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## BROADCAST AM/FM RADIO RECEIVER

### Features

- Worldwide FM band support (64–108 MHz)
- Worldwide AM band support (520–1710 kHz)
- SW band support (2.3–26.1 MHz)
- LW band support (153–279 kHz)
- Excellent real-world performance
- Freq synthesizer with integrated VCO
- Advanced seek tuning
- Automatic frequency control (AFC)
- Automatic gain control (AGC)
- Digital FM stereo decoder
- Programmable AVC max gain
- Programmable de-emphasis
- Seven selectable AM channel filters
- AM/FM/SW/LW digital tuning
- EN55020 compliant
- No manual alignment necessary
- Programmable reference clock
- Volume control
- Adjustable soft mute control
- RDS/RBDS processor (Si4735)
- Optional digital audio out (Si4735)
- 2-wire and 3-wire control interface
- Integrated LDO regulator
- 2.0 to 5.5 V supply voltage (SSOP)
- 2.7 to 5.5 V supply voltage (QFN)
- Wide range of ferrite loop sticks and air loop antennas supported
- QFN and SSOP packages
  - RoHS compliant

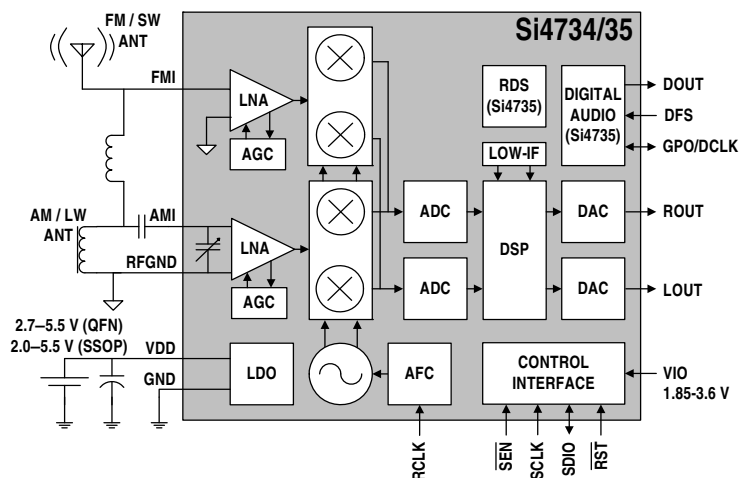
### Applications

- Table and portable radios
- Stereos
- Mini/micro systems
- CD/DVD players
- Boom boxes
- Modules
- Clock radios
- Mini HiFi
- Entertainment systems

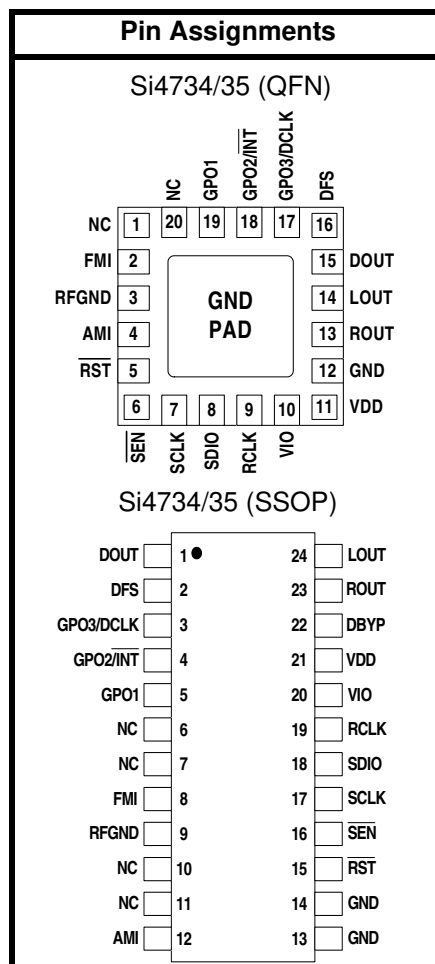
### Description

The Si4734/35 is the first digital CMOS AM/FM radio receiver IC that integrates the complete tuner function from antenna input to audio output.

### Functional Block Diagram



**Ordering Information:**  
See page 32.



This product, its features, and/or its architecture is covered by one or more of the following patents, pending and issued, both foreign and domestic: 7,127,217; 7,272,373; 7,272,375; 7,321,324; 7,355,476; 7,426,376; 7,471,940; 7,339,503; 7,339,504.



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## 1. Electrical Specifications

**Table 1. Recommended Operating Conditions<sup>1</sup>**

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Supply Voltage <sup>2</sup>	$V_{DD}$		2.7	—	5.5	V
Interface Supply Voltage	$V_{IO}$		1.85	—	3.6	V
Power Supply Powerup Rise Time	$V_{DDRISE}$		10	—	—	$\mu$ s
Interface Power Supply Powerup Rise Time	$V_{IORISE}$		10	—	—	$\mu$ s
Ambient Temperature	$T_A$		-20	25	85	$^{\circ}$ C

**Note:**

1. All minimum and maximum specifications apply across the recommended operating conditions. Typical values apply at  $V_{DD} = 3.3$  V and 25  $^{\circ}$ C unless otherwise stated.
2. SSOP devices operate down to  $V_{DD} = 2$  V at 25  $^{\circ}$ C.

**Table 2. Absolute Maximum Ratings<sup>1,2</sup>**

Parameter	Symbol	Value	Unit
Supply Voltage	$V_{DD}$	-0.5 to 5.8	V
Interface Supply Voltage	$V_{IO}$	-0.5 to 3.9	V
Input Current <sup>3</sup>	$I_{IN}$	10	mA
Input Voltage <sup>3</sup>	$V_{IN}$	-0.3 to ( $V_{IO} + 0.3$ )	V
Operating Temperature	$T_{OP}$	-40 to 95	$^{\circ}$ C
Storage Temperature	$T_{STG}$	-55 to 150	$^{\circ}$ C
RF Input Level <sup>4</sup>		0.4	$V_{PK}$

**Notes:**

1. Permanent device damage may occur if the above Absolute Maximum Ratings are exceeded. Functional operation should be restricted to the conditions as specified in the operational sections of this data sheet. Exposure beyond recommended operating conditions for extended periods may affect device reliability.
2. The Si4734/35 devices are high-performance RF integrated circuits with certain pins having an ESD rating of < 2 kV HBM. Handling and assembly of these devices should only be done at ESD-protected workstations.
3. For input pins SCLK, SEN, SDIO, RST, RCLK, DCLK, DFS, GPO1, GPO2, and GPO3.
4. At RF input pins, FMI and AMI.

**Table 3. DC Characteristics**(V<sub>DD</sub> = 2.7 to 5.5 V, V<sub>IO</sub> = 1.85 to 3.6 V, T<sub>A</sub> = -20 to 85 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
<b>FM Mode</b>						
Supply Current <sup>1</sup>	I <sub>FM</sub>		—	19.2	22	mA
Supply Current <sup>2</sup>	I <sub>FM</sub>	Low SNR level	—	19.9	23	mA
RDS Supply Current <sup>1</sup>	I <sub>FM</sub>		—	19.2	23	mA
<b>AM/SW/LW Mode</b>						
Supply Current <sup>1</sup>	I <sub>AM</sub>	Analog Output Mode	—	15.4	20.5	mA
<b>Supplies and Interface</b>						
Interface Supply Current	I <sub>IO</sub>		—	320	600	μA
V <sub>DD</sub> Powerdown Current	I <sub>DDPD</sub>		—	10	20	μA
V <sub>IO</sub> Powerdown Current	I <sub>IOPD</sub>	SCLK, RCLK inactive	—	1	10	μA
High Level Input Voltage <sup>3</sup>	V <sub>IH</sub>		0.7 x V <sub>IO</sub>	—	V <sub>IO</sub> + 0.3	V
Low Level Input Voltage <sup>3</sup>	V <sub>IL</sub>		-0.3	—	0.3 x V <sub>IO</sub>	V
High Level Input Current <sup>3</sup>	I <sub>IH</sub>	V <sub>IN</sub> = V <sub>IO</sub> = 3.6 V	-10	—	10	μA
Low Level Input Current <sup>3</sup>	I <sub>IL</sub>	V <sub>IN</sub> = 0 V, V <sub>IO</sub> = 3.6 V	-10	—	10	μA
High Level Output Voltage <sup>4</sup>	V <sub>OH</sub>	I <sub>OUT</sub> = 500 μA	0.8 x V <sub>IO</sub>	—	—	V
Low Level Output Voltage <sup>4</sup>	V <sub>OL</sub>	I <sub>OUT</sub> = -500 μA	—	—	0.2 x V <sub>IO</sub>	V
<b>Notes:</b>						
1. Specifications are guaranteed by characterization.						
2. LNA is automatically switched to higher current mode for optimum sensitivity in weak signal conditions.						
3. For input pins SCLK, SEN, SDIO, RST, RCLK, DCLK, DFS, GPO1, GPO2, and GPO3.						
4. For output pins SDIO, DOUT, GPO1, GPO2, and GPO3.						

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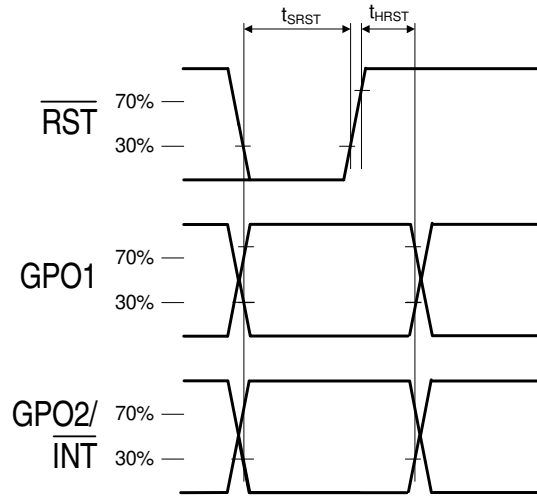
**Table 4. Reset Timing Characteristics<sup>1,2,3</sup>**

( $V_{DD} = 2.7$  to  $5.5$  V,  $V_{IO} = 1.85$  to  $3.6$  V,  $T_A = -20$  to  $85$  °C)

Parameter	Symbol	Min	Typ	Max	Unit
$\overline{RST}$ Pulse Width and GPO1, GPO2/ $\overline{INT}$ Setup to $\overline{RST}\uparrow^4$	$t_{SRST}$	100	—	—	$\mu s$
GPO1, GPO2/ $\overline{INT}$ Hold from $\overline{RST}\uparrow$	$t_{HRST}$	30	—	—	ns

**Important Notes:**

1. When selecting 2-wire mode, the user must ensure that a 2-wire start condition (falling edge of SDIO while SCLK is high) does not occur within 300 ns before the rising edge of  $\overline{RST}$ .
2. When selecting 2-wire mode, the user must ensure that SCLK is high during the rising edge of  $\overline{RST}$ , and stays high until after the first start condition.
3. When selecting 3-wire or SPI modes, the user must ensure that a rising edge of SCLK does not occur within 300 ns before the rising edge of  $\overline{RST}$ .
4. If GPO1 and GPO2 are actively driven by the user, then minimum  $t_{SRST}$  is only 30 ns. If GPO1 or GPO2 is hi-Z, then minimum  $t_{SRST}$  is 100  $\mu s$ , to provide time for on-chip 1 M $\Omega$  devices (active while  $\overline{RST}$  is low) to pull GPO1 high and GPO2 low.



**Figure 1. Reset Timing Parameters for Busmode Select**

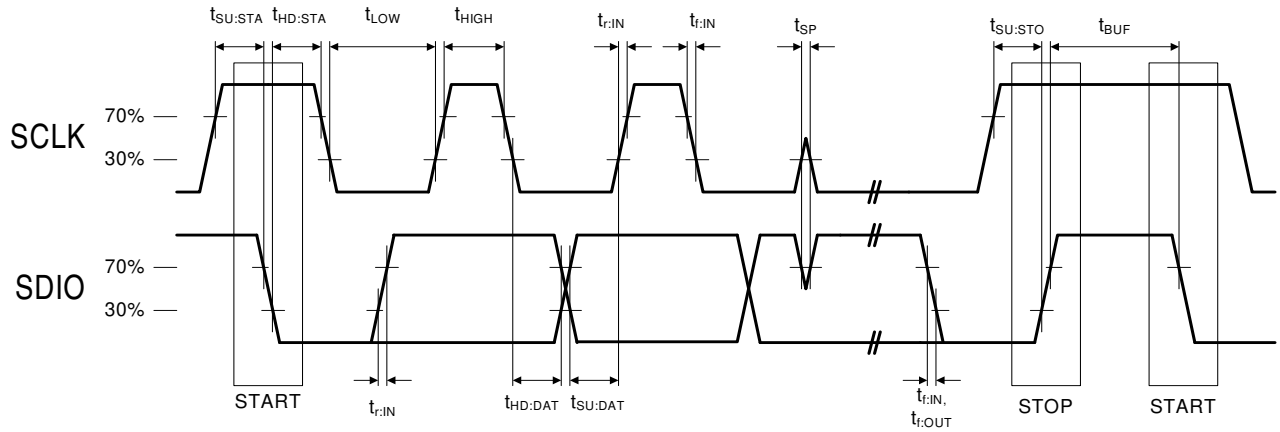
**Table 5. 2-Wire Control Interface Characteristics<sup>1,2,3</sup>**(V<sub>DD</sub> = 2.7 to 5.5 V, V<sub>IO</sub> = 1.85 to 3.6 V, T<sub>A</sub> = -20 to 85 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
SCLK Frequency	f <sub>SCL</sub>		0	—	400	kHz
SCLK Low Time	t <sub>LOW</sub>		1.3	—	—	μs
SCLK High Time	t <sub>HIGH</sub>		0.6	—	—	μs
SCLK Input to SDIO ↓ Setup (START)	t <sub>SU:STA</sub>		0.6	—	—	μs
SCLK Input to SDIO ↓ Hold (START)	t <sub>HD:STA</sub>		0.6	—	—	μs
SDIO Input to SCLK ↑ Setup	t <sub>SU:DAT</sub>		100	—	—	ns
SDIO Input to SCLK ↓ Hold <sup>4,5</sup>	t <sub>HD:DAT</sub>		0	—	900	ns
SCLK input to SDIO ↑ Setup (STOP)	t <sub>SU:STO</sub>		0.6	—	—	μs
STOP to START Time	t <sub>BUF</sub>		1.3	—	—	μs
SDIO Output Fall Time	t <sub>f:OUT</sub>		$20 + 0.1 \frac{C_b}{1\text{pF}}$	—	250	ns
SDIO Input, SCLK Rise/Fall Time	t <sub>f:IN</sub> t <sub>r:IN</sub>		$20 + 0.1 \frac{C_b}{1\text{pF}}$	—	300	ns
SCLK, SDIO Capacitive Loading	C <sub>b</sub>		—	—	50	pF
Input Filter Pulse Suppression	t <sub>SP</sub>		—	—	50	ns

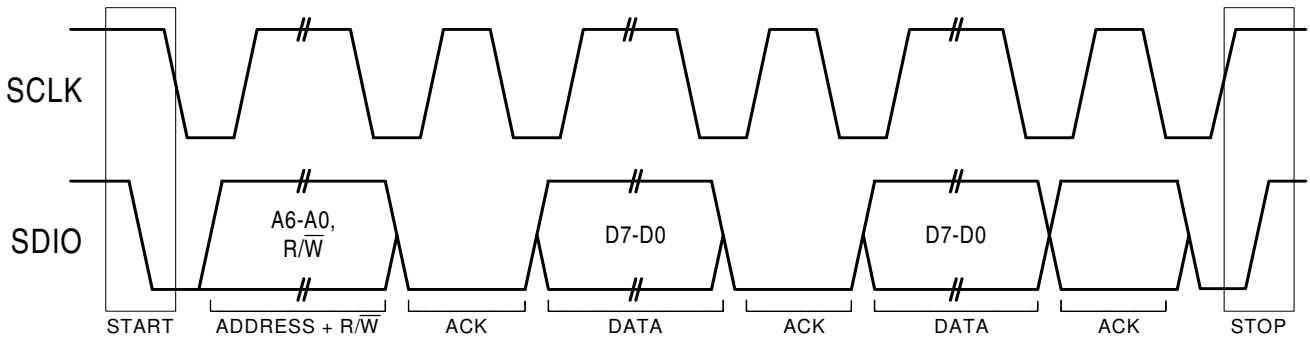
**Notes:**

1. When V<sub>IO</sub> = 0 V, SCLK and SDIO are low impedance.
2. When selecting 2-wire mode, the user must ensure that a 2-wire start condition (falling edge of SDIO while SCLK is high) does not occur within 300 ns before the rising edge of  $\overline{\text{RST}}$ .
3. When selecting 2-wire mode, the user must ensure that SCLK is high during the rising edge of  $\overline{\text{RST}}$ , and stays high until after the first start condition.
4. The Si4734/35 delays SDIO by a minimum of 300 ns from the V<sub>IH</sub> threshold of SCLK to comply with the minimum t<sub>HD:DAT</sub> specification.
5. The maximum t<sub>HD:DAT</sub> has only to be met when f<sub>SCL</sub> = 400 kHz. At frequencies below 400 KHz, t<sub>HD:DAT</sub> may be violated as long as all other timing parameters are met.





**Figure 2. 2-Wire Control Interface Read and Write Timing Parameters**



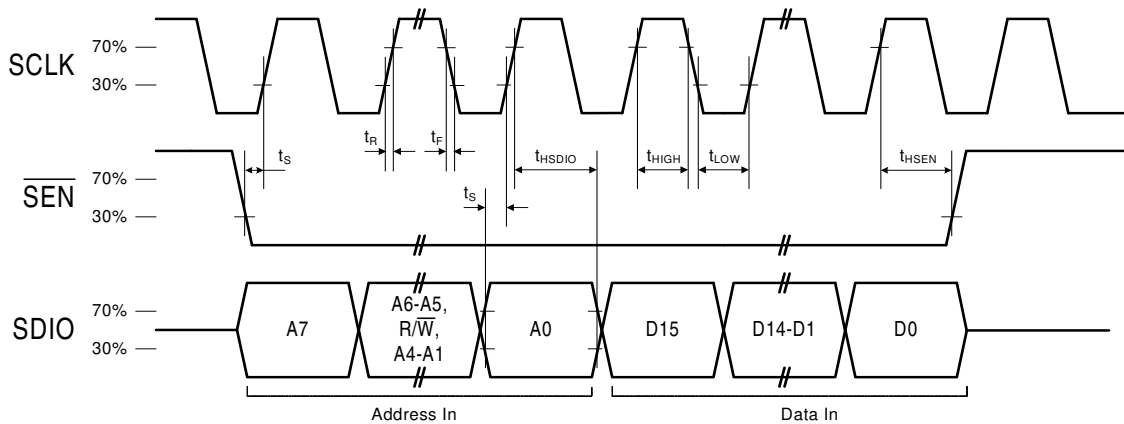
**Figure 3. 2-Wire Control Interface Read and Write Timing Diagram**

**Table 6. 3-Wire Control Interface Characteristics**

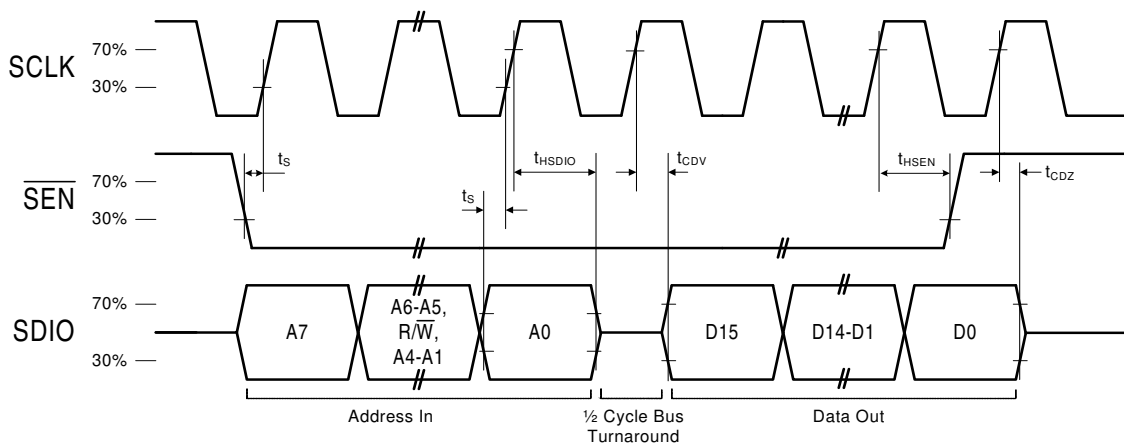
( $V_{DD} = 2.7$  to  $5.5$  V,  $V_{IO} = 1.85$  to  $3.6$  V,  $T_A = -20$  to  $85$  °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
SCLK Frequency	$f_{CLK}$		0	—	2.5	MHz
SCLK High Time	$t_{HIGH}$		25	—	—	ns
SCLK Low Time	$t_{LOW}$		25	—	—	ns
SDIO Input, $\overline{SEN}$ to SCLK $\uparrow$ Setup	$t_S$		20	—	—	ns
SDIO Input to SCLK $\uparrow$ Hold	$t_{HSDIO}$		10	—	—	ns
$\overline{SEN}$ Input to SCLK $\downarrow$ Hold	$t_{HSEN}$		10	—	—	ns
SCLK $\uparrow$ to SDIO Output Valid	$t_{CDV}$	Read	2	—	25	ns
SCLK $\uparrow$ to SDIO Output High Z	$t_{CDZ}$	Read	2	—	25	ns
SCLK, $\overline{SEN}$ , SDIO, Rise/Fall time	$t_R, t_F$		—	—	10	ns

**Note:** When selecting 3-wire mode, the user must ensure that a rising edge of SCLK does not occur within 300 ns before the rising edge of RST.



**Figure 4. 3-Wire Control Interface Write Timing Parameters**



**Figure 5. 3-Wire Control Interface Read Timing Parameters**

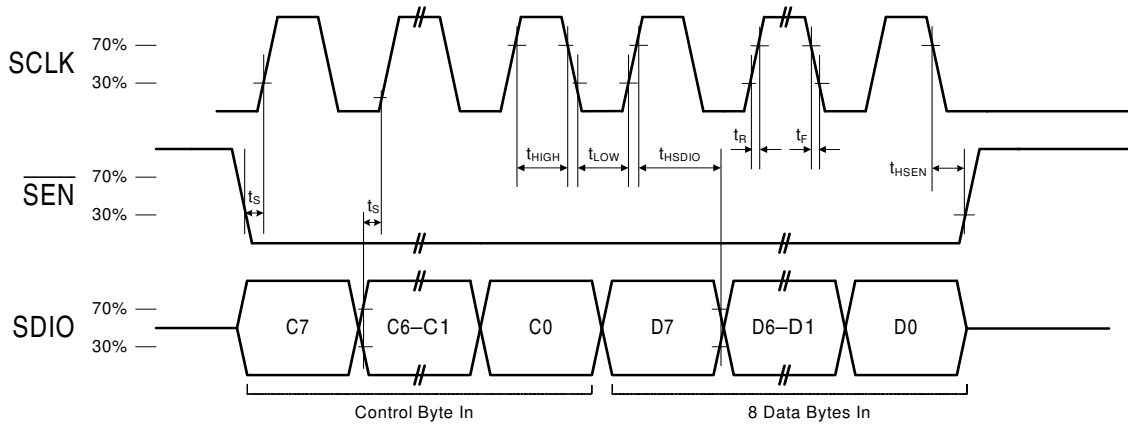
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**Table 7. SPI Control Interface Characteristics**

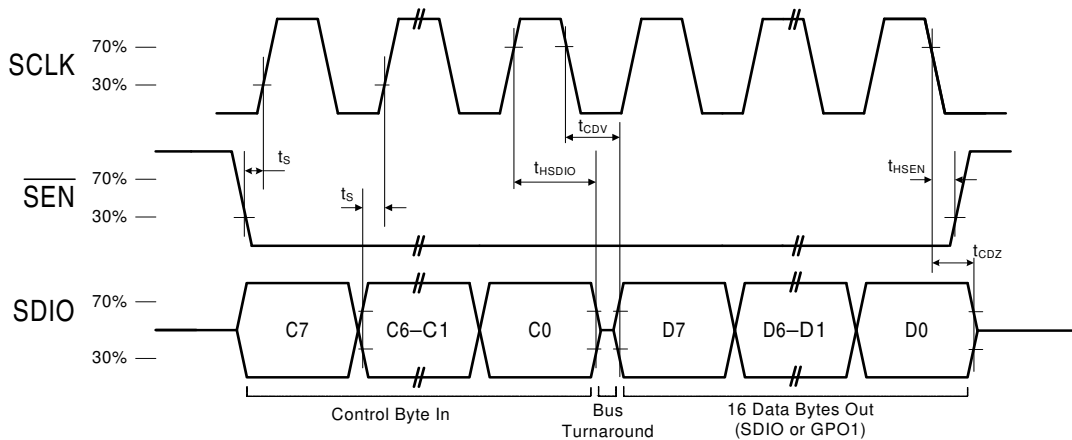
( $V_{DD} = 2.7$  to  $5.5$  V,  $V_{IO} = 1.85$  to  $3.6$  V,  $T_A = -20$  to  $85$  °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
SCLK Frequency	$f_{CLK}$		0	—	2.5	MHz
SCLK High Time	$t_{HIGH}$		25	—	—	ns
SCLK Low Time	$t_{LOW}$		25	—	—	ns
SDIO Input, $\overline{SEN}$ to SCLK $\uparrow$ Setup	$t_S$		15	—	—	ns
SDIO Input to SCLK $\uparrow$ Hold	$t_{HSDIO}$		10	—	—	ns
$\overline{SEN}$ Input to SCLK $\downarrow$ Hold	$t_{HSEN}$		5	—	—	ns
SCLK $\downarrow$ to SDIO Output Valid	$t_{CDV}$	Read	2	—	25	ns
SCLK $\downarrow$ to SDIO Output High Z	$t_{CDZ}$	Read	2	—	25	ns
SCLK, $\overline{SEN}$ , SDIO, Rise/Fall time	$t_R, t_F$		—	—	10	ns

**Note:** When selecting SPI mode, the user must ensure that a rising edge of SCLK does not occur within 300 ns before the rising edge of RST.



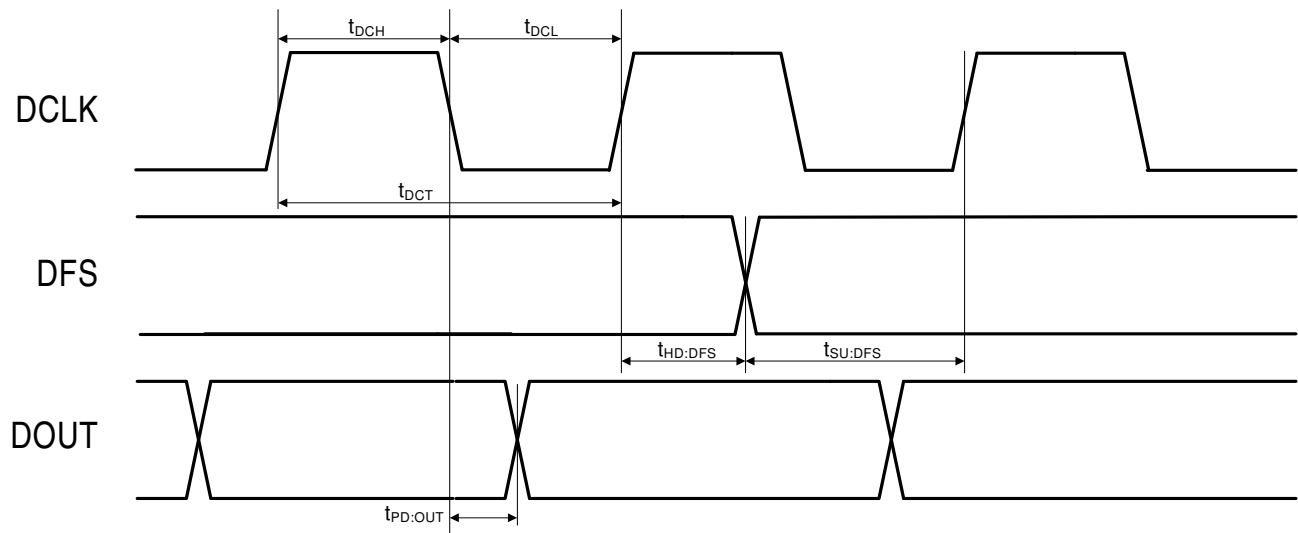
**Figure 6. SPI Control Interface Write Timing Parameters**



**Figure 7. SPI Control Interface Read Timing Parameters**

**Table 8. Digital Audio Interface Characteristics** $(V_{DD} = 2.7 \text{ to } 5.5 \text{ V}, V_{IO} = 1.85 \text{ to } 3.6 \text{ V}, T_A = -20 \text{ to } 85 \text{ }^\circ\text{C})$ 

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
DCLK Cycle Time	$t_{DCT}$		26	—	1000	ns
DCLK Pulse Width High	$t_{DCH}$		10	—	—	ns
DCLK Pulse Width Low	$t_{DCL}$		10	—	—	ns
DFS Set-up Time to DCLK Rising Edge	$t_{SU:DFS}$		5	—	—	ns
DFS Hold Time from DCLK Rising Edge	$t_{HD:DFS}$		5	—	—	ns
DOUT Propagation Delay from DCLK Falling Edge	$t_{PD:DOUT}$		0	—	12	ns

**Figure 8. Digital Audio Interface Timing Parameters, I<sup>2</sup>S Mode**

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**Table 9. FM Receiver Characteristics<sup>1,2</sup>**

( $V_{DD} = 2.7$  to  $5.5$  V,  $V_{IO} = 1.85$  to  $3.6$  V,  $T_A = -20$  to  $85$  °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Input Frequency	$f_{RF}$		76	—	108	MHz
Sensitivity with Headphone Network <sup>3,4,5</sup>		(S+N)/N = 26 dB	—	2.2	3.5	$\mu$ V EMF
Sensitivity with 50 $\Omega$ Network <sup>3,4,5,6</sup>		(S+N)/N = 26 dB	—	1.1	—	$\mu$ V EMF
RDS Sensitivity <sup>6</sup>		$\Delta f = 2$ kHz, RDS BLER < 5%	—	15	—	$\mu$ V EMF
LNA Input Resistance <sup>6,7</sup>			3	4	5	k $\Omega$
LNA Input Capacitance <sup>6,7</sup>			4	5	6	pF
Input IP3 <sup>6,8</sup>			100	105	—	dB $\mu$ V EMF
AM Suppression <sup>3,4,6,7</sup>		m = 0.3	40	50	—	dB
Adjacent Channel Selectivity		$\pm 200$ kHz	35	50	—	dB
Alternate Channel Selectivity		$\pm 400$ kHz	60	70	—	dB
Spurious Response Rejection <sup>6</sup>		In-band	35	—	—	dB
Audio Output Voltage <sup>3,4,7</sup>			72	80	90	mV <sub>RMS</sub>
Audio Output L/R Imbalance <sup>3,7,9</sup>			—	—	1	dB
Audio Frequency Response Low <sup>6</sup>		-3 dB	—	—	30	Hz
Audio Frequency Response High <sup>6</sup>		-3 dB	15	—	—	kHz
Audio Stereo Separation <sup>7,9</sup>			32	42	—	dB
Audio Mono S/N <sup>3,4,5,7,10</sup>			55	63	—	dB
Audio Stereo S/N <sup>4,5,6,7,10,11</sup>			—	58	—	dB
Audio THD <sup>3,7,9</sup>			—	0.1	0.5	%
De-emphasis Time Constant <sup>6</sup>		FM_DEEMPHASIS = 2	70	75	80	$\mu$ s
		FM_DEEMPHASIS = 1	45	50	54	$\mu$ s

**Notes:**

1. Additional testing information is available in “AN388: Si470x/1x/2x/3x/4x Evaluation Board Test Procedure.” Volume = maximum for all tests. Tested at RF = 98.1 MHz.
2. To ensure proper operation and receiver performance, follow the guidelines in “AN383: Si47xx Antenna, Schematic, Layout, and Design Guidelines.” Silicon Laboratories will evaluate schematics and layouts for qualified customers.
3.  $F_{MOD} = 1$  kHz, 75  $\mu$ s de-emphasis, MONO = enabled, and L = R unless noted otherwise.
4.  $\Delta f = 22.5$  kHz.
5.  $B_{AF} = 300$  Hz to 15 kHz, A-weighted.
6. Guaranteed by characterization.
7.  $V_{EMF} = 1$  mV.
8.  $|f_2 - f_1| > 2$  MHz,  $f_0 = 2 \times f_1 - f_2$ . AGC is disabled.
9.  $\Delta f = 75$  kHz.
10. At  $L_{OUT}$  and  $R_{OUT}$  pins.
11. Analog audio output mode.
12. Blocker Amplitude = 100 dB $\mu$ V
13. Sensitivity measured at (S+N)/N = 26 dB.
14. At temperature (25°C).

**Table 9. FM Receiver Characteristics<sup>1,2</sup> (Continued)**(V<sub>DD</sub> = 2.7 to 5.5 V, V<sub>IO</sub> = 1.85 to 3.6 V, T<sub>A</sub> = -20 to 85 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Blocking Sensitivity <sup>3,4,5,6,12,13</sup>		$\Delta f = \pm 400$ kHz	—	32	—	dB $\mu$ V
		$\Delta f = \pm 4$ MHz	—	38	—	dB $\mu$ V
Intermode Sensitivity <sup>3,4,5,6,12,13</sup>		$\Delta f = \pm 400$ kHz, $\pm 800$ kHz	—	40	—	dB $\mu$ V
		$\Delta f = \pm 4$ MHz, $\pm 8$ MHz	—	35	—	dB $\mu$ V
Audio Output Load Resistance <sup>6,10</sup>	R <sub>L</sub>	Single-ended	10	—	—	k $\Omega$
Audio Output Load Capacitance <sup>6,10</sup>	C <sub>L</sub>	Single-ended	—	—	50	pF
Seek/Tune Time <sup>6</sup>		RCLK tolerance = 100 ppm	—	—	60	ms/channel
Powerup Time <sup>6</sup>		From powerdown	—	—	110	ms
RSSI Offset <sup>14</sup>		Input levels of 8 and 60 dB $\mu$ V at RF Input	-3	—	3	dB

**Notes:**

1. Additional testing information is available in "AN388: Si470x/1x/2x/3x/4x Evaluation Board Test Procedure." Volume = maximum for all tests. Tested at RF = 98.1 MHz.
2. To ensure proper operation and receiver performance, follow the guidelines in "AN383: Si47xx Antenna, Schematic, Layout, and Design Guidelines." Silicon Laboratories will evaluate schematics and layouts for qualified customers.
3. F<sub>MOD</sub> = 1 kHz, 75  $\mu$ s de-emphasis, MONO = enabled, and L = R unless noted otherwise.
4.  $\Delta f = 22.5$  kHz.
5. B<sub>AF</sub> = 300 Hz to 15 kHz, A-weighted.
6. Guaranteed by characterization.
7. V<sub>EMF</sub> = 1 mV.
8.  $|f_2 - f_1| > 2$  MHz,  $f_0 = 2 \times f_1 - f_2$ . AGC is disabled.
9.  $\Delta f = 75$  kHz.
10. At L<sub>OUT</sub> and R<sub>OUT</sub> pins.
11. Analog audio output mode.
12. Blocker Amplitude = 100 dB $\mu$ V
13. Sensitivity measured at (S+N)/N = 26 dB.
14. At temperature (25°C).

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**Table 10. 64–75.9 MHz Input Frequency FM Receiver Characteristics<sup>1,2,6</sup>**

( $V_{DD} = 2.7$  to  $5.5$  V,  $V_{IO} = 1.85$  to  $3.6$  V,  $T_A = -20$  to  $85$  °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Input Frequency	$f_{RF}$		64	—	75.9	MHz
Sensitivity with Headphone Network <sup>3,4,5</sup>		(S+N)/N = 26 dB	—	4.0	—	$\mu$ V EMF
LNA Input Resistance <sup>7</sup>			3	4	5	k $\Omega$
LNA Input Capacitance <sup>7</sup>			4	5	6	pF
Input IP3 <sup>8</sup>			100	105	—	dB $\mu$ V EMF
AM Suppression <sup>3,4,7</sup>		m = 0.3	40	50	—	dB
Adjacent Channel Selectivity		$\pm$ 200 kHz	—	50	—	dB
Alternate Channel Selectivity		$\pm$ 400 kHz	—	70	—	dB
Audio Output Voltage <sup>3,4,7</sup>			72	80	90	mV <sub>RMS</sub>
Audio Output L/R Imbalance <sup>3,7,9</sup>			—	—	1	dB
Audio Frequency Response Low		–3 dB	—	—	30	Hz
Audio Frequency Response High		–3 dB	15	—	—	kHz
Audio Mono S/N <sup>3,4,5,7,10</sup>			55	63	—	dB
Audio THD <sup>3,7,9</sup>			—	0.1	0.5	%
De-emphasis Time Constant		FM_DEEMPHASIS = 2	70	75	80	$\mu$ s
		FM_DEEMPHASIS = 1	45	50	54	$\mu$ s
Audio Output Load Resistance <sup>10</sup>	$R_L$	Single-ended	10	—	—	k $\Omega$
Audio Output Load Capacitance <sup>10</sup>	$C_L$	Single-ended	—	—	50	pF
Seek/Tune Time		RCLK tolerance = 100 ppm	—	—	60	ms/channel
Powerup Time		From powerdown	—	—	110	ms
RSSI Offset <sup>11</sup>		Input levels of 8 and 60 dB $\mu$ V EMF	–3	—	3	dB

**Notes:**

1. Additional testing information is available in “AN388: Si470x/1x/2x/3x/4x Evaluation Board Test Procedure.” Volume = maximum for all tests. Tested at RF = 98.1 MHz.
2. To ensure proper operation and receiver performance, follow the guidelines in “AN383: Si47xx Antenna, Schematic, Layout, and Design Guidelines.” Silicon Laboratories will evaluate schematics and layouts for qualified customers.
3.  $F_{MOD} = 1$  kHz, 75  $\mu$ s de-emphasis, MONO = enabled, and L = R unless noted otherwise.
4.  $\Delta f = 22.5$  kHz.
5.  $B_{AF} = 300$  Hz to 15 kHz, A-weighted.
6. Guaranteed by characterization.
7.  $V_{EMF} = 1$  mV.
8.  $|f_2 - f_1| > 2$  MHz,  $f_0 = 2 \times f_1 - f_2$ . AGC is disabled.
9.  $\Delta f = 75$  kHz.
10. At  $L_{OUT}$  and  $R_{OUT}$  pins.
11. At temperature (25 °C).

**Table 11. AM/SW/LW Receiver Characteristics<sup>1,2</sup>**(V<sub>DD</sub> = 2.7 to 5.5 V, V<sub>IO</sub> = 1.85 to 3.6 V, TA = -20 to 85 °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Input Frequency	f <sub>RF</sub>	Long Wave (LW)	153	—	279	kHz
		Medium Wave (AM)	520	—	1710	kHz
		Short Wave (SW)	2.3	—	26.1	MHz
Sensitivity <sup>3,4,5,6</sup>		(S+N)/N = 26 dB	—	25	35	μV EMF
Large Signal Voltage Handling <sup>4,6,7</sup>		THD < 8%	—	300	—	mV <sub>RMS</sub>
Power Supply Rejection Ratio <sup>6</sup>		ΔV <sub>DD</sub> = 100 mV <sub>RMS</sub> , 100 Hz	—	40	—	dB
Audio Output Voltage <sup>3,4,8</sup>			54	60	67	mV <sub>RMS</sub>
Audio S/N <sup>3,4,5,8</sup>			50	56	—	dB
Audio THD <sup>3,4,8</sup>			—	0.1	0.5	%
Antenna Inductance <sup>6,9</sup>		Long Wave (LW)	—	2800	—	μH
		Medium Wave (AM)	180	—	450	μH
Powerup Time <sup>6</sup>		From powerdown	—	—	110	ms

**Notes:**

1. Additional testing information is available in “AN388: Si470x/1x/2x/3x/4x Evaluation Board Test Procedure.” Volume = maximum for all tests. Tested at RF = 520 kHz.
2. To ensure proper operation and receiver performance, follow the guidelines in “AN383: Si47xx Antenna, Schematic, Layout, and Design Guidelines.” Silicon Laboratories will evaluate schematics and layouts for qualified customers.
3. F<sub>MOD</sub> = 1 kHz, 30% modulation, 2 kHz channel filter.
4. Analog audio output mode.
5. B<sub>AF</sub> = 300 Hz to 15 kHz, A-weighted.
6. Guaranteed by characterization.
7. See “AN388: Si470x/1x/2x/3x/4x Evaluation Board Test Procedure” for evaluation method.
8. V<sub>IN</sub> = 5 mV<sub>rms</sub>.
9. Stray capacitance on antenna and board must be < 10 pF to achieve full tuning range at higher inductance levels.



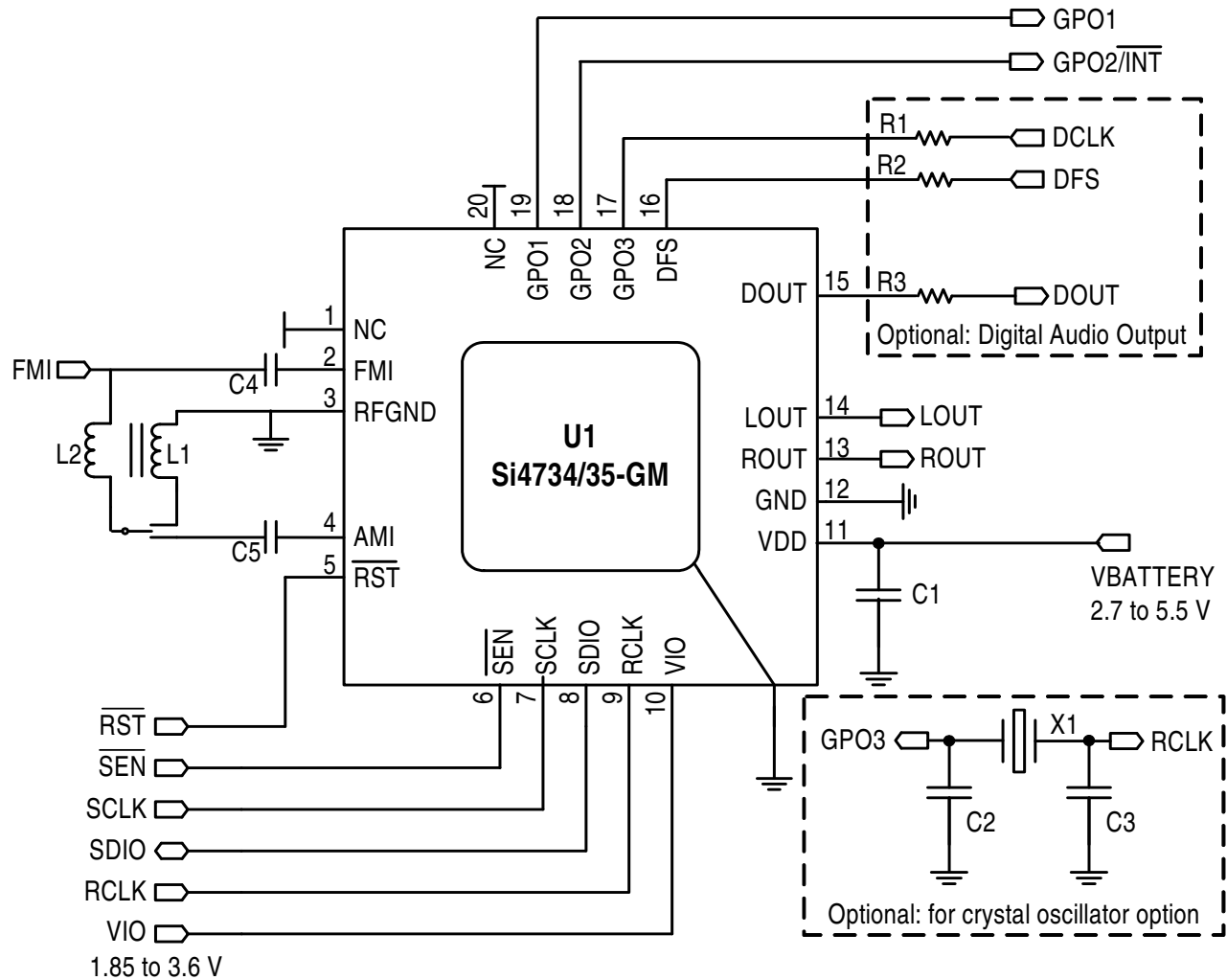
# Si4734/35-C40

**Table 12. Reference Clock and Crystal Characteristics**

( $V_{DD} = 2.7$  to  $5.5$  V,  $V_{IO} = 1.85$  to  $3.6$  V,  $T_A = -20$  to  $85$  °C)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
<b>Reference Clock</b>						
RCLK Supported Frequencies <sup>1</sup>			31.130	32.768	40,000	kHz
RCLK Frequency Tolerance <sup>2</sup>			-100	—	100	ppm
REFCLK_PRESCALE			1	—	4095	
REFCLK			31.130	32.768	34.406	kHz
<b>Crystal Oscillator</b>						
Crystal Oscillator Frequency			—	32.768	—	kHz
Crystal Frequency Tolerance <sup>2</sup>			-100	—	100	ppm
Board Capacitance			—	—	3.5	pF
<b>Notes:</b>						
1. The Si4734/35 divides the RCLK input by REFCLK_PRESCALE to obtain REFCLK. There are some RCLK frequencies between 31.130 kHz and 40 MHz that are not supported. For more details, see Table 6 of “AN332: Si47xx Programming Guide”.						
2. A frequency tolerance of $\pm 50$ ppm is required for FM seek/tune using 50 kHz channel spacing and AM seek/tune in SW frequencies.						

## 2. Typical Application Schematic (QFN)

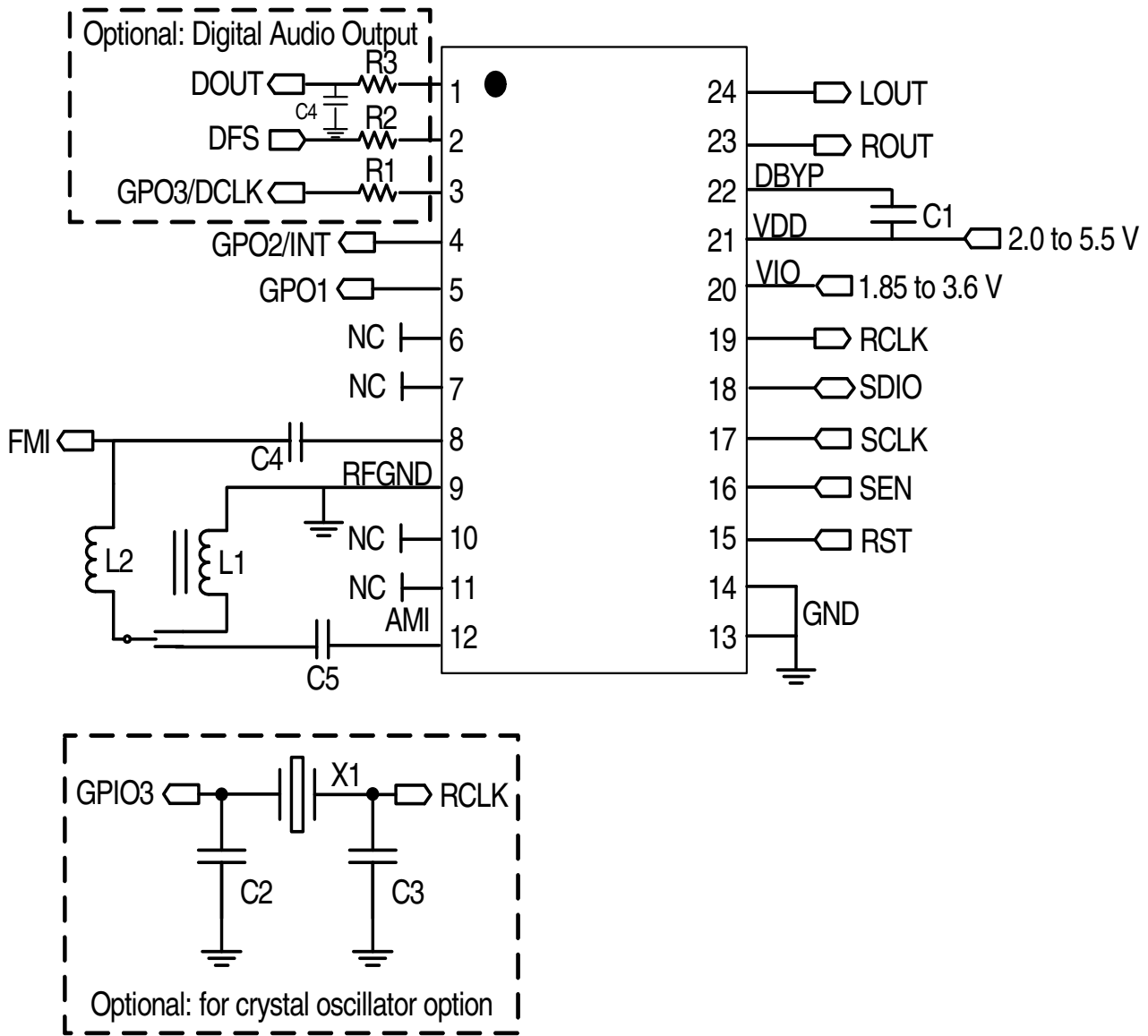


### Notes:

1. Place C1 close to V<sub>DD</sub> pin.
2. All grounds connect directly to GND plane on PCB.
3. Pins 1 and 20 are no connects, leave floating.
4. To ensure proper operation and receiver performance, follow the guidelines in "AN383: Si47xx Antenna, Schematic, Layout, and Design Guidelines." Silicon Laboratories will evaluate schematics and layouts for qualified customers.
5. Pin 2 connects to the FM antenna interface, and pin 4 connects to the AM antenna interface.
6. Place Si4734/35 as close as possible to antenna and keep the FMI and AMI traces as short as possible.

# Si4734/35-C40

## 3. Typical Application Schematic (SSOP)



### Notes:

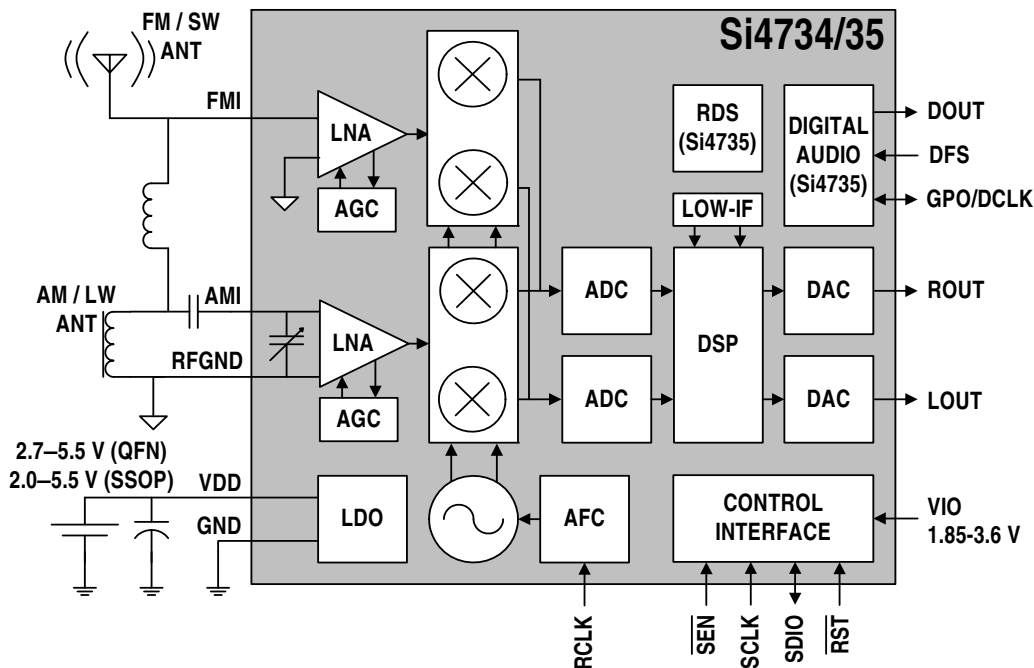
1. Place C1 close to V<sub>DD</sub> and DBYP pins.
2. All grounds connect directly to GND plane on PCB.
3. Pins 6 and 7 are no connects, leave floating.
4. Pins 10 and 11 are unused. Tie these pins to GND.
5. To ensure proper operation and receiver performance, follow the guidelines in "AN383: Si47xx Antenna, Schematic, Layout, and Design Guidelines." Silicon Laboratories will evaluate schematics and layouts for qualified customers.
6. Pin 8 connects to the FM antenna interface, and pin 12 connects to the AM antenna interface.
7. Place Si4734/35 as close as possible to antenna and keep the FMI and AMI traces as short as possible.

#### 4. Bill of Materials (QFN/SSOP)

Component(s)	Value/Description	Supplier
C1	Supply bypass capacitor, 22 nF, $\pm 20\%$ , Z5U/X7R	Murata
C5	Coupling capacitor, 0.47 $\mu$ F, $\pm 20\%$ , Z5U/X7R	Murata
L1	Ferrite loop stick, 180–450 $\mu$ H	Jiaxin
L2	4.7 $\mu$ H	Coilcraft
U1	Si4734/35 AM/FM Radio Tuner	Silicon Laboratories
<b>Optional Components</b>		
C2, C3	Crystal load capacitors, 22 pF, $\pm 5\%$ , COG (Optional for crystal oscillator option)	Venkel
C4	Noise mitigating capacitor, 2~5 pF (Optional for digital audio)	Murata
X1	32.768 kHz crystal (Optional for crystal oscillator option)	Epson
R1	Resistor, 2 k $\Omega$ (Optional for digital audio)	Venkel
R2	Resistor, 2 k $\Omega$ (Optional for digital audio)	Venkel
R3	Resistor, 600 $\Omega$ (Optional for digital audio)	Venkel

## 5. Functional Description

### 5.1. Overview



**Figure 9. Functional Block Diagram**

The Si4734/35 is the industry's first fully integrated, 100% CMOS AM/FM/SW/LW radio receiver IC. Offering unmatched integration and PCB space savings, the Si4734/35 requires minimal external components and less than 20 mm<sup>2</sup> of board area, excluding the antenna inputs. The Si4734/35 AM/FM/SW/LW radio provides the space savings and low power consumption necessary for portable devices while delivering the high performance and design simplicity desired for all AM/FM/SW/LW solutions.

Leveraging Silicon Laboratories' proven and patented Si4700/01 FM tuner's digital low intermediate frequency (low-IF) receiver architecture, the Si4734/35 delivers superior RF performance and interference rejection in AM, FM, and short wave and long wave bands. The high integration and complete system production test simplifies design-in, increases system quality, and improves manufacturability.

The Si4734/35 is a feature-rich solution including advanced seek algorithms, soft mute, auto-calibrated digital tuning, and FM stereo processing. In addition, the Si4734/35 provides analog or digital audio output and a programmable reference clock. The device supports

I<sup>2</sup>C-compatible 2-wire control interface, SPI, and a Si4700/01 backwards-compatible 3-wire control interface.

The Si4734/35 utilizes digital processing to achieve high fidelity, optimal performance, and design flexibility. The chip provides excellent pilot rejection, selectivity, and unmatched audio performance, and offers both the manufacturer and the end-user extensive programmability and flexibility in the listening experience.

The Si4735 incorporates a digital processor for the European Radio Data System (RDS) and the North American Radio Broadcast Data System (RBDS), including all required symbol decoding, block synchronization, error detection, and error correction functions. Using RDS, the Si4735 enables broadcast data such as station identification and song name to be displayed to the user.

## 5.2. Operating Modes

The Si4734/35 operates in either an FM receive or an AM/SW/LW receive mode. In FM mode, radio signals are received on FMI and processed by the FM front-end circuitry. In AM/SW/LW mode, radio signals are received on AMI and processed by the AM front-end circuitry. In addition to the receiver mode, there is a clocking mode to choose to clock the Si4734/35 from a reference clock or crystal. On the Si4735, there is an audio output mode to choose between an analog and/or digital audio output. In the analog audio output mode, ROUT and LOU are used for the audio output pins. In the digital audio mode, DOUT, DFS, and DCLK pins are used. Concurrent analog/digital audio output mode is also available requiring all five pins. The receiver mode and the audio output mode are set by the POWER\_UP command listed in Table 14, “Selected Si473x Commands,” on page 27.

## 5.3. FM Receiver

The Si4734/35 FM receiver is based on the proven Si4700/01 FM tuner. The receiver uses a digital low-IF architecture allowing the elimination of external components and factory adjustments. The Si4734/35 integrates a low noise amplifier (LNA) supporting the worldwide FM broadcast band (64 to 108 MHz). An AGC circuit controls the gain of the LNA to optimize sensitivity and rejection of strong interferers. An image-reject mixer downconverts the RF signal to low-IF. The quadrature mixer output is amplified, filtered, and digitized with high resolution analog-to-digital converters (ADCs). This advanced architecture allows the Si4734/35 to perform channel selection, FM demodulation, and stereo audio processing to achieve superior performance compared to traditional analog architectures.

## 5.4. AM Receiver

The highly-integrated Si4734/35 supports worldwide AM band reception from 520 to 1710 kHz using a digital low-IF architecture with a minimum number of external components and no manual alignment required. This digital low-IF architecture allows for high-precision filtering offering excellent selectivity and SNR with minimum variation across the AM band. The DSP also provides adjustable channel step sizes in 1 kHz increments, AM demodulation, soft mute, seven different channel bandwidth filters, and additional features, such as a programmable automatic volume control (AVC) maximum gain allowing users to adjust the level of background noise. Similar to the FM receiver, the integrated LNA and AGC optimize sensitivity and rejection of strong interferers allowing better reception of weak stations.

The Si4734/35 provides highly-accurate digital AM tuning without factory adjustments. To offer maximum flexibility, the receiver supports a wide range of ferrite loop sticks from 180–450  $\mu$ H. An air loop antenna is supported by using a transformer to increase the effective inductance from the air loop. Using a 1:5 turn ratio inductor, the inductance is increased by 25 times and easily supports all typical AM air loop antennas which generally vary between 10 and 20  $\mu$ H.

## 5.5. SW Receiver

The Si4734/35 is the first fully integrated IC to support AM and FM, as well as short wave (SW) band reception from 2.3 to 26.1 MHz fully covering the 120 meter to 11 meter bands. The Si4734/35 offers extensive shortwave features such as continuous digital tuning with minimal discrete components and no factory adjustments. Other SW features include adjustable channel step sizes in 1 kHz increments, adjustable channel bandwidth settings, advanced seek algorithm, and soft mute.

The Si4734/35 uses the FM antenna to capture short wave signals. These signals are then fed directly into the AMI pin in a wide band configuration. See “AN382: Si4734/35 Designer’s Guide” for more details.

## 5.6. LW Receiver

The Si4734/35 supports the long wave (LW) band from 153 to 279 kHz. The highly integrated Si4734/35 offers continuous digital tuning with minimal discrete components and no factory adjustments. The Si4734/35 also offers adjustable channel step sizes in 1 kHz increments, adjustable channel bandwidth settings, advanced seek algorithm, and soft mute.

The Si4734/35 uses a separate ferrite bar antenna to capture long wave signals.

## 5.7. Digital Audio Interface (Si4735 Only)

The digital audio interface operates in slave mode and supports three different audio data formats:

- I<sup>2</sup>S
- Left-Justified
- DSP Mode

### 5.7.1. Audio Data Formats

In I<sup>2</sup>S mode, by default the MSB is captured on the second rising edge of DCLK following each DFS transition. The remaining bits of the word are sent in order, down to the LSB. The left channel is transferred first when the DFS is low, and the right channel is transferred when the DFS is high.

In Left-Justified mode, by default the MSB is captured on the first rising edge of DCLK following each DFS transition. The remaining bits of the word are sent in

order, down to the LSB. The left channel is transferred first when the DFS is high, and the right channel is transferred when the DFS is low.

In DSP mode, the DFS becomes a pulse with a width of 1DCLK period. The left channel is transferred first, followed right away by the right channel. There are two options in transferring the digital audio data in DSP mode: the MSB of the left channel can be transferred on the first rising edge of DCLK following the DFS pulse or on the second rising edge.

In all audio formats, depending on the word size, DCLK frequency and sample rates, there may be unused DCLK cycles after the LSB of each word before the next DFS transition and MSB of the next word. In addition, if preferred, the user can configure the MSB to be captured on the falling edge of DCLK via properties.

The number of audio bits can be configured for 8, 16, 20, or 24 bits.

## 5.7.2. Audio Sample Rates

The device supports a number of industry-standard sampling rates including 32, 40, 44.1, and 48 kHz. The digital audio interface enables low-power operation by eliminating the need for redundant DACs on the audio baseband processor.

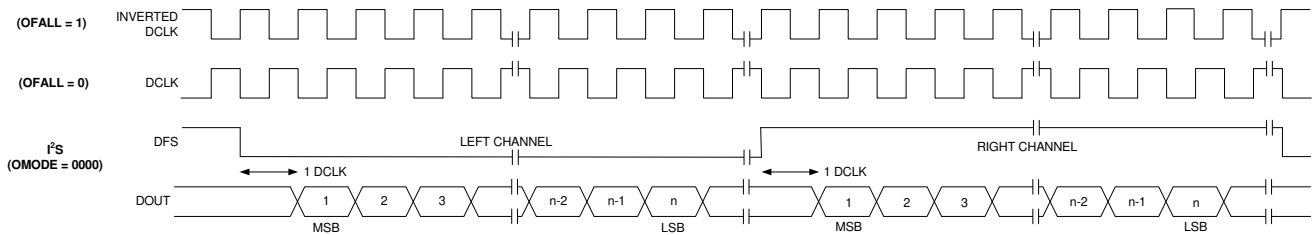


Figure 10. I²S Digital Audio Format

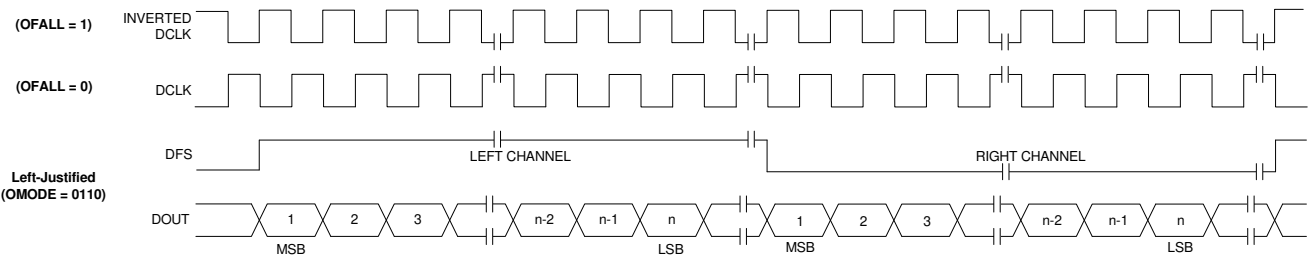


Figure 11. Left-Justified Digital Audio Format

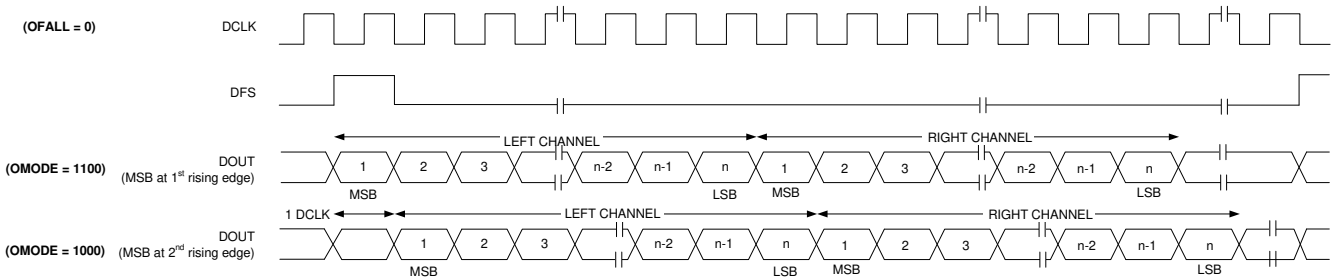


Figure 12. DSP Digital Audio Format

### 5.8. Stereo Audio Processing

The output of the FM demodulator is a stereo multiplexed (MPX) signal. The MPX standard was developed in 1961, and is used worldwide. Today's MPX signal format consists of left + right (L+R) audio, left – right (L–R) audio, a 19 kHz pilot tone, and RDS/RBDS data as shown in Figure 13 below.

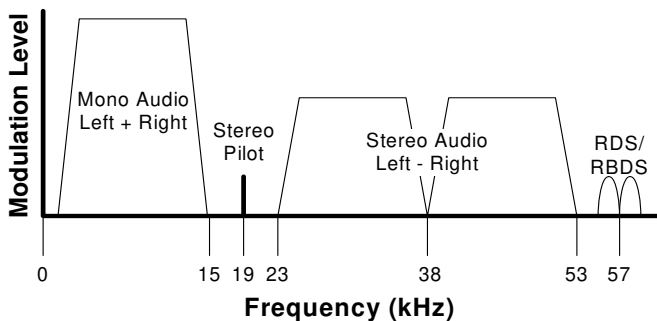


Figure 13. MPX Signal Spectrum

#### 5.8.1. Stereo Decoder

The Si4734/35's integrated stereo decoder automatically decodes the MPX signal using DSP techniques. The 0 to 15 kHz (L+R) signal is the mono output of the FM tuner. Stereo is generated from the (L+R), (L–R), and a 19 kHz pilot tone. The pilot tone is used as a reference to recover the (L–R) signal. Output left and right channels are obtained by adding and subtracting the (L+R) and (L–R) signals respectively. The Si4735 uses frequency information from the 19 kHz stereo pilot to recover the 57 kHz RDS/RBDS signal.

#### 5.8.2. Stereo-Mono Blending

Adaptive noise suppression is employed to gradually combine the stereo left and right audio channels to a mono (L+R) audio signal as the signal quality degrades to maintain optimum sound fidelity under varying reception conditions. Stereo/mono status can be monitored with the FM\_RSQ\_STATUS command. Mono operation can be forced with the FM\_BLEND\_MONO\_THRESHOLD property.



## 5.9. De-emphasis

Pre-emphasis and de-emphasis is a technique used by FM broadcasters to improve the signal-to-noise ratio of FM receivers by reducing the effects of high-frequency interference and noise. When the FM signal is transmitted, a pre-emphasis filter is applied to accentuate the high audio frequencies. The Si4734/35 incorporates a de-emphasis filter which attenuates high frequencies to restore a flat frequency response. Two time constants are used in various regions. The de-emphasis time constant is programmable to 50 or 75  $\mu$ s and is set by the FM\_DEEMPHASIS property.

## 5.10. Stereo DAC

High-fidelity stereo digital-to-analog converters (DACs) drive analog audio signals onto the LOUT and ROUT pins. The audio output may be muted. Volume is adjusted digitally with the RX\_VOLUME property.

## 5.11. Soft Mute

The soft mute feature is available to attenuate the audio outputs and minimize audible noise in very weak signal conditions. The softmute attenuation level is adjustable using the FM\_SOFT\_MUTE\_MAX\_ATTENUATION and AM\_SOFT\_MUTE\_MAX\_ATTENUATION properties.

## 5.12. RDS/RBDS Processor (Si4735 Only)

The Si4735 implements an RDS/RBDS\* processor for symbol decoding, block synchronization, error detection, and error correction.

The Si4735 device is user configurable and provides an optional interrupt when RDS is synchronized, loses synchronization, and/or the user configurable RDS FIFO threshold has been met.

The Si4735 reports RDS decoder synchronization status and detailed bit errors in the information word for each RDS block with the FM\_RDS\_STATUS command. The range of reportable block errors is 0, 1–2, 3–5, or 6+. More than six errors indicates that the corresponding block information word contains six or more non-correctable errors or that the block checkword contains errors.

**\*Note:** RDS/RBDS is referred to only as RDS throughout the remainder of this document.

## 5.13. Tuning

The tuning frequency is directly programmed using the FM\_TUNE\_FREQ and AM\_TUNE\_FREQ commands. The Si4734/35 supports channel spacing steps of 10 kHz in FM mode and 1 kHz in AM/SW/LW mode.

## 5.14. Seek

Seek tuning will search up or down for a valid channel. Valid channels are found when the receive signal strength indicator (RSSI) and the signal-to-noise ratio (SNR) values exceed the set threshold. Using the SNR qualifier rather than solely relying on the more traditional RSSI qualifier can reduce false stops and increase the number of valid stations detected. Seek is initiated using the FM\_SEEK\_START and AM\_SEEK\_START commands. The RSSI and SNR threshold settings are adjustable using properties (see Table 15).

## 5.15. Reference Clock

The Si4734/35 reference clock is programmable, supporting RCLK frequencies in Table 12. Refer to Table 3, "DC Characteristics," on page 5 for switching voltage levels and Table 9, "FM Receiver Characteristics," on page 12 for frequency tolerance information.

An onboard crystal oscillator is available to generate the 32.768 kHz reference when an external crystal and load capacitors are provided. Refer to "2. Typical Application Schematic (QFN)" on page 17. This mode is enabled using the POWER\_UP command. Refer to Table 14, "Selected Si473x Commands," on page 27.

The Si4734/35 performance may be affected by data activity on the SDIO bus when using the integrated internal oscillator. SDIO activity results from polling the tuner for status or communicating with other devices that share the SDIO bus. If there is SDIO bus activity while the Si4734/35 is performing the seek/tune function, the crystal oscillator may experience jitter, which may result in mistunes, false stops, and/or lower SNR.

For best seek/tune results, Silicon Laboratories recommends that all SDIO data traffic be suspended during Si4734/35 seek and tune operations. This is achieved by keeping the bus quiet for all other devices on the bus, and delaying tuner polling until the tune or seek operation is complete. The seek/tune complete (STC) interrupt should be used instead of polling to determine when a seek/tune operation is complete.

## 5.16. Control Interface

A serial port slave interface is provided, which allows an external controller to send commands to the Si4734/35 and receive responses from the device. The serial port can operate in three bus modes: 2-wire mode, 3-wire mode, or SPI mode. The Si4734/35 selects the bus mode by sampling the state of the GPO1 and GPO2 pins on the rising edge of RST. The GPO1 pin includes an internal pull-up resistor, which is connected while

$\overline{\text{RST}}$  is low, and the GPO2 pin includes an internal pull-down resistor, which is connected while  $\overline{\text{RST}}$  is low. Therefore, it is only necessary for the user to actively drive pins which differ from these states. See Table 13.

**Table 13. Bus Mode Select on Rising Edge of  $\overline{\text{RST}}$**

Bus Mode	GPO1	GPO2
2-Wire	1	0
SPI	1	1 (must drive)
3-Wire	0 (must drive)	0

After the rising edge of  $\overline{\text{RST}}$ , the pins GPO1 and GPO2 are used as general purpose output (O) pins, as described in Section “5.17. GPO Outputs”. In any bus mode, commands may only be sent after  $V_{\text{IO}}$  and  $V_{\text{DD}}$  supplies are applied.

In any bus mode, before sending a command or reading a response, the user must first read the status byte to ensure that the device is ready (CTS bit is high).

#### 5.16.1. 2-Wire Control Interface Mode

When selecting 2-wire mode, the user must ensure that SCLK is high during the rising edge of  $\overline{\text{RST}}$ , and stays high until after the first start condition. Also, a start condition must not occur within 300 ns before the rising edge of  $\overline{\text{RST}}$ .

The 2-wire bus mode uses only the SCLK and SDIO pins for signaling. A transaction begins with the START condition, which occurs when SDIO falls while SCLK is high. Next, the user drives an 8-bit control word serially on SDIO, which is captured by the device on rising edges of SCLK. The control word consists of a 7-bit device address, followed by a read/write bit (read = 1, write = 0). The Si4734/35 acknowledges the control word by driving SDIO low on the next falling edge of SCLK.

Although the Si4734/35 will respond to only a single device address, this address can be changed with the  $\overline{\text{SEN}}$  pin (note that the  $\overline{\text{SEN}}$  pin is not used for signaling in 2-wire mode). When  $\overline{\text{SEN}} = 0$ , the 7-bit device address is 0010001b. When  $\overline{\text{SEN}} = 1$ , the address is 1100011b.

For write operations, the user then sends an 8-bit data byte on SDIO, which is captured by the device on rising edges of SCLK. The Si4734/35 acknowledges each data byte by driving SDIO low for one cycle, on the next falling edge of SCLK. The user may write up to 8 data bytes in a single 2-wire transaction. The first byte is a command, and the next seven bytes are arguments.

For read operations, after the Si4734/35 has acknowledged the control byte, it will drive an 8-bit data byte on SDIO, changing the state of SDIO on the falling edge of SCLK. The user acknowledges each data byte by driving SDIO low for one cycle, on the next falling edge of SCLK. If a data byte is not acknowledged, the transaction will end. The user may read up to 16 data bytes in a single 2-wire transaction. These bytes contain the response data from the Si4734/35.

A 2-wire transaction ends with the STOP condition, which occurs when SDIO rises while SCLK is high.

For details on timing specifications and diagrams, refer to Table 5, “2-Wire Control Interface Characteristics” on page 7; Figure 2, “2-Wire Control Interface Read and Write Timing Parameters,” on page 8, and Figure 3, “2-Wire Control Interface Read and Write Timing Diagram,” on page 8.

#### 5.16.2. 3-Wire Control Interface Mode

When selecting 3-wire mode, the user must ensure that a rising edge of SCLK does not occur within 300 ns before the rising edge of  $\overline{\text{RST}}$ .

The 3-wire bus mode uses the SCLK, SDIO, and  $\overline{\text{SEN}}$  pins. A transaction begins when the user drives  $\overline{\text{SEN}}$  low. Next, the user drives a 9-bit control word on SDIO, which is captured by the device on rising edges of SCLK. The control word consists of a 9-bit device address ( $A7:A5 = 101b$ ), a read/write bit (read = 1, write = 0), and a 5-bit register address ( $A4:A0$ ).

For write operations, the control word is followed by a 16-bit data word, which is captured by the device on rising edges of SCLK.

For read operations, the control word is followed by a delay of one-half SCLK cycle for bus turn-around. Next, the Si4734/35 will drive the 16-bit read data word serially on SDIO, changing the state of SDIO on each rising edge of SCLK.

A transaction ends when the user sets  $\overline{\text{SEN}}$  high, then pulses SCLK high and low one final time. SCLK may either stop or continue to toggle while  $\overline{\text{SEN}}$  is high.

In 3-wire mode, commands are sent by first writing each argument to register(s) 0xA1–0xA3, then writing the command word to register 0xA0. A response is retrieved by reading registers 0xA8–0xAF.

For details on timing specifications and diagrams, refer to Table 6, “3-Wire Control Interface Characteristics,” on page 9; Figure 4, “3-Wire Control Interface Write Timing Parameters,” on page 9, and Figure 5, “3-Wire Control Interface Read Timing Parameters,” on page 9.