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Wireless Components



Edition 2007-02-26

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TDA5255 E1 ASK/FSK 434MHz Wireless Transceiver

Wireless Components



Data Sheet

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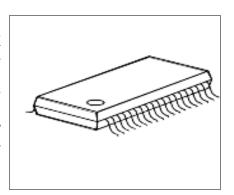
ASK/FSK 434MHz Wireless Transceiver TDA5255 E1

Version 1.2

Product Info

General Description

The IC is a low power consumption single chip FSK/ASK Transceiver for half duplex low datarate communication in the 433-435MHz band. The IC offers a very high level of integration and needs only a few external components. It contains a highly efficient power amplifier, a low noise amplifier (LNA) with AGC, a double balanced mixer, a complex direct conversion stage, I/Q limiters with RSSI generation, an FSK demodulator, a fully integrated VCO and PLL synthesizer, a tuneable crystal oscillator, an onboard data filter, a data comparator (slicer), positive and negative peak detectors, a data rate detection circuit and a 2/3-wire bus interface. Additionally there is a power down feature to save battery power.



Features

- Low supply current (I_s = 9mA typ. receive, I_s
 = 13mA typ. transmit mode)
- Supply voltage range 2.1 5.5V
- Power down mode with very low supply current consumption
- FSK and ASK modulation and demodulation capability
- Fully integrated VCO and PLL synthesizer and loop filter on-chip with on chip crystal oscillator tuning
- I²C/3-wire µController Interface

- On-chip low pass channel select filter and data filter with tuneable bandwidth
- Data slicer with self-adjusting threshold and 2 peak detectors
- FSK sensitivity <-109dBm, ASK sensitivity < -109dBm
- Transmit power up to +13dBm
- Datarates up to 100kBit/s Manchester Encoded
- Self-polling logic with ultra fast data rate detection

Application

- Low Bitrate Communication
 Systems
- Keyless Entry Systems
- Remote Control Systems
- Alarm Systems

- Telemetry Systems
- Electronic Metering
- Home Automation Systems

Туре	Ordering Code	Package
TDA5255 E1	SP000014555	PG-TSSOP-38



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Product Description

1 Product Description

1.1 Overview

The IC is a low power consumption single chip FSK/ASK Transceiver for the frequency band 433-435 MHz. The IC combines a very high level of integration and minimum external part count. The device contains a low noise amplifier (LNA), a double balanced mixer, a fully integrated VCO, a PLL synthesizer, a crystal oscillator with FSK modulator, a limiter with RSSI generator, an FSK demodulator, a data filter, a data comparator (slicer), a positive and a negative data peak detector, a highly efficient power amplifier and a complex digital timing and control unit with I²C/3-wire microcontroller interface. Additionally there is a power down feature to save battery power.

The transmit section uses direct ASK modulation by switching the power amplifier, and crystal oscillator detuning for FSK modulation. The necessary detuning load capacitors are external. The capacitors for fine tuning are integrated. The receive section is using a novel single-conversion/direct-conversion scheme that is combining the advantages of both receive topologies. The IF is contained on the chip, no RF channel filters are necessary as the channel filter is also on the chip.

The self-polling logic can be used to let the device operate autonomously as a master for a decoding microcontroller.

1.2 Features

- Low supply current (I_s = 9 mA typ. receive, I_s = 13mA typ. transmit mode, both at 3 V supply voltage, 25°C)
- Supply voltage range 2.1 V to 5.5 V
- Operating temperature range -40°C to +85°C
- Power down mode with very low supply current consumption
- FSK and ASK modulation and demodulation capability without external circuitry changes, FM demodulation capability
- Fully integrated VCO and PLL synthesizer and loop filter on-chip with on-chip crystal oscillator tuning, therefore no additional external components necessary
- Differential receive signal path completely on-chip, therefore no external filters are necessary
- On-chip low pass channel select and data filter with tuneable bandwith
- Data slicer with self-adjusting threshold and 2 peak detectors
- Self-polling logic with adjustable duty cycle and ultrafast data rate detection and timer mode providing periodical interrupt
- FSK and ASK sensitivity < -109 dBm
- Adjustable LNA gain
- Digital RSSI and Battery Voltage Readout
- Provides Clock Out Pin for external microcontroller
- Transmit power up to +13 dBm in 50Ω load at 5V supply voltage
- Maximum datarate up to 100 kBaud Manchester Encoded
- I²C/3-wire microcontroller interface, working at max. 400kbit/s
- meets the ETSI EN300 220 regulation and CEPT ERC 7003 recommendation

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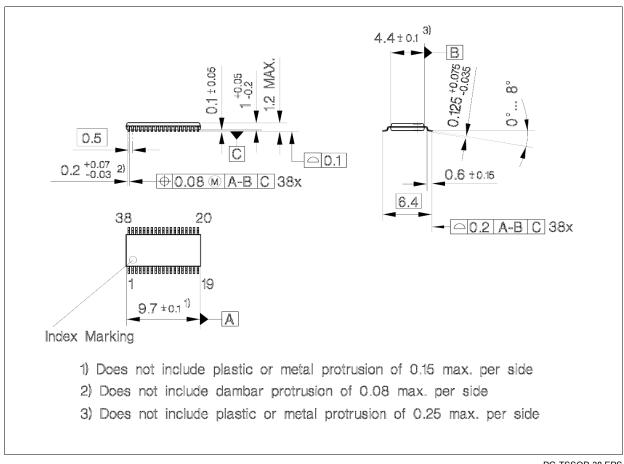


Product Description

1.3 Application

- Low Bitrate Communication Systems
- Keyless Entry Systems
- Remote Control Systems
- Alarm Systems
- Telemetry Systems
- Electronic Metering
- Home Automation Systems

1.4 Package Outlines



PG-TSSOP-38.EPS

Figure 1-1 PG-TSSOP-38 package outlines

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2 Functional Description

2.1 Pin Configuration

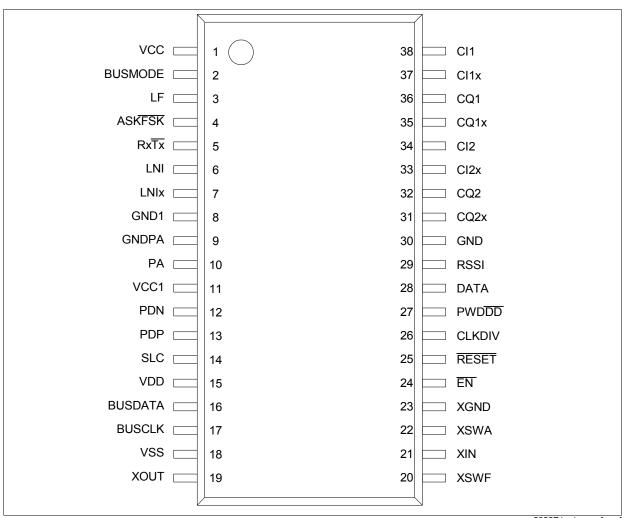


Figure 2-1 Pin Configuration

5255E1_pin_conf.wmf



2.2 Pin Definitions and Functions

Table 2	Table 2-1 Pin Definition and Function				
	Symbol	Equivalent I/O-Schematic	Function		
1	VCC	111	Analog supply (antiparallel diodes between VCC, VCC1, VDD)		
2	BUSMODE	2 350	Bus mode selection (I ² C/3 wire bus mode selection)		
3	LF	3 - 200	Loop filter and VCO control voltage		
4	ASKFSK	350	ASK/FSK- mode switch input		



5	RXTX	5 - 350 TX	RX/TX-mode switch input/output
6	LNI	5k 1.1V 5k 7	RF input to differential Low Noise Amplifier (LNA))
7	LNIX	see Pin 6	Complementary RF input to differential LNA
8	GND1	8 ————————————————————————————————————	Ground return for LNA and Power Amplifier (PA) dirver stage
9	GNDPA	see Pin 8	Ground return for PA output stage
10	PA	10 Ω 10 Ω 9 GndPA	PA output stage
11	VCC1	see Pin 1	Supply for LNA and PA

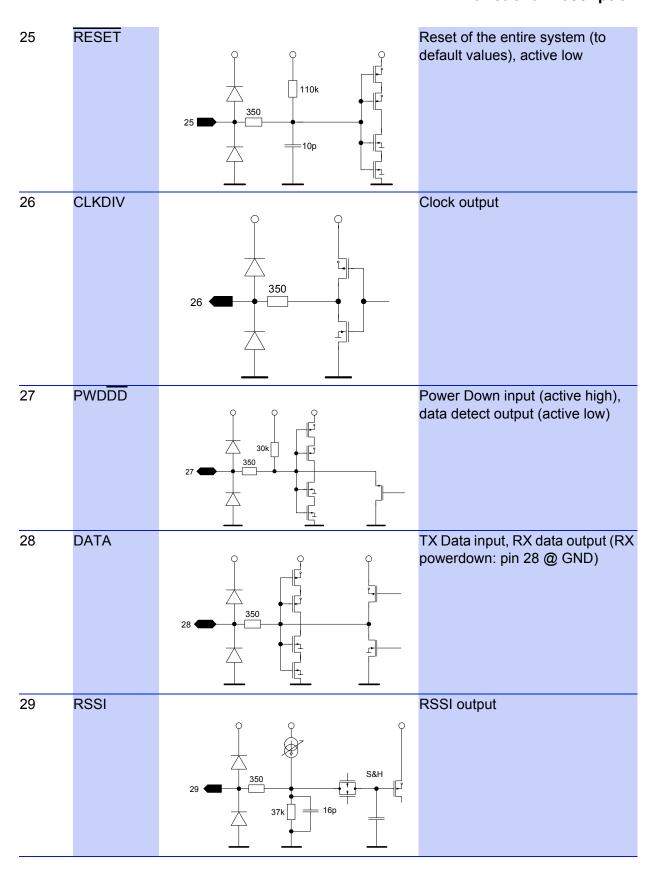


12	PDN	12 PWDN 50k 50k 50k	Output of the negative peak detector
13	PDP	350 PWDN	Output of the positive peakdetector
14	SLC	1.2uA 50k 50k 50k 50k 50k 50k 50k 50k 50k	Slicer level for the data slicer
15	VDD	see Pin 1	Digital supply; A 10Ω serial resistor in the VDD supply line is strongly recommended; see also Section 4.4
16	BUSDATA	15k	Bus data in/output
17	BUSCLK	350	Bus clock input



18	VSS	see Pin 8	Ground for digital section
19	XOUT	Vcc-860mV 150μΑ	Crystal oscillator output, can also be used as external reference frequency input.
20	XSWF	21 125fF 4pF 20 250fF 8pF 23	FSK modulation switch
21	XIN	see Pin 20	
22	XSWA	22	ASK modulation/FSK center frequency switch
23	XGND	see Pin 22	Crystal oscillator ground return
24	ĒN	350	3-wire bus enable input







30	GND	see Pin 8	Analog ground
31	CQ2x	Stage1:Vcc-630mV Stage2: Vcc-560mV	Pin for external Capacitor Q-channel, stage 2
32	CQ2	II	Q-channel, stage 2
33	CI2x	II	I-channel, stage 2
34	CI2	II	I-channel, stage 2
35	CQ1x	II	Q-channel, stage 1
36	CQ1	II	Q-channel, stage 1
37	CI1x	II	I-channel, stage 1
38	CI1	II	I-channel, stage 1



2.3 **Functional Block Diagram**

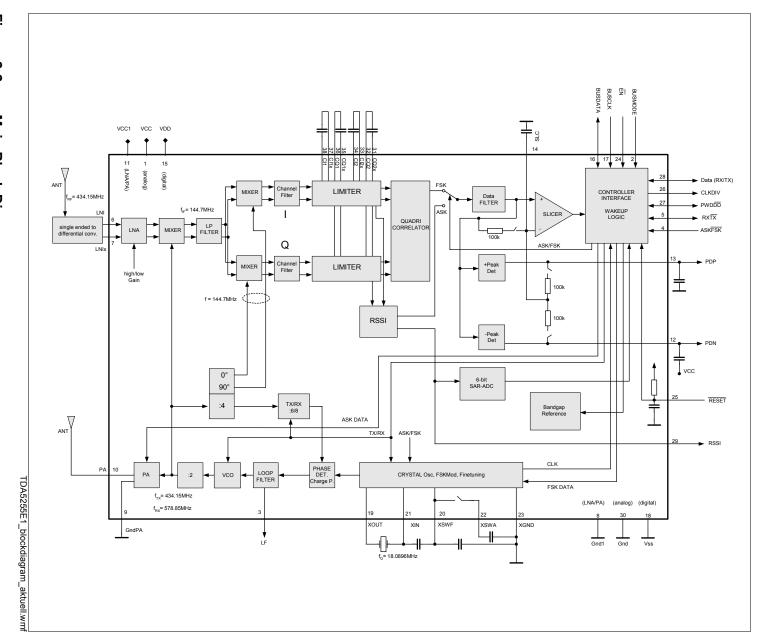


Figure 2-2 Main Block Diagram



2.4 Functional Block Description

2.4.1 Power Amplifier (PA)

The power amplifier is operating in C-mode. It can be used in either high or low power mode. In high-power mode the transmit power is approximately +13dBm into 50 Ohm at 5V and +6dBm at 2.1V supply voltage. In low power mode the transmit power is approximately +10dBm at 5V and -32dBm at 2.1V supply voltage using the same matching network. The transmit power is controlled by the **D0**-bit of the **CONFIG** register (subaddress 00H) as shown in the following **Table 2-2**. The default output power mode is high power mode.

Table 2	Table 2-2 Sub Address 00H: CONFIG				
Bit	Function	Description	Default		
D0	PA_PWR	0= low TX Power, 1= high TX Power	1		

In case of ASK modulation the power amplifier is turned fully on and off by the transmit baseband data, i.e. 100% On-Off-Keying.

2.4.2 Low Noise Amplifier (LNA)

The LNA is an on-chip cascode amplifier with a voltage gain of 15 to 20dB and symmetrical inputs. It is possible to reduce the gain to 0 dB via logic.

Table 2-3 Sub Address 00H: CONFIG				
Bit	Function	Description	Default	
D4	LNA_GAIN	0= low Gain, 1= high Gain	1	

2.4.3 Downconverter 1st Mixer

The Double Balanced 1st Mixer converts the input frequency (RF) in the range of 434-435 MHz down to the intermediate frequency (IF) at approximately 144MHz. The local oscillator frequency is generated by the PLL synthesizer that is fully implemented on-chip as described in **Section 2.4.5**. This local oscillator operates at approximately 578MHz in receive mode providing the above mentioned IF frequency of 144MHz. The mixer is followed by a low pass filter with a corner frequency of approximately 175MHz in order to prevent RF and LO signals from appearing in the 144MHz IF signal.

2.4.4 Downconverter 2nd I/Q Mixers

The Low pass filter is followed by 2 mixers (inphase I and quadrature Q) that convert the 144MHz IF signal down to zero-IF. These two mixers are driven by a signal that is generated by dividing the local oscillator signal by 4, thus equalling the IF frequency.

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2.4.5 PLL Synthesizer

The Phase Locked Loop synthesizer consists of two VCOs (i.e. transmit and receive VCO), a divider by 4, an asynchronous divider chain with selectable overall division ratio, a phase detector with charge pump and a loop filter and is fully implemented on-chip. The VCOs are including spiral inductors and varactor diodes. The center frequency of the transmit VCO is 868MHz, the center frequency of the receive VCO is 1156MHz.

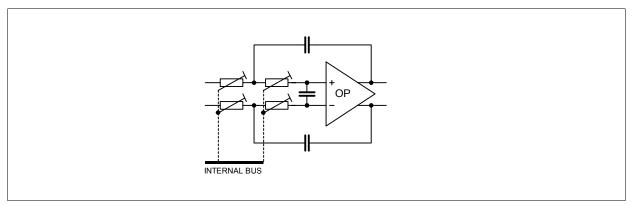
Generally in receive mode the relationship between local oscillator frequency f_{OSC} , the receive RF frequency f_{RF} and the IF frequency f_{IF} and thus the frequency that is applied to the I/Q Mixers is given in the following formula:

$$\frac{f_{osc}}{2} = 4/3 f_{RF} = 4 f_{IF}$$
 [2 – 1]

The VCO signal is applied to a divider by 2 and afterwards by 4 which is producing approximately 144MHz signals in quadrature. The overall division ratio of the divider chain following the divider by 2 and 4 is 6 in transmit mode and 8 in receive mode <u>as</u> the nominal crystal oscillator frequency is 18.083MHz. The division ratio is controlled by the **RxTx** pin (pin 5) and the **D10** bit in the **CONFIG** register.

2.4.6 I/Q Filters

The I/Q IF to zero-IF mixers are followed by baseband 6th order low pass filters that are used for RF-channel filtering.



iq_filter.wmf

Figure 2-3 One I/Q Filter stage

The bandwidth of the filters is controlled by the values set in the filter-register. It can be adjusted between 50 and 350kHz in 50kHz steps via the bits D1 to D3 of the **LPF** register (subaddress 03H).

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2.4.7 I/Q Limiters

The I/Q Limiters are DC coupled multistage amplifiers with offset-compensating feedback circuit and an overall gain of approximately 80dB each in the frequency range of 100Hz up to 350kHz. Receive Signal Strength Indicator (RSSI) generators are included in both limiters which produce DC voltages that are directly proportional to the input signal level in the respective channels. The resulting I- and Q-channel RSSI-signals are summed to the nominal RSSI signal.

2.4.8 FSK Demodulator

The output differential signals of the I/Q limiters are fed to a quadrature correlator circuit that is used to demodulate frequency shift keyed (FSK) signals. The demodulator gain is 2.4mV/kHz, the maximum frequency deviation is ±300kHz as shown in **Figure 2-4** below.

The demodulated signal is applied to the ASK/FSK mode switch which is connected to the input of the data filter. The switch can be controlled by the **ASKFSK** pin (pin 4) and via the D11 bit in the CONFIG register.

The modulation index *m* must be significantly larger than 2 and the deviation at least larger than 25kHz for correct demodulation of the signal.

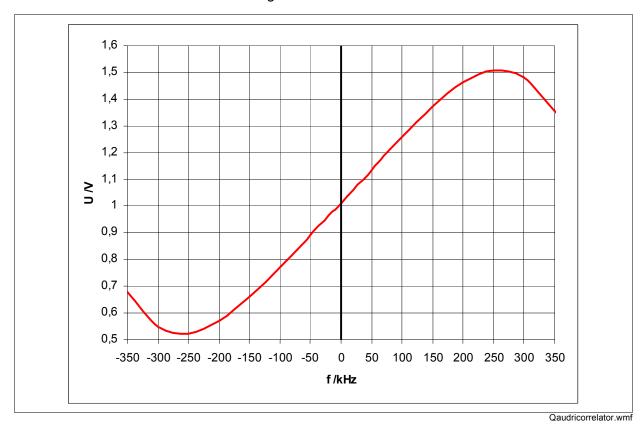


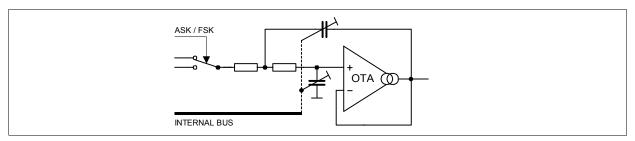
Figure 2-4 Quadricorrelator Demodulation Characteristic

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2.4.9 Data Filter

The 2-pole data filter has a Sallen-Key architecture and is implemented fully on-chip. The bandwidth can be adjusted between approximately 5kHz and 102kHz via the bits **D4** to **D7** of the **LPF** register as shown in **Table 3-10**.



data_filter.wmf

Figure 2-5 Data Filter architecture

2.4.10 Data Slicer

The data slicer is a fast comparator with a bandwidth of 100kHz. The self-adjusting threshold is generated by a RC-network (LPF) or by use of one or both peak detectors depending on the baseband coding scheme as described in **Section 3.6**. This can be controlled by the **D15** bit of the **CONFIG** register as shown in the following table.

Table 2	Table 2-4 Sub Address 00H: CONFIG				
Bit	Function	Description	Default		
D15	SLICER	0= Lowpass Filter, 1= Peak Detector	0		

2.4.11 Peak Detectors

Two separate Peak Detectors are available. They are generating DC voltages in a fast-attack and slow-release manner that are proportional to the positive and negative peak voltages appearing in the data signal. These voltages may be used to generate a threshold voltage for non-Manchester encoded signals, for example. The time-constant of the fast-attack/slow-release action is determined by the RC network with external capacitor.

2.4.12 Crystal Oscillator

The reference oscillator is an NIC oscillator type (Negative Impedance Converter) with a crystal operating in serial resonance. The nominal operating frequency of 18.083MHz and the frequencies for FSK modulation can be adjusted via 3 external capacitors. Via microcontroller and bus interface the chip-internal capacitors can be used for finetuning of the nominal and the FSK modulation frequencies. This finetuning of the crystal oscillator allows to eliminate frequency errors due to crystal or component tolerances.

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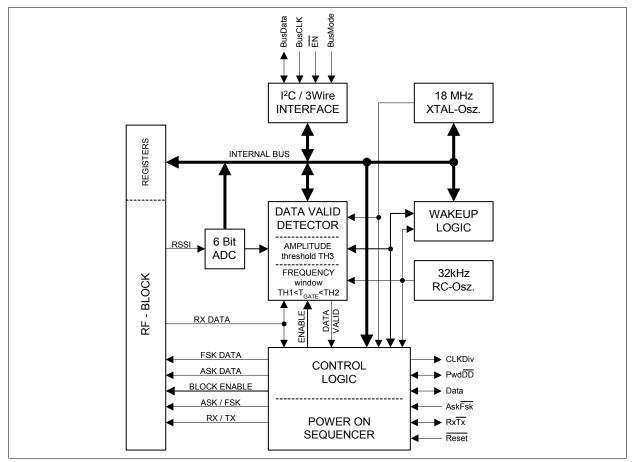
2.4.13 Bandgap Reference Circuitry and Powerdown

A Bandgap Reference Circuit provides a temperature stable 1.2V reference voltage for the device. A power down mode is available to switch off all subcircuits that are controlled by the bidirectional Powerdown&DataDetect **PwdDD** pin (pin 27) as shown in the following table. Power down mode can either be activated by pin 27 or bit D14 in Register 00h. In power down mode also pin 28 (DATA) is affected (see **Section 2.4.17**).

Table 2-5 PwdDD Pin Operating States	
PwdDD	Operating State
VDD	Powerdown Mode
Ground/VSS	Device On

2.4.14 Timing and Data Control Unit

The timing and data control unit contains a wake-up logic unit, an I²C/3-wire microcontroller interface, a "data valid" detection unit and a set of configuration registers as shown in the subsequent figure.



logic.wmf

Figure 2-6 Timing and Data Control Unit

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The I²C / 3-wire Bus Interface gives an external microcontroller full control over important system parameters at any time.

It is possible to set the device in three different modes: Slave Mode, Self Polling Mode and Timer Mode. This is done by a state machine which is implemented in the WAKEUP LOGIC unit. A detailed description is given in **Section 2.4.16**.

The DATA VALID DETECTOR contains a frequency window counter and an RSSI threshold comparator. The window counter uses the incoming data signal from the data slicer as the gating signal and the crystal oscillator frequency as the timebase to determine the actual datarate. The result is compared with the expected datarate.

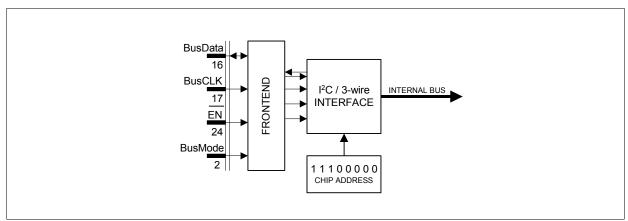
The threshold comparator compares the actual RSSI level with the expected RSSI level.

If both conditions are true the **PwdDD** pin is set to LOW in self polling mode as you can see in **Section 2.4.16**. This signal can be used as an interrupt for an external μP . Because the **PwdDD** pin is bidirectional and open drain driven by an internal pull-up resistor it is possible to apply an external LOW thus enabling the device.

2.4.15 Bus Interface and Register Definition

The TDA5255 supports the I^2C bus protocol (2 wire) and a 3-wire bus protocol. Operation is selectable by the **BusMode** pin (pin 2) as shown in the following table. All bus pins (BusData, BusCLK, **EN**, BusMode) have a Schmitt-triggered input stage. The BusData pin is bidirectional where the output is open drain driven by an internal $15k\Omega$ pull up resistor.

Table 2-6 Bus I	nterface Format			
Function	BusMode	EN	BusCLK	BusData
I ² C Mode	Low	High= inactive,	Clock input	Data in/out
3-wire Mode	High	Low= active		



i2c 3w bus.wmf

Figure 2-7 Bus Interface

Note: The Interface is able to access the internal registers at any time, even in POWER DOWN mode. There is no internal clock necessary for Interface operation.

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I²C Bus Mode

In this mode the **BusMode** pin (pin 2) = LOW and the \overline{EN} pin (pin 24) = LOW.

Data Transition:

Data transition on the pin BusData can only occur when BusCLK is LOW. BusData transitions while BusCLK is HIGH will be interpreted as start or stop condition.

Start Condition (STA):

A start condition is defined by a HIGH to LOW transition of the BusData line while BusCLK is HIGH. This start condition must precede any command and initiate a data transfer onto the bus.

Stop Condition (STO):

A stop condition is defined by a LOW to HIGH transition of the BusData line while BusCLK is HIGH. This condition terminates the communication between the devices and forces the bus interface into the initial state.

Acknowledge (ACK):

Indicates a successful data transfer. The transmitter will release the bus after sending 8 bit of data. During the 9th clock cycle the receiver will set the SDA line to LOW level to indicate it has received the 8 bits of data correctly.

Data Transfer Write Mode:

To start the communication, the bus master must initiate a start condition (STA), followed by the 8bit chip address. The chip address for the TDA5255 is fixed as "1110000" (MSB at first). The last bit (LSB=A0) of the chip address byte defines the type of operation to be performed:

A0=0, a write operation is selected and A0=1 a read operation is selected.

After this comparison the TDA5255 will generate an ACK and awaits the desired sub address byte (00H...0FH) and data bytes. At the end of the data transition the master has to generate the stop condition (STO).

Data Transfer Read Mode:

To start the communication in the read mode, the bus master must initiate a start condition (STA), followed by the 8 bit chip address (write: A0=0), followed by the sub address to read (80H, 81H), followed by the chip address (read: A0=1). After that procedure the data of the selected register (80H, 81H) is read out. During this time the data line has to be kept in HIGH state and the chip sends out the data. At the end of data transition the master has to generate the stop condition (STO).

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Bus Data Format in I²C Mode

Table 2-7	LSB Function													
MSB					Function									
1	1	1	0	0	0	0	0	Chip Address Write						
1	1	1	0	0	0	0	1	Chip Address Read						

Tab	ole 2-	-8		l ²	C I	3u:	s V	Vrite	Mod	le 8	Bit																	
	MSB		СНІ		DDR	ESS	i	LSB		MSB		UB A		•		,	LSB		MSB			D	ATA	IN		LSB		
STA	1	1	1	0	0	0	0	0	ACK	S7	S6	S5	S4	S3	S2	S1	S0	ACK	D7	D6	D5	D4	D3	D2	D1	D0	ACK	STO

Tak	ole 2	2-9		l ²	C	Bus	W	rite	Мо	de '	16 E	3it														
	MSB	CH	IIP AI	DDRI	ESS	(WRI	TE)	LSB		MSB			ADDRI 08H. 0	•		,	LSB		MSB		DA	TA II	N	LSB		
STA	1	1	1	0	0	0	0	0	ACK	S7	S6	S5	S4	S3	S2	S1	S0	ACK	D15	 D8 AC	K D	7 D	6	D0	ACK	STO

Tal	ble	2	-10)		² C	Вι	ıs	Rea	ıd N	/lod	е																	
	MS	SB (CHIE	P A [DDRI	ESS	(WR	ITE)	LSB		MSB	5	SUB A		ESS (, 81H	REAI	D)	LSB			MSB	CH	IIP A	DDI	RESS	(RE	AD)	LSB	
STA	. 1		1	1	0	0	0	0	0	ACK	S7	S6	S5	S4	S3	S2	S1	S0	ACK	STA	1	1	1	0	0	0	0	1	ACK
Tal	ble	2	-10)		² C	Вι	ıs	Rea	id N	/lod	e (d	con	tinı	ıed)														
	MS	SB		DATA OUT FROM SUB ADDRESS LSB																									
	R7	7			R6	,		F	R5		R	4		R3	3		R2			R1			R0			ACK	*	5	STO

^{*} mandatory HIGH

3-wire Bus Mode

In this mode pin 2 (BusMode)= HIGH and Pin 16 (BusData) is in the data input/output pin. Pin 24 $(\overline{\text{EN}})$ is used to activate the bus interface to allow the transfer of data to / from the device. When pin 24 $(\overline{\text{EN}})$ is inactive (HIGH), data transfer is inhibited.

Data Transition:

Data transition on pin 16 (BusData) can only occur if the clock BusCLK is LOW. To perform a data transfer the interface has to be enabled. This is done by setting the $\overline{\text{EN}}$ line to LOW. A serial transfer is done via BusData, BusCLK and $\overline{\text{EN}}$. The bit stream needs no chip address.

Data Transfer Write Mode:

To start the communication the $\overline{\text{EN}}$ line has to be set to LOW. The desired sub address byte and data bytes have to follow. The subaddress (00H...0FH) determines which of the data bytes are transmitted. At the end of data transition the $\overline{\text{EN}}$ must be HIGH.

Data transfer Read Mode:

To start the communication in the read mode, the $\overline{\text{EN}}$ line has to be set to LOW followed by the sub address to read (80H, 81H). Afterwards the device is ready to read out data. At the end of data transition $\overline{\text{EN}}$ must be HIGH.

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