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Unipolar 2-Phase Stepper Motor Driver ICs SLA7070MS Series Data Sheet

March, 2018 Rev.1.6

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1. General Description

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This document describes summary of the SLA7070MS series.

2. Features and Benefits

- (1) Main power supply voltages, V_{BB} : 46 V (max.), 10 to 44 V normal operating range
- (2) Logic supply voltages, V_{DD} : 3.0 to 5.5 V
- (3) Maximum output currents, Io(max): 2.0 A, 3.0 A
- (4) Clock-in stepping control (built-in sequencer)
- (5) Full-, half-, and microstep products are available
 → Microstepping options are capable of full-, half-, quarter-, eighth-, and sixteenth-stepping
- (6) Built-in "sense resistor" detects motor current
- (7) All variants are pin-compatible for enhanced design flexibility
- (8) ZIP type 23-pin molded package (SLA package)
- (9) Self-excitation PWM current control with fixed OFF-time
 - → For microstepping variants, OFF-time adjusted automatically by step reference current ratio (3 levels)
- (10) Built-in synchronous rectifying circuit reduces power dissipation at PWM-OFF
- (11) Synchronous PWM chopping function prevents motor noise in the Hold mode
- (12) The Sleep mode to reduce IC input current in stand-by state
- (13) Built-in protection circuitry against motor coil opens/shorts and thermal shutdown protection
- (14) The following are the product variants and optional features available:
 - Blanking Time
 - Full/Half step products:
 - Microstep product:
- $3.2 \ \mu s$ (standard), $5.2 \ \mu s$ (optional type B) $1.7 \ \mu s$ (standard), $3.2 \ \mu s$ (optional type B)

- Input Clock Edge
 - Standard type:
 - Optional type W:
- POS (positive) edge POS/NEG (positive and negative) edge

NOTE: The optional types listed above, "type B" and "type W", are abbreviated and referred to "B" and "W" as the letters for product branding codes, respectively. These terms and abbreviations are also used throughout this document. See also Section 6 for more details.

3. Part Numbers and Functional Characteristics

Table 3-1 provides the product variants available in the SLA7070MS series.

Table				00	
		aracteristics			
			Rated Current		
	Seque	ncer	(Maximum Se	etting Value)	
Part Number	Full/Half Step	Microstep	2.0 A	3.0 A	
SLA7072MS	Х		Х		
SLA7073MS	X			X	
SLA7078MS		Х		X	

Table 3-1 Part Nu	mbers and Functiona	I Charactoristics
	inders and i uncliona	

In addition, the following functional options are available in the SLA7070MS series: • Blanking Time

• Full/Half step products:

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3.2 μs (standard), 5.2 μs (optional type B)

1.7 μs (standard), 3.2 μs (optional type B)

- Microstep product:Input Clock Edge
- POS (positive) edge
- Standard type:Optional type W:

POS/NEG (positive and negative) edge

NOTE: The optional types listed above, "type B" and "type W", are abbreviated and referred to "B" and "W" as the letters for product branding codes, respectively. These terms and abbreviations are also used throughout this document. See also Section 6 for more details.

4. Specifications

	1		Unle	ess specifically not	ted, T _A = 25 °C
Characteristic	Symbol	Rating	Unit	Rema	rks
Load (Motor Supply) Voltage	V _M	46	V		
Main Power Supply Voltage	V _{BB}	46	V		
Supply Voltage	V	6	V	Power supply	to DC
Supply Voltage	V _{DD}	7	V	≤1 µs (5% dut	y)
		2.0	Α	SLA7072MS	Control
Output Current	I _O	3.0	Α	SLA7073MS SLA7078MS	current value
Logic Input Voltage	V _{IN}	-0.3 to V _{DD} + 0.3	V		
REF Input Voltage	V _{REF}	-0.3 to V _{DD} + 0.3	V		
Sense Voltage	V _{RS}	±1	V		
Allowable Power Dissipation	PD	4.7	W	Without heats	ink
Junction Temperature	TJ	150	°C		
Operating Ambient Temperature	T _A	-20 to 85	°C		
Storage Temperature	T _{stg}	-30 to 150	°C		

Table 4-1. Absolute Maximum Ratings

NOTE: Output current ratings may be limited by duty cycles, ambient temperatures, and heat sinking conditions. Do not exceed the maximum output currents and the maximum junction temperature (T_J) given above, under any conditions of use.

Table 4-2.	Recommended	Operating	Conditions
			Liniaga an

Tabl	e 4-2. Rec	ommendeo	d Operating	•	ons ss specifically noted. $T_A = 25 \ ^{\circ}C$
		Standard Value			
Characteristic	Symbol	Min.	Max.	Unit	Remarks
Load (Motor Supply) Voltage	V _M		44	V	
Main Power Supply Voltage	V _{BB}	10	44	V	
Logic Supply Voltage	V _{DD}	3.0	5.5	v	Surge voltage at VDD pin should be less than ±0.5 V to avoid malfunctioning in operation
Case Temperature	T _C		90	°C	Measured at Pin 12 (lead portion), without heatsink

NOTE: As the motor supply voltage, V_M , becomes higher, it also approaches the breakdown voltage of the OUTx pins (100 V min.); and breakdown will be more likely to happen. Even if one of the OUTx pins breaks down (due to surge noise or other factors), the SLA7070MS series will recognize it as abnormality (coil open) and will run appropriate protection functions. Therefore, a thorough evaluation is recommended.

	Ia			aracteristic ly noted, T_A		C, $V_{BB} = 24$ V, and $V_{DD} = 5$ V
			Rating			
Characteristic	Symbol	Min.	Тур.	Max.	Unit	Conditions
Main Power Supply	I _{BB}			15	mA	Normal mode
Current	I _{BBS}			100	μA	Sleep1, Sleep2
Logic Power Current	I _{DD}			5	mA	
MOSFET Breakdown Voltage	V _{DSS}	100			V	$V_{BB} = 44V, I_D = 1 \text{ mA}$
Maximum Response Frequency	f _{clk}	250			kHz	Clock duty cycle = 50%
Logic Input Voltage	V _{IL}			0.25 × V _{DD}	V	
	V _{IH}	0.75 × V _{DD}			V	
Logic Input Current	IL		±1		μΑ	
Logic input ourient	I _{IH}		±1		μA	
REF Input Voltage	V _{REF}	0.04		→	V	Table 4-5, Figure 4-1
	V _{REFS}	2.0		V _{DD}	V	Output OFF, Sleep1 ¹⁾
REF Input Current	I _{REF}		±10		μΑ	
SENSE Detection Voltage	V _{SENSE}	V _{REF} – 0.03	V _{REF}	V _{REF} + 0.03	V	V _{REF} = 0 to1.5 V, Step reference ratio: 100%
Sleep-Enable Recovery Time	t _{SE}	100			μs	Sleep1, Sleep2
Switching Time	t _{con}		2.0		μs	Clock \rightarrow Output ON
Switching fille	t _{coff}		1.5		μs	Clock \rightarrow Output OFF
Overcurrent Detection Voltage ²⁾	V _{CCP}	0.65	0.7	0.75	V	Motor coils shorted
Overeurrent Detection			2.3		A	1.0 A and 1.5 A devices
Overcurrent Detection Current (V _{OCP} / R _S)	ICCP		3.5		A	2.0 A devices
			4.6		Α	3.0 A devices
Load Disconnection Undetected Time	t _{opp}		2		μs	From PWM-OFF
Overheat Protection Temperature	T _{tsd}		140		°C	Measured at back of device case (after heat has saturated)
	V _{FlagL}			1.25	V	I _{FlagL} = 1.25 mA
FLAG Output Voltage	V _{FlagH}	V _{DD} – 1.25			V	I _{FlagH} = −1.25 mA
FLAG Output Current	I _{FlagL}			1.25	mA	
	I _{FlagH}	-1.25			mA	

Table 4-3. Electrical Characteristics 1

NOTE: Unless specifically noted, negative current is defined as output current flow from a specified pin. ¹⁾ In a state of: IBBS, output OFF, and sequencer <u>enabled</u>. ²⁾ Protection circuit operates when $V_{SENSE} \ge V_{OCP}$.



Table 4-4. Electrical Characteristics 2 (Varying with Step Sequence)

(1) Full-/Half-step products: SLA7072MS, SLA7073MS

		mess spec		$100, T_A =$	<u>23 0,</u>	$V_{BB} = 24$ V, and $V_{DD} = 5$
Characteristic	Symbol	Min.	Rating Typ.	Max.	Unit	Conditions
Step Reference Current	Mode F	101111.	100	IVIAN.	%	V ~V 100%
Ratio	Mode 1 Mode 8		70		%	$V_{\text{REF}} \approx V_{\text{SENSE}} = 100\%,$ $V_{\text{REF}} = 0 \text{ to } 1.0 \text{ V}$
			3.2		μs	Standard type (w/o branding codes)
Minimum PWM ON-Time	t _{on(min)}		5.2		μs	Optional type B (w/ branding codes)
PWM OFF-Time	t _{off}		12		μs	_
(2) Microstep product: SLA7		iless speci Min.	<u>fically note</u> Rating Typ.	ed, T _A = 2 Max.	25 °C, V	$V_{BB} = 24$ V, and VDD = 5 Conditions
	Mode F		100	IVIAN.	%	
	Mode F		98.1		%	
	Mode D		95.7		%	r.
	Mode C		92.4		%	
	Mode B		88.2		%	
	Mode A		83.1		%	
	Mode 9		77.3		%	
Step Reference Current	Mode 8		70.7		%	$V_{\text{REF}} \approx V_{\text{SENSE}} = 100\%,$
Ratio	Mode 7		63.4		%	V _{REF} = 0 to 1.0 V
	Mode 6		55.5		%	
	Mode 5		47.1		%	
	Mode 4		38.2		%	
	Mode 3		29		%	
	Mode 2		19.5		%	
	Mode 1		9.8		%	
	V _{MOL}			1.25	V	I _{MOL} = 1.25 mA
MO Output Voltage	V _{MOH}	V _{DD} – 1.25			V	I _{MOH} = −1.25 mA
MO Output Current	I _{MOL}			1.25	mA	
	I _{MOH}	-1.25			mA	
Minimum PWM ON-Time	t _{on(min)}		1.7		μs	Standard type (w/o branding codes)
	•on(min)		3.2		μs	Optional type B (w/ branding codes)
	t _{off 1}		12		μs	Mode 8 to Mode F
PWM OFF-time	t _{off 2}		9		μs	Mode 4 to Mode 7
	t _{off 3}		7		μs	Mode 1 to Mode 3

Table 4-5. Electrical Characteristics 3 (Varying with Output Current Range)

(1) $I_0 = 2.0 A$

() <u>-</u>		Unless sp	ecifically r	noted, T _A :	= 25 °C	, $V_{BB} = 24$ V, and $V_{DD} = 5$ V
Characteristic	Cumbal		Rating			Conditiono
Characteristic	Symbol	Min.	Тур.	Max.	Unit	Conditions
Output MOSFET ON-Voltage	R _{DS(on)}		0.25	0.4	Ω	I _D = 2.0 A
Output MOSFET Body Diodes Forward Voltage	V _F		0.95	1.2	V	I _F = 2.0 A
Sense Resistor ¹⁾	Rs	0.199	0.205	0.211	Ω	Tolerance: ±3%
REF Input Voltage	V _{REF}	0.04		0.4	V	Within specified current limit

¹⁾ Includes approximately 5 m Ω circuit resistance in addition to the resistance of the built-in resistor itself.

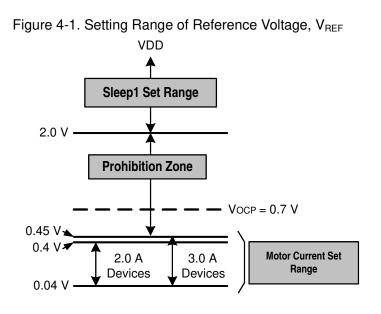
(2) $I_0 = 3.0 \text{ A}$

Unless specifically noted, $T_A = 25 \text{ °C}$, $V_{BB} = 24 \text{ V}$, and $V_{DD} = 5 \text{ V}$

Characteristic	Symbol		Rating		Unit	Conditions
Characteristic	Symbol	Min.	Тур.	Max.	Unit	Conditions
Output MOSFET ON-Resistance	R _{DS(on)}		0.18	0.24	Ω	I _D = 3.0 A
Output MOSFET Body Diodes Forward Voltage	V _F		0.95	2.1	V	I _F = 3.0 A
Sense Resistor ¹⁾	R _S	0.150	0.155	0.160	Ω	Tolerance: ±3%
REF Input Voltage	V _{REF}	0.04		0.45	V	Within specified current limit

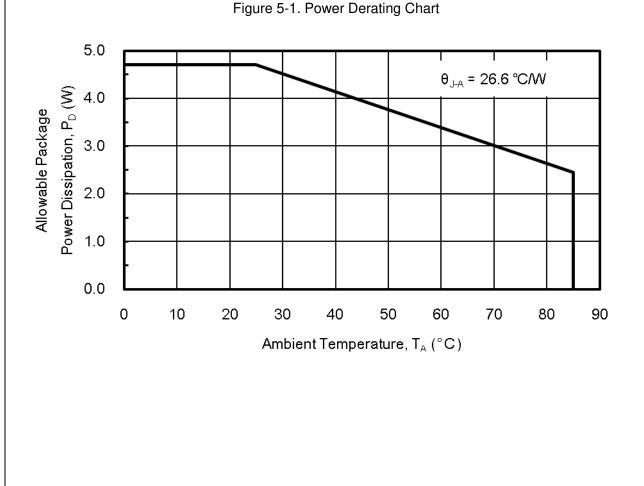
¹⁾ Includes approximately 5 m Ω circuit resistance in addition to the resistance of the built-in resistor itself.

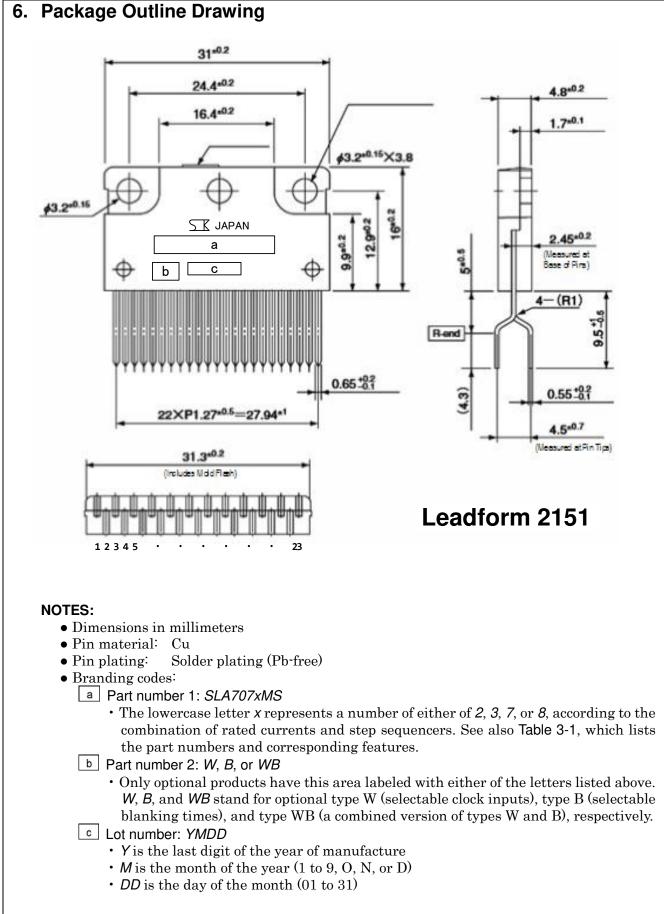




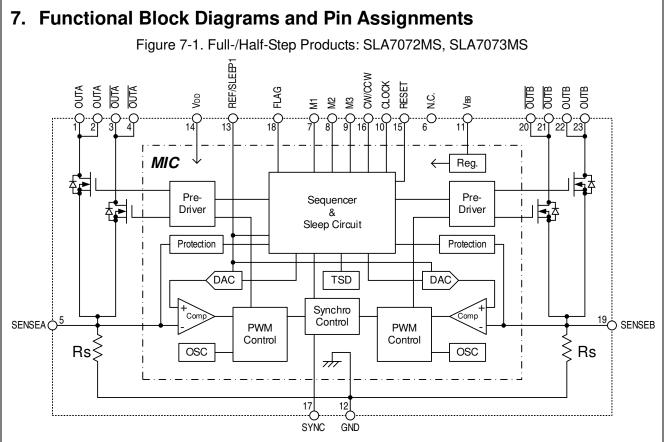
NOTE: Extra attentions should be paid to the changeover between the motor current specification range and the Sleep1 set range. Vocp falls on the "prohibition zone" threshold. If the changeover takes too long, OCP operation will start when $V_{SENSE} > V_{OCP}$.

5. Power Derating Chart



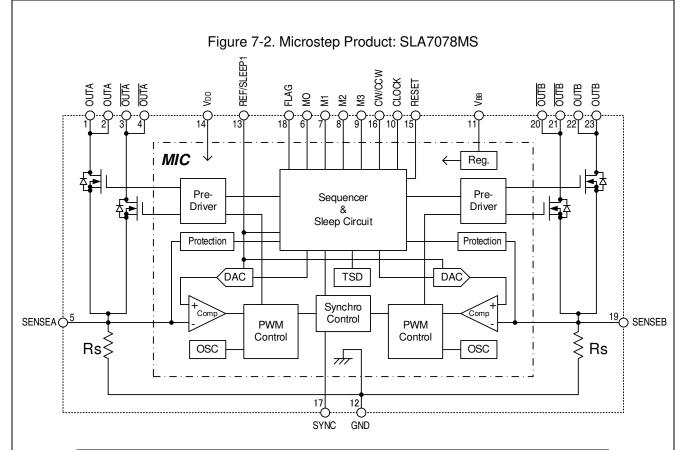




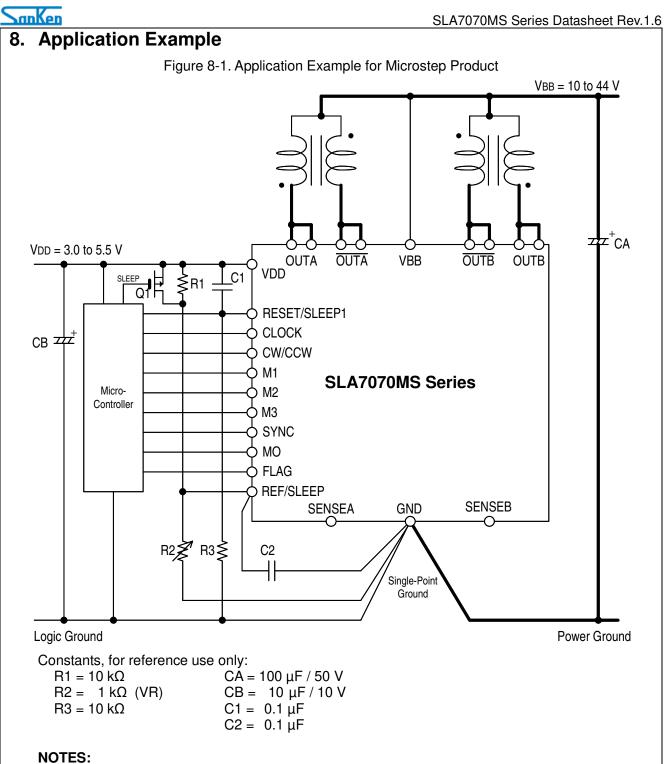


Pin No.	Symbol	Function			
1					
2	OUTA	Phase A output			
3	OUTA	Phase A output			
4	OUTA				
5	SENSEA	Phase A current sensing			
6	NC	No connection			
7	M1				
8	M2	Input for excitation mode & Sleep2 setting			
9	M3				
10	CLOCK	Step clock input			
11	V _{BB}	Main power supply voltage (for motor)			
12	GND	Ground			
13	REF/SLEEP1	Input for control current / Sleep1 setting			
14	V _{DD}	Power supply to logic			
15	RESET	Reset input for internal logic			
16	CW/CCW	Forward / reverse input			
17	SYNC	Synchronous PMW control switch input			
18	FLAG	Output from protection circuit monitor			
19	SENSEB	Phase B current sensing			
20	OUTB				
21	OUTB	Phase B output			
22		Dhace D evitevit			
23	OUTB	Phase B output			





Pin No.	Symbol	Function				
1	OUTA	Phase A output				
2	OUTA	Filase A oulput				
3	OUTA	Phase A output				
4	UUIA	Filase A oulput				
5	SENSEA	Phase A current sensing				
6	MO	Output from 2-phase excitation status monitor				
7	M1					
8	M2	Input for excitation mode & Sleep2 setting				
9	M3					
10	CLOCK	Step clock input				
11	V _{BB}	Main power supply voltage (for motor)				
12	GND	Ground				
13	REF/SLEEP1	Input for control current / Sleep1 setting				
14	V _{DD}	Power supply to logic				
15	RESET	Reset input for internal logic				
16	CW/CCW	Forward / reverse input				
17	SYNC	Synchronous PMW control switch input				
18	FLAG	Output from protection circuit monitor for				
10	I LAG	detecting coil opens/shorts				
19	SENSEB	Phase B current sensing				
20	OUTB	Phase B output				
21						
22	OUTB	Phase P output				
23	OUTB	Phase B output				



- Take precautions to avoid noise on the V_{DD} line; noise levels greater than 0.5 V on the V_{DD} line may cause device malfunction. Noise can be reduced by separating the logic ground and the power ground on a PCB from the GND pin (Pin 12).
- Unused logic input pins (CW/CCW, M1, M2, M3, RESET, and SYNC) <u>must be pulled up or</u> <u>down to VDD or ground.</u> If those unused pins are left open, the device may malfunction.
- Unused logic output pins (MO, FLAG) <u>must be kept open</u>.

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9. Truth Tables

(1) Common Input Pins

Table 9-1 shows the truth table for the input pins common to both full-/half-stepping and microstepping products available in the SLA7070MS series.

			Clock	
			POS Edge	POS/NEG Edge
Pin Name	Low Level	High Level	(Standard)	(Optional Type W)
RESET	Normal operation	Logic reset		—
CW/CCW	Forward (CW)	Reverse (CCW)		
M1				
M2	Commutation (Slee	ep2 is not included)		
M3				
REF/SLEEP1	Normal operation Sleep1 function			
OVNO	Non-sync PWM control Sync PWM control			
SYNC				

Table 9-1.Truth Table for Common Input Pins

The Reset function is asynchronous. If an input on the RESET pin is high, the internal logic circuit is reset. At this point, if the REF/SLEEP1 pin stays low, outputs turn on at the starting point of excitation. Note that a signal on the RESET pin cannot control an output disable command.

Voltage across the REF/SLEEP1 pin controls PWM currents and the Sleep1 function.

- When $V_{REF} \le 1.5$ V (low level), the REF/SLEEP1 pin functions as the reference voltage input for normal operation.
- When $V_{REF} \ge 2.0 V$ (high level), the REF/SLEEP1 pin disables all outputs. This is the Sleep1 mode that disables the internal linear circuitry and minimizes the main power supply current, IBB. Although much of the internal circuitry is disabled, the logic circuit is still active. If an input signal on the CLOCK pin is asserted, the internal sequencer/translator circuit reacts and sets a step starting point for the next operation.

The Sleep2 function operates in the same way as the Sleep1 function does, except that the internal logic circuit enters the Hold mode. Therefore, in the Sleep2 mode, the internal sequencer/translator circuit is not activated even if a step command signal occurs on the CLOCK input pin.

The Sync function runs only at "2-phase excitation timing." If this function is used at other than the 2-phase excitation timing, an overall balance might collapse because PWM OFF-times and setting currents are different in each of phase A and phase B control scenario. (If this function is used at a point of 1-phase excitation, it does not react as the Sync function does. But there is no problem.) The 2-phase excitation timing is a point where the step reference current ratio of both phases A and B is either of Mode 8 or F.

(2) Commutation/Sleep2 Function Setting

Table 9-2 provides the logic for the pins (M1, M2, and M3) which set commutation.

Funct	tion (Pin N	Name)		
M1	M2	M3	Full/Half Step	Microstep
L	L	L	Full step (Mode 8 fixed)	Full step (Mode 8 fixed)
Н	L	L	Full step (Mode F fixed)	Full step (Mode F fixed)
L	Н	L	Half step (1-2 Phase)	Half step (1-2 Phase)
Н	Н	L	Half step (Mode F fixed)	Half step (Mode F fixed)
L	L	Н		Quarter step (W1-2 Phase)
Н	L	Н	Clear of typetion	Eighth step (2W1-2 Phase)
L	H	Н	Sleep2 function	Sixteenth step (4W1-2 Phase)
Н	H	Н		Sleep2 function

Table 9-2.	Truth Table for	Commutation/Sleep	2 Functions
	indin idolo ioi	oonninatation, oloopi	

NOTE: The Sleep2 function disables outputs and reduces the driver supply current (IBB) in the same way as the Sleep1 function does. However, unlike the Sleep1 function, the logic circuitry is put into a "standby" state in the Sleep2 function. Therefore, the sequencer/translator is not activated even if a step command signal occurs on the CLOCK input pin. When awaking from the Sleep2 mode, a delay of 100 µs or longer before sending a clock pulse is recommended.

(3) Monitor Output Pins

The SLA7070MS series provides two device status monitor outputs:

- MO pin (microstepping product only) Step sequence
- FLAG pin Protection feature operation

Table 9-3 shows the logic for the monitor output pins.

Table 9-3. Truth Table for Monitor Output Pins				
Pin Name	Low Level	High Level		
МО	Other than 2-phase excitation timing	2-phase excitation timing		
FLAG	Normal operation	Protection circuit operation		

NOTE: The outputs turn off at the point where the protection circuit starts operating. To release the protection state, cycle the logic supply voltage, VDD.

10. Logic Input Pins

The low pass filter (LPF) incorporated with the logic input pins (CLOCK, RESET, CW/CCW, M1, M2, M3, and SYNC) improves noise rejection.

The logic inputs are CMOS input compatible; therefore, they are in a high impedance state. Note that the IC should be used at a fixed input level, either low or high.

If there is a possibility that signals from the microcontroller are in high impedance, add a pull-up/-down resistor. Since outputs from the logic input pins, which function as output ON/OFF controllers, may result in abnormal oscillation, leading to MOSFET breakdown as the worst-case scenario.

11. Logic Input Timing

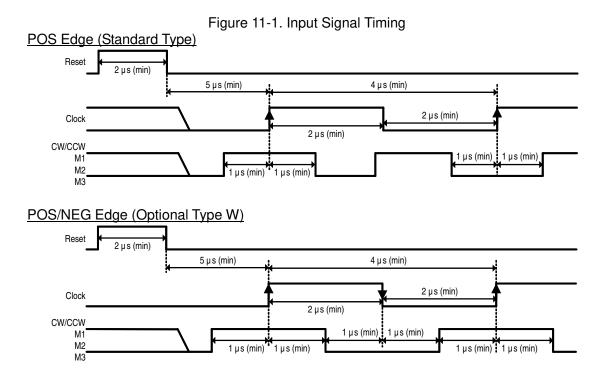
(1) Clock Signal

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a. A low-to-high transition (rising, or POS edge) or a low-to-high then high-to-low transition (rising and falling, or POS/NEG edge) on the CLOCK input signal advances the sequencer/translator. Clock pulse width should be set at 2 µs or longer in both positive and negative polarities. Therefore, clock response frequency is set to 250 kHz.

b. Clock Edge Timing

With regard to the input logic of the CW/CCW, M1, M2, and M3 pins, a 1 µs delay should occur both before and after a pulse edge, as setup and hold times (see Figure 11-1). The sequencer logic circuitry might malfunction if the logic polarity is changed during these setup and hold times.



NOTE: When awaking from the Sleep1 or Sleep2 mode, a delay of <u> $100 \ \mu s \ or \ longer$ </u> before sending a clock pulse is recommended.

(2) Reset Signal

a. Reset Signal Pulse Width

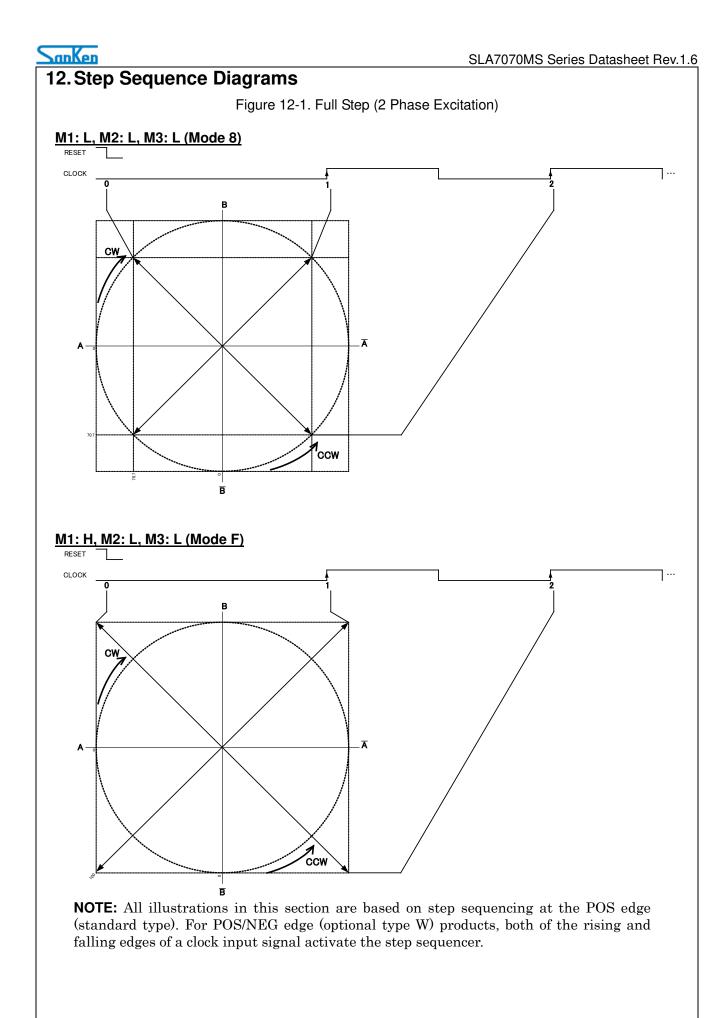
Reset pulse width is equivalent to the hold time of a high level input. It should be 2 μs or longer, same as the clock pulse width.

b. Reset Release and Clock Input Timing

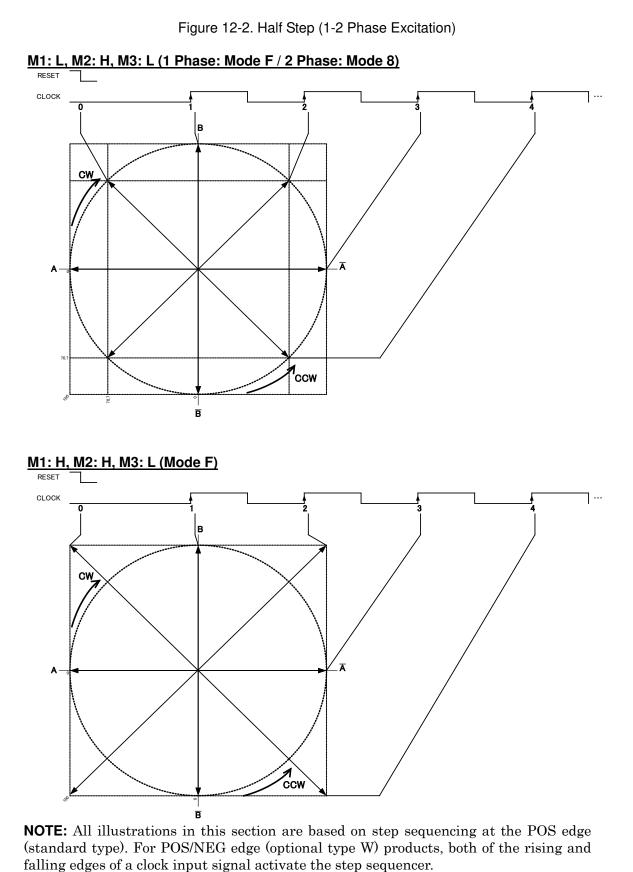
When the timing of a reset release (falling edge) and a clock edge is simultaneous, the internal logic might result in an unexpected operation. Therefore, a greater than 5 μ s delay is required between the falling edge of the RESET input signal and the next rising edge of the CLOCK input signal (see Figure 11-1).

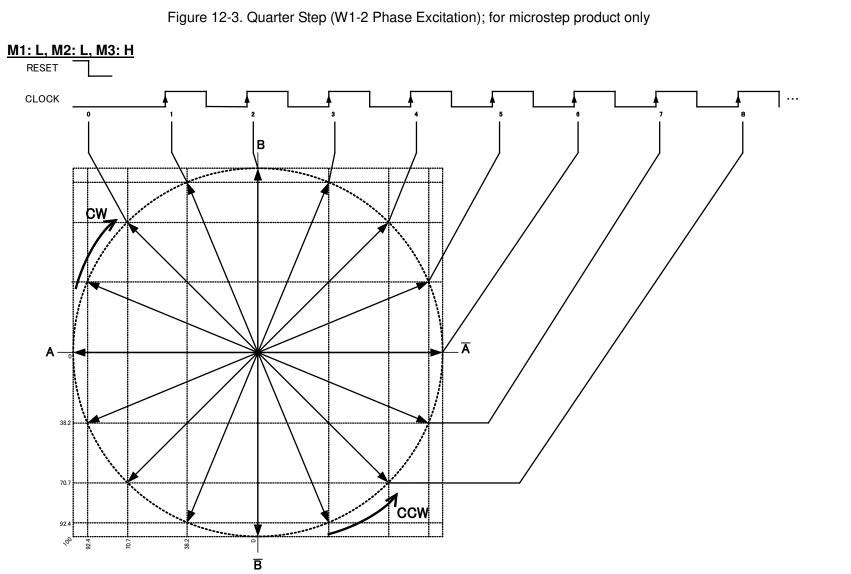
(3) Logic Level Change

Logic level inputs on CW/CCW, M1, M2, and M3 set the translator step direction (CW/CCW) and step mode (M1, M2, and M3; see also Table 9-2, the commutation truth table). Changes to those inputs do not take effect until the rising edge of an input signal on the CLOCK pin. However, depending on the type and state of a motor, there may be errors in motor operation such as step-out. A thorough evaluation on the changes of sequence should be carried out.

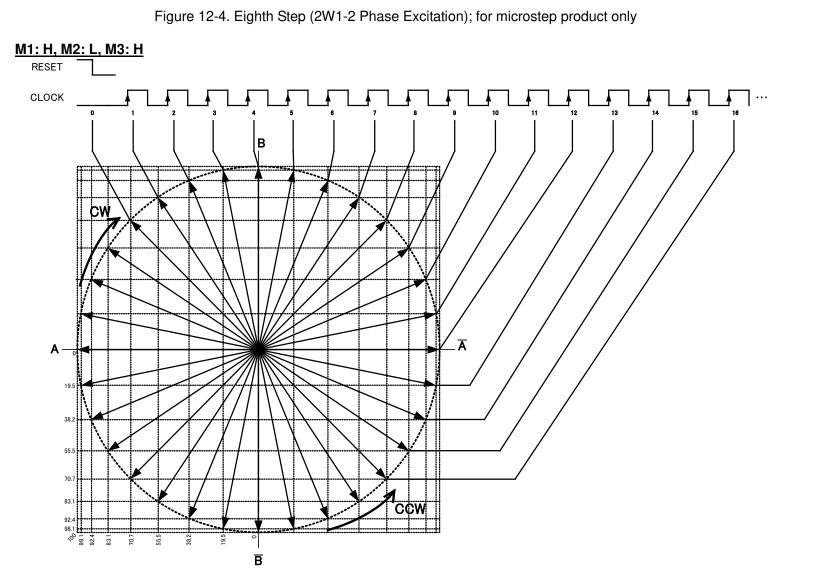






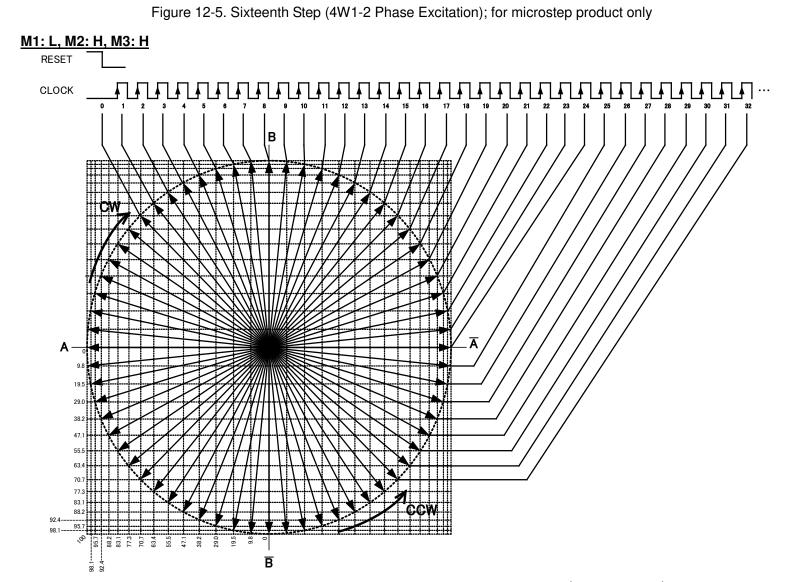


NOTE: All illustrations in this section are based on step sequencing at the POS edge (standard type). For POS/NEG edge (optional type W) products, both of the rising and falling edges of a clock input signal activate the step sequencer.



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Excitation Change Sequence

The change of excitation modes is determined by the settings of the excitation pins (M1, M2, and M3) before and after a step signal. Table 12-1 shows each excitation mode state setting.

	Internal Sequence State ¹⁾ Step Sequencing ²⁾										
Direction	Phase A Phase B			2 Phase (Full Stap) 1.2 Phase (Half Stap)			W1-2 Phase	2W1-2 4W1-2			
	PWM	Mode	PWM	Mode	Mode 8	Mode F	Mode 8/F	Mode F	(1/4 Step)	Phase (1/8 Step)	Phase (1/16 Step
	Α	8	В	8	Х	ХХ	Х	ХХ	Х	X	Х
	А	7	В	9							Х
	Α	6	В	A						Х	Х
	A	5	В	B					~	V	X
	A	4	B	C D					Х	Х	X
	A	2	B	E					1 1	Х	X
	A	1	B	F					1	~	X
	-	-	В	F			Х	Х	Х	Х	Х
	/A	1	В	F					1		Х
	/A	2	В	Е						Х	Х
	/A	3	В	D							Х
CCW	/A	4	В	С					Х	Х	Х
	/A	5	B	B						v	X
Ť	/A /A	6 7	B	A 9					1 1	Х	X
	/A	8	B	8	Х	ХХ	Х	ХХ	Х	Х	X
	/A	9	B	7							X
	/A	A	В	6						Х	Х
	/A	В	В	5							Х
	/A	С	В	4					Х	Х	Х
	/A	D	В	3						V	Х
	/A	E F	B	2						Х	X
	/A /A	F	B -	-			Х	Х	Х	Х	X
	/A /A	F	/B	1			~	^	~	~	X
	/A	E	/B	2						Х	X
	/A	D	/B	3							Х
	/A	С	/B	4					Х	Х	Х
	/A	В	/B	5							Х
	/A	A	/B	6						Х	Х
	/A	9	/B	7	v	~~~	v	~~	~	v	X
	/A /A	8	/B /B	8 9	Х	XX	Х	XX	Х	Х	X
	/A /A	6	/B	A						Х	X
	/A	5	/B	B							Х
	/A	4	/B	С					Х	Х	Х
	/A	3	/B	D							Х
	/A	2	/B	Е						Х	Х
	/A	1	/B	F			v	V	× ×	V	Х
	A	- 1	/B /B	F			Х	Х	Х	Х	X
	A	2	/в /В	E					+	Х	X
	A	3	/B	D					1 1	~	X
	A	4	/B	C					Х	Х	X
	Α	5	/B	В							Х
	Α	6	/B	Α						Х	Х
	A	7	/B	9	V	~~~		~~~		V	X
	A	8	/B /B	8	Х	XX	Х	XX	Х	Х	X
•	A	9 A	/B /B	76						Х	X
CW	A	B	/B	5						Λ	X
	A	C	/B	4					Х	Х	X
	Α	D	/B	3							Х
	А	Е	/B	2						Х	Х
	А	F	/B	1							Х
	A	F	•	-			Х	Х	Х	Х	Х
	A	F	В	1						v	X
	A	E D	B	2					+	Х	X
	A	C	B	3			-		Х	Х	X
	A	B	B	5					^	~	X
	A	A	B	6						Х	X
	А	9	В	7					1		Х



¹⁾ Each mode is defined accordingly to the SLA7070M series.

²⁾ XX indicates that sequence state is Mode 8; but step reference current ratio is Mode F. Mode F has a step reference current ratio of 100%, and a PWM OFF-time of 12 µs.

13. Individual Circuit Descriptions

(1) Monolithic IC (MIC)

Sequencer Logic

A single clock strategy is employed for step timing. An input on the CW/CCW pin determines the direction of motor rotation. Excitation mode is controlled by the combination of the M1, M2, and M3 input logic levels. See Section 9 for truth tables, and Section 11 for input timings.

• DAC (D-to-A Converter)

DACs that generate the reference voltage for controlling current. In microstep sequencing, the current at each step is set by the values of a sense resistor (R_s), a reference voltage (V_{REF}), the output voltage of the DACs, controlled by the output of the sequencer/translator circuit. For the step reference current ratios, see the electrical characteristics tables given in Section 4.

• PWM Control

Circuits that allow self-excitation PWM current controlling with a fixed OFF-time are used in this series. Each built-in oscillator (OSC) determines an OFF-time and a blanking time for proper PWM operation. The operation mechanism of the PWM control circuitry is identical to that of the SLA7070M family. For more detailed functional descriptions, see Section 14.

Synchronous Control

A synchronous chopping circuit that prevents occasional motor noise during a hold state which normally results from the asynchronous PWM operation of both motor phases. When the SYNC input pin is set to logic high, the circuit sends a timing signal that simultaneously turns off the chopping of phases A and B.

This function adopts the same operation mechanism applied to the SLA7070M series. Therefore, the use of the synchronous control during normal stepping is not recommended because, it produces less motor torque or may cause motor vibration due to staircase current.

The use of the synchronous control when the motor is not in operation is only allowed in 2-phase excitation timing, because the differences in current control values and PWM OFF-times between phases A and B exist at other excitation timings; otherwise, these two phases may not be synchronized or may be greatly disrupted in their current control values.

• Regulator Circuit

An integrated regulator circuit is used for powering the output MOSFET gate drive circuit (pre-driver) and other internal linear circuits.

• Protection Circuit

Built-in protection circuits against motor coil opens or shorts are provided. This protection is activated by sensing the voltage across internal sense resistors, R_S. Therefore, an overcurrent condition cannot be detected which results from the OUTx pins or SENSEx pins, or both, shorting to GND. The protection against motor coil opens is available only during PWM operation; therefore, it does not work at constant voltage driving, when the motor is rotating at a high speed.

The operation of the protection circuit disables all outputs. To come out of the Protection mode, cycle the logic supply, V_{DD} . For more details, see the next section.

• TSD Circuit

A TSD circuit that protects a driver by shifting an output to the Disable mode is incorporated. When the temperature of the product control IC (MIC) rises and becomes higher than its threshold, the circuit starts operating. To reset the function, perform the same steps as described in the Protection Circuit description.

(2) Output MOSFET Chip

The type of MOSFET chips to be mounted varies according to which of the two different output current ratings has been selected. For specifications, see Table 4-5.

Rated Current (A)	Resistance (ΩTyp.)	
2.0	0.25	
3.0	0.18	

NOTE: Each resistance shown above includes approximately 5 m Ω circuit resistance in addition to the resistance of the built-in resistor itself.

(3) Sense Resistor

Sense resistors are incorporated in this series to detect motor current. The resistance of these varies according to which of the two different output current ratings has been selected. For specifications, see Table 4-5.

Rated Current (A)	Resistance (ΩTyp.)		
2.0	0.205		
3.0	0.155		

NOTE: Each resistance shown above includes approximately 5 m Ω circuit resistance in addition to the resistance of the built-in resistor itself.

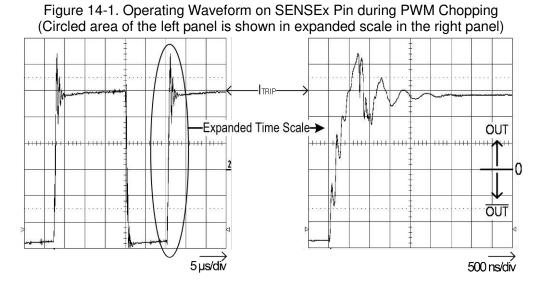
14. Functional Descriptions

SanKen

(1) PWM Current Control

[1] Blanking Time

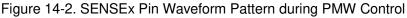
An actual operating waveform on the SENSEx pin when driving a motor is shown in Figure 14-1.

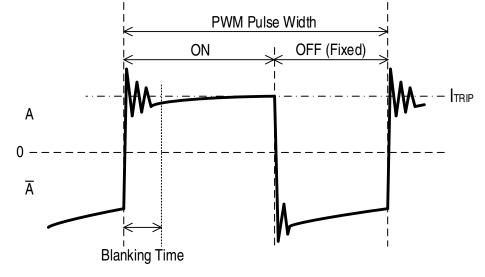


Immediately after a PWM turns off, ringing (or spike) noise on the SENSEx pin is observed for a period of a few microseconds. Ringing noise can be generated by various causes, such as capacitance between motor coils or inappropriate motor wiring.

Each pair of outputs is controlled by a fixed OFF-time PWM current-control circuit that limits the load current to a target value, ITRIP. Initially, an output is enabled and then currents flow through the motor winding and the current sense resistors. When the voltage across the current sense resistors equals the DAC output voltage, V_{TRIP} , the current sense comparator resets a PWM latch. This turns off the driver for the fixed OFF-time, during which the load inductance causes the current to recirculate for the OFF-time period. Therefore, if the ringing noise on the current sense resistor(s) equals and surpasses V_{TRIP} , the PWM turns off (i.e., a hunting phenomenon).

To prevent this phenomenon, a blanking time is set to override signals from the current sense comparator for a certain period immediately after the PWM turns on (Figure 14-2).

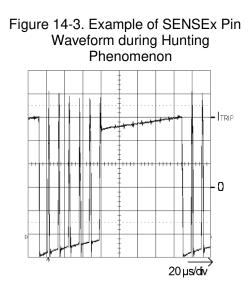




[2] Blanking Time and Hunting Phenomenon

Although current control can be improved by shortening a blanking time, the degree of margin to a ringing noise decreases simultaneously. For this reason, when a motor is driven by the device, a hunting phenomenon may occur. Figure 14-3 shows an example of the waveform pattern when the phenomenon occurs.

In order to overcome this problem, Sanken has released a new option, "type B", which offers a longer blanking time. Having the longer blanking time, the optional type B can improve problems such as torque reduction and huge motor noise that are occasionally found during the hunting phenomenon.



[3] Blanking Time Difference

Table 14-1 shows characteristic differences between two blanking times, shorter and longer blanking periods.

This comparison is based on the case where drive conditions, such as a motor, motor power supply voltage, REF input voltage, and circuit constant were kept the same while only the indicated parameters were changed.

Table 14-1. Onaracteristic Companson of Difference in Dianking Time					
Parameter	Better Performance				
Internal blanking time	Short	Long			
Minimum PWM ON-time	Small 🗲				
Ringing noise suppression		→ Large			
Minimum coil current	Small 🗲				
Coil current waveform distortion at a high rotation (mainly microstep)		→ Large			

Table 14-1. Characteristic Comparison of Difference in Blanking Time

Brief descriptions for each parameter are as follows:

• Minimum PWM ON-time, t_{ON(min)}

This series has a blanking time that is effectively selected and fixed by the PWM control. Therefore, even if an application attempts to shorten its ON-time for limiting currents, it would not go below the fixed blanking time. Minimum PWM ON-time refers to the time when an output is on during this blanking time period, that is, when an output MOSFET is actually turned on. In other words, a blanking time determines a minimum ON-time ("Small" in Table 14-1).

Minimum Coil Current

This refers to the coil current when the PWM control is performed during a minimum PWM ON-time. In other words, the device with a shorter blanking time can reduce more coil current.

Coil Current Waveform Distortion during High-Velocity Revolution

While a microstep drive is active, the I_{TRIP} value changes to a predetermined value in accordance with a clock input. The I_{TRIP} value (internal reference voltage splitting ratio) is then set up to be a sine wave. Because the PWM control of motor coil current is set according to the I_{TRIP} value, (the envelope of) the motor coil current will also be controlled to be sine wave-like.