this reason, a low leakage tantalum or ceramic capacitor is the best choice. When a polarized capacitor is used, the positive side of the capacitor should face the amplifier input in most applications as the DC level is held at V_{DD}/2, which is likely higher than the source DC level. Please note that it is important to confirm the capacitor polarity in the application.

Power Supply Decoupling (C$_S$)

The PAM8301 is a high-performance CMOS audio amplifier that requires adequate power supply decoupling to ensure the output THD and PSRR as low as possible. Power supply decoupling affects low frequency response. Optimum decoupling is achieved by using two capacitors of different types that target different types of noise on the power supply leads. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-series-resistance (ESR) ceramic capacitor, typically 1.0µF is good, placing it as close as possible to the device V_{DD} terminal. For filtering lower-frequency noise signals, a capacitor of 10µF or larger, closely located to near the audio power amplifier is recommended.

Shutdown Operation

In order to reduce shutdown power consumption, the PAM8301 contains shutdown circuitry for turn off the amplifier. This shutdown feature turns the amplifier off when a logic low is applied on the SHDOWN pin. By switching the shutdown pin over to GND, the PAM8301 supply current draw will be minimized in idle mode.

For the best power on/off pop performance, the amplifier should be set in the shutdown mode prior to power on/off operation.

Under Voltage Lock-Out (UVLO)

The PAM8301 incorporates circuitry to detect low on or off voltage. When the supply voltage drops to 2.1V or below, the PAM8301 goes into a state of shutdown, and the device comes out of its shutdown state and starts to normal operation by reset the power supply or S$\overline{D}$ pin.

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Application Information (cont.)

How to Reduce EMI (Electro Magnetic Interference)

A simple solution is to put an additional capacitor 1000μF at power supply terminal for power line coupling if the traces from amplifier to speakers are short (< 20cm).

Most applications require a ferrite bead filter as shown at Figure 1. The ferrite filter depresses EMI of around 1MHz and higher. When selecting a ferrite bead, choose one with high impedance at high frequencies and low impedance at low frequencies.
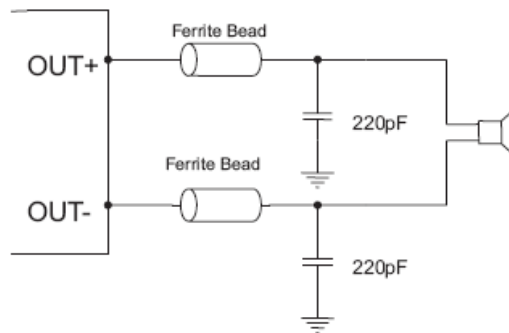

Figure 1: Ferrite Bead Filter to Reduce EMI

Ordering Information

PAM8301 <u>X</u> <u>X</u> <u>X</u>

Pin Configuration

A:
  1: OUT-
  2: GND
  3: IN
  4: SD
  5: VDD
  6: OUT+

Package Type

A: TSOT26

Number of Pins

F: 6
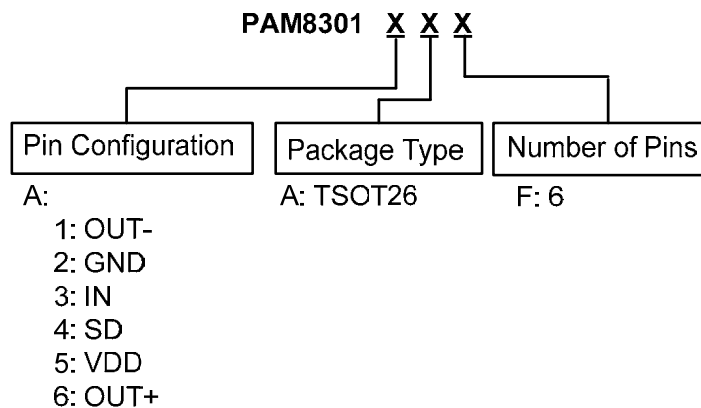
Part Number	Part Marking	Package Type	Standard Package
PAM8301AAF	FPXYW	TSOT26	3000 Unit/Tape&Reel

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Marking Information

Top View
TSOT26
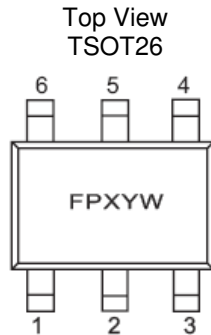
FP: Product Code of PAM8301
X: Internal Code
Y: Year
W: Week

Package Outline Dimensions (All dimensions in mm.)

TSOT26

SEE VIEWS

WITH PLANTING
BASE METAL
SECTION A-A

GAUGE PLANE
SEATING PLANE

VIEW B

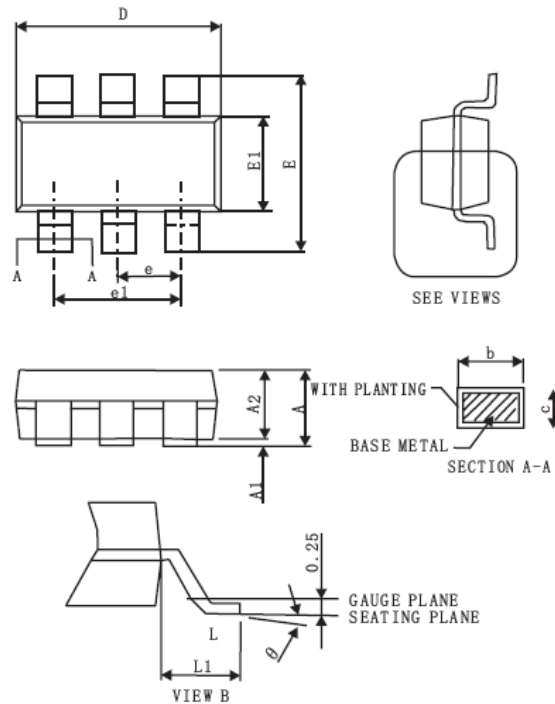| Symbol	A	A1	A2	b	c	D	E
Spec	1.20±0.25	0.10±0.05	1.10±0.2	0.40±0.1	0.15±0.07	2.90±0.1	2.80±0.2
Symbol	E1	e	e 1	L	L1	θ	
Spec	1.60±0.1	0.95BSC	1.90BSC	0.55±0.25	0.60REF	4°±4°	

Unit: Millimeter

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