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# SSTUP32866

1.8 V 25-bit 1 : 1 or 14-bit 1 : 2 configurable registered buffer with parity and programmable output for DDR2-800 RDIMMs

Rev. 02 — 14 September 2006

Product data sheet

## 1. General description

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The SSTUP32866 is a 1.8 V configurable register specifically designed for use on DDR2 memory modules requiring a parity checking function. It is defined in accordance with the JEDEC standard for the SSTUA32866 and SSTUB32866 registered buffers. The register is configurable (using configuration pins C0 and C1) to two topologies: 25-bit 1 : 1 or 14-bit 1 : 2, and in the latter configuration can be designated as Register A or Register B on the DIMM. It offers added features over the JEDEC standard register in that it can be configured for high or normal output drive strength, as well as for operation to 667 MT/s or 800 MT/s, simply by tying two input pins HIGH or LOW as needed.

The SSTUP32866 accepts a parity bit from the memory controller on its parity bit (PAR\_IN) input, compares it with the data received on the DIMM-independent D-inputs and indicates whether a parity error has occurred on its open-drain  $\overline{QERR}$  pin (active LOW). The convention is even parity, that is, valid parity is defined as an even number of ones across the DIMM-independent data inputs combined with the parity input bit.

The SSTUP32866 is packaged in a 96-ball, 6 × 16 grid, 0.8 mm ball pitch LFBGA package (13.5 mm × 5.5 mm).

## 2. Features

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- Configurable register supporting DDR2 up to 667 MT/s or 800 MT/s Registered DIMM applications
- Configurable to 25-bit 1 : 1 mode or 14-bit 1 : 2 mode
- Programmable for normal or high output drive
- Controlled multi-impedance output drivers enable optimal signal integrity and speed
- Programmable for 667 MT/s or 800 MT/s speed
- Excellent propagation delay performance
- Supports up to 450 MHz clock frequency of operation
- Optimized pinout for high-density DDR2 module design
- Chip-selects minimize power consumption by gating data outputs from changing state
- Supports SSTL\_18 data inputs
- Checks parity on the DIMM-independent data inputs
- Partial parity output and input allows cascading of two SSTUP32866s for correct parity error processing
- Differential clock (CK and  $\overline{CK}$ ) inputs
- Supports LVCMOS switching levels on the control and  $\overline{RESET}$  inputs
- Single 1.8 V supply operation (1.7 V to 2.0 V)

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- Available in 96-ball, 13.5 mm × 5.5 mm, 0.8 mm ball pitch LFBGA package

### 3. Applications

- 667 MT/s to 800 MT/s DDR2 registered DIMMs desiring parity checking functionality

### 4. Ordering information

Table 1. Ordering information

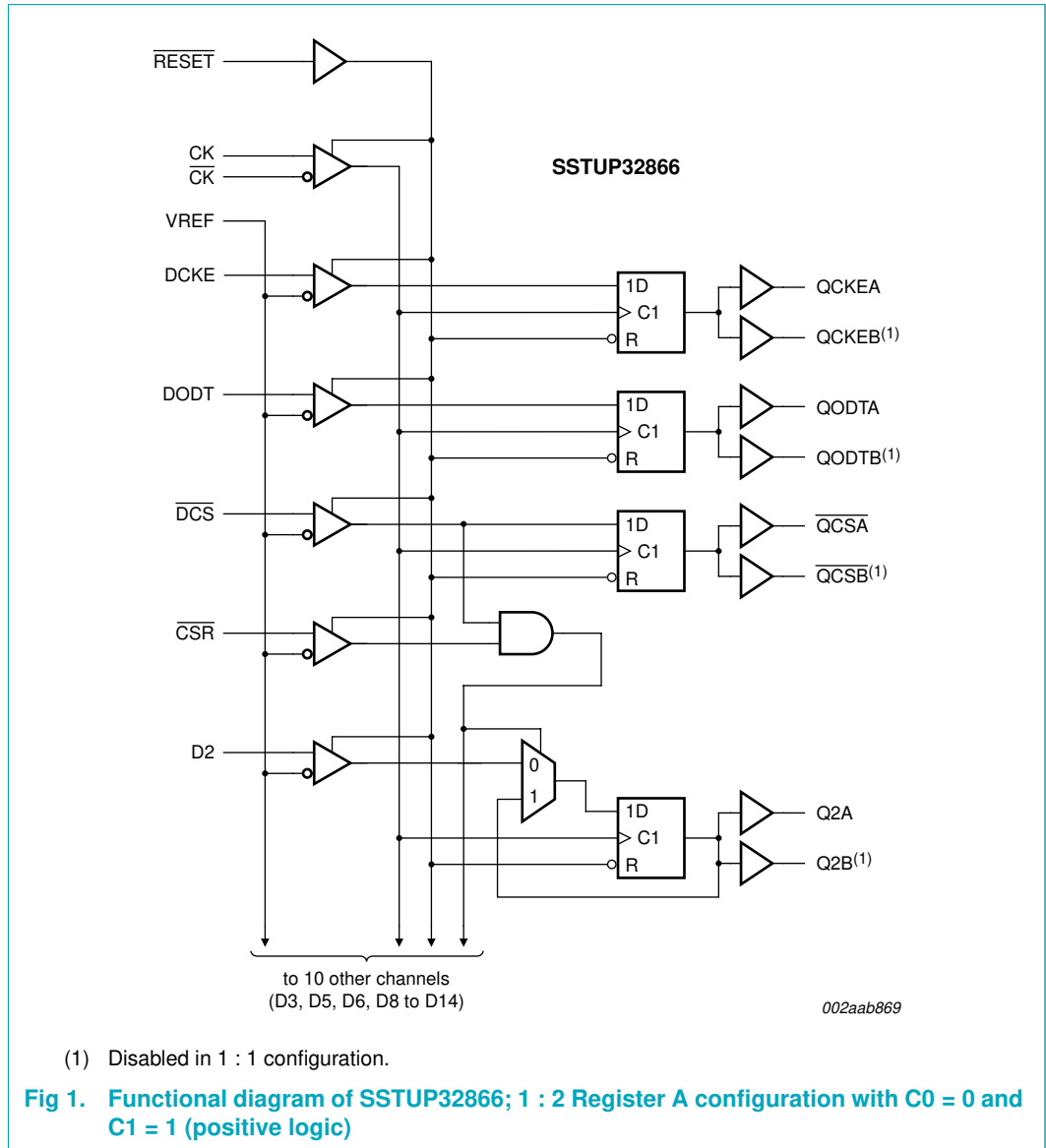
Type number	Solder process	Package		
		Name	Description	Version
SSTUP32866EC/G	Pb-free (SnAgCu solder ball compound)	LFBGA96	plastic low profile fine-pitch ball grid array package; 96 balls; body 13.5 × 5.5 × 1.05 mm	SOT536-1
SSTUP32866EC/S	Pb-free (SnAgCu solder ball compound)	LFBGA96	plastic low profile fine-pitch ball grid array package; 96 balls; body 13.5 × 5.5 × 1.05 mm	SOT536-1

#### 4.1 Ordering options

Table 2. Ordering options

Type number	Temperature range
SSTUP32866EC/G	T <sub>amb</sub> = 0 °C to +70 °C
SSTUP32866EC/S	T <sub>amb</sub> = 0 °C to +85 °C

5. Functional diagram



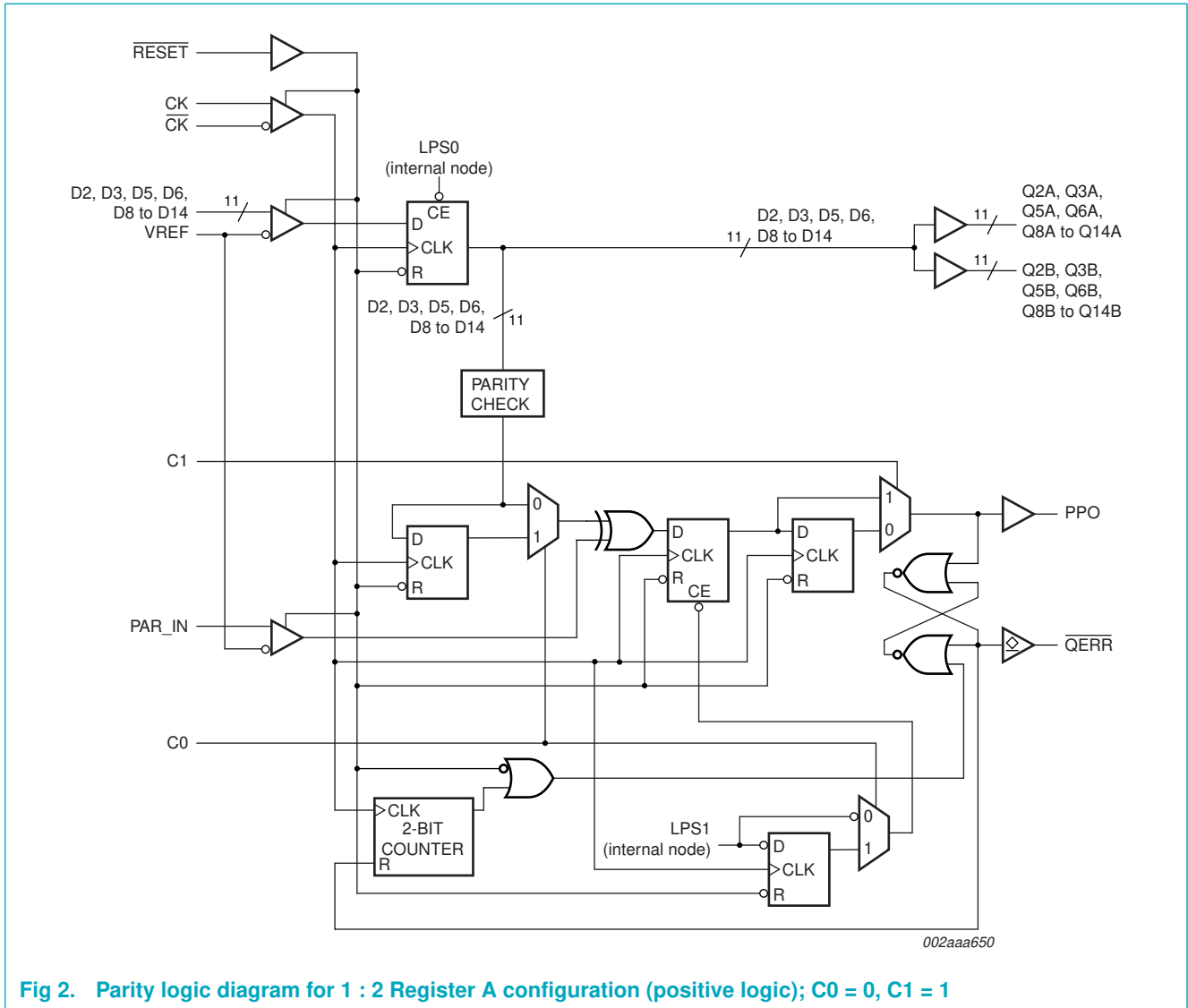
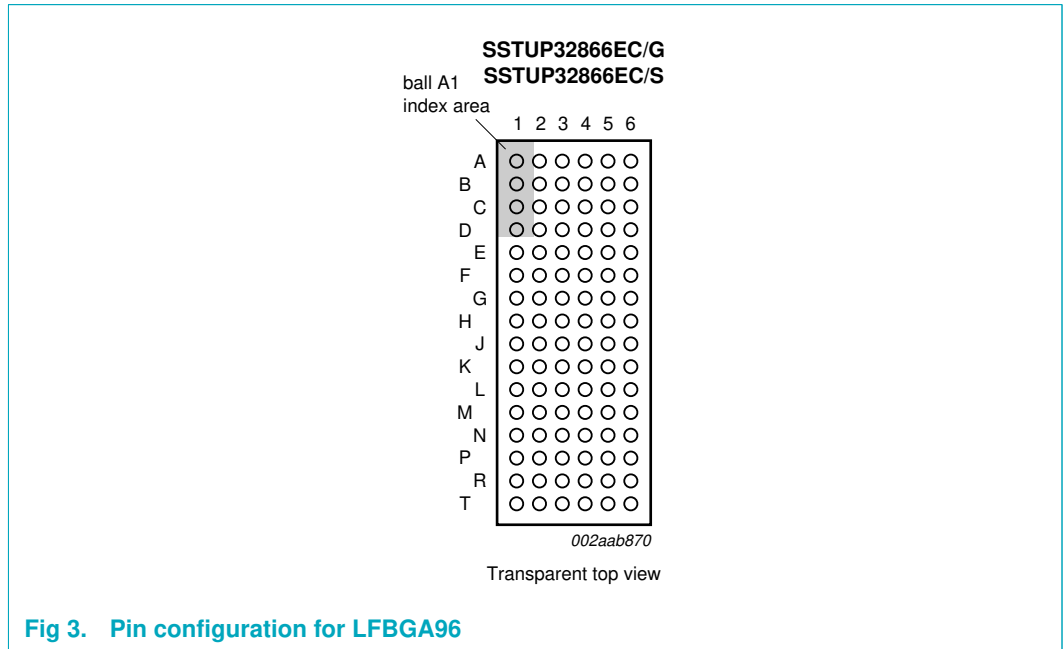


Fig 2. Parity logic diagram for 1 : 2 Register A configuration (positive logic); C0 = 0, C1 = 1

## 6. Pinning information

### 6.1 Pinning



	1	2	3	4	5	6
A	DCKE	PPO	VREF	V <sub>DD</sub>	QCKE	DNU
B	D2	D15	GND	GND	Q2	Q15
C	D3	D16	V <sub>DD</sub>	V <sub>DD</sub>	Q3	Q16
D	DODT	$\overline{QERR}$	GND	GND	QODT	DNU
E	D5	D17	V <sub>DD</sub>	V <sub>DD</sub>	Q5	Q17
F	D6	D18	GND	GND	Q6	Q18
G	PAR_IN	$\overline{RESET}$	V <sub>DD</sub>	V <sub>DD</sub>	C1	C0
H	CK	$\overline{DCS}$	GND	GND	$\overline{QCS}$	DNU
J	$\overline{CK}$	$\overline{CSR}$	V <sub>DD</sub>	V <sub>DD</sub>	SELAB	SELDLDR
K	D8	D19	GND	GND	Q8	Q19
L	D9	D20	V <sub>DD</sub>	V <sub>DD</sub>	Q9	Q20
M	D10	D21	GND	GND	Q10	Q21
N	D11	D22	V <sub>DD</sub>	V <sub>DD</sub>	Q11	Q22
P	D12	D23	GND	GND	Q12	Q23
R	D13	D24	V <sub>DD</sub>	V <sub>DD</sub>	Q13	Q24
T	D14	D25	VREF	V <sub>DD</sub>	Q14	Q25

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**Fig 4. Ball mapping, 1 : 1 register (C0 = 0, C1 = 0)**

	1	2	3	4	5	6
A	DCKE	PPO	VREF	V <sub>DD</sub>	QCKEA	QCKEB
B	D2	DNU	GND	GND	Q2A	Q2B
C	D3	DNU	V <sub>DD</sub>	V <sub>DD</sub>	Q3A	Q3B
D	DODT	$\overline{QERR}$	GND	GND	QODTA	QODTB
E	D5	n.c.	V <sub>DD</sub>	V <sub>DD</sub>	Q5A	Q5B
F	D6	n.c.	GND	GND	Q6A	Q6B
G	PAR_IN	$\overline{RESET}$	V <sub>DD</sub>	V <sub>DD</sub>	C1	C0
H	CK	$\overline{DCS}$	GND	GND	$\overline{QCSA}$	$\overline{QCSB}$
J	$\overline{CK}$	$\overline{CSR}$	V <sub>DD</sub>	V <sub>DD</sub>	SELAB	SELDR
K	D8	DNU	GND	GND	Q8A	Q8B
L	D9	DNU	V <sub>DD</sub>	V <sub>DD</sub>	Q9A	Q9B
M	D10	DNU	GND	GND	Q10A	Q10B
N	D11	DNU	V <sub>DD</sub>	V <sub>DD</sub>	Q11A	Q11B
P	D12	DNU	GND	GND	Q12A	Q12B
R	D13	DNU	V <sub>DD</sub>	V <sub>DD</sub>	Q13A	Q13B
T	D14	DNU	VREF	V <sub>DD</sub>	Q14A	Q14B

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**Fig 5. Ball mapping, 1 : 2 Register A (C0 = 0, C1 = 1)**

	1	2	3	4	5	6
A	D1	PPO	VREF	V <sub>DD</sub>	Q1A	Q1B
B	D2	DNU	GND	GND	Q2A	Q2B
C	D3	DNU	V <sub>DD</sub>	V <sub>DD</sub>	Q3A	Q3B
D	D4	$\overline{QERR}$	GND	GND	Q4A	Q4B
E	D5	DNU	V <sub>DD</sub>	V <sub>DD</sub>	Q5A	Q5B
F	D6	DNU	GND	GND	Q6A	Q6B
G	PAR_IN	$\overline{RESET}$	V <sub>DD</sub>	V <sub>DD</sub>	C1	C0
H	CK	$\overline{DCS}$	GND	GND	$\overline{QCSA}$	$\overline{QCSB}$
J	$\overline{CK}$	$\overline{CSR}$	V <sub>DD</sub>	V <sub>DD</sub>	SELAB	SELDR
K	D8	DNU	GND	GND	Q8A	Q8B
L	D9	DNU	V <sub>DD</sub>	V <sub>DD</sub>	Q9A	Q9B
M	D10	DNU	GND	GND	Q10A	Q10B
N	DODT	DNU	V <sub>DD</sub>	V <sub>DD</sub>	QODTA	QODTB
P	D12	DNU	GND	GND	Q12A	Q12B
R	D13	DNU	V <sub>DD</sub>	V <sub>DD</sub>	Q13A	Q13B
T	DCKE	DNU	VREF	V <sub>DD</sub>	QCKEA	QCKEB

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**Fig 6. Ball mapping, 1 : 2 Register B (C0 = 1, C1 = 1)**

## 6.2 Pin description

Table 3. Pin description

Symbol	Pin	Type	Description
GND	B3, B4, D3, D4, F3, F4, H3, H4, K3, K4, M3, M4, P3, P4	ground input	ground
V <sub>DD</sub>	A4, C3, C4, E3, E4, G3, G4, J3, J4, L3, L4, N3, N4, R3, R4, T4	1.8 V nominal	power supply voltage
VREF	A3, T3	0.9 V nominal	input reference voltage
CK	H1	differential input	positive master clock input
$\overline{CK}$	J1	differential input	negative master clock input
C0	G6	LVC MOS inputs	Configuration control inputs; Register A or Register B and 1 : 1 mode or 1 : 2 mode select.
C1	G5		
SELDR	J6	LVC MOS input	Selects output drive strength: HIGH for normal drive, LOW for high drive. This pin will default HIGH if left open-circuit (built-in weak pull-up resistor).
SELAB	J5	LVC MOS input	Selects speed grade: HIGH for DDR2-667, LOW for DDR2-800. This pin will default HIGH if left open-circuit (built-in weak pull-up resistor).
$\overline{RESET}$	G2	LVC MOS input	Asynchronous reset input (active LOW). Resets registers and disables VREF data and clock.
CSR	J2	SSTL <sub>18</sub> input	Chip select inputs (active LOW). Disables D1 to D25 <sup>[1]</sup> outputs switching when both inputs are HIGH.
$\overline{DCS}$	H2		
D1 to D25	<a href="#">[2]</a>	SSTL <sub>18</sub> input	Data input. Clocked in on the crossing of the rising edge of CK and the falling edge of CK.
DODT	<a href="#">[2]</a>	SSTL <sub>18</sub> input	The outputs of this register bit will not be suspended by the $\overline{DCS}$ and $\overline{CSR}$ control.
DCKE	<a href="#">[2]</a>	SSTL <sub>18</sub> input	The outputs of this register bit will not be suspended by the $\overline{DCS}$ and $\overline{CSR}$ control.
PAR_IN	G1	SSTL <sub>18</sub> input	Parity input. Arrives one clock cycle after the corresponding data input.
Q1 to Q25, Q2A to Q14A, Q1B to Q14B	<a href="#">[2]</a>	1.8 V CMOS outputs	Data outputs that are suspended by the $\overline{DCS}$ and $\overline{CSR}$ control <sup>[3]</sup> .
PPO	A2	1.8 V CMOS output	Partial parity out. Indicates odd parity of inputs D1 to D25 <sup>[1]</sup> .
$\overline{QCS}$ , $\overline{QCSA}$ , $\overline{QCSB}$	<a href="#">[2]</a>	1.8 V CMOS output	Data output that will not be suspended by the $\overline{DCS}$ and $\overline{CSR}$ control.
QODT, QODTA, QODTB	<a href="#">[2]</a>	1.8 V CMOS output	Data output that will not be suspended by the $\overline{DCS}$ and $\overline{CSR}$ control.



Table 3. Pin description ...continued

Symbol	Pin	Type	Description
QCKE, QCKEA, QCKEB	[2]	1.8 V CMOS output	Data output that will not be suspended by the $\overline{DCS}$ and $\overline{CSR}$ control.
$\overline{QERR}$	D2	open-drain output	Output error bit (active LOW). Generated one clock cycle after the corresponding data output
DNU	[2]	-	Do not use. Inputs are in standby-equivalent mode and outputs are driven LOW.

- [1] Data inputs = D2, D3, D5, D6, D8 to D25 when C0 = 0 and C1 = 0.  
 Data inputs = D2, D3, D5, D6, D8 to D14 when C0 = 0 and C1 = 1.  
 Data inputs = D1 to D6, D8 to D10, D12, D13 when C0 = 1 and C1 = 1.
- [2] Depends on configuration. See [Figure 4](#), [Figure 5](#), and [Figure 6](#) for ball number.
- [3] Data outputs = Q2, Q3, Q5, Q6, Q8 to Q25 when C0 = 0 and C1 = 0.  
 Data outputs = Q2, Q3, Q5, Q6, Q8 to Q14 when C0 = 0 and C1 = 1.  
 Data outputs = Q1 to Q6, Q8 to Q10, Q12, Q13 when C0 = 1 and C1 = 1.

## 7. Functional description

The SSTUP32866 is a 25-bit 1 : 1 or 14-bit 1 : 2 configurable registered buffer with parity, designed for 1.7 V to 2.0 V  $V_{DD}$  operation. Additionally, the SSTUP32866 can be programmed to deliver either normal or high output drive, and either 600 MT/s or 800 MT/s speeds.

Two programming pins, SELAB and SELDR, allow the user to respectively select speed and drive strength options by tying these pins either LOW or HIGH on the DIMM. The truth table for these options is shown in [Table 6](#).

All clock and data inputs are compatible with the JEDEC standard for SSTL\_18. The control and reset ( $\overline{RESET}$ ) inputs are LVCMOS. All data outputs are 1.8 V CMOS drivers that have been optimized to drive the DDR2 DIMM load, and meet SSTL\_18 specifications. The error ( $\overline{QERR}$ ) output is 1.8 V open-drain driver.

The SSTUP32866 operates from a differential clock (CK and  $\overline{CK}$ ). Data are registered at the crossing of CK going HIGH, and CK going LOW.

The C0 input controls the pinout configuration for the 1 : 2 pinout from A configuration (when LOW) to B configuration (when HIGH). The C1 input controls the pinout configuration from 25-bit 1 : 1 (when LOW) to 14-bit 1 : 2 (when HIGH).

The SSTUP32866 accepts a parity bit from the memory controller on its parity bit (PAR\_IN) input, compares it with the data received on the DIMM-independent D-inputs and indicates whether a parity error has occurred on its open-drain  $\overline{QERR}$  pin (active LOW). The convention is even parity, that is, valid parity is defined as an even number of ones across the DIMM-independent data inputs combined with the parity input bit.

When used as a single device, the C0 and C1 inputs are tied LOW. In this configuration, parity is checked on the PAR\_IN input which arrives one cycle after the input data to which it applies. The Partial-Parity-Out (PPO) and  $\overline{QERR}$  signals are produced three cycles after the corresponding data inputs.

When used in pairs, the C0 input of the first register is tied LOW and the C0 input of the second register is tied HIGH. The C1 input of both registers are tied HIGH. Parity, which arrives one cycle after the data input to which it applies, is checked on the PAR\_IN input of the first device. The PPO and  $\overline{QERR}$  signals are produced on the second device three clock cycles after the corresponding data inputs. The PPO output of the first register is cascaded to the PAR\_IN of the second register. The  $\overline{QERR}$  output of the first register is left floating and the valid error information is latched on the  $\overline{QERR}$  output of the second register.

If an error occurs and the  $\overline{QERR}$  output is driven LOW, it stays latched LOW for two clock cycles or until  $\overline{RESET}$  is driven LOW. The DIMM-dependent signals ( $\overline{DCKE}$ ,  $\overline{DCS}$ ,  $\overline{DODT}$ , and  $\overline{CSR}$ ) are not included in the parity check computation.

The device supports low-power standby operation. When  $\overline{RESET}$  is LOW, the differential input receivers are disabled, and undriven (floating) data, clock and reference voltage ( $V_{REF}$ ) inputs are allowed. In addition, when  $\overline{RESET}$  is LOW all registers are reset, and all outputs are forced LOW. The LVCMOS  $\overline{RESET}$  input must always be held at a valid logic HIGH or LOW level.

The device also supports low-power active operation by monitoring both system chip select ( $\overline{DCS}$  and  $\overline{CSR}$ ) inputs and will gate the Qn and PPO outputs from changing states when both  $\overline{DCS}$  and  $\overline{CSR}$  inputs are HIGH. If either  $\overline{DCS}$  or  $\overline{CSR}$  input is LOW, the Qn and PPO outputs will function normally. The  $\overline{RESET}$  input has priority over the  $\overline{DCS}$  and  $\overline{CSR}$  control and when driven LOW will force the Qn and PPO outputs LOW, and the  $\overline{QERR}$  output HIGH. If the  $\overline{DCS}$  control functionality is not desired, then the  $\overline{CSR}$  input can be hard-wired to ground, in which case, the setup time requirement for  $\overline{DCS}$  would be the same as for the other Dn data inputs. To control the low-power mode with  $\overline{DCS}$  only, then the  $\overline{CSR}$  input should be pulled up to  $V_{DD}$  through a pull-up resistor.

To ensure defined outputs from the register before a stable clock has been supplied,  $\overline{RESET}$  must be held in the LOW state during power-up.

In the DDR2 RDIMM application,  $\overline{RESET}$  is specified to be completely asynchronous with respect to CK and  $\overline{CK}$ . Therefore, no timing relationship can be guaranteed between the two. When entering reset, the register will be cleared and the Qn outputs will be driven LOW quickly, relative to the time to disable the differential input receivers. However, when coming out of reset, the register will become active quickly, relative to the time to enable the differential input receivers. As long as the data inputs are LOW, and the clock is stable during the time from the LOW-to-HIGH transition of  $\overline{RESET}$  until the input receivers are fully enabled, the design of the SSTUP32866 must ensure that the outputs will remain LOW, thus ensuring no glitches on the output.

## 7.1 Function table

**Table 4. Function table (each flip-flop)**

L = LOW voltage level; H = HIGH voltage level; X = don't care;  $\uparrow$  = LOW-to-HIGH transition;  $\downarrow$  = HIGH-to-LOW transition.

Inputs						Outputs <sup>[1]</sup>		
RESET	DCS	CSR	CK	$\overline{CK}$	Dn, DODTn, DCKEn	Qn	$\overline{QCS}$	QODT, QCKE
H	L	L	$\uparrow$	$\downarrow$	L	L	L	L
H	L	L	$\uparrow$	$\downarrow$	H	H	L	H
H	L	L	L or H	L or H	X	Q <sub>0</sub>	Q <sub>0</sub>	Q <sub>0</sub>
H	L	H	$\uparrow$	$\downarrow$	L	L	L	L
H	L	H	$\uparrow$	$\downarrow$	H	H	L	H
H	L	H	L or H	L or H	X	Q <sub>0</sub>	Q <sub>0</sub>	Q <sub>0</sub>
H	H	L	$\uparrow$	$\downarrow$	L	L	H	L
H	H	L	$\uparrow$	$\downarrow$	H	H	H	H
H	H	L	L or H	L or H	X	Q <sub>0</sub>	Q <sub>0</sub>	Q <sub>0</sub>
H	H	H	$\uparrow$	$\downarrow$	L	Q <sub>0</sub>	H	L
H	H	H	$\uparrow$	$\downarrow$	H	Q <sub>0</sub>	H	H
H	H	H	L or H	L or H	X	Q <sub>0</sub>	Q <sub>0</sub>	Q <sub>0</sub>
L	X or floating	X or floating	X or floating	X or floating	X or floating	L	L	L

[1] Q<sub>0</sub> is the previous state of the associated output.

**Table 5. Parity and standby function table**

L = LOW voltage level; H = HIGH voltage level; X = don't care;  $\uparrow$  = LOW-to-HIGH transition;  $\downarrow$  = HIGH-to-LOW transition.

Inputs						Outputs <sup>[1]</sup>		
RESET	DCS	CSR	CK	$\overline{CK}$	$\Sigma$ of inputs = H (D1 to D25)	PAR_IN <sup>[2]</sup>	PPO <sup>[3]</sup>	$\overline{QERR}$ <sup>[4]</sup>
H	L	X	$\uparrow$	$\downarrow$	even	L	L	H
H	L	X	$\uparrow$	$\downarrow$	odd	L	H	L
H	L	X	$\uparrow$	$\downarrow$	even	H	H	L
H	L	X	$\uparrow$	$\downarrow$	odd	H	L	H
H	H	L	$\uparrow$	$\downarrow$	even	L	L	H
H	H	L	$\uparrow$	$\downarrow$	odd	L	H	L
H	H	L	$\uparrow$	$\downarrow$	even	H	H	L
H	H	L	$\uparrow$	$\downarrow$	odd	H	L	H
H	H	H	$\uparrow$	$\downarrow$	X	X	PPO <sub>0</sub>	$\overline{QERR}_0$
H	X	X	L or H	L or H	X	X	PPO <sub>0</sub>	$\overline{QERR}_0$
L	X or floating	X or floating	X or floating	X or floating	X or floating	X or floating	L	H

[1] PPO<sub>0</sub> is the previous state of output PPO;  $\overline{QERR}_0$  is the previous state of output  $\overline{QERR}$ .

[2] Data inputs = D2, D3, D5, D6, D8 to D25 when C0 = 0 and C1 = 0.  
Data inputs = D2, D3, D5, D6, D8 to D14 when C0 = 0 and C1 = 1.  
Data inputs = D1 to D6, D8 to D10, D12, D13 when C0 = 1 and C1 = 1.

[3] PAR\_IN arrives one clock cycle (C0 = 0), or two clock cycles (C0 = 1), after the data to which it applies.

[4] This condition assumes  $\overline{QERR}$  is HIGH at the crossing of CK going HIGH and  $\overline{CK}$  going LOW. If  $\overline{QERR}$  is LOW, it stays latched LOW for two clock cycles or until RESET is driven LOW.

Table 6. Speed and drive programmability

Inputs		Mode
SELDR	SELAB	
L	L	DDR2-800; high output drive
L	H	DDR2-667; high output drive
H	L	DDR2-800; normal output drive
H	H	DDR2-667; normal output drive

## 8. Limiting values

Table 7. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{DD}$	supply voltage		-0.5	+2.5	V
$V_I$	input voltage	receiver	-0.5 <sup>[1]</sup>	+2.5 <sup>[2]</sup>	V
$V_O$	output voltage	driver	-0.5 <sup>[1]</sup>	$V_{DD} + 0.5$ <sup>[2]</sup>	V
$I_{IK}$	input clamping current	$V_I < 0\text{ V}$ or $V_I > V_{DD}$	-	-50	mA
$I_{OK}$	output clamping current	$V_O < 0\text{ V}$ or $V_O > V_{DD}$	-	$\pm 50$	mA
$I_O$	output current	continuous; $0\text{ V} < V_O < V_{DD}$	-	$\pm 50$	mA
$I_{CCC}$	continuous current through each $V_{DD}$ or GND pin		-	$\pm 100$	mA
$T_{stg}$	storage temperature		-65	+150	°C
$V_{esd}$	electrostatic discharge voltage	Human Body Model (HBM); 1.5 k $\Omega$ ; 100 pF	2	-	kV
		Machine Model (MM); 0 $\Omega$ ; 200 pF	200	-	V

[1] The input and output negative voltage ratings may be exceeded if the input and output clamping current ratings are observed.

[2] This value is limited to 2.5 V maximum.

## 9. Recommended operating conditions

**Table 8. Recommended operating conditions**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{DD}$	supply voltage		1.7	-	2.0	V
$V_{ref}$	reference voltage		$0.49 \times V_{DD}$	$0.50 \times V_{DD}$	$0.51 \times V_{DD}$	V
$V_T$	termination voltage		$V_{ref} - 0.040$	$V_{ref}$	$V_{ref} + 0.040$	V
$V_I$	input voltage		0	-	$V_{DD}$	V
$V_{IH(AC)}$	AC HIGH-level input voltage	data (Dn), $\overline{CSR}$ , and PAR_IN inputs	$V_{ref} + 0.250$	-	-	V
$V_{IL(AC)}$	AC LOW-level input voltage	data (Dn), $\overline{CSR}$ , and PAR_IN inputs	-	-	$V_{ref} - 0.250$	V
$V_{IH(DC)}$	DC HIGH-level input voltage	data (Dn), $\overline{CSR}$ , and PAR_IN inputs	$V_{ref} + 0.125$	-	-	V
$V_{IL(DC)}$	DC LOW-level input voltage	data (Dn), $\overline{CSR}$ , and PAR_IN inputs	-	-	$V_{ref} - 0.125$	V
$V_{IH}$	HIGH-level input voltage	$\overline{RESET}$ , Cn, SELAB, SELDR	[1] $0.65 \times V_{DD}$	-	-	V
$V_{IL}$	LOW-level input voltage	$\overline{RESET}$ , Cn, SELAB, SELDR	[1] -	-	$0.35 \times V_{DD}$	V
$V_{ICR}$	common mode input voltage range	CK, $\overline{CK}$	[2] 0.675	-	1.125	V
$V_{ID}$	differential input voltage	CK, $\overline{CK}$	[2] 600	-	-	mV
$I_{OH}$	HIGH-level output current	SELDR either HIGH or LOW	-	-	-8	mA
$I_{OL}$	LOW-level output current	SELDR either HIGH or LOW	-	-	8	mA
$T_{amb}$	ambient temperature	operating in free air				
		SSTUP32866EC/G	0	-	70	°C
		SSTUP32866EC/S	0	-	85	°C

[1] The  $\overline{RESET}$  and Cn inputs of the device must be held at valid levels (not floating) to ensure proper device operation.

[2] The differential inputs must not be floating, unless  $\overline{RESET}$  is LOW.

## 10. Characteristics

**Table 9. Characteristics**

At recommended operating conditions (see [Table 8](#)); unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{OH}$	HIGH-level output voltage	$I_{OH} = -6 \text{ mA}$ ; $V_{DD} = 1.7 \text{ V}$	1.2	-	-	V
$V_{OL}$	LOW-level output voltage	$I_{OL} = 6 \text{ mA}$ ; $V_{DD} = 1.7 \text{ V}$	-	-	0.5	V
$I_I$	input current	all inputs; $V_I = V_{DD}$ or GND; $V_{DD} = 2.0 \text{ V}$	-	-	$\pm 5$	$\mu\text{A}$
$I_{DD}$	supply current	static Standby mode; $\overline{\text{RESET}} = \text{GND}$ ; $I_O = 0 \text{ mA}$ ; $V_{DD} = 2.0 \text{ V}$	-	-	100	$\mu\text{A}$
		static Operating mode; $\text{RESET} = V_{DD}$ ; $I_O = 0 \text{ mA}$ ; $V_{DD} = 2.0 \text{ V}$ ; $V_I = V_{IH(AC)}$ or $V_{IL(AC)}$	-	-	40	mA
$I_{DD}$	dynamic operating current per MHz	clock only; $\overline{\text{RESET}} = V_{DD}$ ; $V_I = V_{IH(AC)}$ or $V_{IL(AC)}$ ; CK and $\overline{\text{CK}}$ switching at 50 % duty cycle; $I_O = 0 \text{ mA}$ ; $V_{DD} = 1.8 \text{ V}$	-	16	-	$\mu\text{A}$
		per each data input, 1 : 1 mode; $\overline{\text{RESET}} = V_{DD}$ ; $V_I = V_{IH(AC)}$ or $V_{IL(AC)}$ ; CK and $\overline{\text{CK}}$ switching at 50 % duty cycle; one data input switching at half clock frequency, 50 % duty cycle; $I_O = 0 \text{ mA}$ ; $V_{DD} = 1.8 \text{ V}$	-	11	-	$\mu\text{A}$
		per each data input, 1 : 2 mode; $\overline{\text{RESET}} = V_{DD}$ ; $V_I = V_{IH(AC)}$ or $V_{IL(AC)}$ ; CK and $\overline{\text{CK}}$ switching at 50 % duty cycle; one data input switching at half clock frequency, 50 % duty cycle; $I_O = 0 \text{ mA}$ ; $V_{DD} = 1.8 \text{ V}$	-	19	-	$\mu\text{A}$
$C_i$	input capacitance	data and $\overline{\text{CSR}}$ inputs; $V_I = V_{ref} \pm 250 \text{ mV}$ ; $V_{DD} = 1.8 \text{ V}$	2.5	-	3.5	pF
		CK and $\overline{\text{CK}}$ inputs; $V_{ICR} = 0.9 \text{ V}$ ; $V_{i(p-p)} = 600 \text{ mV}$ ; $V_{DD} = 1.8 \text{ V}$	2	-	3	pF
		$\overline{\text{RESET}}$ input; $V_I = V_{DD}$ or GND; $V_{DD} = 1.8 \text{ V}$	3	-	4	pF
$Z_o$	output impedance	normal drive; instantaneous	[1]	-	15	$\Omega$
		normal drive; steady-state	-	53	-	$\Omega$
		high drive; instantaneous	[1]	-	7	$\Omega$
		high drive; steady-state	-	53	-	$\Omega$

[1] Instantaneous is defined as within < 2 ns following the output data transition edge.

**Table 10. Timing requirements**

At recommended operating conditions (see [Table 8](#)), unless otherwise specified. See [Section 11.1](#).

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{\text{clock}}$	clock frequency		-	-	450	MHz
$t_{\text{W}}$	pulse width	CK, $\overline{\text{CK}}$ HIGH or LOW	1	-	-	ns
$t_{\text{ACT}}$	differential inputs active time		[1][2]	-	10	ns
$t_{\text{INACT}}$	differential inputs inactive time		[1][3]	-	15	ns
$t_{\text{su}}$	setup time	$\overline{\text{DCS}}$ before CK $\uparrow$ , $\overline{\text{CK}}\downarrow$ , $\overline{\text{CSR}}$ HIGH; $\overline{\text{CSR}}$ before CK $\uparrow$ , $\overline{\text{CK}}\downarrow$ , $\overline{\text{DCS}}$ HIGH	0.6	-	-	ns
		$\overline{\text{DCS}}$ before CK $\uparrow$ , $\overline{\text{CK}}\downarrow$ , $\overline{\text{CSR}}$ LOW	0.5	-	-	ns
		DODT, DCKE and data (Dn) before CK $\uparrow$ , CK $\downarrow$	0.5	-	-	ns
		PAR_IN before CK $\uparrow$ , $\overline{\text{CK}}\downarrow$	0.5	-	-	ns
$t_{\text{h}}$	hold time	DCS, DODT, DCKE and data (Dn) after CK $\uparrow$ , $\overline{\text{CK}}\downarrow$	0.4	-	-	ns
		PAR_IN after CK $\uparrow$ , $\overline{\text{CK}}\downarrow$	0.4	-	-	ns

[1] This parameter is not necessarily production tested.

[2] VREF must be held at a valid input voltage level and data inputs must be held LOW for a minimum time of  $t_{\text{ACT(max)}}$  after  $\overline{\text{RESET}}$  is taken HIGH.

[3] VREF, data and clock inputs must be held at valid levels (not floating) a minimum time of  $t_{\text{INACT(max)}}$  after  $\overline{\text{RESET}}$  is taken LOW.

**Table 11. Switching characteristics (667 mode, SELAB = HIGH)**

At recommended operating conditions (see [Table 8](#)), unless otherwise specified. See [Section 11.1](#).

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{\text{max}}$	maximum input clock frequency		450	-	-	MHz
$t_{\text{PDM}}$	peak propagation delay	single bit switching; from CK $\uparrow$ and $\overline{\text{CK}}\downarrow$ to Qn	[1]	1.2	1.8	ns
$t_{\text{PD}}$	propagation delay	from CK $\uparrow$ and $\overline{\text{CK}}\downarrow$ to PPO	0.5	-	1.8	ns
$t_{\text{LH}}$	LOW-to-HIGH delay	from CK $\uparrow$ and $\overline{\text{CK}}\downarrow$ to $\overline{\text{QERR}}$	1.2	-	3	ns
$t_{\text{HL}}$	HIGH-to-LOW delay	from CK $\uparrow$ and $\overline{\text{CK}}\downarrow$ to $\overline{\text{QERR}}$	1	-	2.4	ns
$t_{\text{PDMSS}}$	simultaneous switching peak propagation delay	from CK $\uparrow$ and $\overline{\text{CK}}\downarrow$ to Qn	[1][2]	-	2.0	ns
$t_{\text{PHL}}$	HIGH-to-LOW propagation delay	from $\overline{\text{RESET}}\downarrow$ to Qn $\downarrow$	-	-	3	ns
		from $\overline{\text{RESET}}\downarrow$ to PPO $\downarrow$	-	-	3	ns
$t_{\text{PLH}}$	LOW-to-HIGH propagation delay	from $\overline{\text{RESET}}\downarrow$ to $\overline{\text{QERR}}\uparrow$	-	-	3	ns

[1] Includes 350 ps of test load transmission line delay.

[2] This parameter is not necessarily production tested.

**Table 12. Switching characteristics (800 mode, SELAB = LOW)**At recommended operating conditions (see [Table 8](#)), unless otherwise specified. See [Section 11.1](#).

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{\max}$	maximum input clock frequency		450	-	-	MHz
$t_{\text{PDM}}$	peak propagation delay	single bit switching; from CK $\uparrow$ and CK $\downarrow$ to Qn	[1] 1.1	-	1.5	ns
$t_{\text{PD}}$	propagation delay	from CK $\uparrow$ and CK $\downarrow$ to PPO	0.5	-	1.7	ns
$t_{\text{LH}}$	LOW-to-HIGH delay	from CK $\uparrow$ and CK $\downarrow$ to QERR	1.2	-	3	ns
$t_{\text{HL}}$	HIGH-to-LOW delay	from CK $\uparrow$ and CK $\downarrow$ to QERR	1	-	2.4	ns
$t_{\text{PDMSS}}$	simultaneous switching peak propagation delay	from CK $\uparrow$ and CK $\downarrow$ to Qn	[1][2] -	-	1.6	ns
$t_{\text{PHL}}$	HIGH-to-LOW propagation delay	from RESET $\downarrow$ to Qn $\downarrow$	-	-	3	ns
		from RESET $\downarrow$ to PPO $\downarrow$	-	-	3	ns
$t_{\text{PLH}}$	LOW-to-HIGH propagation delay	from RESET $\downarrow$ to QERR $\uparrow$	-	-	3	ns

[1] Includes 350 ps of test load transmission line delay.

[2] This parameter is not necessarily production tested.

**Table 13. Data output edge rates**At recommended operating conditions (see [Table 8](#)), unless otherwise specified. See [Section 11.2](#).

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$dV/dt_r$	rising edge slew rate	from 20 % to 80 %	1	-	4	V/ns
$dV/dt_f$	falling edge slew rate	from 80 % to 20 %	1	-	4	V/ns
$dV/dt_{\Delta}$	absolute difference between $dV/dt_r$ and $dV/dt_f$	from 20 % or 80 % to 80 % or 20 %	-	-	1	V/ns



10.1 Timing diagrams

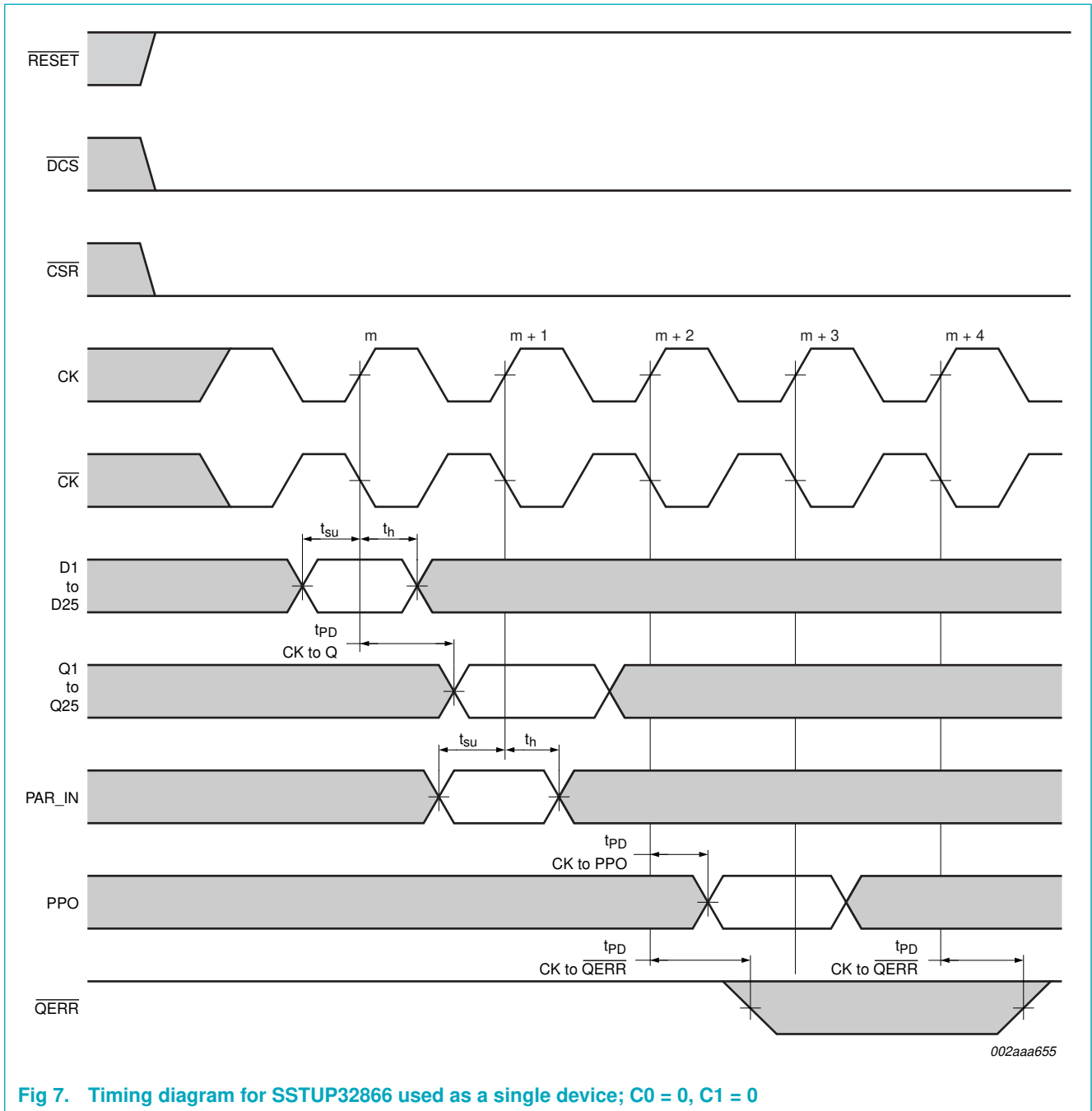
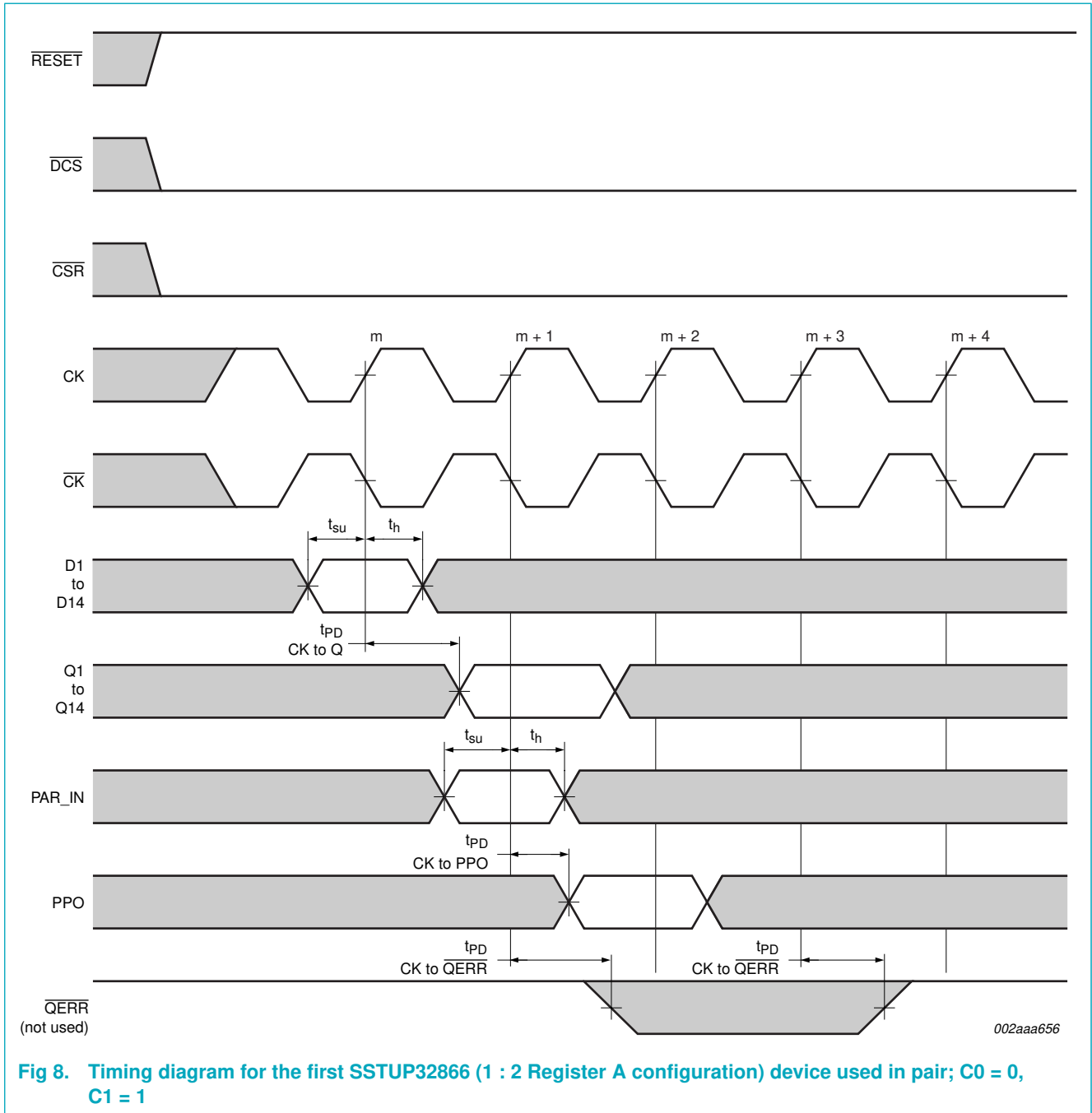
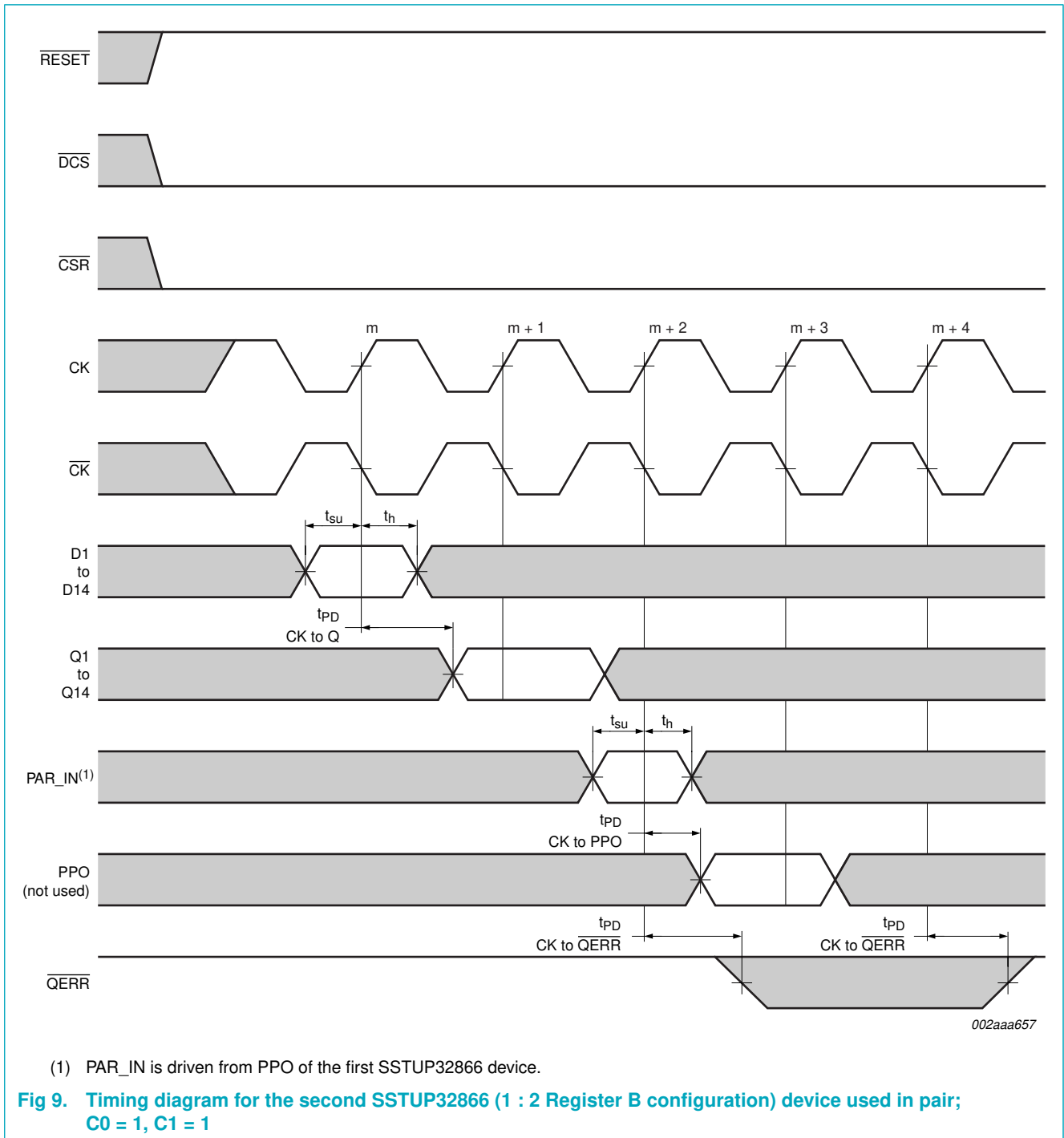


Fig 7. Timing diagram for SSTUP32866 used as a single device; C0 = 0, C1 = 0





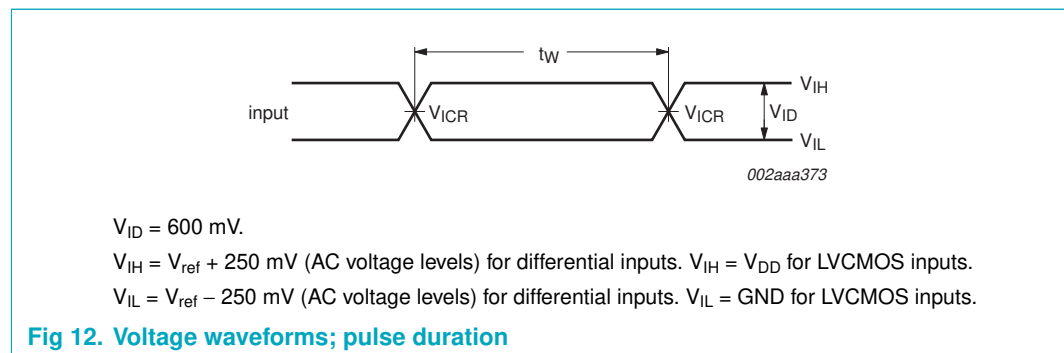
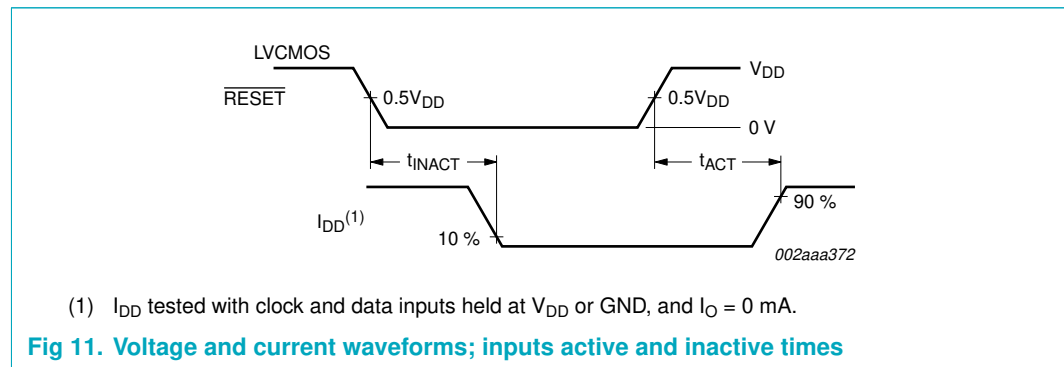
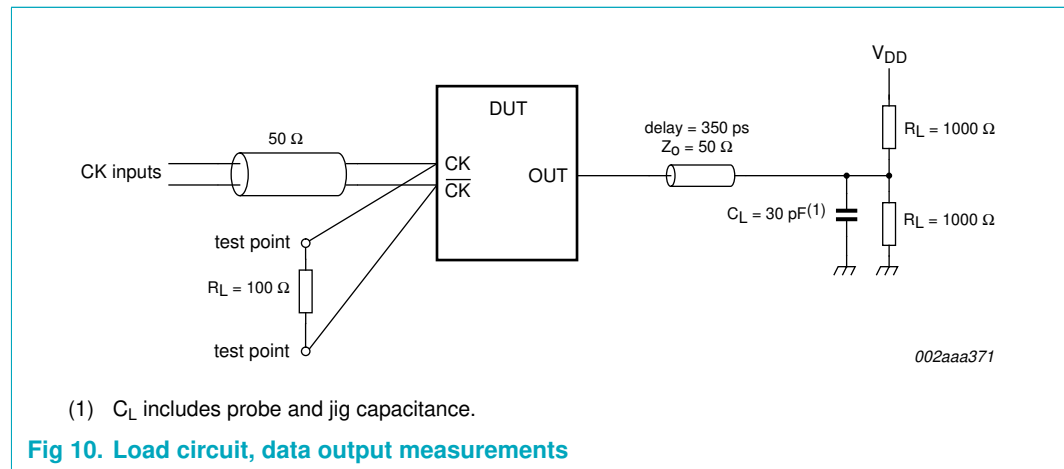
## 11. Test information

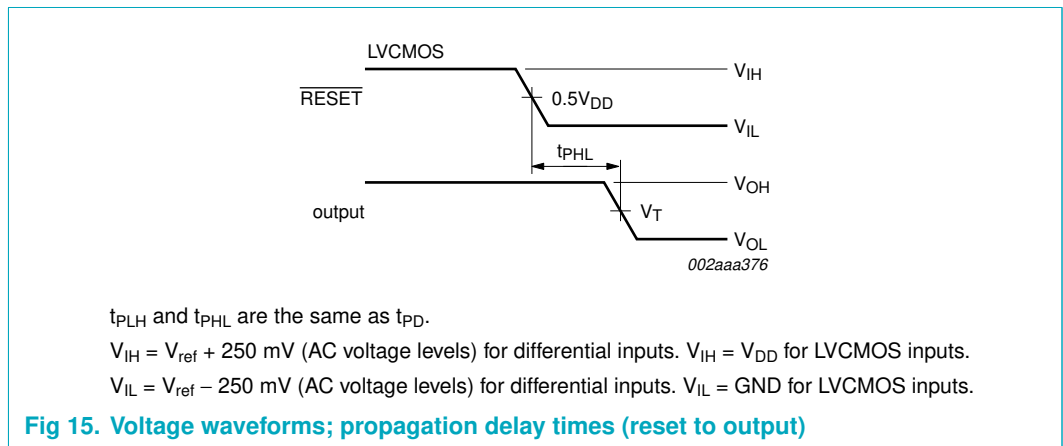
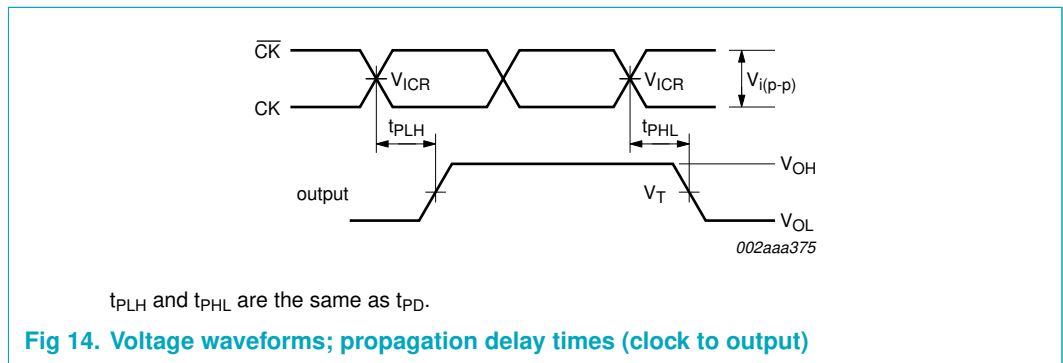
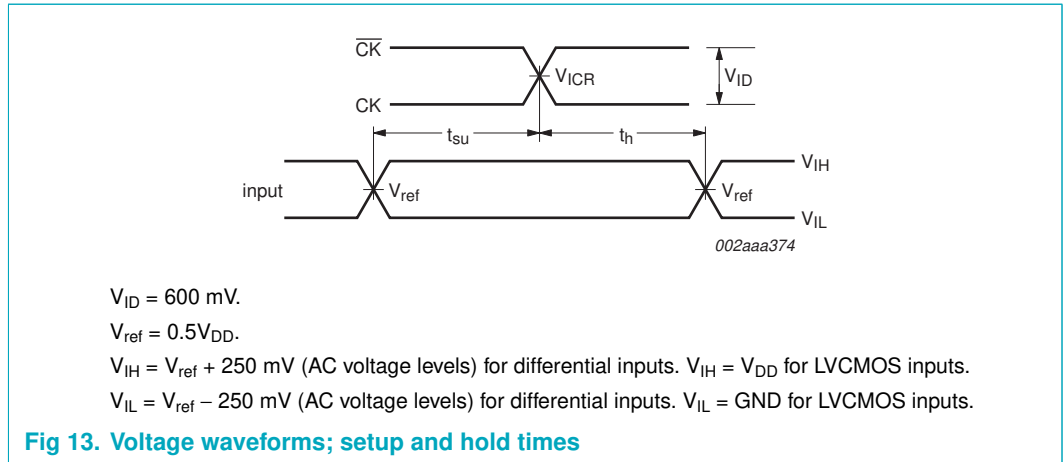
### 11.1 Parameter measurement information for data output load circuit

$V_{DD} = 1.8 \text{ V} \pm 0.1 \text{ V}$ .

All input pulses are supplied by generators having the following characteristics:  
 PRR  $\leq 10 \text{ MHz}$ ;  $Z_0 = 50 \Omega$ ; input slew rate =  $1 \text{ V/ns} \pm 20 \%$ , unless otherwise specified.

The outputs are measured one at a time with one transition per measurement.

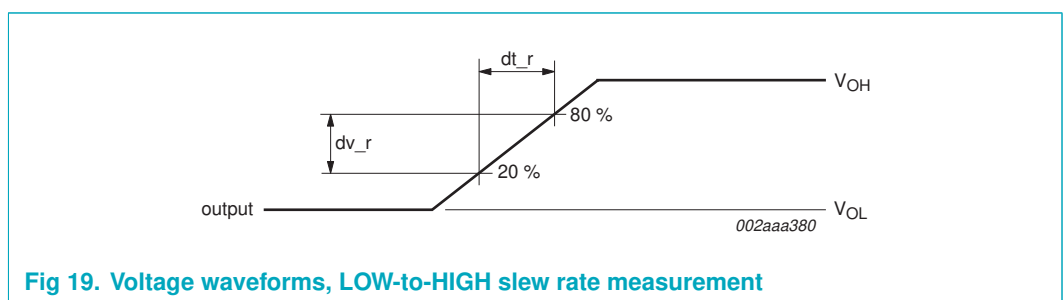
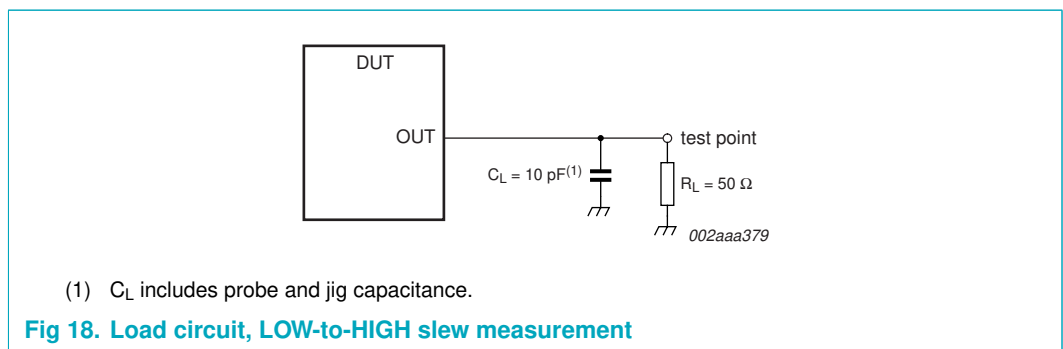
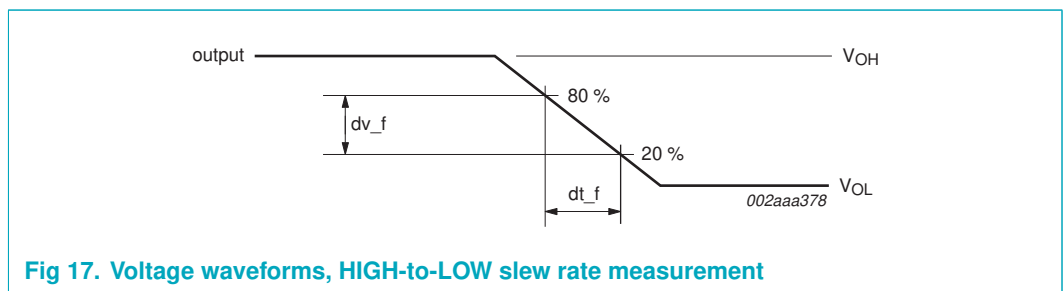
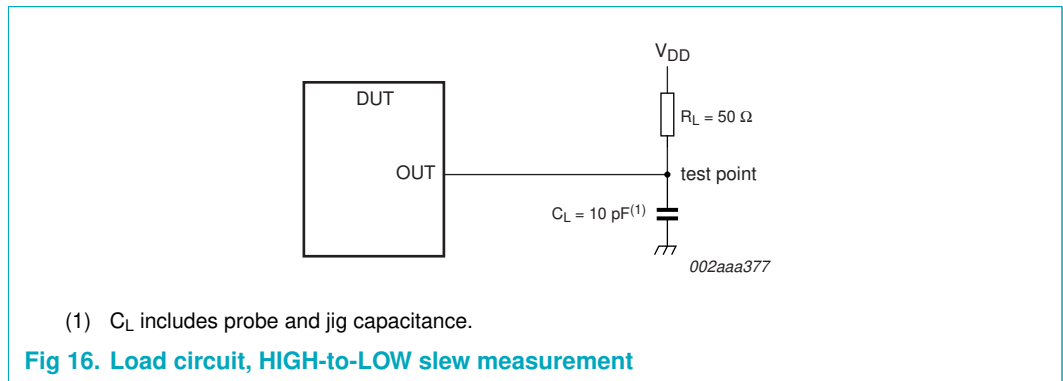




11.2 Data output slew rate measurement information

$V_{DD} = 1.8 V \pm 0.1 V$ .

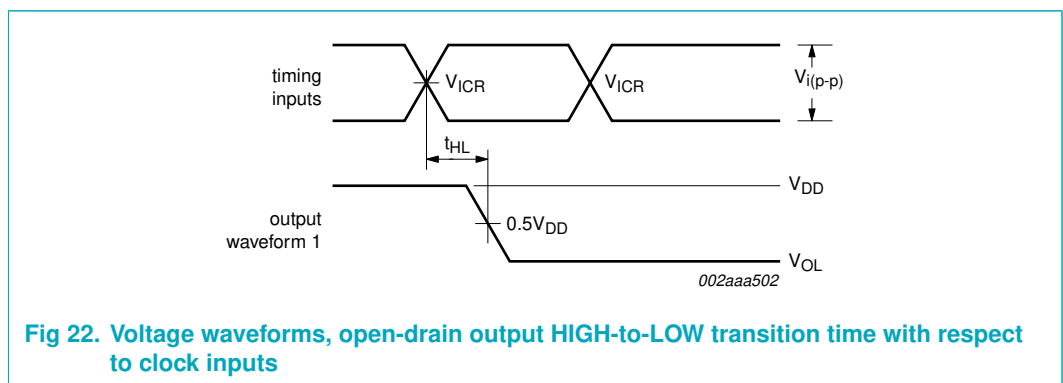
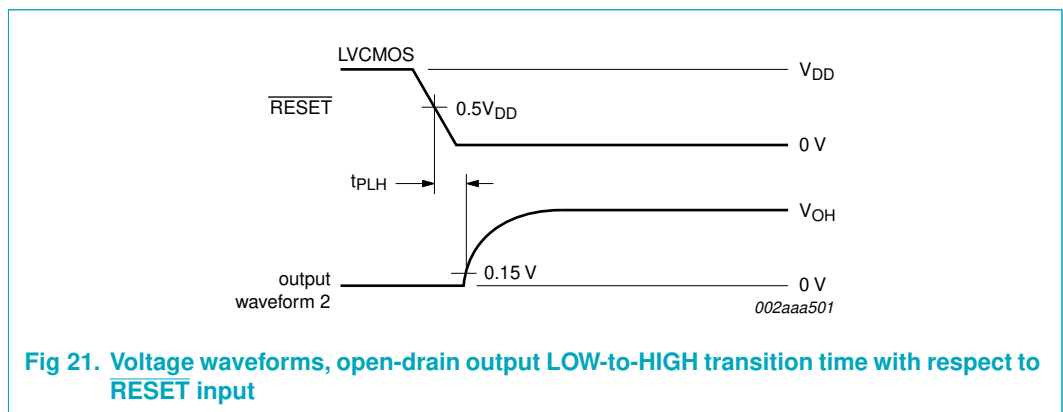
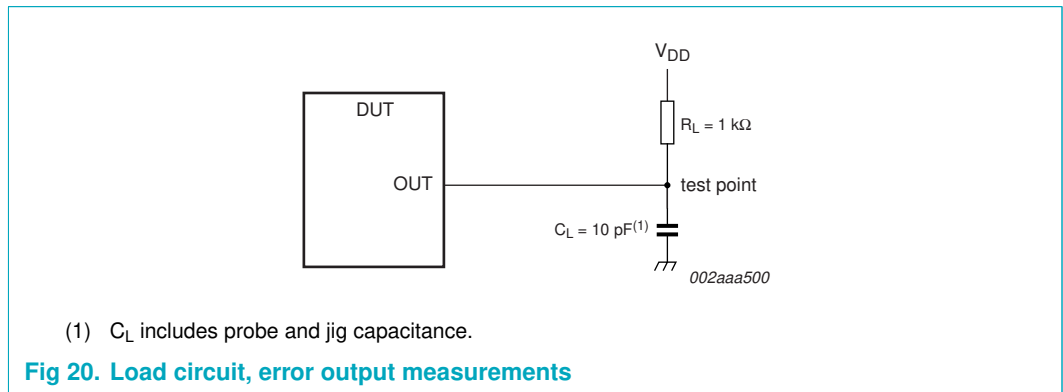
All input pulses are supplied by generators having the following characteristics:  $PRR \leq 10 \text{ MHz}$ ;  $Z_0 = 50 \Omega$ ; input slew rate =  $1 \text{ V/ns} \pm 20 \%$ , unless otherwise specified.

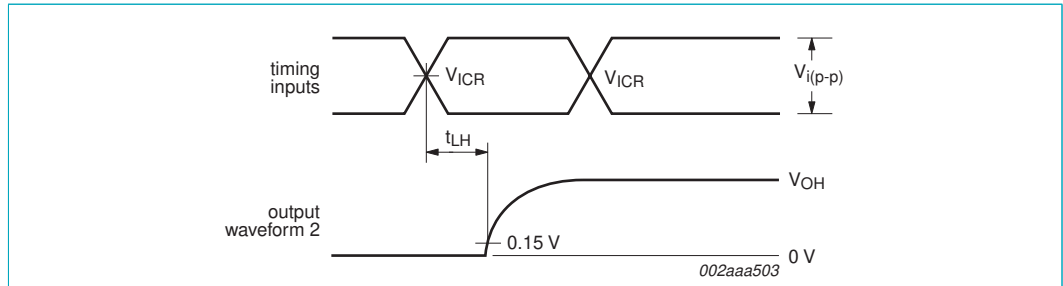


### 11.3 Error output load circuit and voltage measurement information

$V_{DD} = 1.8\text{ V} \pm 0.1\text{ V}$ .

All input pulses are supplied by generators having the following characteristics:  $PRR \leq 10\text{ MHz}$ ;  $Z_0 = 50\ \Omega$ ; input slew rate =  $1\text{ V/ns} \pm 20\%$ , unless otherwise specified.



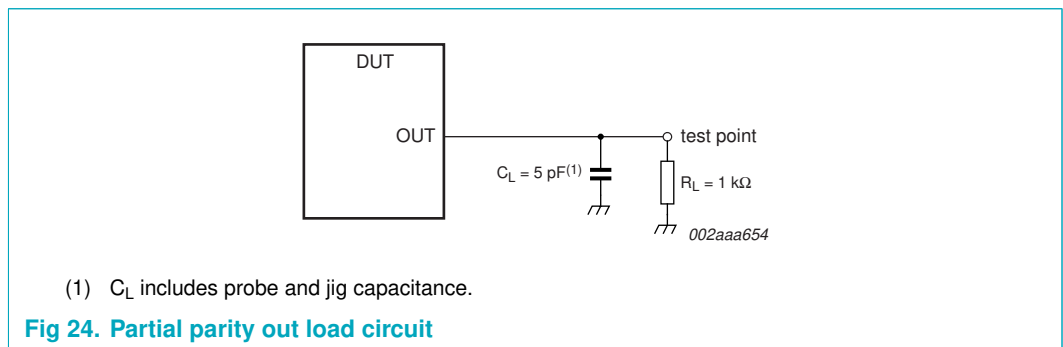


**Fig 23. Voltage waveforms, open-drain output LOW-to-HIGH transition time with respect to clock inputs**

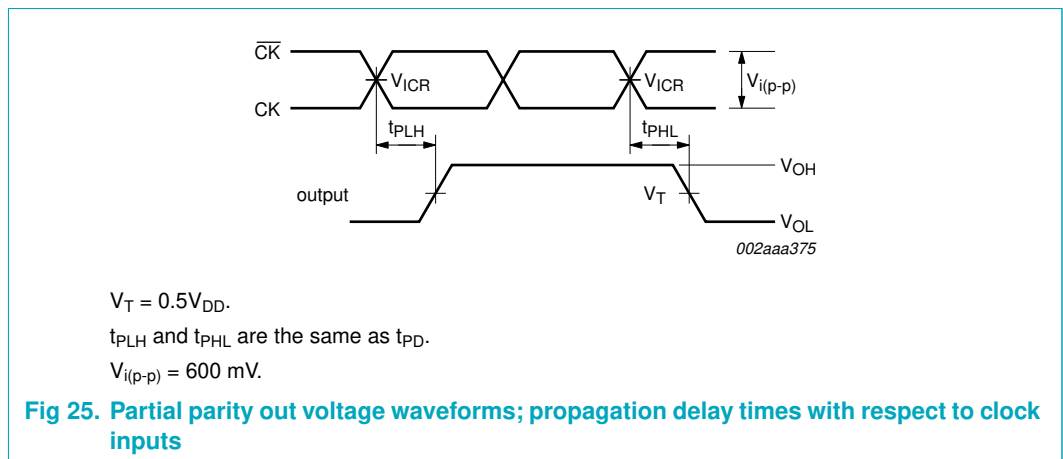
### 11.4 Partial parity out load circuit and voltage measurement information

$V_{DD} = 1.8 \text{ V} \pm 0.1 \text{ V}$ .

All input pulses are supplied by generators having the following characteristics:  
 $PRR \leq 10 \text{ MHz}$ ;  $Z_0 = 50 \Omega$ ; input slew rate =  $1 \text{ V/ns} \pm 20 \%$ , unless otherwise specified.

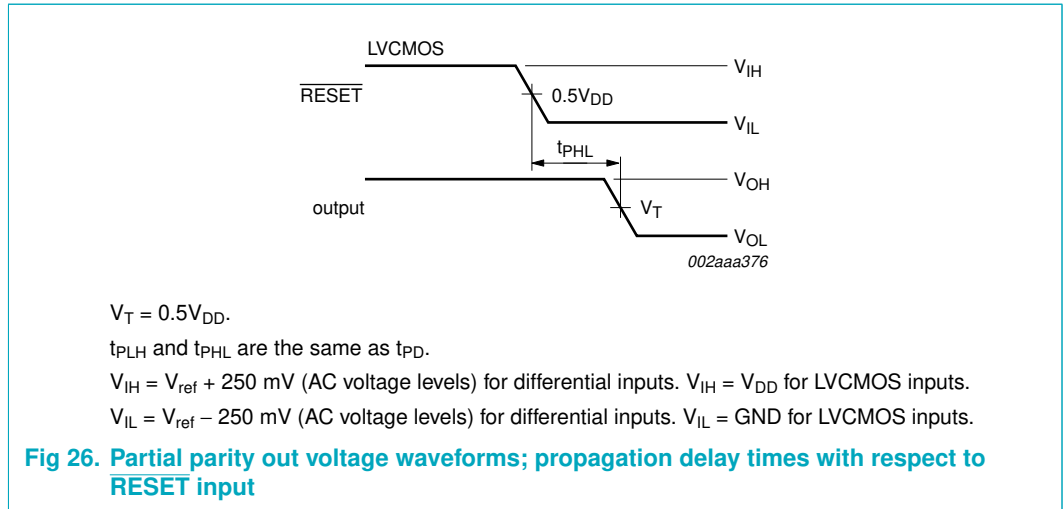


**Fig 24. Partial parity out load circuit**



**Fig 25. Partial parity out voltage waveforms; propagation delay times with respect to clock inputs**





12. Package outline

LFBGA96: plastic low profile fine-pitch ball grid array package; 96 balls; body 13.5 x 5.5 x 1.05 mm SOT536-1

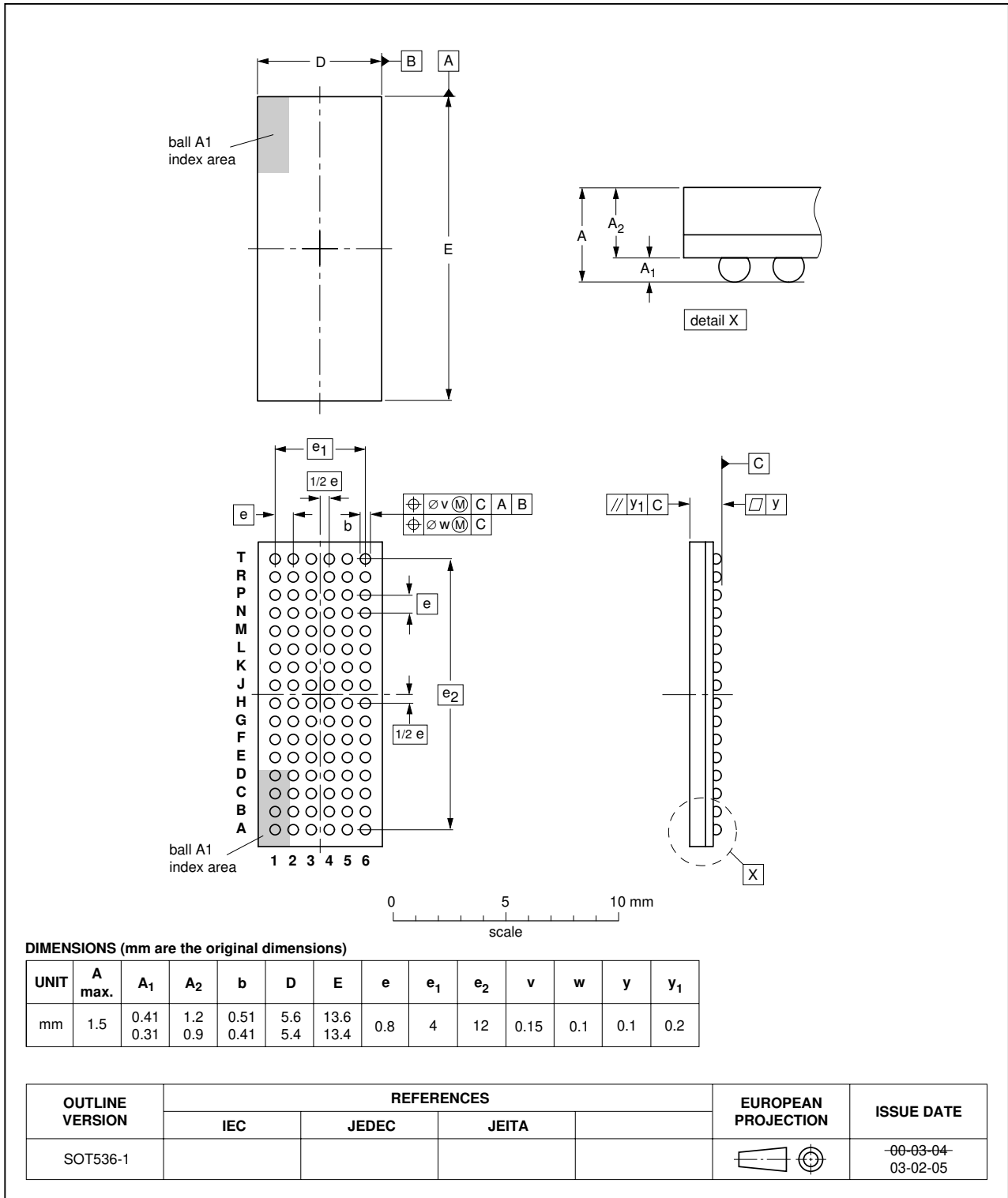


Fig 27. Package outline SOT536-1 (LFBGA96)