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Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

Email & Skype: info@chipsmall.com Web: www.chipsmall.com

Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China



DATA SHEET



TEA6843HL New In Car Entertainment (NICE) extended car radio

Product specification

2003 Dec 19

New In Car Entertainment (NICE) extended car radio

TEA6843HL

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1 FEATURES

- FM mixer for conversion of FM RF from 65 to 108 MHz to IF of 10.7 MHz; the mixer provides inherent image rejection
- FM RF mixer can be set to receive weather band radio up to 162.55 MHz
- Buffer output for weather band flag
- AM mixer 1 for conversion of AM RF to AM IF1 of 10.7 MHz
- LC tuner oscillator providing mixer frequencies for FM mixer and AM mixer 1
- AM mixer 2 for conversion of AM IF1 to AM IF2 of 450 kHz
- Crystal oscillator providing mixer frequencies for AM mixer 2 and reference for synthesizer PLL, IF count, timing for inaudible Radio Data System (RDS) update and reference frequency for car audio signal processor ICs
- Fast synthesizer PLL tuning system with local control for inaudible RDS updating
- Timing function for RDS update algorithm and control signal output for car audio signal processor ICs (TEA688x, SAA77xx) or car radio integrated signal processor IC (TEF689x)
- Digital auto alignment circuit for conversion of LC oscillator tuning voltage to controlled alignment voltage of FM antenna tank circuit
- AGC PIN diode drive circuit for FM RF AGC; AGC detection at FM mixer input; the AGC PIN diode current can be set to a fixed value to allow local function for search tuning; AGC threshold is a programmable and keyed function switchable via the I²C-bus
- FM IF linear amplifiers with high dynamic input range
- FM quadrature demodulator with automatic centre frequency adjustment and Total Harmonic Distortion (THD) compensation and level driven soft mute; soft mute characteristic is adjustable via the I²C-bus



- Level detector for AM and FM with temperature compensated output voltage; starting point and slope of level output is programmable via the I²C-bus
- AM RF PIN diode drive circuit; AGC threshold detection at AM mixer 1 and IF2 AGC input; threshold is programmable via the I²C-bus; AM IF2 AGC and demodulator
- AM AF output switchable to provide AM IF2 for AM stereo decoder
- AM noise blanker with blanking at AM IF2
- Several test modes available for fast IC and system tests.

2 GENERAL DESCRIPTION

The TEA6843HL is a single IC with car radio tuner for AM, FM and Weather Band (WB) intended for microcontroller tuning with the I²C-bus. It provides the following functions:

- AM double conversion receiver for LW, MW and SW (31 m, 41 m and 49 m bands) with IF1 = 10.7 MHz and IF2 = 450 kHz
- FM single conversion receiver with integrated image rejection for IF = 10.7 MHz capable of selecting US FM, US weather, Europe FM, East Europe FM and Japan FM bands
- The tuning system includes FM mixer, AM mixer 1, AM mixer 2, crystal oscillator, VCO and PLL synthesizer on one chip.

3 ORDERING INFORMATION

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
TEA6843HL	LQFP80	plastic low profile quad flat package; 80 leads; body 12 × 12 × 1.4 mm	SOT315-1

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4 QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$V_{DDA(n)}$	analog supply voltage 1 to 4 and 6		8	8.5	9	V
$I_{DDA(tot)}$	total analog supply currents 1 to 4 and 6 and mixer bias currents	FM mode; data byte 5: bit TMS3 = 1	45	60	69	mA
		AM mode; data byte 5: bit TMS3 = 1	39	50	59	mA
V_{DDA5}	analog supply voltage 5		4.75	5	5.25	V
I_{DDA5}	analog supply current 5	FM mode; data byte 5: bit 5 = 0, bit 6 = 0	6.5	9.3	11.2	mA
		AM mode; data byte 5: bit 5 = 1, bit 6 = 1	12.7	17.4	22.1	mA
V_{DDD}	digital supply voltage		4.75	5	5.25	V
I_{DDD}	digital supply current	FM mode	18	23	28	mA
		AM mode	18	23	28	mA
$f_{AM(ant)}$	AM input frequency	LW	0.144	–	0.288	MHz
		MW	0.522	–	1.710	MHz
		SW	5.85	–	9.99	MHz
$f_{FM(ant)}$	FM input frequency		65	–	108	MHz
$f_{FM(WB)(ant)}$	FM weather band input frequency		162.4	–	162.55	MHz
T_{amb}	ambient temperature		–40	–	+85	°C

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5 BLOCK DIAGRAM

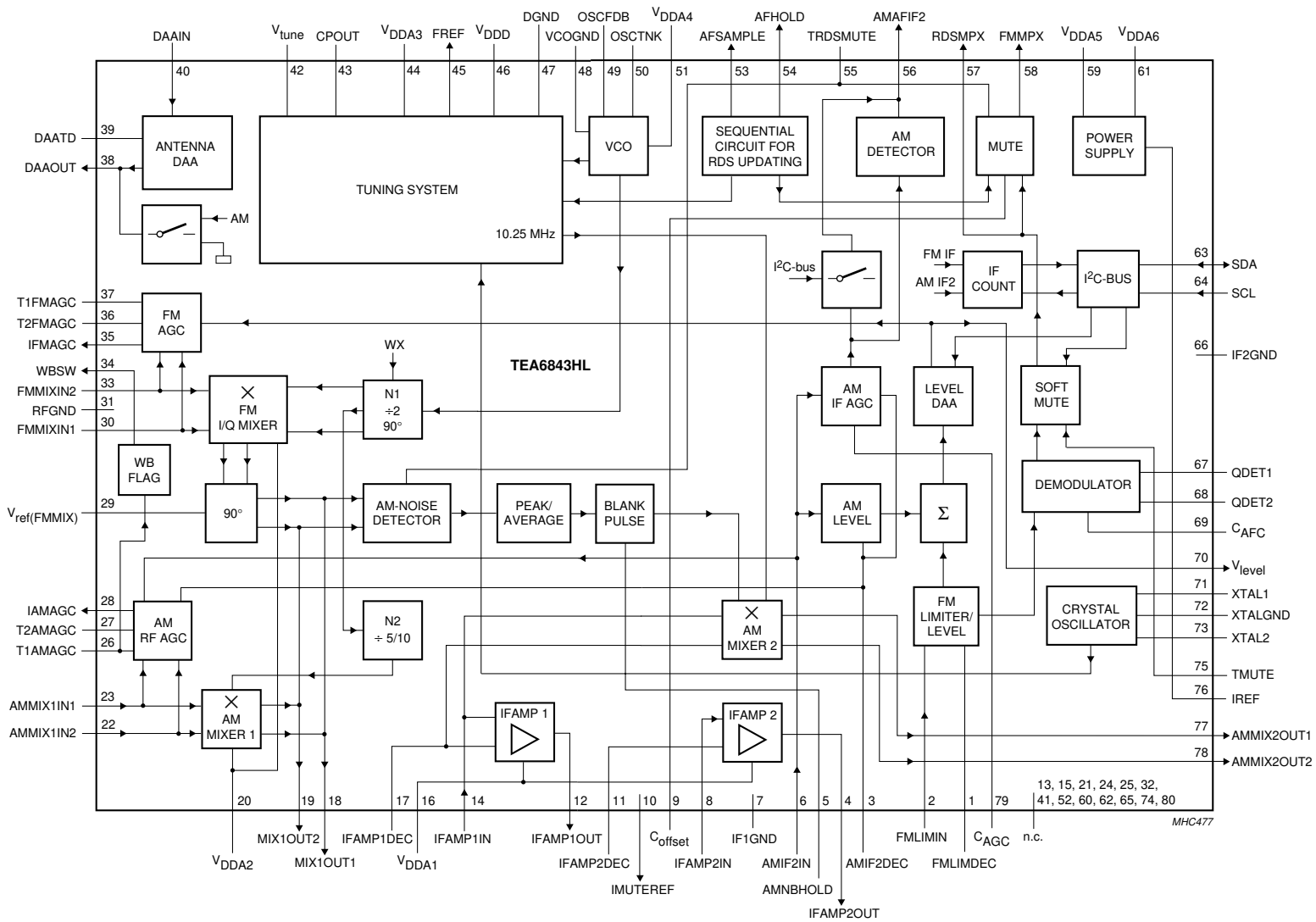


Fig.1 Block diagram

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6 PINNING

SYMBOL	PIN	DESCRIPTION
FMLIMDEC	1	FM limiter decoupling
FMLIMIN	2	FM limiter input (10.7 MHz)
AMIF2DEC	3	decoupling for AM IF2 input
IFAMP2OUT	4	IF amplifier 2 output (10.7 MHz)
AMNBHOLD	5	AM noise blanker threshold
AMIF2IN	6	AM IF2 input (450 kHz) for demodulator AGC and AM level detector
IF1GND	7	AM IF1 ground
IFAMP2IN	8	IF amplifier 2 input (10.7 MHz)
C _{offset}	9	DC feedback for offset compensation RDS mute
IMUTEREF	10	mute reference current output for testing
IFAMP2DEC	11	IF amplifier 2 decoupling and AGC capacitor for AM noise blanker
IFAMP1OUT	12	IF amplifier 1 output (10.7 MHz)
n.c.	13	not connected
IFAMP1IN	14	IF amplifier 1 and AM mixer 2 input (10.7 MHz)
n.c.	15	not connected
V _{DDA1}	16	analog supply voltage 1 (8.5 V) for IF amplifier 1 and 2
IFAMP1DEC	17	AM mixer 2 and FM IF amplifier 1 decoupling
MIX1OUT1	18	FM mixer and AM mixer 1 IF output 1 (10.7 MHz)
MIX1OUT2	19	FM mixer and AM mixer 1 IF output 2 (10.7 MHz)
V _{DDA2}	20	analog supply voltage 2 (8.5 V) for FM mixer and AM mixer 1
n.c.	21	not connected
AMMIX1IN2	22	AM mixer 1 input 2
AMMIX1IN1	23	AM mixer 1 input 1
n.c.	24	not connected
n.c.	25	not connected
T1AMAGC	26	1st time constant of AM front-end AGC
T2AMAGC	27	2nd time constant of AM front-end AGC
IAMAGC	28	PIN diode drive current output of AM front-end AGC
V _{ref(FMMIX)}	29	reference voltage for FM I/Q mixer
FMMIXIN1	30	FM RF mixer input 1
RFGND	31	RF ground
n.c.	32	not connected
FMMIXIN2	33	FM RF mixer input 2
WBSW	34	buffered weather band flag output
IFMAGC	35	PIN diode drive current output of FM front-end AGC
T2FMAGC	36	2nd time constant of FM front-end AGC
T1FMAGC	37	1st time constant of FM front-end AGC
DAAOUT	38	output of digital auto alignment circuit for antenna tank circuit
DAATD	39	temperature compensation diode for digital auto alignment circuit for antenna tank circuit
DAAIN	40	input of digital auto alignment circuit for antenna tank circuit

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SYMBOL	PIN	DESCRIPTION
n.c.	41	not connected
V _{tune}	42	tuning voltage
CPOUT	43	charge pump output
V _{DDA3}	44	analog supply voltage 3 (8.5 V) for tuning PLL
FREF	45	reference frequency output for signal processor IC
V _{DDD}	46	digital supply voltage (5 V)
DGND	47	digital ground
VCOGND	48	VCO ground
OSCFDB	49	VCO feedback
OSCTNK	50	VCO tank circuit
V _{DDA4}	51	analog supply voltage 4 (8.5 V) for VCO
n.c.	52	not connected
AFSAMPLE	53	AF sample flag output for car audio signal processor IC
AFHOLD	54	AF hold flag output for car audio signal processor IC
TRDSMUTE	55	time constant for RDS update mute
AMAFIF2	56	AM demodulator AF output or IF2 output for AM stereo (multiplexed by I ² C-bus)
RDSMPX	57	MPX output for RDS decoder and signal processor (not muted)
FMMPX	58	FM demodulator MPX output
V _{DDA5}	59	analog supply voltage 5 (5 V) for on-chip power supply
n.c.	60	not connected
V _{DDA6}	61	analog supply voltage 6 (8.5 V) for on-chip power supply
n.c.	62	not connected
SDA	63	I ² C-bus data line input and output
SCL	64	I ² C-bus clock line input
n.c.	65	not connected
IF2GND	66	AM IF2 ground
QDET1	67	quadrature demodulator tank 1
QDET2	68	quadrature demodulator tank 2
C _{AFC}	69	FM demodulator AFC capacitor
V _{level}	70	level voltage output for AM and FM
XTAL1	71	crystal oscillator 1
XTALGND	72	crystal oscillator ground
XTAL2	73	crystal oscillator 2
n.c.	74	not connected
TMUTE	75	time constant for mute control voltage
IREF	76	reference current for power supply
AMMIX2OUT1	77	AM mixer 2 output 1 (450 kHz)
AMMIX2OUT2	78	AM mixer 2 output 2 (450 kHz)
C _{AGC}	79	AM IF AGC capacitor
n.c.	80	not connected

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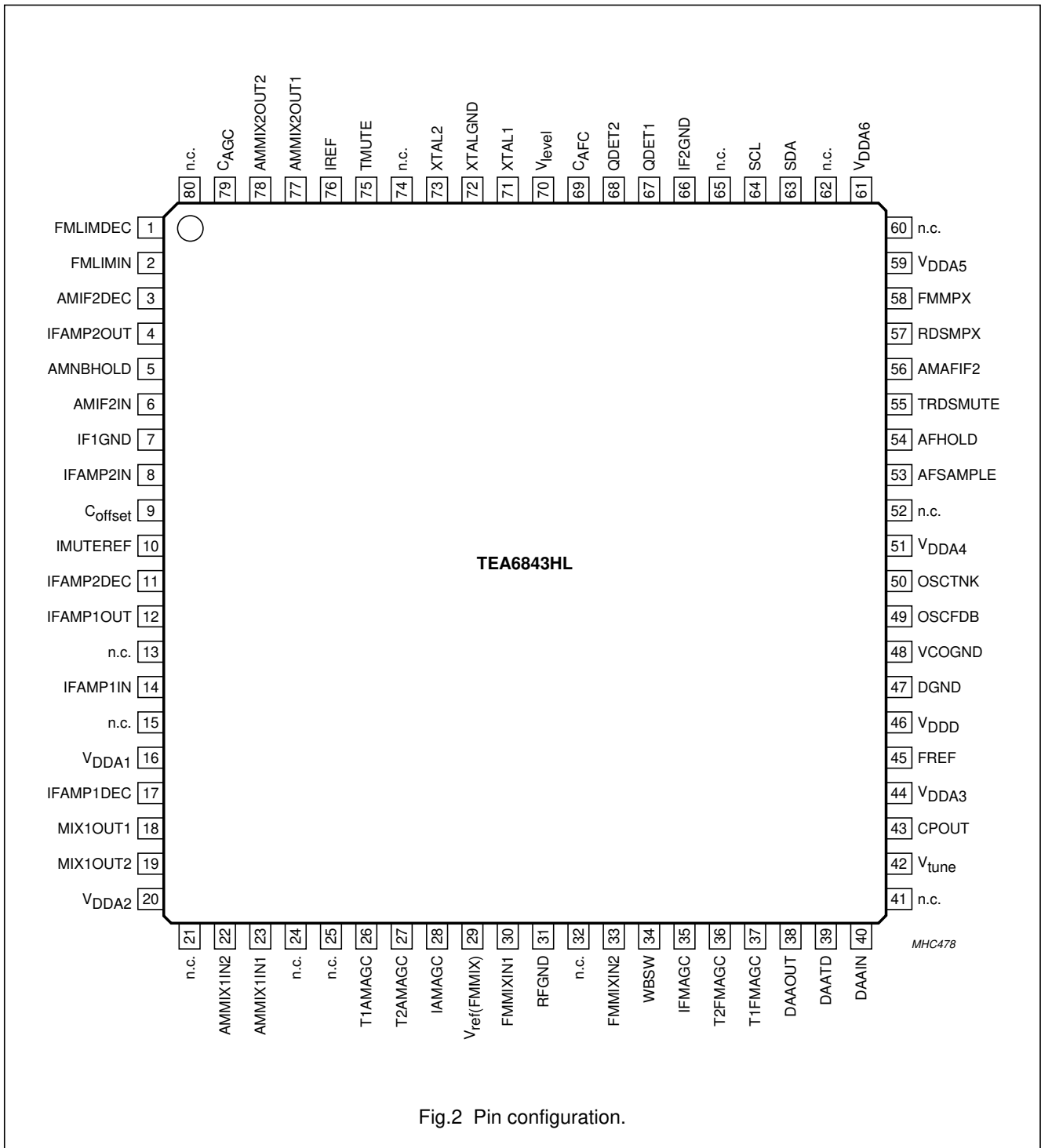


Fig.2 Pin configuration.

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7 FUNCTIONAL DESCRIPTION

7.1 Oscillators

7.1.1 VCO

The L and C tuned VCO provides the local oscillator signal for both FM and AM mixer 1. It has a frequency range of 151.2 to 248.2 MHz.

7.1.2 CRYSTAL OSCILLATOR

The crystal oscillator provides a 20.5 MHz signal that is used for:

- Reference frequency for frequency synthesizer PLL
- Local oscillator for AM mixer 2
- Reference frequency for the IF counter
- Timing signal for the RDS update algorithm
- Reference frequency (75.368 kHz) for the TEA688x (car audio signal processor - CASP) or TEF689x (car radio integrated signal processor - CRISP).

7.1.3 PLL

Fast synthesizer PLL tuning system with local control for inaudible RDS updating.

7.2 FM signal channel

7.2.1 DAA

FM RF Digital Auto Alignment (DAA) circuitry for the conversion of the VCO tuning voltage to a controlled alignment voltage for the FM antenna tank circuit.

7.2.2 FM I/Q MIXER

FM quadrature mixer converts FM RF (65 to 162.55 MHz) to IF of 10.7 MHz. The FM mixer provides inherent image rejection and high RF sensitivity.

It is capable of tuning the US FM, US weather, Europe FM, Japan FM and East Europe FM bands:

- US FM = 87.9 to 107.9 MHz
- US weather FM = 162.4 to 162.55 MHz
- Europe FM = 87.5 to 108 MHz
- Japan FM = 76 to 91 MHz
- East Europe FM = 65.8 to 74 MHz.

7.2.3 FM KEYED AGC

The AGC threshold is programmable and the keyed AGC function is switchable via the I²C-bus.

If the keyed AGC function is activated, the AGC function is keyed by the narrow band level (IF).

The AGC PIN diode drive can be set via the I²C-bus to local function for search tuning.

7.2.4 FM IF AMPLIFIERS

The two FM IF amplifiers provide 10 dB and 8.5 dB amplification with high linearity and dynamic range.

7.2.5 FM DEMODULATOR

The FM quadrature demodulator includes automatic centre frequency adjustment and THD compensation.

7.2.6 FM MPX SOFT MUTE

Muting depth and start of muting are adjustable via the I²C-bus.

7.3 AM signal channel

7.3.1 AM TUNER

The AM tuner is realized in a double conversion technique and is capable of selecting LW, MW and SW bands.

AM mixer 1 converts AM RF to IF1 of 10.7 MHz, while AM mixer 2 converts IF1 of 10.7 MHz to IF2 of 450 kHz:

- LW = 144 to 288 kHz
- MW = 522 to 1710 kHz (European and US AM band)
- SW = 5.85 to 9.99 MHz (including the 31 m, 41 m and 49 m bands).

7.3.2 AM RF AGC AND IF2 AGC

The AM RF includes a PIN diode drive circuit. The threshold detection points for AM AGC are performed at AM mixer 1 and AM IF2. AGC thresholds are programmable at AM mixer 1 input via the I²C-bus.

7.3.3 AM AF OR IF2 SWITCH

The AM output provides either a demodulated AM Audio Frequency (AF) or the corresponding AM IF2 signal. The IF2 signal can be used for AM stereo decoder processing.

7.3.4 AM SOFT MUTE

The AM detector output is either not muted or signal strength dependent soft muted audio. The soft mute function can be switched off via the I²C-bus.

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7.3.5 AM NOISE BLANKER

The detection point for the AM noise blanker is the output stage of AM mixer 1, while blanking is realized at the output of the mixer 2.

Trigger sensitivity can be modified by changing the resistor value at pin AMNBHOLD.

7.4 FM and AM level detector

FM and AM level detectors provide the temperature compensated output voltage. The starting points and slopes of the level detector outputs are programmable via the I²C-bus.

7.5 Buffer output for weather band flag

In weather band mode the output is HIGH.

7.6 Test mode

The test mode of the IC is activated by connecting pin FREF through a 100 k Ω resistor to V_{DDA1}.

The test mode is intended for IC test during IC and radio manufacturing.

In test mode the settling time of the AM IF2 AGC is reduced to less than 100 ms in the nominal application.

8 LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 60134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V _{DDA1}	analog supply voltage 1 for IF amplifier 1 and 2		-0.3	+10	V
V _{DDA2}	analog supply voltage 2 for FM mixer and AM mixer 1		-0.3	+10	V
V _{DDA3}	analog supply voltage 3 for tuning PLL		-0.3	+10	V
V _{DDA4}	analog supply voltage 4 for VCO		-0.3	+10	V
V _{DDA5}	analog supply voltage 5 for on-chip power supply		-0.3	+6.5	V
V _{DDA6}	analog supply voltage 6 for on-chip power supply		-0.3	+10	V
V _{DDD}	digital supply voltage		-0.3	+6.5	V
$\Delta V_{DD8.5-DD5}$	difference between any 8.5 V supply voltage and any 5 V supply voltage	note 1	-0.3	-	V
T _{stg}	storage temperature		-55	+150	°C
T _{amb}	ambient temperature		-40	+85	°C
V _{es}	electrostatic handling voltage	note 2	-200	+200	V
		note 3	-2000	+2000	V

Notes

- To avoid damages and wrong operation it is necessary to keep all 8.5 V supply voltages at a higher level than any 5 V supply voltage. This is also necessary during power-on and power-down sequences. Precautions have to be provided in such a way that interferences can not pull down the 8.5 V supply below the 5 V supply.
- Machine model (R = 0 Ω , C = 200 pF).
- Human body model (R = 1.5 k Ω , C = 100 pF).

9 THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
R _{th(j-a)}	thermal resistance from junction to ambient	in free air	54	K/W

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10 DC CHARACTERISTICS

 $V_{MIX1OUT} = V_{AMMIX2OUT} = V_{DDA(n)} = 8.5 \text{ V}; V_{DDA5} = 5 \text{ V}; V_{DDD} = 5 \text{ V}; T_{amb} = 25 \text{ }^\circ\text{C};$ unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply voltage						
$V_{DDA(n)}$	analog supply voltages 1 to 4 and 6		8	8.5	9	V
$V_{MIX1OUT};$ $V_{AMMIX2OUT}$	mixer supply voltages		8	8.5	9	V
V_{DDA5}	analog supply voltage 5		4.75	5	5.25	V
V_{DDD}	digital supply voltage		4.75	5	5.25	V
Supply current in FM mode						
I_{DDD}	digital supply current		18	23	28	mA
I_{DDA1}	analog supply current 1 for IF amplifier 1 and 2		11	13	15	mA
I_{DDA2}	analog supply current 2 for FM mixer and AM mixer 1		4.2	5.2	6.2	mA
I_{DDA3}	analog supply current 3 for tuning PLL	data byte 5: bit TMS3 = 1	2.3	3	3.6	mA
I_{DDA4}	analog supply current 4 for VCO		5.2	6.5	7.8	mA
I_{DDA5}	analog supply current 5 for on-chip power supply		6.5	9.3	11.2	mA
I_{DDA6}	analog supply current 6 for on-chip power supply		16	20	24	mA
$I_{MIX1OUT1}$	bias current of FM mixer output 1		4.8	6	7.2	mA
$I_{MIX1OUT2}$	bias current of FM mixer output 2		4.8	6	7.2	mA
Supply current in AM mode						
I_{DDD}	digital supply current		18	23	28	mA
I_{DDA1}	analog supply current 1 for AM mixer 2		–	240	–	μA
I_{DDA2}	analog supply current 2 for FM mixer and AM mixer 1		1.6	2	2.4	mA
I_{DDA3}	analog supply current 3 for tuning PLL	data byte 5: bit TMS3 = 1	1.3	1.7	2.1	mA
I_{DDA4}	analog supply current 4 for VCO		5	6.5	8	mA
I_{DDA5}	analog supply current 5 for on-chip power supply		12.7	17.4	22.1	mA
I_{DDA6}	analog supply current 6 for on-chip power supply		15	19	23	mA
$I_{MIX1OUT1}$	bias current of AM mixer 1 output 1		4.8	6	7.2	mA
$I_{MIX1OUT2}$	bias current of AM mixer 1 output 2		4.8	6	7.2	mA
$I_{AMMIX2OUT1}$	bias current of AM mixer 2 output 1		3.6	4.5	5.4	mA
$I_{AMMIX2OUT2}$	bias current of AM mixer 2 output 2		3.6	4.5	5.4	mA

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Power supply reference current generator: pin IREF						
$V_{o(ref)}$	output reference voltage		4	4.25	4.5	V
R_o	output resistance		8	11	13	k Ω
$I_{o(max)}$	maximum output current		-100	-	+100	nA

11 AC CHARACTERISTICS

$V_{MIX1OUT} = V_{AMMIX2OUT} = V_{DDA(n)} = 8.5$ V; $V_{DDA5} = 5$ V; $V_{DDD} = 5$ V; $T_{amb} = 25$ °C; see Figs 14 and 15; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Voltage controlled oscillator						
f_{osc}	oscillator frequency		151.2	-	248.2	MHz
C/N	carrier-to-noise ratio	$f_{osc} = 200$ MHz; $\Delta f = 10$ kHz; $B = 1$ Hz	-	101	-	dBc
RR	ripple rejection $\frac{\Delta f_{osc}}{f_{osc}}$	$f_{ripple} = 100$ Hz; $V_{DDA4(ripple)} = 100$ mV (RMS) $f_{osc} = 250$ MHz $f_{osc} = 200$ MHz	90 92	97 99	- -	dB dB
Crystal oscillator: pins XTAL1, XTALGND and XTAL2						
f_{osc}	oscillator frequency		-	20.5	-	MHz
R_i	real part of input impedance	$V_{XTAL1} - V_{XTAL2} = 1$ mV	-500	-	-	Ω
C_i	input capacitance		8	10	12	pF
$V_{o(osc)(rms)}$	oscillator output voltage (RMS value)		240	350	500	mV
C/N	carrier-to-noise ratio	$f_{osc} = 20.5$ MHz (10.25 MHz); $\Delta f = 10$ kHz	-	112	-	$\frac{dBc}{\sqrt{Hz}}$
Oscillator divider N1						
N1	oscillator divider ratio	FM mode standard, East Europe and local weather band (WX)	- -	2 1	- -	
Oscillator divider N2						
N2	oscillator divider ratio	AM mode LW and MW SW	- -	10 5	- -	
Synthesizer						
PROGRAMMABLE DIVIDER						
N_{prog}	programmable divider ratio		512	-	32767	

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
ΔN_{step}	programmable divider step size		–	1	–	
REFERENCE FREQUENCY DIVIDER						
N_{ref}	crystal oscillator divider ratio	$f_{\text{osc}} = 20.5 \text{ MHz}$ $f_{\text{ref}} = 100 \text{ kHz}$ $f_{\text{ref}} = 50 \text{ kHz}$ $f_{\text{ref}} = 25 \text{ kHz}$ $f_{\text{ref}} = 20 \text{ kHz}$ $f_{\text{ref}} = 10 \text{ kHz}$	–	205	–	
			–	410	–	
			–	820	–	
			–	1025	–	
			–	2050	–	
CHARGE PUMP: PIN CPOUT						
$I_{\text{sink(cp1)l}}$	low charge pump 1 sink current	$0.4 \text{ V} < V_{\text{CPOUT}} < 7.6 \text{ V}$; data byte 4: bit 0 = 0, bit 1 = 1, bit 2 = 1 for FM weather band; $f_{\text{VCO}} > f_{\text{ref}} \times \text{divider ratio}$	–	300	–	μA
$I_{\text{source(cp1)l}}$	low charge pump 1 source current	$0.4 \text{ V} < V_{\text{CPOUT}} < 7.6 \text{ V}$; data byte 4: bit 0 = 0, bit 1 = 1, bit 2 = 1 for FM weather band; $f_{\text{VCO}} < f_{\text{ref}} \times \text{divider ratio}$	–	–300	–	μA
$I_{\text{sink(cp1)h}}$	high charge pump 1 sink current	$0.4 \text{ V} < V_{\text{CPOUT}} < 7.6 \text{ V}$; data byte 4: bit 0 = 1, bit 1 = 1, bit 2 = 1 for AM IF2 output; $N2 = 10$ (LW and MW); $f_{\text{VCO}} > f_{\text{ref}} \times \text{divider ratio}$	–	1	–	mA
$I_{\text{source(cp1)h}}$	high charge pump 1 source current	$0.4 \text{ V} < V_{\text{CPOUT}} < 7.6 \text{ V}$; data byte 4: bit 0 = 1, bit 1 = 1, bit 2 = 1 for AM IF2 output; $N2 = 10$ (LW and MW); $f_{\text{VCO}} < f_{\text{ref}} \times \text{divider ratio}$	–	–1	–	mA
$I_{\text{sink(cp2)}}$	charge pump 2 sink current	$0.3 \text{ V} < V_{\text{CPOUT}} < 7.1 \text{ V}$; data byte 4: bit 0 = 0, bit 1 = 0, bit 2 = 0; FM standard mode; $f_{\text{VCO}} > f_{\text{ref}} \times \text{divider ratio}$	–	130	–	μA
$I_{\text{source(cp2)}}$	charge pump 2 source current	$0.3 \text{ V} < V_{\text{CPOUT}} < 7.1 \text{ V}$; data byte 4: bit 0 = 0, bit 1 = 0, bit 2 = 0; FM standard mode; $f_{\text{VCO}} < f_{\text{ref}} \times \text{divider ratio}$	–	–130	–	μA
CHARGE PUMP: PIN V_{tune}						
$I_{\text{sink(cp3)}}$	charge pump 3 sink current	$0.4 \text{ V} < V_{\text{tune}} < 7.6 \text{ V}$; data byte 4: bit 0 = 0, bit 1 = 0, bit 2 = 0; FM standard mode; $f_{\text{VCO}} > f_{\text{ref}} \times \text{divider ratio}$	–	3	–	mA
$I_{\text{source(cp3)}}$	charge pump 3 source current	$0.4 \text{ V} < V_{\text{tune}} < 7.6 \text{ V}$; data byte 4: bit 0 = 0, bit 1 = 0, bit 2 = 0; FM standard mode; $f_{\text{VCO}} < f_{\text{ref}} \times \text{divider ratio}$	–	–3	–	mA

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Antenna Digital Auto Alignment (DAA)						
DAA INPUT: PIN DAAIN						
$I_{bias(cp)}$	charge pump buffer input bias current	$V_{DAAIN} = 0.4$ to 8 V	-10	-	+10	nA
$V_{i(cp)}$	charge pump buffer input voltage		0	-	8.5	V
DAA OUTPUT: PIN DAAOUT						
$V_{o(min)}$	minimum DAA output voltage	data byte 3 = 10000000; FM mode; $V_{DAAIN} = 0.5$ V; $V_{DAAATD} = 0.45$ V	-	-	0.5	V
$V_{o(max)}$	maximum DAA output voltage	data byte 3 = 11111111; FM mode; $V_{DAAIN} = 4$ V; $V_{DAAATD} = 0.45$ V	8	-	8.5	V
V_o	DAA output voltage	$I_{DAAOUT} < 100$ μ A; AM mode	-	-	0.3	V
		data byte 3 = 10000000; FM mode; $V_{DAAIN} = 4$ V; $V_{DAAATD} = 0.45$ V; note 1	-	1.7	-	V
		data byte 3 = 11111111; FM mode; $V_{DAAIN} = 3$ V; $V_{DAAATD} = 0.45$ V; note 1	6.2	6.5	6.8	V
$V_{o(n)}$	DAA output noise voltage	data byte 3 = 11000000; FM mode; $V_{DAAIN} = 4$ V; $V_{DAAATD} = 0.45$ V; B = 400 Hz to 30 kHz	-	30	100	μ V
$\Delta V_{o(T)}$	DAA output voltage variation with temperature	$T_{amb} = -40$ to $+85$ °C; data byte 3 = 10101011	-8	-	+8	mV
$\Delta V_{o(step)}$	DAA step accuracy $V_{DAAOUT(n+1)} - V_{DAAOUT(n)}$	$n = 0$ to 127 ; FM mode; $V_{DAAOUT} < 8$ V; $V_{DAAIN} = 2$ V; $V_{DAAATD} = 0.45$ V	$0.5V_{LSB}$	V_{LSB}	$1.5V_{LSB}$	mV
$\Delta V_{o(sink)}$	DAA output variation caused by sink current	$V_{DAAIN} = 4$ V; $I_L = 50$ μ A	$-V_{LSB}$	-	$+V_{LSB}$	
$\Delta V_{o(source)}$	DAA output variation caused by source current	$V_{DAAIN} = 4$ V; $I_L = -50$ μ A	$-V_{LSB}$	-	$+V_{LSB}$	
t_{st}	DAA output settling time	$V_{DAAOUT} = 0.2$ to 8.25 V; $C_L = 270$ pF	-	20	30	μ s
RR	ripple rejection $\frac{V_{DAAOUT}}{V_{DDA3}}$	data byte 3 = 10101011; FM mode; $V_{DAAIN} = 4$ V; $V_{DAAATD} = 0.45$ V; $f_{ripple} = 100$ Hz; $V_{DDA3(ripple)} = 100$ mV	-	50	-	dB
C_L	DAA output load capacitance	$V_{DAAOUT} < 8$ V; FM mode	-	-	270	pF

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
DAA TEMPERATURE COMPENSATION: PIN DAATD						
I_{source}	compensation diode source current	$V_{\text{DAATD}} = 0.2 \text{ to } 1.2 \text{ V}$	-50	-40	-30	μA
T_{Csource}	temperature coefficient of compensation diode source current	$V_{\text{DAATD}} = 0.2 \text{ to } 1.2 \text{ V};$ $T_{\text{amb}} = -40 \text{ to } +85 \text{ }^\circ\text{C}$	-300	-	+300	$\frac{10^{-6}}{\text{K}}$
IF counter (FM IF1 or AM IF2 counter)						
N_{IF}	IF counter length for AM and FM		-	8	-	bit
$T_{\text{count(IF)}}$	IF counter period	data byte 4: bit 7 = 1	-	2	-	ms
		data byte 4: bit 7 = 0	-	20	-	ms
R_{precount}	FM IF counter prescaler ratio	data byte 4: bit 3 = 1	-	10	-	
		data byte 4: bit 3 = 0	-	100	-	
$V_{\text{FMLIMIN(sens)(rms)}}$	FM sensitivity voltage (RMS value)	FM mode	-	30	100	μV
N	counter result (decimal)	period = 2 ms; $f_{\text{IF1}} = 10.7 \text{ MHz};$ $V_{\text{FMLIMIN}} = 100 \mu\text{V (RMS)}$ prescaler ratio = 10	-	92	-	
		prescaler ratio = 100	-	214	-	
N	counter result (decimal)	period = 20 ms; $f_{\text{IF1}} = 10.7 \text{ MHz};$ $V_{\text{FMLIMIN}} = 100 \mu\text{V (RMS)}$ prescaler ratio = 10	-	152	-	
		prescaler ratio = 100	-	92	-	
$V_{\text{AMIF2IN(sens)(rms)}}$	AM sensitivity voltage (RMS value)	AM mode; $m = 0$	-	50	100	μV
N	counter result (decimal)	period = 2 ms; $f_{\text{IF2}} = 450 \text{ kHz};$ $V_{\text{AMIF2IN}} = 200 \mu\text{V (RMS)}$	-	133	-	
		period = 20 ms; $f_{\text{IF2}} = 450 \text{ kHz};$ $V_{\text{AMIF2IN}} = 200 \mu\text{V (RMS)}$	-	40	-	
Reference frequency for car signal processor IC; note 2						
REFERENCE FREQUENCY DIVIDER						
N_{ref}	crystal oscillator divider ratio		-	272	-	
f_{ref}	reference frequency at pin FREF	$f_{\text{osc}} = 20.5 \text{ MHz}$	-	75.368	-	kHz
VOLTAGE GENERATOR: PIN FREF						
$V_{\text{O(p-p)}}$	AC output voltage (peak-to-peak value)	$R_{\text{L}} = \infty$	80	130	200	mV
V_{O}	DC output voltage		3.2	3.4	3.7	V
R_{O}	output resistance		-	-	50	k Ω

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$R_{L(\min)}$	minimum load resistance for first I ² C-bus address		1	–	–	M Ω
AM signal channel						
RF AGC STAGE INPUTS						
<i>Pins AMMIX1IN1 and AMMIX1IN2</i>						
$V_{i(\text{RF})}$	RF input voltage for AGC start level	data byte 5: bit 5 = 0, bit 6 = 0	–	375	–	mV
		data byte 5: bit 5 = 1, bit 6 = 0	–	500	–	mV
		data byte 5: bit 5 = 0, bit 6 = 1	–	625	–	mV
		data byte 5: bit 5 = 1, bit 6 = 1	–	750	–	mV
<i>Pin AMIF2IN</i>						
$V_{i(\text{IF2})}$	IF2 input voltage	AGC start level	280	380	500	mV
<i>PIN diode drive: pin IAMAGC</i>						
$I_{\text{sink}(\max)}$	maximum AGC sink current	$V_o = 2.8 \text{ V}$	11	15	19	mA
I_{sink}	AGC sink current	FM mode	1	–	–	mA
R_o	output resistance	$I_o = 1 \mu\text{A}$	1	–	–	M Ω
C_o	AM AGC current generator output capacitance		–	5	7	pF
AM MIXER 1 (IF1 = 10.7 MHz)						
<i>Mixer inputs: pins AMMIX1IN1 and AMMIX1IN2</i>						
R_i	input resistance		15	25	35	k Ω
C_i	input capacitance		2.5	4	5.5	pF
V_I	DC input voltage		2.3	2.7	3.1	V
$V_{i(\max)}$	maximum input voltage	1 dB compression point of AM mixer 1 output (peak-to-peak value)	500	–	–	mV
$V_{i(n)(\text{eq})}$	equivalent input noise voltage	band limited noise; $R_{\text{gen}} = 750 \Omega$; $R_L = 2.8 \text{ k}\Omega$	–	5.8	8	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
<i>Mixer outputs: pins MIX1OUT1 and MIX1OUT2</i>						
R_o	output resistance		100	–	–	k Ω
C_o	output capacitance		–	5	7	pF
$V_{o(\max)(\text{p-p})}$	maximum output voltage (peak-to-peak value)		12	15	–	V
$g_{m(\text{conv})}$	conversion transconductance		2.0	2.55	3.2	$\frac{\text{mA}}{\text{V}}$
	$\frac{I_{\text{MIX1OUT}}}{V_{\text{AMMIX1IN}}}$					

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$g_{m(\text{conv})(T)}$	conversion transconductance variation with temperature $\frac{\Delta g_{m(\text{conv})}}{g_{m(\text{conv})} \times \Delta T}$		–	-9×10^{-4}	–	K^{-1}
IP3	3rd-order intermodulation	$R_L = 2.8 \text{ k}\Omega$; $\Delta f = 300 \text{ kHz}$	135	138	–	$\text{dB}\mu\text{V}$
IP2	2nd-order intermodulation	$R_L = 2.8 \text{ k}\Omega$	–	170	–	$\text{dB}\mu\text{V}$
F	noise figure of AM mixer 1		–	4.5	7.1	dB
I_{bias}	mixer bias current	AM mode	4.8	6	7.2	mA
AM MIXER 2 (IF2 = 450 kHz)						
<i>Mixer inputs: pins IFAMP1IN and IFAMP1DEC</i>						
R_i	input resistance		270	330	390	Ω
C_i	input capacitance		–	5	7	pF
V_i	DC input voltage		2.4	2.8	3.2	V
$V_{i(\text{max})(p)}$	maximum input voltage (peak value)	1 dB compression point of AM mixer 2 output (peak-to-peak value)	1.1	1.4	–	V
$V_{i(n)(\text{eq})}$	equivalent input noise voltage	$R_{\text{gen}} = 330 \Omega$; $R_L = 4 \text{ k}\Omega$	–	15	22	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
<i>Mixer outputs: pins AMMIX2OUT1 and AMMIX2OUT2</i>						
R_o	output resistance		100	–	–	$\text{k}\Omega$
C_o	output capacitance		–	5	7	pF
$V_{o(\text{max})(p-p)}$	maximum output voltage (peak-to-peak value)	$V_{\text{DDA1}} = 8.5 \text{ V}$	12	15	–	V
$g_{m(\text{conv})}$	conversion transconductance $\frac{I_{\text{AMMIX2OUT}}}{V_{\text{IFAMP1IN}}}$		1.3	1.6	1.9	$\frac{\text{mA}}{\text{V}}$
$g_{m(\text{conv})(T)}$	conversion transconductance variation with temperature $\frac{\Delta g_{m(\text{conv})}}{g_{m(\text{conv})} \times \Delta T}$		–	-9×10^{-4}	–	K^{-1}
IP3	3rd-order intermodulation	$R_L = 4 \text{ k}\Omega$; $\Delta f = 300 \text{ kHz}$	134	137	–	$\text{dB}\mu\text{V}$
IP2	2nd-order intermodulation	$R_L = 4 \text{ k}\Omega$	–	170	–	$\text{dB}\mu\text{V}$

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
F	noise figure of AM mixer 2		–	16	19.5	dB
I_{bias}	mixer bias current	AM mode	3.6	4.5	5.4	mA
I_L	mixer leakage current	FM mode	–	–	50	μ A
IF2 AGC STAGE: PINS AMIF2IN AND AMIF2DEC						
$V_{AGC(start)}$	AGC start voltage	input carrier voltage	–	14	30	μ V
$V_{AGC(stop)}$	AGC stop voltage	maximum input peak voltage	1	–	–	V
$V_{AGC(ctrl)}$	AGC control voltage	$V_{AMIF2IN} = 1$ mV	4.1	4.3	4.7	V
ΔAGC	AGC range	between start and stop of AGC	–	89	–	dB
R_i	input resistance		1.8	2	2.2	k Ω
C_i	input capacitance		–	–	5	pF
V_i	input voltage	audio attenuation $\alpha = -10$ dB data byte 5: bit 4 = 1; mute on data byte 5: bit 4 = 0; mute off	10 –	22 6	37 12	μ V μ V
$V_{sens(rms)}$	sensitivity voltage (RMS value)	$m = 0.3$; $f_{mod} = 400$ Hz; $B_{AF} = 2.5$ kHz; $R_{source} = 2$ k Ω (S+N)/N = 26 dB (S+N)/N = 46 dB	– –	45 600	65 900	μ V μ V
AM DETECTOR OUTPUT: PIN AMAFIF2; see Fig.3						
$V_{o(min)(rms)}$	minimum AM IF2 output level (RMS value)	data byte 4: bit 0 = 1, bit 1 = 1; $m = 0$; $V_{AMIF2IN} = 14$ μ V	1.5	3	4.5	mV
$V_{o(max)(rms)}$	maximum AM IF2 output level (RMS value)	data byte 4: bit 0 = 1, bit 1 = 1; $m = 0$; $V_{AMIF2IN} = 5$ mV	130	180	230	mV
$V_{o(rms)}$	AM AF output voltage level (RMS value)	data byte 4: bit 0 = 1, bit 1 = 0; $m = 0.3$; $f_{mod} = 400$ Hz; $V_{AMIF2IN} = 100$ μ V to 500 mV (RMS)	235	285	340	mV
R_o	output resistance	data byte 4: bit 0 = 1, bit 1 = 1	–	–	500	Ω
C_o	output capacitance	data byte 4: bit 0 = 1, bit 1 = 0	–	5	7	pF
Z_L	load impedance	data byte 4: bit 0 = 1, bit 1 = 0	100	–	–	k Ω
		data byte 4: bit 0 = 1, bit 1 = 1	10	–	–	k Ω
RR	ripple rejection	$f_{ripple} = 100$ Hz; $V_{DDA5(ripple)} = 100$ mV (RMS)	30	40	–	dB
AM DETECTOR; see Fig.3						
(S+N)/N	maximum signal plus noise-to-noise ratio	$m = 0.3$; $f_{mod} = 400$ Hz; $B_{AF} = 2.5$ kHz; $R_{source} = 2$ k Ω	54	60	70	dB

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
THD	total harmonic distortion	$B_{AF} = 2.5 \text{ kHz}; C_{AGC} = 22 \mu\text{F}; V_{AMIF2IN} = 100 \mu\text{V to } 250 \text{ mV (RMS)}$				
		$m = 0.8; f_{mod} = 400 \text{ Hz}$	–	0.5	1	%
		$m = 0.9; f_{mod} = 400 \text{ Hz}$	–	1	2	%
		$m = 0.8; f_{mod} = 100 \text{ Hz}$	–	1.25	2.5	%
		$m = 0.9; f_{mod} = 100 \text{ Hz}$	–	1.75	3.5	%
t_{sw}	FM to AM switching time	$V_{AMIF2IN} = 100 \mu\text{V}; C_{AGC} = 22 \mu\text{F}$	–	1000	1500	ms
t_{st}	AM demodulator AGC settling time	$C_{AGC} = 22 \mu\text{F}$				
		$V_{AMIF2IN} = 100 \mu\text{V to } 100 \text{ mV}$	–	800	1200	ms
		$V_{AMIF2IN} = 100 \text{ mV to } 100 \mu\text{V}$	–	1200	1800	ms
AM IF2 LEVEL DETECTOR: PIN V_{level} ; see Fig.4						
V_{level}	level output DC voltage	$V_{AMIF2IN} = 10 \mu\text{V to } 1 \text{ V}$	0	–	5	V
		$V_{AMIF2IN} < 1 \mu\text{V};$ standard setting of level DAA	0.1	0.35	0.8	V
		$V_{AMIF2IN} = 1.4 \text{ mV};$ standard setting of level DAA	–	2.2	–	V
ΔV_{level}	step size of starting point adjustment	standard setting of level slope	40	53	70	mV
$V_{level(slope)}$	slope of level voltage $\frac{\Delta V_{level}}{\Delta V_{AMIF2IN}}$	$V_{AMIF2IN} = 140 \mu\text{V to } 140 \text{ mV};$ standard setting of level slope	800	1000	1200	$\frac{\text{mV}}{20 \text{ dB}}$
ΔV_{step}	step size of slope adjustment	$V_{AMIF2IN} = 1.4 \text{ mV}$	55	75	95	$\frac{\text{mV}}{20 \text{ dB}}$
B_{level}	bandwidth of level output voltage		200	300	–	kHz
R_o	output resistance		–	–	500	Ω
RR	ripple rejection $\frac{V_{level}}{V_{DDA6}}$	$f_{ripple} = 100 \text{ Hz}; V_{DDA6(ripple)} = 100 \text{ mV (RMS)}$	–	40	–	dB
AM NOISE BLANKER; see Fig.5						
<i>Threshold: pin AMNBHOLD</i>						
V_o	DC output voltage		4.4	4.75	5.25	V
t_{sup}	suppression time		6	7.5	10	μs
$f_{trigger}$	trigger sensitivity	$V_{pulse} = 200 \text{ mV}; V_{level} < 1.8 \text{ V}$	–	1000	–	Hz
		$V_{pulse} = 200 \text{ mV}; V_{level} > 2.2 \text{ V}$	–	–	100	Hz
		$V_{pulse} = 20 \text{ mV}; V_{level} < 1.8 \text{ V}$	–	–	100	Hz
<i>Noise detector output: pin TRDSMUTE</i>						
$I_{sink(AGC)}$	AM noise blanker AGC sink current	$V_{TRDSMUTE} = 3 \text{ V}$	30	45	60	μA

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_{AGC}	AM noise blanker AGC voltage	AM mixer 1 input $V_i = 0$ V	2.1	2.45	2.85	V
FM signal channel						
RF AGC (FM DISTANCE MODE)						
<i>FM mixer inputs: pins FMMIXIN1 and FMMIXIN2</i>						
$V_{i(RF)(rms)}$	RF input voltage for start of wideband AGC (RMS value)	data byte 5: bit 5 = 1, bit 6 = 1	–	4	–	mV
		data byte 5: bit 5 = 0, bit 6 = 1	–	8	–	mV
		data byte 5: bit 5 = 1, bit 6 = 0	–	12	–	mV
		data byte 5: bit 5 = 0, bit 6 = 0	–	16	–	mV
<i>AGC control: pin T2FMAGC</i>						
R_{source}	source resistance		4	5	6	k Ω
$V_{O(ref)}$	DC output reference voltage	data byte 5: bit 5 = 0, bit 6 = 0; $V_{FMMIXIN1-FMMIXIN2} = 0$ V	3.8	4.1	4.4	V
<i>PIN diode drive output: pin IFMAGC</i>						
$I_{sink(AGC)(max)}$	maximum AGC sink current	$V_{IFMAGC} = 2.5$ V; $V_{T2FMAGC} = V_{O(ref)} - 0.5$ V; data byte 5: bit 5 = 0, bit 6 = 0, bit 7 = 0; data byte 4: bit 1 = 0	8	10.5	14	mA
$I_{source(AGC)(max)}$	maximum AGC source current	$V_{IFMAGC} = 2.5$ V; $V_{T2FMAGC} = V_{O(ref)} + 0.5$ V; data byte 5: bit 5 = 0, bit 6 = 0, bit 7 = 0; data byte 4: bit 1 = 0	–14	–10.5	–8	mA
$I_{source(AGC)}$	AGC source current	AM mode	–19	–14	–10	mA
		$V_{IFMAGC} = 2.5$ V; data byte 4: bit 0 = 0, bit 1 = 1, bit 2 = 0 (FM local)	–4.4	–3.7	–2.7	mA
<i>Level voltage: pin V_{level}</i>						
V_{th}	threshold voltage for narrow-band AGC	data byte 5: bit 5 = 0, bit 6 = 0, bit 7 = 1	600	750	900	mV
FM RF MIXER						
<i>FM mixer inputs: pins FMMIXIN1 and FMMIXIN2</i>						
$V_{i(RF)(max)}$	maximum RF input voltage	1 dB compression point of FM mixer output voltage (peak-to-peak value)	70	140	–	mV
$V_{i(n)(eq)}$	equivalent input noise voltage	$R_{gen} = 200$ Ω ; $R_L = 2.8$ k Ω	–	2.6	3.1	$\frac{nV}{\sqrt{Hz}}$
R_i	input resistance		1.4	2.8	4.2	k Ω
C_i	input capacitance		–	5	7	pF
<i>FM mixer outputs: pins MIX1OUT1 and MIX1OUT2</i>						
R_o	output resistance		100	–	–	k Ω
C_o	output capacitance		2	3.5	5	pF

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
I_{bias}	mixer bias current	FM mode	4.8	6	7.2	mA
$V_{o(max)(p-p)}$	maximum output voltage (peak-to-peak value)		3	–	–	V
<i>FM mixer reference voltage: pin $V_{ref(FMMIX)}$</i>						
V_{ref}	reference voltage	FM mode	6.5	7.1	7.9	V
		AM mode	2.7	3.1	3.4	V
<i>FM mixer</i>						
$g_{m(conv)}$	conversion transconductance $\frac{I_{MIX1OUT}}{V_{FMMIXIN}}$		8.5	13	18	$\frac{mA}{V}$
$g_{m(conv)(T)}$	conversion transconductance variation with temperature $\frac{\Delta g_{m(conv)}}{g_{m(conv)} \times \Delta T}$		–	-1×10^{-3}	–	K ⁻¹
F	noise figure		–	3	4.6	dB
$R_{gen(opt)}$	optimum generator resistance		180	240	300	Ω
IP3	3rd-order intermodulation		116	119	–	dB μ V
IRR	image rejection ratio $\frac{V_{MIX1OUTwanted}}{V_{MIX1OUTimage}}$	$f_{RFwanted} = 87.5$ MHz; $f_{RFimage} = 108.9$ MHz	25	30	–	dB
		data byte 4 = X010X110; $f_{RFwanted} = 162.475$ MHz; $f_{RFimage} = 183.875$ MHz; weather band mode	22	30	–	dB
<i>IF AMPLIFIER 1</i>						
G	gain $\frac{V_{IFAMP1OUT}}{V_{IFAMP1IN}}$	$R_L = 330 \Omega$; $V_{IFAMP1IN} = 1$ mV	8.5	10.5	12.5	dB
F	noise figure		–	10	13	dB
IP3	3rd-order intermodulation		117	120	–	dB μ V
<i>Inputs: pins IFAMP1IN and IFAMP1DEC</i>						
$V_{i(max)(p)}$	maximum input voltage (peak value)	1 dB compression point of IF amplifier 1 output voltage (peak value)	250	–	–	mV
$V_{i(n)(eq)}$	equivalent input noise voltage	$R_{gen} = 330 \Omega$; $R_L = 330 \Omega$	–	8	10	$\frac{nV}{\sqrt{Hz}}$

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
R_i	input resistance		270	330	390	Ω
C_i	input capacitance		–	5	7	pF
<i>Output: pin IFAMP1OUT</i>						
$V_{o(max)(p)}$	maximum output voltage (peak value)		1.2	1.5	–	V
R_o	output resistance		270	330	390	Ω
C_o	output capacitance		–	5	7	pF
IF AMPLIFIER 2						
G	gain $\frac{V_{IFAMP2OUT}}{V_{IFAMP2IN}}$	$R_L = 330 \Omega$; $V_{IFAMP2IN} = 1 \text{ mV}$	6.5	8.5	10.5	dB
F	noise figure		–	13	15	dB
IP3	3rd-order intermodulation		127	130	–	dB μ V
<i>Inputs: pins IFAMP2IN and IFAMP2DEC</i>						
$V_{i(max)(p)}$	maximum input voltage (peak value)	1 dB compression point of IF amplifier 2 output voltage (peak value)	500	–	–	mV
$V_{i(n)(eq)}$	equivalent input noise voltage	$R_{gen} = 330 \Omega$; $R_L = 330 \Omega$	–	10	13	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
R_i	input resistance		270	330	390	Ω
C_i	input capacitance		–	5	7	pF
<i>Output: pin IFAMP2OUT</i>						
$V_{o(max)(p)}$	maximum output voltage (peak value)		1.2	1.5	–	V
R_o	output resistance		270	330	390	Ω
C_o	output capacitance		–	5	7	pF
FM demodulator and level detector; see Figs 6 and 7						
FM DEMODULATOR						
<i>FM limiter inputs: pins FMLIMIN and FMLIMDEC</i>						
R_i	input resistance		10	12	14	k Ω
C_i	input capacitance		–	5	7	pF
$V_{start(lim)(rms)}$	start of limiting of RDS MPX output voltage (RMS value)	$\alpha_{AF} = -3 \text{ dB}$; $R_{gen} = 165 \Omega$	–	10	15	μ V
$V_{o(sens)(rms)}$	sensitivity for RDS MPX output voltage (RMS value)	$\Delta f = 22.5 \text{ kHz}$; $f_{mod} = 1 \text{ kHz}$; de-emphasis = 75 μ s $R_{gen} = 165 \Omega$; (S+N)/N = 26 dB (S+N)/N = 46 dB	–	10	15	μ V
			–	50	75	μ V

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$V_{\text{start(lim)(rms)}}$	start of limiting of FM MPX output voltage (RMS value)	$\alpha_{\text{AF}} = -3 \text{ dB}$; $R_{\text{gen}} = 165 \Omega$	–	10	15	μV
$V_{\text{o(sens)(rms)}}$	sensitivity for FM MPX output voltage (RMS value)	$\Delta f = 22.5 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; de-emphasis = $75 \mu\text{s}$ $R_{\text{gen}} = 165 \Omega$; (S+N)/N = 26 dB (S+N)/N = 46 dB	–	10	15	μV
			–	50	75	μV
<i>RDS MPX output: pin RDSMPX</i>						
$V_{\text{o(rms)}}$	RDS MPX output voltage (RMS value)	$V_{\text{FMLIMIN}} = 20 \mu\text{V}$ to 1 V $\Delta f = 5 \text{ kHz}$; $f_{\text{mod}} = 57 \text{ kHz}$ $\Delta f = 22.5 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$	40 180	50 230	60 280	mV mV
(S+N)/N	maximum signal plus noise-to-noise ratio	$\Delta f = 22.5 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; de-emphasis = $75 \mu\text{s}$; $V_{\text{FMLIMIN}} = 10 \text{ mV}$	70	73	–	dB
THD	total harmonic distortion	$\Delta f = 75 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; de-emphasis = $75 \mu\text{s}$; $V_{\text{FMLIMIN}} = 200 \mu\text{V}$ to 800 mV	–	0.4	0.75	%
α_{AM}	AM suppression $\frac{V_{\text{o(rms)}}}{V_{\text{o(AM)(rms)}}$	FM: $\Delta f = 22.5 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; AM: $m = 0.3$; $f_{\text{mod}} = 1 \text{ kHz}$; de-emphasis = $75 \mu\text{s}$ $V_{\text{FMLIMIN}} = 30$ to $70 \mu\text{V}$ $V_{\text{FMLIMIN}} = 70$ to $500 \mu\text{V}$ $V_{\text{FMLIMIN}} = 500 \mu\text{V}$ to 300 mV $V_{\text{FMLIMIN}} = 300 \text{ mV}$ to 1 V	20 30 50 30	30 40 60 40	– – – –	dB dB dB dB
$I_{\text{o(max)}}$	maximum output current		–	–	100	μA
R_{o}	output resistance		–	–	500	Ω
R_{L}	load resistance		20	–	–	k Ω
C_{L}	load capacitance		–	–	50	pF
B	bandwidth	$C_{\text{L}} = 0$; $R_{\text{L}} > 20 \text{ k}\Omega$	200	300	–	kHz
PSRR	power supply ripple rejection	$f_{\text{ripple}} = 100 \text{ Hz}$ to 20 kHz	–	40	–	dB
<i>FM MPX output: pin FMMPX</i>						
$V_{\text{o(rms)}}$	FM MPX output voltage (RMS value)	$\Delta f = 22.5 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; $V_{\text{FMLIMIN}} = 20 \mu\text{V}$ to 1 V	180	230	280	mV
		data byte 4 = X010X110; $\Delta f = 1.5 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; $V_{\text{FMLIMIN}} = 20 \mu\text{V}$ to 1 V; weather band mode	150	230	310	mV
(S+N)/N	maximum signal plus noise-to-noise ratio	$\Delta f = 22.5 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; de-emphasis = $75 \mu\text{s}$; $V_{\text{FMLIMIN}} = 10 \text{ mV}$	66	70	–	dB

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
THD	total harmonic distortion	$\Delta f = 75 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; de-emphasis = $75 \mu\text{s}$; $V_{\text{FMLIMIN}} = 200 \mu\text{V}$ to 800 mV	–	0.4	0.75	%
α_{AM}	AM suppression $\frac{V_{\text{o(rms)}}}{V_{\text{o(AM)(rms)}}$	FM: $\Delta f = 22.5 \text{ kHz}$; $f_{\text{mod}} = 1 \text{ kHz}$; AM: $m = 0.3$; $f_{\text{mod}} = 1 \text{ kHz}$; de-emphasis = $75 \mu\text{s}$ $V_{\text{FMLIMIN}} = 30$ to $70 \mu\text{V}$ $V_{\text{FMLIMIN}} = 70$ to $500 \mu\text{V}$ $V_{\text{FMLIMIN}} = 500 \mu\text{V}$ to 300 mV $V_{\text{FMLIMIN}} = 300 \text{ mV}$ to 1 V	20 30 50 30	30 40 60 40	– – – –	dB dB dB dB
$I_{\text{o(max)}}$	maximum output current		–	–	100	μA
$R_{\text{o(max)}}$	maximum output resistance		–	–	500	Ω
$R_{\text{L(min)}}$	minimum load resistance		20	–	–	$\text{k}\Omega$
$C_{\text{L(max)}}$	maximum load capacitance		–	–	50	pF
B	bandwidth	$C_{\text{L}} = 0$; $R_{\text{L}} > 20 \text{ k}\Omega$	200	–	–	kHz
PSRR	power supply ripple rejection	$f_{\text{ripple}} = 100 \text{ Hz}$ to 20 kHz	–	40	–	dB
t_{sw}	AM to FM switching time	$V_{\text{FMLIMIN}} = 100 \mu\text{V}$	–	100	150	ms
FM MPX SOFT MUTE						
<i>Soft mute</i>						
α_{mute}	mute attenuation	$V_{\text{FMLIMIN}} = 10 \text{ mV}$; $V_{\text{TMUTE}} = 300 \text{ mV}$ data byte 7 = XX000100 data byte 7 = XX001100 data byte 7 = XX010100 data byte 7 = XX011100 data byte 7 = XX100100 data byte 7 = XX101100 data byte 7 = XX110100 data byte 7 = XX111100	– – – – – – – –	7 8 9.5 11 14 18 25 39	– – – – – – – –	dB dB dB dB dB dB dB dB
$\alpha_{\text{mute(off)}}$	mute off	$V_{\text{FMLIMIN}} = 10 \text{ mV}$; $V_{\text{TMUTE}} = 300 \text{ mV}$; data byte 7 = XXXXX111	–1	0	+1	dB
$V_{\text{offset(DC)}}$	DC offset created by soft mute on pin FMMPX $\Delta V = V_{\text{muted}} - V_{\text{notmuted}}$		–150	–	+150	mV

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<i>Input: pin TMUTE</i>						
V_{TMUTE}	mute voltage	$V_{FMLIMIN} = 0$ to 1 V	0	–	5	V
V_{offset}	voltage offset with V_{level}	$V_{FMLIMIN} = 0$ to 1 V	–200	–	+200	mV
I_{dch}	discharge current	$V_{level} < V_{TMUTE}$	2.5	3.5	4.5	μ A
I_{ch}	charge current	$V_{level} > V_{TMUTE}$	–4.5	–3.5	–2.5	μ A
$V_{TMUTE(start)}$	mute start voltage	$\alpha_{mute} = 3$ dB; $V_{FMLIMIN} = 10$ mV data byte 7 = XX111000 data byte 7 = XX111001 data byte 7 = XX111010 data byte 7 = XX111011 data byte 7 = XX111100 data byte 7 = XX111101 data byte 7 = XX111110	300 400 450 520 580 640 680	450 550 630 720 800 890 980	600 700 810 920 1020 1140 1280	mV mV mV mV mV mV mV
<i>Output: pin V_{level}</i>						
R_o	output resistance of level output		–	–	500	Ω
RR	ripple rejection of level output	$f_{ripple} = 100$ Hz; $V_{DDA1(ripple)} = 100$ mV (RMS)	–	40	–	dB
RDS update; see Figs 10 and 11						
<i>RDS mute</i>						
α_{mute}	muting depth	data byte 3: bit 7 = 1 (mute)	60	80	–	dB
$V_{offset(DC)}$	DC offset during RDS update mute on pin FMMPX $\Delta V = V_{muted} - V_{notmuted}$		–30	–	+30	mV
<i>Capacitor: pin TRDSMUTE</i>						
$V_{TRDSMUTE}$	voltage on pin TRDSMUTE	no mute	–	5.7	–	V
		mute	–	1.1	–	V
I_{dch}	discharge current	$V_o = 3$ V; data byte 3: bit 7 = 1	24	32	38	μ A
I_{ch}	charge current	$V_o = 3$ V; data byte 3: bit 7 = 0	–38	–32	–24	μ A
I_{LI}	input leakage current	mute	–10	–	+10	nA
<i>Output: pin AFHOLD</i>						
$I_{sink(max)}$	maximum sink current	after first bus transmission with data byte 1: bit 7 = 1 (start of RDS update); $V_o = 0.5$ V	1.0	1.2	1.4	mA
<i>Output: pin AFSAMPLE</i>						
$I_{sink(max)}$	maximum sink current	no RDS update in progress; $V_o = 0.5$ V	1.0	1.2	1.4	mA