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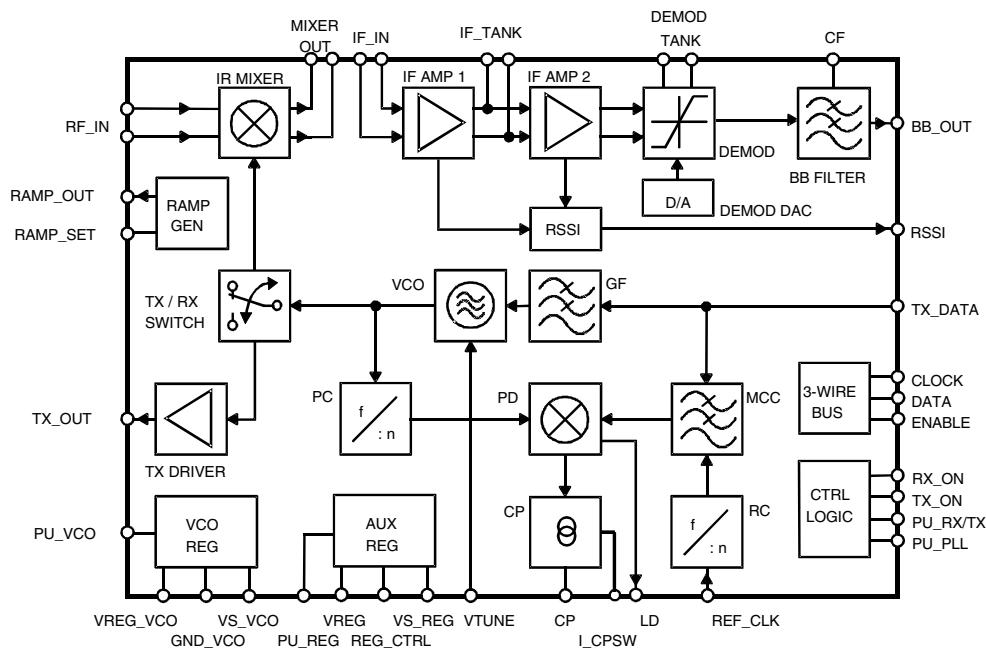
Features

- Supply Voltage Range 3V to 4.6V (Unregulated)
- Auxiliary Voltage Regulator On-chip
- Low Current Consumption
- Few Low Cost External Components
- No Mechanical Tuning Required
- Non-blindslot and Blindsight Operation
- Unlimited Multislot Operation with Advanced Closed-loop Modulation
- Supports Multiple Reference Clocks (10.368 MHz/13.824 MHz/20.736 MHz)
- TX Preamplifier with 0 dBm Output Power at 1.9 GHz and Ramp-signal Generator for SiGe Power Amplifier

1. Description

The T2801 is an RF IC for low-power DECT applications. The QFN48 packaged IC is a complete transceiver including image rejection mixer, IF amplifier, FM demodulator, baseband filter, RSSI, TX preamplifier, power-ramping generator for power amplifiers, integrated synthesizer, fully integrated VCO, TX filter and modulation compensation circuit for advanced closed-loop modulation concept. No mechanical tuning is necessary in production.

Figure 1-1. Block Diagram



DECT Single-chip Transceiver

T2801

Table 1-1. Functional Block Description

Name	Description
AUX REG	Auxiliary voltage regulator
BBF	Baseband filter
CP	Charge pump
DAC	D/A converter for demodulator tuning
DEMOD	Demodulator
GF	Gaussian filter for transmit data
IF AMP1	1st intermediate frequency amplifier
IF AMP2	2nd intermediate frequency amplifier
IR MIXER	Image rejection mixer
MCC	Modulation compensation circuit
PC	Programmable counter
PD	Phase detector
RAMP GEN	Ramp-signal generator
RC	Reference counter
RSSI	Received signal-strength indicator
TX DRIVER	Buffer amplifier for TX_OUT
TX/RX SWITCH	Switches VCO signal to IR mixer resp. TX driver
VCO	Voltage-controlled oscillator
VCO REG	Voltage regulator for VCO

2. Pin Configuration

Figure 2-1. Pinning QFN48

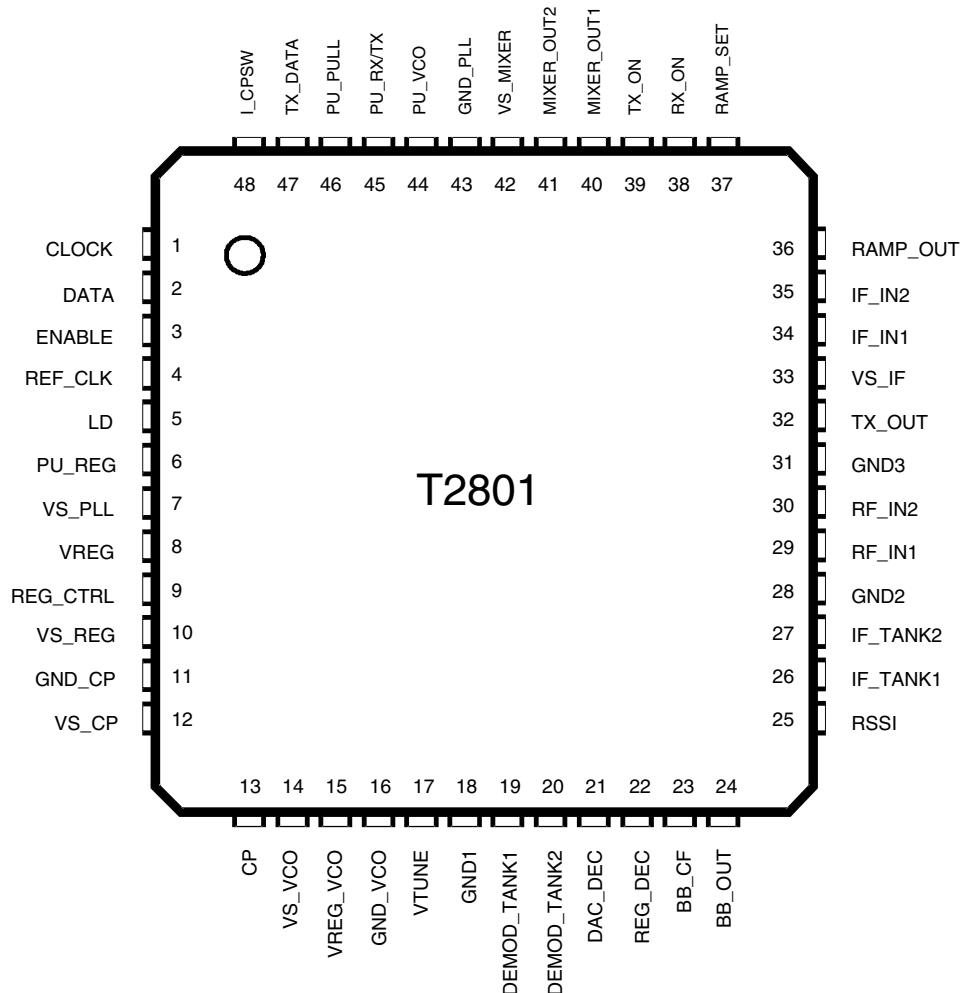


Table 2-1. Pin Description

Pin	Symbol	Function	Configuration
1 2 3	CLOCK DATA ENABLE	3-wire-bus: Clock input 3-wire-bus: Data input 3-wire-bus: Enable input	
4	REF_CLK	Reference-frequency input	
5	LD	Lock-detect output	
6	PU_REG	Power-up input for auxiliary voltage regulator	

Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
7	VS_PLL	PLL supply voltage	
8 9 10	VREG REG_CTRL VS_REG	Auxiliary voltage-regulator output Auxiliary voltage-regulator control output Auxiliary voltage-regulator supply voltage	
11 12 13	GND_CP VS_CP CP	Charge-pump ground Charge-pump supply voltage Charge-pump output	

Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
14 15 16	VS_VCO VREG_VCO GND_VCO	VCO voltage-regulator supply voltage VCO voltage-regulator control output VCO ground	
17	VTUNE	VCO tuning voltage input	
18	GND1	Ground	

Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
19 20	DEMOD_TANK1 DEMOD_TANK2	Demodulator tank circuit Demodulator tank circuit	<p>The diagram shows a differential input stage with two transistors. The outputs of this stage feed into two parallel tank circuits, each consisting of two series resistors (10k ohms) and two shunt capacitors. The outputs of these tanks are connected to a common-emitter stage with two transistors. The collector of the bottom transistor is connected to GND1 (pin 18). The collector of the top transistor is connected to VS_IF (pin 33) through a diode. The base of the top transistor is connected to VS_MIXER (pin 42) through a diode. The base of the bottom transistor is connected to DEMOD_TANK1 (pin 19) through a diode. The output of the common-emitter stage is connected to DEMOD_TANK2 (pin 20) through a diode.</p>
21	DAC_DEC	Decoupling pin for VCO_DAC	<p>The diagram shows a circuit for decoupling the VCO_DAC. It consists of a resistor (10k ohm) connected between VREG_VCO (pin 15) and VS_PLL (pin 7). The output of this stage is connected to DAC_DEC (pin 21) through a diode. A ground connection (GND_PLL, pin 43) is also shown. A capacitor (400 pF) is connected between DAC_DEC (pin 21) and GND_VCO (pin 16).</p>
22	REG_DEC	Decoupling pin for VCO_REG	<p>The diagram shows a circuit for decoupling the VCO_REG. It consists of a resistor (2k ohm) connected between VREG_VCO (pin 15) and VS_IF (pin 33). The output of this stage is connected to REG_DEC (pin 22) through a diode. A ground connection (GND2, pin 28) is also shown. A capacitor is connected between REG_DEC (pin 22) and GND_VCO (pin 16), and a resistor (42k ohm) is connected between REG_DEC (pin 22) and GND_VCO (pin 16).</p>

Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
23	BB_CF	Baseband filter corner-frequency control input	<p>VS_IF 33</p> <p>BB_CF 23</p> <p>GND2 28</p> <p>GND1 18</p>
24	BB_OUT	Baseband filter output	<p>VS_IF 33</p> <p>BB_OUT 24</p> <p>GND1 18</p> <p>GND2 28</p>
25	RSSI	Received signal-strength indicator output	<p>VS_IF 33</p> <p>RSSI 25</p> <p>13k</p> <p>GND2 28</p>
26 27	IF_TANK1 IF_TANK2	IF tank circuit IF tank circuit	<p>VS_IF 33</p> <p>RSSI 25</p> <p>13k</p> <p>GND2 28</p>

Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
28	GND2	Ground	<p>VS_PLL VS_REG VS_CP VS_VCO VS_IF VS_MIXER</p> <p>GND1 GND2 GND3 GND_VCO GND_CP GND_PLL</p>
29 30	RF_IN1 RF_IN2	RF input of image reject mixer RF input of image reject mixer	<p>VS_MIXER RF_IN1 RF_IN2 GND2</p>
31	GND3	Ground	<p>VS_PLL VS_REG VS_CP VS_VCO VS_IF VS_MIXER</p> <p>GND1 GND2 GND3 GND_VCO GND_CP GND_PLL</p>

Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
32	TX_OUT	TX driver amplifier output for PA	
33	VS_IF	IF amplifier supply voltage	
34 35	IF_IN1 IF_IN2	IF input of IF amplifier IF input of IF amplifier	
36	RAMP_OUT	Ramp-generator output for PA power ramping	

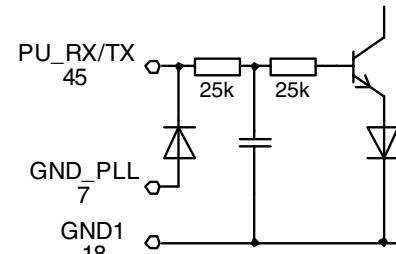
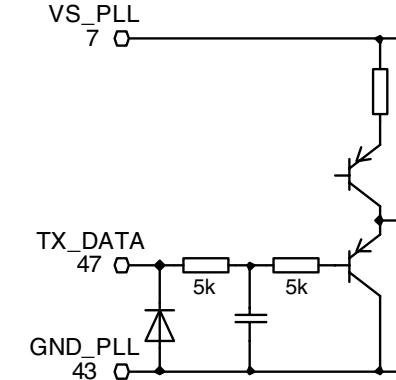
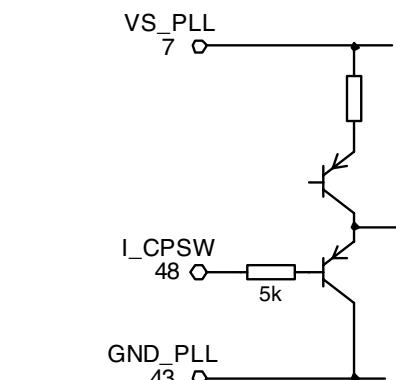
Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
37	RAMP_SET	Slew-rate setting of ramping signal	
38 39	RX_ON TX_ON	RX control input TX control input	
40 41	MIXER_OUT1 MIXER_OUT2	Mixer output to SAW filter Mixer output to SAW filter	

Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
42 43	VS_MIXER GND_PLL	Mixer supply voltage PLL ground	<p>The diagram shows the connection of pins 42 and 43. Pin 42 is connected to VS_MIXER and GND_Mixer. Pin 43 is connected to VS_REG, VS_CP, VS_VCO, VS_IF, and GND_VCO. These signals are connected through diodes to a central bus. From this bus, signals are distributed to various components: VS_REG connects to pin 18 (GND1), VS_CP connects to pin 28 (GND2), VS_VCO connects to pin 31 (GND3), VS_IF connects to pin 16 (GND_VCO), and VS_MIXER connects to pin 11 (GND_CP). A separate ground connection, GND_PLL, is shown at pin 43.</p>
44	PU_VCO	VCO power-up input	<p>The diagram shows the connection of pin 44. Pin 44 is connected to PU_VCO. This signal is connected to a base-emitter junction of a transistor. The collector of this transistor is connected to a resistor (5k) and then to another resistor (5k). The other end of the second resistor is connected to GND_VCO (pin 16) and GND_PLL (pin 7). A diode is also connected between the collector and GND_VCO.</p>
45	PU_RX/TX	RX/TX power-up input	<p>The diagram shows the connection of pin 45. Pin 45 is connected to PU_RX/TX. This signal is connected to a base-emitter junction of a transistor. The collector of this transistor is connected to two resistors (25k each) in series. The other end of these resistors is connected to GND1 (pin 18) and GND_PLL (pin 7). A diode is also connected between the collector and GND1.</p>

Table 2-1. Pin Description (Continued)

Pin	Symbol	Function	Configuration
46	PU_PLL	PLL power-up input	
47	TX_DATA	TX data input of Gaussian filter and modulation-compensation circuit	
48	I_CPSW	Charge pump switch input controls charge pump current	

3. Functional Description

3.1 Receiver

The RF signal at RF_IN is fed to an image rejection mixer IR_MIXER with its differential outputs MIXER_OUT1 and MIXER_OUT2 driving an IF-SAW filter at 110.592 MHz or 112.32 MHz. The IF amplifiers IF_AMP1 and IF_AMP2 with an external IF_TANK and an integrated RSSI function feed the signal to the demodulator DEMOD working at $f = f_{IF}/2$ ([55 MHz]) and finally to an integrated baseband filter BB. For demodulator tuning in production, an integrated 5-bit Digital-to-Analog (D/A) converter is provided to control the on-chip varicap diode.

3.2 Transmitter

The transmit data at TX_DATA is filtered by an integrated Gaussian Filter (GF) and fed to the fully integrated VCO operating at twice the output frequency. After modulation, the signal is frequency-divided by 2 and fed via a TX/RX SWITCH to the TX_DRIVER. This bus-controlled driver amplifier supplies typical +3 dBm output power at TX_OUT. An integrated ramp-signal generator, RAMP_GEN, provides a ramp signal at RAMP_OUT for the external power amplifier. The slope of the ramp signal is controlled by a capacitor at the RAMP_SET pin.

3.3 Synthesizer

The IR_MIXER, the TX_DRIVER and the programmable counter PC are driven by the fully integrated VCO (including on-chip inductors and varactors). An 3-bit digital-to-analog converter is used to pretune the frequency. The output signal is frequency-divided to supply the desired frequency to the TX_DRIVER, 0/90 degree phase shifter for the IR_MIXER and to be used by the PC for the phase detector PD ($f_{PD} = 3.456$ MHz). Unlimited multislots operation is possible by using the integrated advanced closed-loop modulation concept based on the modulation compensation circuit MCC.

3.4 Power Supply

An integrated bandgap-stabilized voltage regulator for use with an external low-cost PNP transistor is implemented. Multiple power-down and current saving modes are provided.

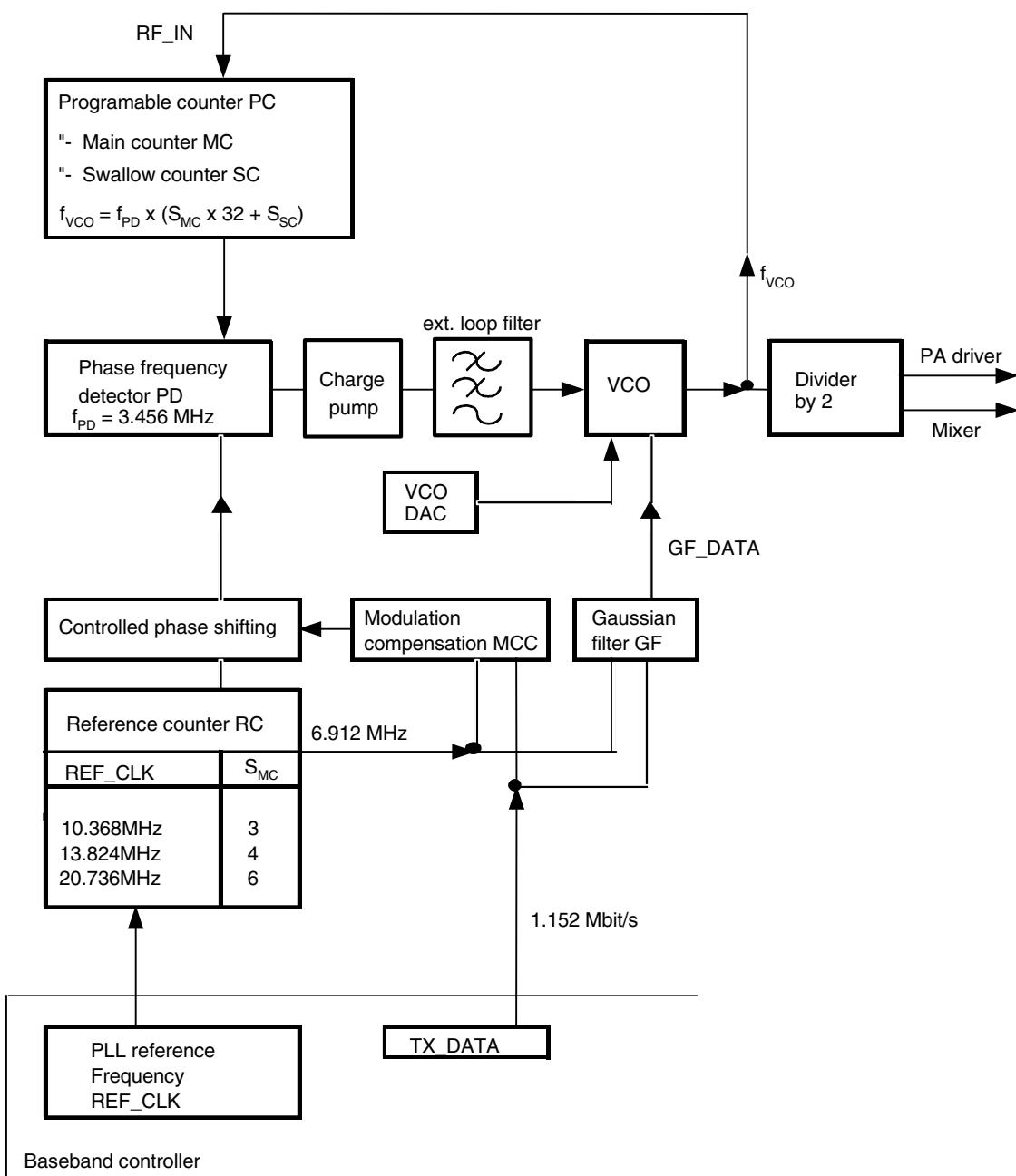
Figure 3-1. PLL Principle

Table 3-1 shows the LO frequencies for RX and TX for the DECT band plus additional channels for the extended DECT band. Intermediate frequencies of 110.592 MHz and 112.32 MHz are supported.

Table 3-1. LO Frequencies

Mode	f _{IF} /MHz	Channel	f _{ANT} /MHz	f _{VCO} /MHz	S _{MC}	S _{SC}
TX		C9	1881.792	1881.792	34	1
TX		C8	1883.520	1883.520	34	2
TX	
TX		C1	1895.616	1895.616	34	9
TX		C0	1897.344	1897.344	34	10
TX		C10	1899.072	1899.072	34	11
TX		C11	1900.800	1900.800	34	12
TX	
TX		C29	1931.904	1931.904	34	30
TX		C30	1933.632	1933.632	34	31
RX	110.592	C9	1881.792	1771.200	32	1
RX	110.592	C8	1883.520	1772.928	32	2
RX	110.592
RX	110.592	C1	1895.616	1785.024	32	9
RX	110.592	C0	1897.344	1786.752	32	10
RX	110.592	C10	1899.072	1788.480	32	11
RX	110.592	C11	1900.800	1790.208	32	12
RX	110.592
RX	110.592	C29	1931.904	1821.312	32	30
RX	110.592	C30	1933.632	1823.040	32	31
RX	112.320	C9	1881.792	1769.472	32	0
RX	112.320	C8	1883.520	1771.200	32	1
RX	112.320
RX	112.320	C1	1895.616	1783.296	32	8
RX	112.320	C0	1897.344	1785.024	32	9
RX	112.320	C10	1899.072	1786.752	32	10
RX	112.320	C11	1900.800	1788.480	32	11
RX	112.320
RX	112.320	C29	1931.904	1819.584	32	29
RX	112.320	C30	1933.632	1821.312	32	30

Formula:

$$\text{TX: } f_{\text{ANT}} = f_{\text{VCO}} = 1.728 \text{ MHz} \times (32 \times S_{\text{MC}} + S_{\text{SC}})$$

$$\text{RX: } f_{\text{ANT}} = 1.728 \text{ MHz} \times (32 \times S_{\text{MC}} + S_{\text{SC}}) + f_{\text{IF}}$$

4. Control Signals

Table 4-1. Control Signals – Functions

Signal	Function
I_CPSW	Controls the charge pump current
PU_REG	Activates AUX voltage regulator supplying the complete transceiver
PU_VCO	Activates VCO voltage regulator which supplies only the VCO
PU_RX/TX	Activates RX/TX blocks
PU_PLL	Activates PLL circuits: PC, PD, CP, RC
RX_ON	Activates RX circuits: BBF, DEMOD, IF AMP, IR MIXER
TX_ON	Activates TX circuits: TX-DRIVER, RAMP GEN. Starts RAMP SIGNAL at RAMP OUT
Data Word 1, Bit D10	Activates GF in TX mode
Data Word 1, Bit D9	Activates MCC in TX mode

Table 4-2. Control Signals – Modes

Mode	TX Mode	RX Mode	RSSI Only
PU_REG	1	1	1
PU_VCO	1	1	1
PU_RX/TX	1	1	1
PU_PLL	1	1	1
RX_ON	0	1	1
TX_ON	1	0	1
BB filter	OFF	ON	OFF
Demodulator	OFF	ON	OFF
IF amplifiers and RSSI	OFF	ON	ON
IR mixer	OFF	ON	ON
RX switch	OFF	ON	ON
TX switch	ON	OFF	OFF
TX driver	ON	OFF	OFF
Ramp generator	ON	OFF	OFF
Programmable counter	ON	ON	ON
Voltage-controlled oscillator	ON	ON	ON
Gaussian filter	ON	OFF	OFF
Phase detector/charge pump	ON	ON	ON
Modulation compensation circuit	ON	OFF	OFF
Reference counter	ON	ON	ON
Typical current consumption/mA at V _S = 3.2 V	54	85	80

5. Serial Programming Bus

The transceiver is programmed by the 3-wire bus (CLOCK, DATA and ENABLE).

After setting enable signal to low condition, on the rising edge of the clock signal, the data is transferred bit by bit into the shift register, starting with the MSB-bit. After enable returning to high condition, the programmed information is loaded into the addressed latches, according to the addressbit condition (last bit). Additional leading bits are ignored and there is no check made on how many pulses arrived during enable-low condition. During enable low condition, the bus current is increased to speed up the bus logic.

The programming of the transceiver is separated into two data words. Data word 1 controls mainly the channel information together with settings, which are closely related with the channel. Data word 2 holds setup information, which is adjusted during production.

5.1 Data Word 1

MSB																					LSB		
Data Bits																					Add. bit		
D22	D21	D20	D19	D18	D17	D16	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	A0
RC	SC				MC	VCOs	1	1	GF	MCC	GFCS			VCODAC			CPCS		GF	1			

5.2 Data Word 2

E10	E9	E8	E7	E6	E5	E4	E3	E2	E1	E0	A0
DEMODDAC				MCCS			TEST			0	

6. Data Word 1 Programs

6.1 PLL Settings

Table 6-1. With the Reference Counter Bits D21-D22

RC (Referene Counter)			
D22	D21	S _{RC}	REF_CLK (MHz)
0	0	3	10.638
0	1	4	13.824
1	0	6	20.736

Table 6-2. With the Main Counter Bits D14-D15

MC (Main Counter)		
D15	D14	S _{RC}
0	0	32
0	1	33
1	0	34
1	1	35

Table 6-3. With the Swallow Counter Bits D16-D20

SC (Swallow Counter)					
D20	D19	D18	D17	D16	S _{SC}
0	0	0	0	0	0
0	0	0	0	1	1
0	0	0	1	0	2
...					...
1	1	1	0	1	29
1	1	1	1	0	30
1	1	1	1	1	31

6.2 VCO Select (RX/TX VCO)

Table 6-4. With bit D13

D13	VCO Select
0	RX-VCO
1	TX-VCO

Note: Used to switch between RX/TX VCO

6.3 Gaussian Filter On/Off

Table 6-5. With bit D10

D10	GF (Gaussian Filter)
0	OFF
1	ON

Note: GF is used only in TX mode

6.4 Modulation Compensation Circuit On/Off

Table 6-6. With bit D9

D9	MCC (Modulation Compensation Circuit)
0	OFF
1	ON

Note: MCC is used only in TX mode

6.5 GFCS Adjustment

Table 6-7. With bit D6 - D8

GFCS(Gaussian Filter Settings)			
D8	D7	D6	GFCS (%)
0	0	0	60
0	0	1	70
0	1	0	80
0	1	1	90
1	0	0	100
1	0	1	110
1	1	0	120
1	1	1	130

Note: Only in TXmode effective for setting the frequency deviation of the modulation

6.6 VCO_DAC Adjustment

Table 6-8. With bit D3 - D5

Pretune DAYC Voltage			
D5	D4	D3	f _{VCO} /%
0	0	0	-5
0	0	1	...
0	1	0	...
0	1	1	...
1	0	0	...
1	0	1	...
1	1	0	...
1	1	1	5

Note: Used to pretune the VCO frequency in case of production tolerances of the device. Tuning voltage in locked condition should be around 1.8V at room temperature. This gives margin for ambient temperature changes

6.7 CPCS Adjustment

Table 6-9. With bit D0 - D2

CPCS (Charge-pump Current Settings)			
D2	D1	D0	CPCS
0	0	0	-4
0	0	1	-3
0	1	0	-2
0	1	1	-1
1	0	0	0
1	0	1	1
1	1	0	2
1	1	1	3

Note: Used to adjust the charge pump current. This can be used to compensate the change of the tuning sensitivity over frequency and device tolerances

7. Data Word 2 Programs

7.1 DEMODDAC Adjustment

Table 7-1. With bits E6 - E10

Demod DAC Voltage					
E10	E9	E8	E7	E6	f _{IFcenter} (%)
0	0	0	0	0	-5
0	0	0	0	1	...
0	0	0	1	0	...
					...
1	1	1	0	1	...
1	1	1	1	0	...
1	1	1	1	1	5

Note: Only in RX mode effective. Used to tune the demodulator center frequency and allows to compensate tolerances of external components and the T2801

7.2 MCCS Adjustment

Table 7-2. With bits E3 - E5

MCCS (Modulation Compensation Settings)			
E5	E4	E3	MCCS (%)
0	0	0	60
0	0	1	70
0	1	0	80
0	1	1	90
1	0	0	100
1	0	1	110
1	1	0	120
1	1	1	130

Note: Only in TX mode effective. Adjusts the modulation compensation circuit for closed loop modulation. This adjustment is done with a test sequence of ,1' - ,0'. The correct setting is achieved, if the modulation is not affected by the PLL

7.3 TEST Mode Settings

Table 7-3. With bit E0 - E2 and D11

D11	E2	E1	E0	Signal at Lock Detect Output	CP Mode
1	0	0	0	Lock detect	Active
0	0	0	1	RC out/2	Active
1	0	1	0	PC out/2	Active
X	0	1	1	MCCTEST: RC out divided by 512	Active
1	1	0	0	Lock detect	High imp.
0	1	0	1	RC out/2	High imp.
1	1	1	0	PC out/2	High imp.
X	1	1	1	GFTEST: RC out	High imp.

Note: In normal operation Lock detect output is used. All other settings are for test only

Figure 7-1. 3-wire Bus Protocol Timing Diagram

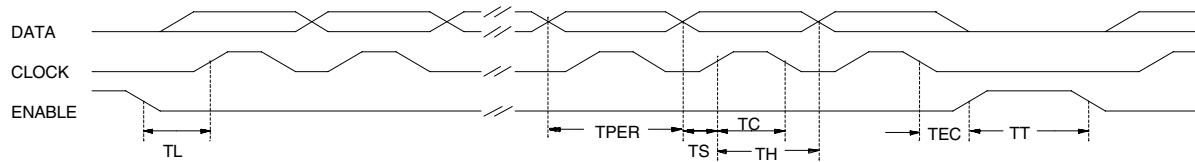


Table 7-4. 3-wire Bus Protocol

Description	Symbol	Minimum Value	Unit
Clock period	TPER	125	ns
Set time data to clock	TS	60	ns
Hold time data to clock	TH	60	ns
Clock pulse width	TC	60	ns
Set time enable to clock	TL	200	ns
Hold time enable to data	TEC	0	ns
Time between two protocols	TT	250	ns

Figure 7-2. TX DATA Timing

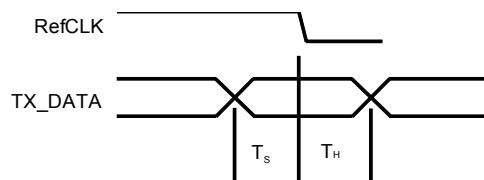


Table 7-5. TX DATA Timing Values

Parameters	Symbol	Value	Remarks
Set-up time TX DATA	TS	10 ns	TS and TH must be considered for both (falling and rising) edges of RefCLK when using REF_CLK = 10.368 MHz.
Hold time TX DATA	TH	10 ns	

8. Absolute Maximum Ratings

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

All voltages refer to GND

Parameters	Symbol	Min.	Max.	Unit
Supply voltage regulator, Pin 10	V_{S_REG}	3.2	4.7	V
Supply voltage, pins 7, 12, 14, 33 and 42	V_S	3.0	4.7	V
Logic input voltage, pins 1, 2, 3, 38, 39, 44, 45, 46, 47 and 48	V_{IN}	-0.3	V_S	V
Junction temperature	T_{jmax}		150	°C
Storage temperature	T_{Stg}	-40	+150	°C

9. Thermal Resistance

Parameters	Symbol	Value	Unit
Junction ambient	R_{thJA}	TBD	K/W

10. Operating Range

Parameters	Symbol	Min.	Typ.	Max.	Unit
Supply voltage regulator, Pins 10	V_{S_REG}	3.2	3.6	4.6	V
Supply voltage, pins 7, 12, 14, 33 and 42	V_S	3.0	3.0	4.6	V
Ambient temperature	T_{amb}	-25		+85	°C

11. Electrical Characteristics

Test conditions (unless otherwise specified): $V_{S_REG} = 3.2V$, $T_{amb} = 25^{\circ}C$

Parameters	Test Conditions/Pins	Symbol	Min.	Typ.	Max.	Unit
IR Mixer, Pins 29, 30, 40 and 41						
Input impedance	Pins 29 and 30	Z_{in}		50		Ω
Input matching	Pins 29 and 30	$VSWR_{in}$		< 2:1		
Image rejection ratio	Pins 40 and 41	IRR		20		dB
DSB noise figure	Pins 40 and 41	$NF_{DSB} = NF_{SSB}$		10		dB
Conversion gain	$R_{load} = 200\Omega$	G_{conv}		11		dB
Input interception point	Pins 40 and 41	IIP3		-10		dBm
IF Amplifier, Pins 26, 27, 34 and 35						
Input impedance	Pins 34 and 35	Z_{in}	200		400	Ω
Lower cut-off frequency		f_{l3dB}		90		MHz
Upper cut-off frequency		f_{u3dB}		130		MHz
Power gain		G_p		85		dB
Bandwidth of external tank circuit	Pins 26 and 27	BW3dB		10		MHz
Noise figure		NF		9		dB
RSSI, Pins 25, 34 and 35						
RSSI sensitivity	At IF_IN1, IF_IN2 Pins 34 and 35	P_{min}		20		$dB\mu V$
RSSI compression	At IF_IN1, IF_IN2 Pins 34 and 35	P_{max}		100		$dB\mu V$
RSSI dynamic range		DR		80		dB
RSSI resolution	Slope of the RSSI has to be steady	Acc		± 2		dB
RSSI rise time	$P_{in} = 30$ to $100 dB\mu V$, pin 25	t_r		1		μs
RSSI fall time	$P_{in} = 100$ to $30 dB\mu V$, pin 25	t_f		1		μs
Quiescent output voltage	At $P_{in} < 20 dB\mu V$ at IF_IN1, IF_IN2, pin 25	I_{out}		0.45		μA
Maximum output voltage	At $P_{in} = 100 dB\mu V$ at IF_IN1, IF_IN2, pin 25	I_{out}		2.25		μA
FM Demodulator, BB-Filter Pins 19, 20, 23 and 24						
Co-channel rejection ratio	At $P_{in} = -75 dBm$ at IR-mixer input	CCRR		10		dB
Sensitivity	Quality factor of external tank circuit approximately 20, $f_{res} = F_{IF}/2$, Pin 24	S		0.5		V/MHz
Amplitude of recovered signal	Nominal deviation of signal $\pm 288 kHz$, Pin 24	A		450		mVss
Corner frequency	Pin 23: C = 68 pF	f_c		680		kHz
Output voltage DC range	Pin 24	V_{outDC}	1		$V_s - 1$	V
DAC for FM Demodulator (Internally Connected)						
DEMOD_DAC range	(see bus protocol E6 ... E10)	$\Delta f_{IFcenter}$		± 5		%

