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5 KV LED EMULATOR INPUT, 4.0 A ISOLATED GATE DRIVERS

Features

- Pin-compatible, drop-in upgrades for popular high speed opto-coupled gate drivers
- Low power diode emulator simplifies design-in process
- 0.6 and 4.0 Amp peak output drive current
- Rail-to-rail output voltage
- Performance and reliability advantages vs. opto-drivers
 - Resistant to temperature and age
 - 10x lower FIT rate for longer service life
 - 14x tighter part-to-part matching
 - Higher common-mode transient immunity: >50 kV/µs typical

- Robust protection features
 - Multiple UVLO ordering options (5, 8, and 12 V) with hysteresis
- 60 ns propagation delay, independent of input drive current
- Wide V_{DD} range: 6.5 to 30 V
- Up to 5000 V_{RMS} isolation
- 10 kV surge withstand capability
- AEC-Q100 qualified
- Wide operating temperature range
 - -40 to +125 °C
- RoHS-compliant packages
 - SOIC-8 (Narrow body)
 - DIP8 (Gull-wing)
 - SDIP6 (Stretched SO-6)
 - I GA8

Applications

- IGBT/ MOSFET gate drives
- Industrial, HEV and renewable energy inverters
- AC, Brushless, and DC motor controls and drives
- Variable speed motor control in consumer white goods
- Isolated switch mode and UPS power supplies

Safety Regulatory Approvals

- UL 1577 recognized
 - Up to 5000 Vrms for 1 minute
- CSA component notice 5A approval
 - IEC 60950-1, 61010-1, 60601-1 (reinforced insulation)
- VDE certification conformity
 - IEC60747-5-2/VDE0884-10 (basic/reinforced insulation)
- CQC certification approval
 - GB4943.1

See page 22 8 VDD ANODE 7 VO CATHODE 6 VO GND 5 NC SOIC-8, DIP8, LGA8 **Industry Standard Pinout** VDD ANODE 6 5 CATHODE 3 GND SDIP6 **Industry Standard Pinout**

Pin Assignments:

Patent pending

Description

The Si826x isolators are pin-compatible, drop-in upgrades for popular opto-coupled gate drivers, such as 0.6 A ACPL-0302/3020, 2.5 A HCPL-3120/ACPL-3130, HCNW3120/3130, and similar opto-drivers. The devices are ideal for driving power MOSFETs and IGBTs used in a wide variety of inverter and motor control applications. The Si826x isolated gate drivers utilize Silicon Laboratories' proprietary silicon isolation technology, supporting up to 5.0 kV $_{\rm RMS}$ withstand voltage per UL1577 and 10kV surge protection per IEC60747. This technology enables higher-performance, reduced variation with temperature and age, tighter part-to-part matching, and superior common-mode rejection compared to opto-coupled gate drivers. While the input circuit mimics the characteristics of an LED, less drive current is required, resulting in higher efficiency. Propagation delay time is independent of input drive current, resulting in consistently short propagation times, tighter unit-to-unit variation, and greater input circuit design flexibility. As a result, the Si826x series offers longer service life and dramatically higher reliability compared to opto-coupled gate drivers.

Functional Block Diagram

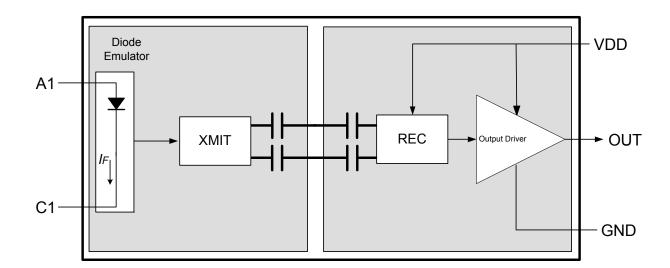


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1. Electrical Specifications

Table 1. Recommended Operating Conditions

Parameter	Symbol	Min	Тур	Max	Unit
Supply Voltage	V_{DD}	6.5	_	30	V
Input Current	I _{F(ON)} (see Figure 1)	6	_	30	mA
Operating Temperature (Ambient)	T _A	-40		125	°C

Table 2. Electrical Characteristics ¹

 V_{DD} = 15 V or 30 V, GND = 0 V, I_F = 6 mA, T_A = -40 to +125 °C; typical specs at 25 °C

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
DC Parameters	1			l	1	
Supply Voltage ²	V _{DD}	(V _{DD} – GND)	6.5	_	30	V
Supply Current (Output High)	I _{DD}	I _F = 10 mA V _{DD} = 15 V V _{DD} = 30 V	_	1.8 2.0	2.4 2.7	mA mA
Supply Current (Output Low)	I _{DD}	$V_F = 0 \text{ V}; I_F = 0 \text{ mA}$ $V_{DD} = 15 \text{ V}$ $V_{DD} = 30 \text{ V}$	_	1.5 1.7	2.1 2.4	mA mA
Input Current Threshold	I _{F(TH)}		_		3.6	mA
Input Current Hysteresis	I _{HYS}		_	0.34	_	mA
Input Forward Voltage (OFF)	V _{F(OFF)}	Measured at ANODE with respect to CATHODE.	_	_	1	V
Input Forward Voltage (ON)	V _{F(ON)}	Measured at ANODE with respect to CATHODE.	1.6	_	2.8	V
Input Capacitance	C _I	f = 100 kHz, $V_F = 0 \text{ V},$ $V_F = 2 \text{ V}$	_	15 15		pF
Output Resistance High	Б	Si826xAxx devices	_	15	_	
(Source) ³	R _{OH}	Si826xBxx devices (I _{OH} = -1 A)	_	2.6	5.1	Ω
Out at Desistance Law (0:41)3	В	Si826xAxx devices	_	5	_	22
Output Resistance Low (Sink) ³	R _{OL}	Si826xBxx devices (I _{OL} = 2 A)	_	0.8	2.0	

Notes:

- **1.** See "8.Ordering Guide" on page 23 for more information.
- 2. Minimum value of (V_{DD} GND) decoupling capacitor is 1 μF .
- 3. Both $V_{\rm O}$ pins are required to be shorted together for 4.0 A compliance.
- **4.** When performing this test, it is recommended that the DUT be soldered down to the PCB to reduce parasitic inductances, which may cause over-stress conditions due to excessive ringing.
- **5.** Guaranteed by characterization.



Table 2. Electrical Characteristics (Continued)¹

 V_{DD} = 15 V or 30 V, GND = 0 V, I_F = 6 mA, T_A = -40 to +125 °C; typical specs at 25 °C

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
		Si826xAxx devices ($I_F = 0$), ($t_{PW_IOH} \le 250 \text{ ns}$) (see Figure 3)	_	0.4	_	
Output High Current (Source) ^{3,4}	I _{OH}	Si826xBxx devices ($I_F = 0$), ($I_{PW_IOH} \le 250 \text{ ns}$), ($I_{DD} - I_{O} = 7.5 \text{ V}$) (see Figure 3)	0.5	1.8	_	A
		Si826xAxx devices $(I_F = 10 \text{ mA}),$ $(t_{PW_IOL} \le 250 \text{ ns})$ (see Figure 2)	_	0.6	_	
Output Low Current (Sink) ^{3,4}	I _{OL}	Si826xBxx devices $(I_F = 10 \text{ mA})$, $(t_{PW_IOL} \le 250 \text{ ns})$, $(V_O - GND = 4.2 \text{ V})$ (see Figure 2)	1.2	4.0	_	A
		Si826xAxx devices (I _{OUT} = -100 mA)	_	V _{DD} - 0.4	_	
High-Level Output Voltage	V _{OH}	Si826xBxx devices (I _{OUT} = -100 mA)	V _{DD} - 0.5	V _{DD} - 0.25	_	V
		Si826xBxx devices (I _{OUT} = 0 mA), (I _F = 0 mA)	_	V _{DD}	_	
Low Lovel Output Voltage	V	Si826xAxx devices (I $_{OUT}$ = 100 mA), (I _F = 10 mA)	_	320	_	mV
Low-Level Output Voltage	V _{OL}	Si826xBxx devices (I $_{OUT}$ = 100 mA), (I _F = 10 mA)	_	80	200	IIIV
UVLO Threshold + (Si826xxAx mode)	VDD _{UV+}	See Figure 11 on page 16. V _{DD} rising	5	5.6	6.3	V
UVLO Threshold – (Si826xxAx mode)	VDD _{UV}	See Figure 11 on page 16. V _{DD} falling	4.7	5.3	6.0	V
UVLO lockout hysteresis (Si826xxAx mode)	VDD _{HYS}		_	300	_	mV

- 1. See "8.Ordering Guide" on page 23 for more information.
- 2. Minimum value of (V_{DD} GND) decoupling capacitor is 1 μ F.
- 3. Both $V_{\rm O}$ pins are required to be shorted together for 4.0 A compliance.
- **4.** When performing this test, it is recommended that the DUT be soldered down to the PCB to reduce parasitic inductances, which may cause over-stress conditions due to excessive ringing.
- 5. Guaranteed by characterization.



Table 2. Electrical Characteristics (Continued)¹

 V_{DD} = 15 V or 30 V, GND = 0 V, I_F = 6 mA, T_A = -40 to +125 °C; typical specs at 25 °C

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
UVLO Threshold + (Si826xxBx mode)	VDD _{UV+}	See Figure 12 on page 16. V _{DD} rising	7.5	8.4	9.4	V
UVLO Threshold – (Si826xxBx mode)	VDD _{UV-}	See Figure 12 on page 16. V _{DD} falling	6.9	7.9	8.9	V
UVLO lockout hysteresis (Si826xxBx mode)	VDD _{HYS}		_	500	_	mV
UVLO Threshold + (Si826xxCx mode)	VDD _{UV+}	See Figure 13 on page 16. V _{DD} rising	10.5	12	13.5	V
UVLO Threshold – (Si826xxCx mode)	VDD _{UV}	See Figure 13 on page 16. V _{DD} falling	9.4	10.7	12.2	V
UVLO lockout hysteresis (Si826xxCx mode)	VDD _{HYS}		_	1.3	_	V
AC Switching Parameters						
Input noise filter cut-off pulse width	t _{NFC}		_	_	15	ns
Minimum pulse width	t _{PMIN}		_	30	_	ns
Propagation delay (Low-to-High)	t _{PLH}	C _L = 200 pF	20	40	60	ns
Propagation delay (High-to-Low)	t _{PHL}	C _L = 200 pF	10	30	50	ns
Pulse Width Distortion	PWD	t _{PLH} – t _{PHL}	_	17	28	ns
Propagation Delay Difference ⁵	PDD	t _{PHLMAX} – t _{PLHMIN}	-1	_	25	ns
Rise time	t _R	C _L = 200 pF	_	5.5	15	ns
Fall time	t _F	C _L = 200 pF	_	8.5	20	ns
Device Startup Time	t _{START}		_	16	30	μs
Common Mode Transient Immunity	CMTI	Output = low or high $(V_{CM} = 1500 \text{ V}), (I_F \ge 6 \text{ mA})$ (See Figure 4)	35	50	_	kV/μs

Notes:

- 1. See "8.Ordering Guide" on page 23 for more information.
- 2. Minimum value of (V $_{DD}$ GND) decoupling capacitor is 1 μF_{\cdot}
- 3. Both V_O pins are required to be shorted together for 4.0 A compliance.
- **4.** When performing this test, it is recommended that the DUT be soldered down to the PCB to reduce parasitic inductances, which may cause over-stress conditions due to excessive ringing.
- **5.** Guaranteed by characterization.



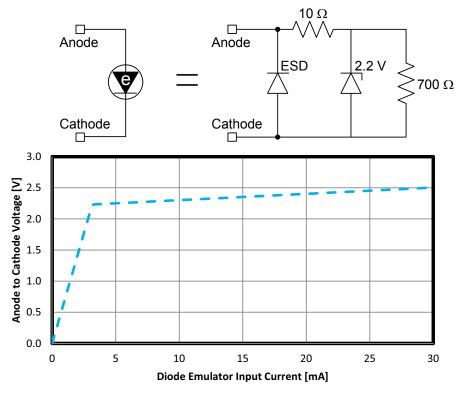


Figure 1. Diode Emulator Model and I-V Curve

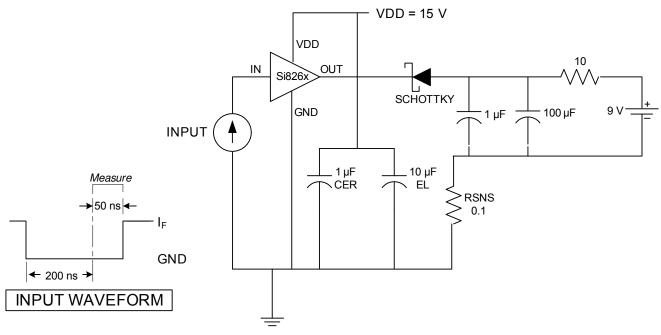


Figure 2. IOL Sink Current Test Circuit



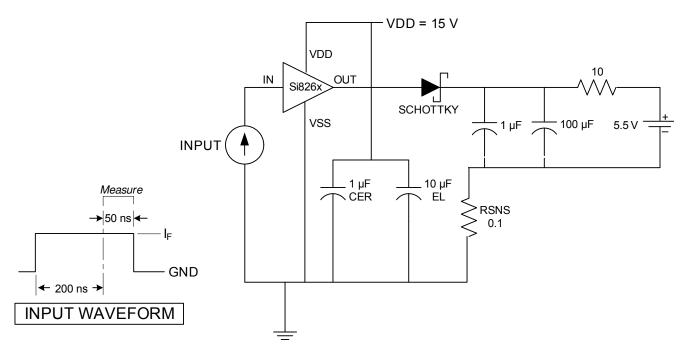


Figure 3. IOH Source Current Test Circuit

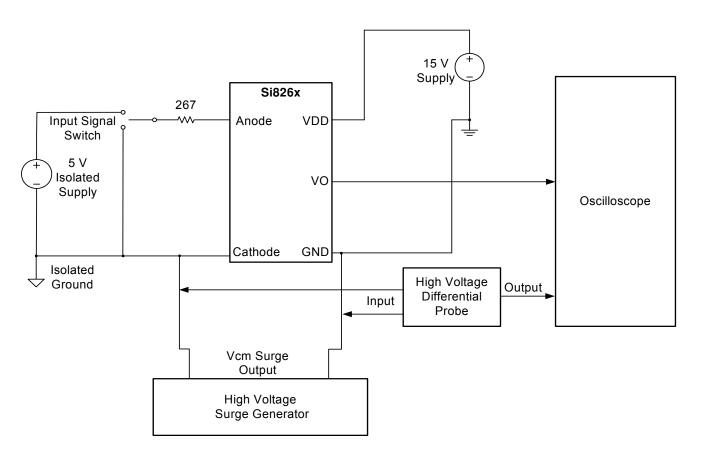


Figure 4. Common Mode Transient Immunity Characterization Circuit



2. Regulatory Information

Table 3. Regulatory Information*

CSA

The Si826x is certified under CSA Component Acceptance Notice 5A. For more details, see File 232873.

61010-1: Up to 600 V_{RMS} reinforced insulation working voltage; up to 600 V_{RMS} basic insulation working voltage.

60950-1: Up to 1000 V_{RMS} reinforced insulation working voltage; up to 1000 V_{RMS} basic insulation working voltage.

60601-1: Up to 250 V_{RMS} reinforced insulation working voltage; up to 500 V_{RMS} basic insulation working voltage.

VDE

The Si826x is certified according to IEC60747 and VDE0884. For more details, see File 5006301-4880-0001.

60747-5-2: Up to 1414 V_{peak} for basic insulation working voltage.

VDE0884-10: Up to 1414 V_{peak} for reinforced insulation working voltage.

UL

The Si826x is certified under UL1577 component recognition program. For more details, see File E257455.

Rated up to 5000 V_{RMS} isolation voltage for basic protection.

CQC

The Si826x is certified under GB4943.1-2011. For more details, see certificates CQC14001104575, CQC15001121282 and CQC15001121283.

Rated up to 1000 V_{RMS} reinforced insulation working voltage; up to 1000 V_{RMS} basic insulation working voltage.

*Note: Regulatory Certifications apply to 3.75 kV_{RMS} rated devices which are production tested to 4.5 kV_{RMS} for 1 sec. Regulatory Certifications apply to 5.0 kV_{RMS} rated devices which are production tested to 6.0 kV_{RMS} for 1 sec. For more information, see "8.Ordering Guide" on page 23.

Table 4. Insulation and Safety-Related Specifications

Parameter	Symbol	Test Condition			Unit		
Faranielei	Symbol Test Condition		SOIC-8	DIP8	SDIP6	LGA8	Ullit
Nominal Air Gap (Clearance)	L(IO1)		4.7 min	7.2 min	9.6 min	10.0 min	mm
Nominal External Tracking (Creepage)	L(IO2)		3.9 min	7.0 min	8.3 min	10.0 min	mm
Minimum Internal Gap (Internal Clearance)			0.016	0.016	0.016	0.016	mm
Tracking Resistance (Proof Tracking Index)	PTI	IEC60112	600	600	600	600	V
Erosion Depth	ED		0.031	0.031	0.057	0.021	mm
Resistance (Input-Output)*	R _{IO}		10 ¹²	10 ¹²	10 ¹²	10 ¹²	Ω
Capacitance (Input-Output)*	C _{IO}	f = 1 MHz	1	1	1	1	pF

*Note: To determine resistance and capacitance, the Si826x is converted into a 2-terminal device. Pins 1–4 (1–3, SDIP6) are shorted together to form the first terminal, and pins 5–8 (4–6, SDIP6) are shorted together to form the second terminal. The parameters are then measured between these two terminals.



Table 5. IEC 60664-1 (VDE 0884) Ratings

Parameter	Test Conditions		Specific	ation	
Parameter	Test Conditions	SOIC-8	DIP8	SDIP6	LGA8
Basic Isolation Group	Material Group	I	I	I	I
	Rated Mains Voltages ≤ 150 V _{RMS}	I-IV	I-IV	I-IV	I-IV
	Rated Mains Voltages ≤ 300 V _{RMS}	I-IV	I-IV	I-IV	I-IV
Installation Classification	Rated Mains Voltages ≤ 450 V _{RMS}	I-III	I-III	I-IV	I-IV
	Rated Mains Voltages ≤ 600 V _{RMS}	I-III	I-III	I-IV	I-IV
	Rated Mains Voltages ≤ 1000 V _{RMS}	1-11	1-11	I-III	1-111

Table 6. IEC 60747-5-2 (VDE 0884-10) Insulation Characteristics*

Parameter	Symbol	Test Condition		Charac	teristic		Unit
Parameter	Syllibol	rest Condition	SOIC-8	DIP8	SDIP6	LGA8	Ollit
Maximum Working Insulation Voltage	V _{IORM}		630	891	1140	1414	V peak
Input to Output Test Voltage	V _{PR}	Method b1 $(V_{IORM} \times 1.875 = V_{PR}, 100\%$ Production Test, $t_m = 1$ sec, Partial Discharge < 5 pC)	1181	1671	2138	2652	V peak
Transient Overvoltage	V _{IOTM}	t = 60 sec	6000	6000	8000	8000	V peak
Pollution Degree (DIN VDE 0110, Table 1)			2	2	2	2	
Insulation Resistance at T_S , V_{IO} = 500 V	R _S		>10 ⁹	>10 ⁹	>10 ⁹	>10 ⁹	Ω

*Note: This isolator is suitable for reinforced electrical isolation only within the safety limit data. Maintenance of the safety data is ensured by protective circuits. The Si826x provides a climate classification of 40/125/21.

Table 7. IEC Safety Limiting Values*

Parameter	Symbol	Test Condition		Ма	X		Unit
Parameter	Symbol	rest Condition	SOIC-8	DIP8	SDIP6	LGA8	Ullit
Case Temperature	T _S		140	140	140	140	°C
Input Current	Is	θ_{JA} = 110 °C/W (SOIC-8), 110 °C/W (DIP8), 105 °C/W (SDIP6), 220 °C (LGA8), V_{F} = 2.8 V, T_{J} = 140 °C, T_{A} = 25 °C	370	370	390	185	mA
Output Power	P _S		1	1	1	0.5	W
*Note: Maximum value	allowed in t	he event of a failure: also see th	e thermal der	ating curve i	n Figures 5.	6. 7. and 8.	•



Table 8. Thermal Characteristics

Parameter	Symbol	Тур			Unit	
Farameter	Syllibol	SOIC-8	DIP8	SDIP6	LGA8	Oilit
IC Junction-to-Air Thermal Resistance	$\theta_{\sf JA}$	110	110	105	220	°C/W

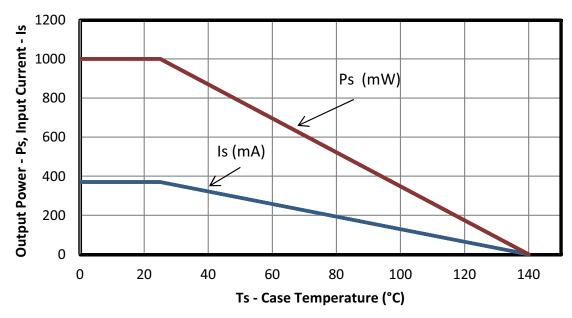


Figure 5. (SOIC-8) Thermal Derating Curve, Dependence of Safety Limiting Values with Case Temperature per DIN EN 60747-5-2 and VDE0884-10

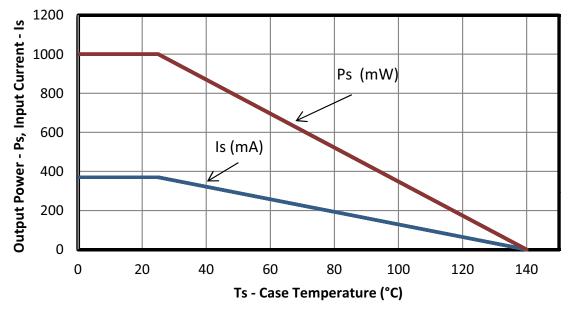


Figure 6. (DIP8) Thermal Derating Curve, Dependence of Safety Limiting Values with Case Temperature per DIN EN 60747-5-2 and VDE0884-10



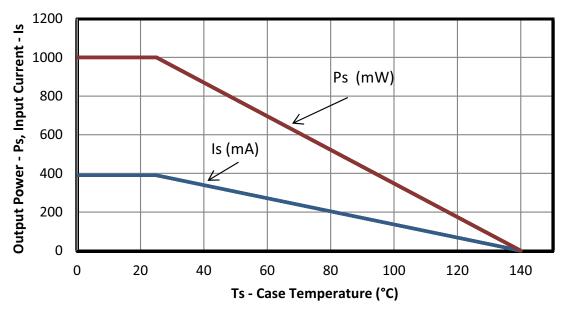


Figure 7. (SDIP6) Thermal Derating Curve, Dependence of Safety Limiting Values with Case Temperature per DIN EN 60747-5-2 and VDE0884-10

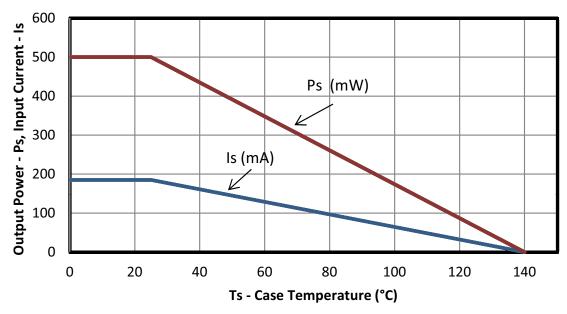


Figure 8. (LGA8) Thermal Derating Curve, Dependence of Safety Limiting Values with Case Temperature per DIN EN 60747-5-2 and VDE0884-10



Table 9. Absolute Maximum Ratings*

Parameter	Symbol	Min	Max	Unit
Storage Temperature	T _{STG}	-65	+150	°C
Operating Temperature	T _A	-40	+125	°C
Junction Temperature	TJ	_	+140	°C
Average Forward Input Current	I _{F(AVG)}	_	30	mA
Peak Transient Input Current (< 1 µs pulse width, 300 ps)	I _{FTR}	_	1	А
Reverse Input Voltage	V_{R}	_	0.3	V
Supply Voltage	VDD	-0.5	36	V
Output Voltage	V _{OUT}	-0.5	36	V
Peak Output Current (t_{PW} = 10 μ s, duty cycle = 0.2%) (0.6 Amp versions)	I _{OPK}	_	0.6	Α
Peak Output Current (t_{PW} = 10 μ s, duty cycle = 0.2%) (4.0 Amp versions)	I _{OPK}	_	4.0	Α
Input Power Dissipation	P _I	_	75	mW
Output Power Dissipation	P _O	_	225	mW
Total Power Dissipation (all packages limited by thermal derating curve)	P _T	_	300	mW
Lead Solder Temperature (10 s)		_	260	°C
HBM Rating ESD		4	_	kV
Machine Model ESD		300	_	V
CDM		2000	_	V
Maximum Isolation Voltage (1 s) SOIC-8		_	4500	V _{RMS}
Maximum Isolation Voltage (1 s) DIP8		_	6500	V _{RMS}
Maximum Isolation Voltage (1 s) SDIP6		_	6500	V _{RMS}
Maximum Isolation Voltage (1 s) LGA8		_	6500	V _{RMS}
		•	•	•

*Note: Permanent device damage may occur if the absolute maximum ratings are exceeded. Functional operation should be restricted to the conditions specified in the operational sections of this data sheet.



3. Functional Description

3.1. Theory of Operation

The Si826x is a functional upgrade for popular opto-isolated drivers, such as the Avago HPCL-3120, HPCL-0302, Toshiba TLP350, and others. The operation of an Si826x channel is analogous to that of an opto coupler, except an RF carrier is modulated instead of light. This simple architecture provides a robust isolated data path and requires no special considerations or initialization at start-up. The Si826x also includes a noise filter that suppresses propagation of any pulse narrower than 15 ns. A simplified block diagram for the Si826x is shown in Figure 9.

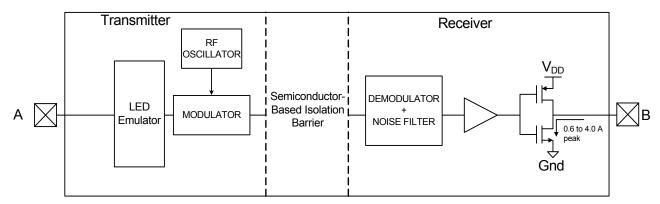


Figure 9. Simplified Channel Diagram



4. Technical Description

4.1. Device Behavior

Truth tables for the Si826x are summarized in Table 10.

Table 10. Si826x Truth Table Summary*

Input	V _{DD}	v _o
OFF	> UVLO	LOW
OFF	< UVLO	LOW
ON	> UVLO	HIGH
ON	< UVLO	LOW

*Note: This truth table assumes VDD is powered. If VDD is below UVLO, see "4.3.Under Voltage Lockout (UVLO)" on page 16 for more information.

4.2. Device Startup

Output V_O is held low during power-up until V_{DD} rises above the UVLO+ threshold for a minimum time period of t_{START} . Following this, the output is high when the current flowing from anode to cathode is > $I_{F(ON)}$. Device startup, normal operation, and shutdown behavior is shown in Figure 10.

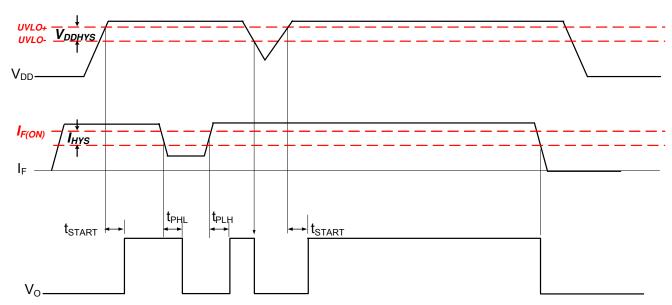


Figure 10. Si826x Operating Behavior $(I_F \ge I_{F(MIN)})$ when $V_F \ge V_{F(MIN)}$



4.3. Under Voltage Lockout (UVLO)

The UVLO circuit unconditionally drives V_O low when V_{DD} is below the lockout threshold. Referring to Figures 11 through 13, upon power up, the Si826x is maintained in UVLO until VDD rises above VDD_{UV+}. During power down, the Si826x enters UVLO when VDD falls below the UVLO threshold plus hysteresis (i.e., VDD \leq VDD_{UV+} – VDD_{HYS}).

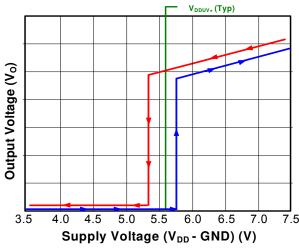


Figure 11. Si826xxAx UVLO Response (5 V)

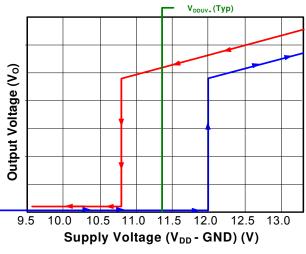


Figure 13. Si826xxCx UVLO Response (12 V)

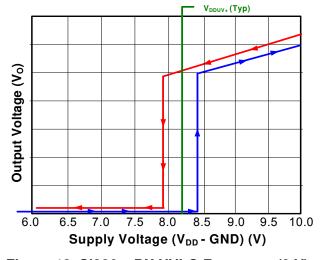


Figure 12. Si826xxBX UVLO Response (8 V)



5. Applications

The following sections detail the input and output circuits necessary for proper operation. Power dissipation and layout considerations are also discussed.

5.1. Input Circuit Design

Opto driver manufacturers typically recommend the circuits shown in Figures 14 and 15. These circuits are specifically designed to improve opto-coupler input common-mode rejection and increase noise immunity.

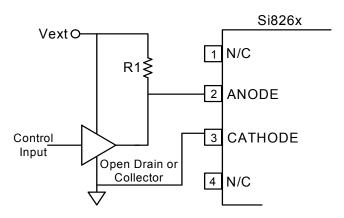


Figure 14. Si826x Input Circuit

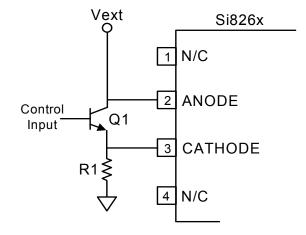


Figure 15. High CMR Si826x Input Circuit

The optically-coupled driver circuit of Figure 14 turns the LED on when the control input is high. However, internal capacitive coupling from the LED to the power and ground conductors can momentarily force the LED into its off state when the anode and cathode inputs are subjected to a high common-mode transient. The circuit shown in Figure 15 addresses this issue by using a value of R1 sufficiently low to overdrive the LED, ensuring it remains on during an input common-mode transient. Q1 shorts the LED off in the low output state, again increasing common-mode transient immunity.

Some opto driver applications recommend reverse-biasing the LED when the control input is off to prevent coupled noise from energizing the LED. The Si826x input circuit requires less current and has twice the off-state noise margin compared to opto couplers. However, high CMR opto coupler designs that overdrive the LED (see Figure 15) may require increasing the value of R1 to limit input current I_F to its maximum rating when using the Si826x. In addition, there is no benefit in driving the Si826x input diode into reverse bias when in the off state.



Si826x

Consequently, opto coupler circuits using this technique should either leave the negative bias circuitry unpopulated or modify the circuitry (e.g., add a clamp diode or current limiting resistor) to ensure that the anode pin of the Si826x is no more than -0.3 V with respect to the cathode when reverse-biased.

New designs should consider the input circuit configurations of Figure 16, which are more efficient than those of Figures 14 and 15. As shown, S1 and S2 represent any suitable switch, such as a BJT or MOSFET, analog transmission gate, processor I/O, etc. Also, note that the Si826x input can be driven from the I/O port of any MCU or FPGA capable of sourcing a minimum of 6 mA (see Figure 16C). Additionally, note that the Si826x propagation delay and output drive do not significantly change for values of I_F between I_{F(MIN)} and I_{F(MAX)}.

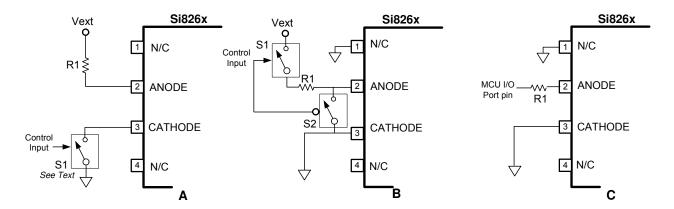


Figure 16. Si826x Other Input Circuit Configurations

5.2. Output Circuit Design

GND can be biased at, above, or below ground as long as the voltage on V_{DD} with respect to GND is a maximum of 30 V. V_{DD} decoupling capacitors should be placed as close to the package pins as possible. The optimum values for these capacitors depend on load current and the distance between the chip and its power source. It is recommended that 0.1 and 10 μ F bypass capacitors be used to reduce high-frequency noise and maximize performance.

5.3. Layout Considerations

It is most important to minimize ringing in the drive path and noise on the V_{DD} lines. Care must be taken to minimize parasitic inductance in these paths by locating the Si826x as close as possible to the device it is driving. In addition, the V_{DD} supply and ground trace paths must be kept short. For this reason, the use of power and ground planes is highly recommended. A split ground plane system having separate ground and V_{DD} planes for power devices and small signal components provides the best overall noise performance.



5.4. Power Dissipation Considerations

Proper system design must assure that the Si826x operates within safe thermal limits across the entire load range. The Si826x total power dissipation is the sum of the power dissipated by bias supply current, internal switching losses, and power delivered to the load, as shown in Equation 1.

$$\mathsf{P}_\mathsf{D} \; = \; \mathsf{I}_\mathsf{F} \times \mathsf{V}_\mathsf{F} \times \mathsf{DC} + \mathsf{V}_\mathsf{DD} \times [\mathsf{I}_\mathsf{DDQ} + (\mathsf{Q}_\mathsf{d} + \mathsf{C}_\mathsf{L} \times \mathsf{V}_\mathsf{DD}) \times \mathsf{f}]$$

where: P_D is the total device power dissipation (W)

I_F is the diode current (30 mA max)

V_F is the diode anode to cathode voltage (2.8 V max)

DC is duty cycle (0.5 typical)

V_{DD} is the driver-side supply voltage (30 V max)

 I_{DDQ} is the driver maximum bias current (2.5 mA)

Q_d is 3 nC

C₁ is the load capacitance

f is the switching frequency (Hz)

Equation 1.

The maximum allowable power dissipation for the Si826x is a function of the package thermal resistance, ambient temperature, and maximum allowable junction temperature, as shown in Equation 2.

$$P_{Dmax} \le \frac{T_{jmax} - T_A}{\theta_{ja}}$$

where:

P_{Dmax} is the maximum allowable power dissipation (W)

T_{imax} is the maximum junction temperature (140 °C)

 T_A is the ambient temperature (°C)

 $\theta_{\rm ja}$ is the package junction-to-air thermal resistance (110 °C/W)

Equation 2.

Substituting values for P_{Dmax} T_{jmax} , T_A , and θ_{ja} into Equation 2 results in a maximum allowable total power dissipation of 1.0 W. Note that the maximum allowable load is found by substituting this limit and the appropriate datasheet values from Table 2 on page 4 into Equation 1 and simplifying. Graphs are shown in Figures 17 and 18. All points along the load lines in these graphs represent the package dissipation-limited value of C_L for the corresponding switching frequency.



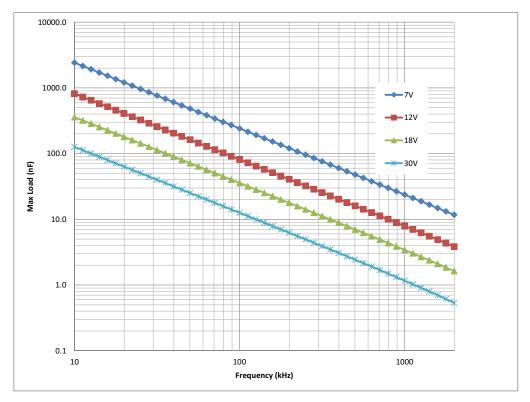


Figure 17. (SOIC-8, DIP8, SDIP6) Maximum Load vs. Switching Frequency (25 °C)

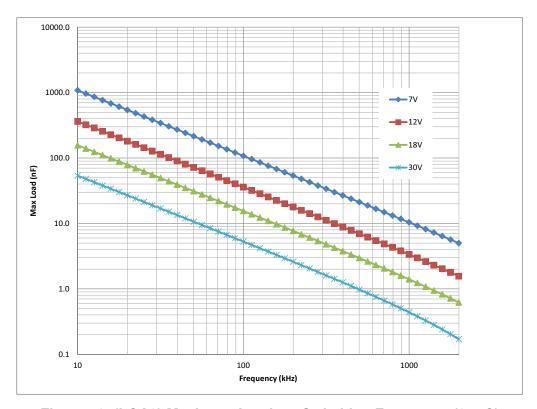


Figure 18. (LGA8) Maximum Load vs. Switching Frequency (25 °C)



6. Pin Descriptions (SOIC-8, DIP8, LGA8)

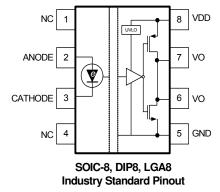


Figure 19. Pin Configuration

Table 11. Pin Descriptions (SOIC-8, DIP8, LGA8)

Pin	Name	Description					
1	NC*	No connect.					
2	ANODE	Anode of LED emulator. $V_{\rm O}$ follows the signal applied to this input with respect to the CATHODE input.					
3	CATHODE	Cathode of LED emulator. $V_{\rm O}$ follows the signal applied to ANODE with respect to this input.					
4	NC*	No connect.					
5	GND	External MOSFET source connection and ground reference for V _{DD} . This terminal is typically connected to ground but may be tied to a negative or positive voltage.					
6	V _O	Output signal. Both $V_{\rm O}$ pins are required to be shorted together for 4.0 A compliance.					
7	V _O	Output signal. Both $V_{\rm O}$ pins are required to be shorted together for 4.0 A compliance.					
8	V_{DD}	Output-side power supply input referenced to GND (30 V max).					

*Note: No Connect. These pins are not internally connected. To maximize CMTI performance, these pins should be connected to the ground plane.



7. Pin Descriptions (SDIP6)

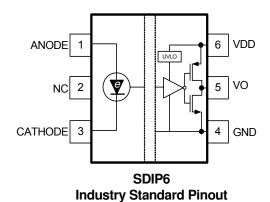


Figure 20. Pin Configuration

Table 12. Pin Descriptions (SDIP6)

Pin	Name	Description					
1	ANODE	Anode of LED emulator. $V_{\rm O}$ follows the signal applied to this input with respect to the CATHODE input.					
2	NC*	No connect.					
3	CATHODE	Cathode of LED emulator. $V_{\rm O}$ follows the signal applied to ANODE with respect to this input.					
4	GND	External MOSFET source connection and ground reference for V _{DD} . This terminal is typically connected to ground but may be tied to a negative or positive voltage.					
5	V _O	Output signal.					
6	V_{DD}	Output-side power supply input referenced to GND (30 V max).					

*Note: No Connect. These pins are not internally connected. To maximize CMTI performance, these pins should be connected to the ground plane.



8. Ordering Guide

Table 13. Si826x Ordering Guide 1,2,3

Now Ordering	Ordering Options						
New Ordering Part Number (OPN)	Output Configuration	Cross Reference	UVLO Voltage	Insulation Rating	Temp Range	Pkg Type	
Si8261AAC-C-IS	0.6 A driver	HCPL-0314	5 V	3.75 kVrms	–40 to +125 °C	SOIC-8	
Si8261BAC-C-IS	4.0 A driver	_	5 V	3.75 kVrms	–40 to +125 °C	SOIC-8	
Si8261AAC-C-IP	0.6 A driver	HCPL-3140	5 V	3.75 kVrms	–40 to +125 °C	DIP8/GW	
Si8261BAC-C-IP	4.0 A driver	TLP 350 HCPL-3120	5 V	3.75 kVrms	–40 to +125 °C	DIP8/GW	
Si8261AAD-C-IS	0.6 A driver	ACPL-W314	5 V	5.0 kVrms	–40 to +125 °C	SDIP6	
Si8261BAD-C-IS	4.0 A driver	TLP 700F	5 V	5.0 kVrms	–40 to +125 °C	SDIP6	
Si8261AAD-C-IM	0.6 A driver	_	5 V	5.0 kVrms	–40 to +125 °C	LGA8	
Si8261BAD-C-IM	4.0 A driver	HCNW-3120	5 V	5.0 kVrms	–40 to +125 °C	LGA8	

- 1. All packages are RoHS-compliant with peak solder reflow temperatures of 260 °C according to the JEDEC industry standard classifications.
- 2. "Si" and "SI" are used interchangeably.
- 3. AEC-Q100 qualified.



Table 13. Si826x Ordering Guide^{1,2,3}

New Ordering	Ordering Options						
Part Number (OPN)	Output Configuration	Cross Reference	UVLO Voltage	Insulation Rating	Temp Range	Pkg Type	
Si8261ABC-C-IS	0.6 A driver	HCPL-0314	8 V	3.75 kVrms	–40 to +125 °C	SOIC-8	
Si8261BBC-C-IS	4.0 A driver	_	8 V	3.75 kVrms	–40 to +125 °C	SOIC-8	
Si8261ABC-C-IP	0.6 A driver	HCPL-3140	8 V	3.75 kVrms	–40 to +125 °C	DIP8/GW	
Si8261BBC-C-IP	4.0 A driver	TLP 350 HCPL-3120	8 V	3.75 kVrms	–40 to +125 °C	DIP8/GW	
Si8261ABD-C-IS	0.6 A driver	ACPL-W314	8 V	5.0 kVrms	–40 to +125 °C	SDIP6	
Si8261BBD-C-IS	4.0 A driver	TLP 700F	8 V	5.0 kVrms	–40 to +125 °C	SDIP6	
Si8261ABD-C-IM	0.6 A driver	_	8 V	5.0 kVrms	–40 to +125 °C	LGA8	
Si8261BBD-C-IM	4.0 A driver	HCNW-3120	8 V	5.0 kVrms	–40 to +125 °C	LGA8	

- 1. All packages are RoHS-compliant with peak solder reflow temperatures of 260 °C according to the JEDEC industry standard classifications.
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Table 13. Si826x Ordering Guide^{1,2,3}

Now Ordering	Ordering Options						
New Ordering Part Number (OPN)	Output Configuration	Cross Reference	UVLO Voltage	Insulation Rating	Temp Range	Pkg Type	
Si8261ACC-C-IS	0.6 A driver	HCPL-0314	12 V	3.75 kVrms	–40 to +125 °C	SOIC-8	
Si8261BCC-C-IS	4.0 A driver	_	12 V	3.75 kVrms	–40 to +125 °C	SOIC-8	
Si8261ACC-C-IP	0.6 A driver	HCPL-3140	12 V	3.75 kVrms	–40 to +125 °C	DIP8/GW	
Si8261BCC-C-IP	4.0 A driver	TLP 350 HCPL-3120	12 V	3.75 kVrms	–40 to +125 °C	DIP8/GW	
Si8261ACD-C-IS	0.6 A driver	ACPL-W314	12 V	5.0 kVrms	–40 to +125 °C	SDIP6	
Si8261BCD-C-IS	4.0 A driver	TLP 700F	12 V	5.0 kVrms	–40 to +125 °C	SDIP6	
Si8261ACD-C-IM	0.6 A driver	_	12 V	5.0 kVrms	–40 to +125 °C	LGA8	
Si8261BCD-C-IM	4.0 A driver	HCNW-3120	12 V	5.0 kVrms	–40 to +125 °C	LGA8	

- 1. All packages are RoHS-compliant with peak solder reflow temperatures of 260 °C according to the JEDEC industry standard classifications.
- 2. "Si" and "SI" are used interchangeably.
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