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TFF11092HN

Low phase noise LO generator for VSAT applications

Rev. 1 — 28 March 2013

Product data sheet

1. General description

The TFF11092HN is a K_u band frequency generator intended for low phase noise Local Oscillator (LO) circuits for K_u band VSAT transmitters and transceivers. The specified phase noise complies with IESS-308 from Intelsat.

CAUTION



This device is sensitive to ElectroStatic Discharge (ESD). Therefore care should be taken during transport and handling.

2. Features and benefits

- Phase noise compliant with IESS-308 (Intelsat) in combination with appropriate source
- LO generator with VCO range from 9.03 GHz to 9.22 GHz
- Input signal 35 MHz to 576 MHz
- Divider settings 16, 32, 64, 128 or 256
- Output level –3 dBm; stability ±2 dB
- Third or fourth order PLL
- Internally stabilized voltage references for loop filter

3. Applications

VSAT up converters

Local oscillator signal generation

4. Quick reference data

Table 1. Quick reference data Operating conditions of Table 10 apply.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{CC}	supply voltage		3.0	3.3	3.6	V
I _{CC}	supply current		-	100	130	mA
f _{o(RF)}	RF output frequency		9.03	-	9.22	GHz
Φn(synth)	synthesizer phase noise	divider value = 64; at 100 kHz offset; reference phase noise is -149 dBc/Hz at 100 kHz offset	-	-97	-92	dBc/Hz
RL _{out}	output return loss	measured at demo board and de-embedded to footprint	-	-10	-	dB
$\alpha_{\text{sup(sp)ref}}$	reference spurious suppression	measured at divider value = 256	-	-	-70	dBc



Low phase noise LO generator for VSAT applications

5. Ordering information

Table 2. Ordering information

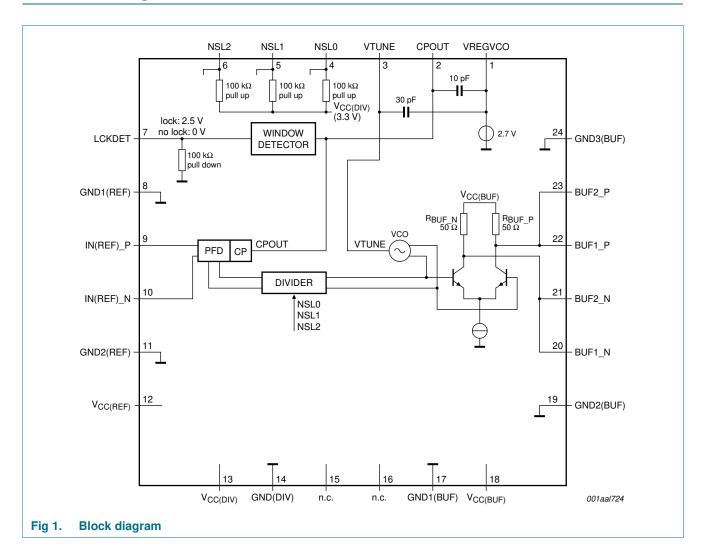
Type number	Package							
	Name	Description	Version					
TFF11092HN	HVQFN24	plastic thermal enhanced very thin quad flat package; no leads; 24 terminals; body $4\times4\times0.85$ mm	SOT616-1					

6. Marking

Table 3. Marking codes

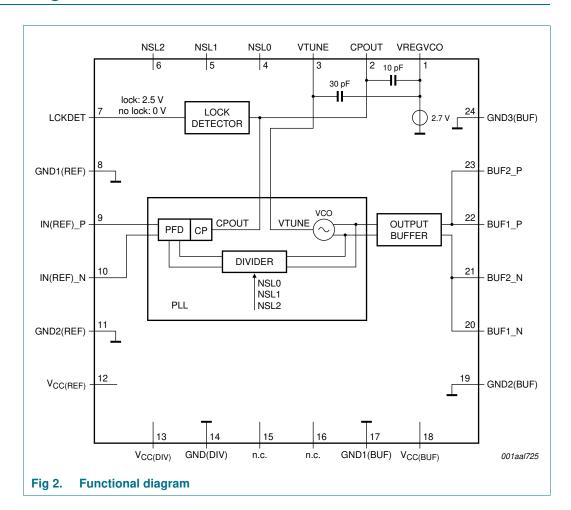
Type number	Marking code
TFF11092HN	T092

7. Block diagram



Low phase noise LO generator for VSAT applications

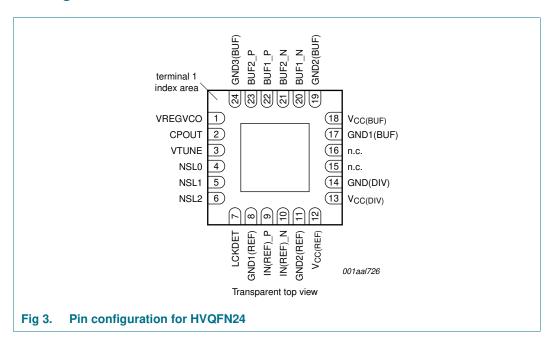
8. Functional diagram



Low phase noise LO generator for VSAT applications

9. Pinning information

9.1 Pinning



9.2 Pin description

Table 4. Pin description

		real Press
Symbol	Pin	Description
VREGVCO	1	Regulated output voltage for VCO loop filter. Connect loop filter to this pin.
CPOUT	2	Charge pump output.
VTUNE	3	Tuning voltage for VCO.
NSL0	4	Divider setting, LSB. Leave open for "1", connect to GND for "0". See <u>Table 8</u> .
NSL1	5	Divider setting. Leave open for "1", connect to GND for "0". See <u>Table 8</u> .
NSL2	6	Divider setting, MSB. Leave open for "1", connect to GND for "0". See <u>Table 8</u> .
LCKDET	7	Lock detect. Lock = 2.5 V ; out of lock = 0 V . See <u>Table 6</u> .
GND1(REF)	8	Ground for REF input. Connect this pin to the exposed diepad landing.
IN(REF)_P	9	Reference signal, non-inverting input. Couple this AC to the source.
IN(REF)_N	10	Reference signal, inverting input. Couple this AC to the source.
GND2(REF)	11	Ground for REF input. Connect this pin to the exposed diepad landing.
V _{CC(REF)}	12	Supply of the internal regulated voltages. Decouple this pin against GND2(REF) (pin 11).
$V_{CC(DIV)}$	13	Supply of the divider and PFD/CP. Decouple this pin against GND(DIV) (pin 14).
GND(DIV)	14	Ground of the divider. Connect this pin to the exposed diepad landing.
n.c.	15	not connected
n.c.	16	not connected
GND1(BUF)	17	Ground for RF output. Connect this pin to the exposed diepad landing.

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Table 4. Pin description ... continued

Symbol	Pin	Description
V _{CC(BUF)}	18	Supply voltage for the RF output buffer. Decouple this pin against GND2(BUF) (pin 19).
GND2(BUF)	19	Ground for RF output. Connect this pin to the exposed diepad landing.
BUF1_N	20	RF output.
BUF2_N	21	RF output.
BUF1_P	22	RF output.
BUF2_P	23	RF output.
GND3(BUF)	24	Ground for RF output. Connect this pin to the exposed diepad landing.

10. Functional description

The TFF11092HN consists of the following blocks:

- PLL
- · Output buffer
- · Lock detector
- · Reference input
- · Divider settings

The functionality of the blocks will be discussed below.

10.1 PLL

The PLL is formed by the VCO, DIVIDER (possible settings: 16, 32, 64, 128 and 256 (see <u>Table 8</u>)) and a PFD/CP. The tune voltage is referred to the band gap regulated voltage: VREGVCO (pin 1).

The loop filter can be set to type 2 or type 3. If a type 2 filter is used, the pins CPOUT (pin 2) and VTUNE (pin 3) must be interconnected. A 10 pF capacitor is placed internally between pins CPOUT (pin 2) and VREGVCO (pin 1), and a 30 pF capacitor is placed between pins VTUNE (pin 3) and VREGVCO (pin 1). See Figure 4 and Figure 4 and Figure 5. Values for the loop filter components are given in Table 5.

The VCO input voltage range is between 0.1 and 0.9 V_{O(reg)VCO}.

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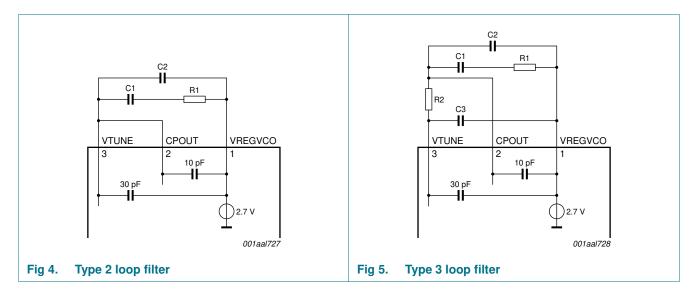


Table 5. Component values used for characterization

f _{i(ref)}	Divider value	C1	C2	C3	R1	R2
(MHz)		(nF)	(pF)	(pF)	(Ω)	(Ω)
35.273 to 36.016	256	27	82	33	470	560
70.547 to 72.031	128	18	82	33	330	560
141.094 to 144.063	64	18	120	33	270	560
282.188 to 288.125	32	33	270	33	120	560
564.375 to 576.250	16	68	560	33	68	560

10.2 Output buffer

The output consists of a differential pair with 50 Ω collector resistors R_{BUF_P} and R_{BUF_N}. If only one output is used, terminate the non used output with the same impedance as the load (see Figure 8)

10.3 Lock detector

The lock detector is the output of a window detector. The window detector compares the output voltage over the charge pump. This voltage is identical to VTUNE when a type 2 loop filter is used (see <u>Figure 4</u>). In case of a type 3 loop filter this voltage is filtered by R2/C3 (see <u>Figure 5</u>). Due to this filtering the attack and decay time will decrease.

The lower window detector threshold voltage is 7 % of the output voltage on VREGVCO (pin 1), the upper window detector threshold voltage is 93 % of the output voltage on VREGVCO (pin 1). The hysteresis is 0.1 V. The output is 2.5 V CMOS compliant. The values are shown in Table 6. The timing diagram is shown in Figure 6.

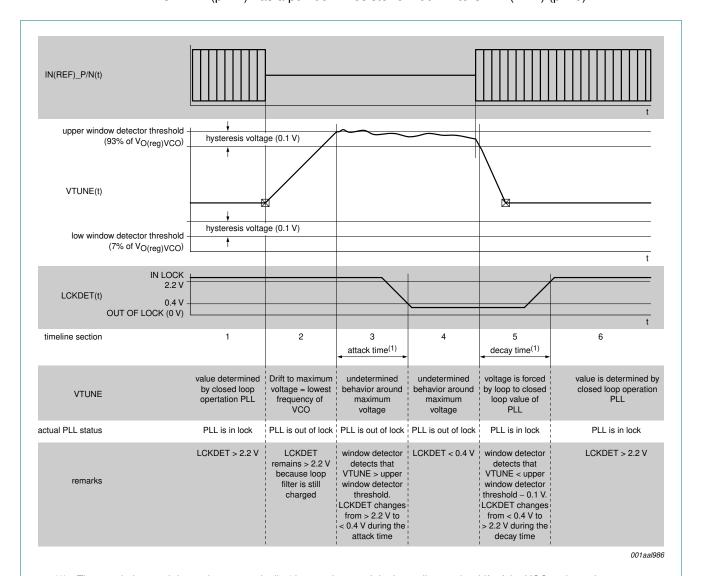
At start-up the LCKDET (pin 7) will be LOW until the circuit has acquired lock.

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Table 6. Logical value and physical value for lock detect (LCKDET)

Logical value	Physical value	Lock detect state
0	0 V	out of lock
1	2.5 V	lock

LCKDET (pin 7) has a pull-down resistor of 100 k Ω to GND1(REF) (pin 8).



(1) The attack time and decay time are typically 10 μs and are mainly depending on the drift of the VCO tuning voltage.

Fig 6. Timing diagram lock detector

10.4 Reference input (IN(REF)_P, IN(REF)_N)

The reference input is a differential pair and is internally biased. The input is high ohmic. The input signal must be AC coupled. If used in a single ended mode, the not used input must be terminated with the same impedance as the driving source.

An example of the differential source and two single ended loads are shown in <u>Figure 7</u>. An example of a single ended application is shown in <u>Figure 8</u>.

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Note that the phase noise of the output signal is also determined by the phase noise of the reference signal. The reference frequency range is equal to the output frequency / division value. Note that the output frequency is guaranteed from 8.20 GHz to 8.60 GHz.

10.5 Divider settings (NSL2, NSL1, NSL0)

The divider can be set to 16, 32, 64, 128 and 256 (See <u>Table 8</u>). The logic levels for NSL0 (pin 4), NSL1 (pin 5) and NSL2 (pin 6) are given in <u>Table 7</u>.

The pins have a pull-up resistor of 100 k Ω to $V_{CC(DIV)}$ (pin 13).

The device is only guaranteed when NSL2, NSL1 and NSL0 are predefined at start-up (no change of divider value is allowed during operation).

Table 7. Logical and physical value for divider setting (NSL2, NSL1, NSL0)

Logical value	Physical value
0	GND
1	open or V _{CC}

The truth table is shown in Table 8.

Table 8. Divider setting as function of NSL2, NSL1 and NSL0

Setting number	NSL2	NSL1	NSL0	Divider value
0	0	0	0	16
1	0	0	1	32
2	0	1	0	64
3	0	1	1	128
4	1	0	0	256
5	1	0	1	[1]
6	1	1	0	[1]
7	1	1	1	[1]

^[1] Test mode, divider output will be disabled.

11. Limiting values

Table 9. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V_{I}	input voltage	on pin NSL0	-0.5	+5	V
		on pin NSL1	-0.5	+5	V
		on pin NSL2	-0.5	+5	V
		on pin IN(REF)_P	-0.5	+5	V
		on pin IN(REF)_N	-0.5	+5	V
		on pin V _{CC(REF)}	-0.5	+5	V
		on pin V _{CC(DIV)}	-0.5	+5	V
		on pin V _{CC(BUF)}	-0.5	+5	V

Low phase noise LO generator for VSAT applications

Table 9. Limiting values ...continued

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
P_{i}	input power	on pin IN(REF)_P	-4	+10	dBm
		on pin IN(REF)_N	-4	+10	dBm
Tj	junction temperature		-40	+125	°C
T _{stg}	storage temperature		-40	+125	°C
V _{ESD}	electrostatic discharge voltage	Human Body Model (HBM); According JEDEC standard 22-A114E	-	2.5	kV
		Charged Device Model (CDM); According to JEDEC standard 22-C101B	-	1	kV

12. Recommended operating conditions

Table 10. Operating conditions

NSL0 (pin 4), NSL1 (pin 5) and NSL2 (pin 6) not changed during operation. Loop filter component values as depicted in Table 5 are used.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
T_{amb}	ambient temperature			-40	+25	+85	°C
Z_0	characteristic impedance			-	50	-	Ω
$\phi_{n(\text{ref})}$	reference phase noise	divider value = 16	[1]	-	-	-134	dBc/Hz
		divider value = 32	[1]	-	-	-143	dBc/Hz
		divider value = 64	[1]	-	-	-149	dBc/Hz
		divider value = 128	[1]	-	-	-150	dBc/Hz
		divider value = 256	[1]	-	-	-151	dBc/Hz
f _{i(ref)}	reference input frequency	$f_{i(ref)} = f_{o(RF)} / divider value$		35	-	576	MHz
P _{i(ref)}	reference input power			-10	-	0	dBm

^[1] Required reference phase noise is set 10 dB below equivalent input phase noise.

13. Thermal characteristics

Table 11. Thermal characteristics

Symbol	Parameter	Conditions	Тур	Unit
$R_{th(j-sp)}$	thermal resistance from junction to solder point		25	K/W

Low phase noise LO generator for VSAT applications

14. Characteristics

Table 12. Characteristics

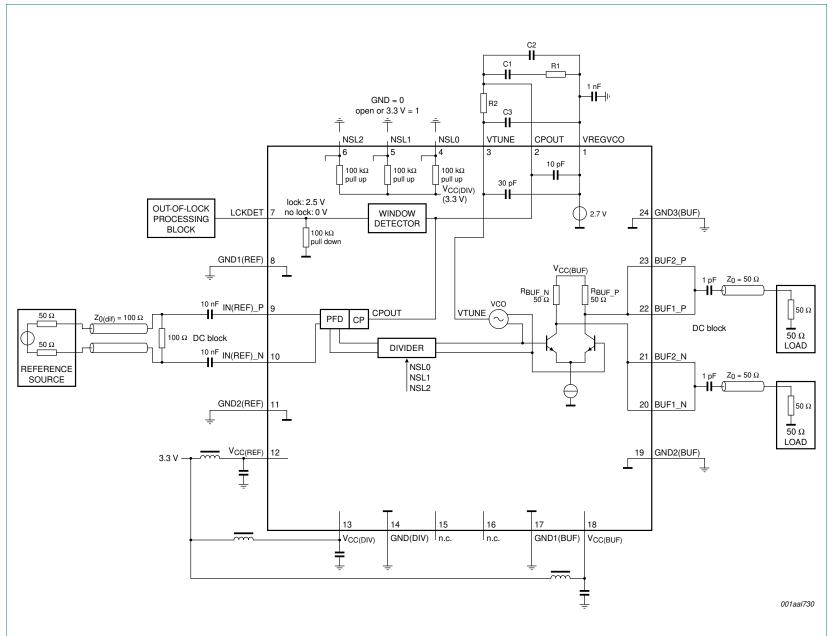
Operating conditions of Table 10 apply.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V_{CC}	supply voltage			3.0	3.3	3.6	V
I _{CC}	supply current			-	100	130	mA
PLL							
f _{o(RF)}	RF output frequency			9.03	-	9.22	GHz
V _{O(reg)} VCO	VCO regulator output voltage			2.5	2.7	2.9	V
I _{cp}	charge pump current			-	1	-	mA
K _O	VCO steepness		[1]	-	0.48	-	GHz/V
Φn(VCO)	VCO phase noise	at 10 MHz offset		-	-130	-	dBc/Hz
Ψn(synth)	synthesizer phase noise	divider value = 64; at 100 kHz offset; reference phase noise is -149 dBc/Hz at 100 kHz offset		-	-97	-92	dBc/Hz
Output buff	fer						
Po	output power	measured single ended	[2]	-5	-	-1	dBm
RL _{out}	output return loss	measured at demo board and de-embedded to footprint		-	-10	-	dB
$\alpha_{\text{sup(sp)ref}}$	reference spurious suppression	measured at divider value = 256		-	-	-70	dBc
$\alpha_{H(LO)}$	LO harmonic rejection			-	-10	-	dBc
Lock detec	tor						
V _{OL}	LOW-level output voltage	I _O = 1 mA		-	-	0.4	V
V _{OH}	HIGH-level output voltage	$I_O = -1 \text{ mA}$		2.2	-	-	V
R _{pd}	pull-down resistance			70	100	130	kΩ
Divider sett	ting (NSL0, NSL1, NSL2)						
R _{pu}	pull-up resistance			70	100	130	kΩ
V_{IL}	LOW-level input voltage			-	-	8.0	V
V _{IH}	HIGH-level input voltage			2.0	-	-	V

^[1] The typical ratio of the maximum K_{O} in relation to the minimum K_{O} is 1.25.

^[2] Output stage is a differential pair with 50 Ω collector impedances. Output power is measured per output pin for the fundamental tone only. Output is DC coupled and is AC coupled in on-board.

15. **Application information**



Application diagram with differential source for IN(REF) and both outputs driving a load, loop filter is type 3 Fig 7.

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Rev. 1 —

Product data sheet

Product data sheet C2 C1 R1 $\dashv \ddot{\vdash} \vdash$ GND = 0R2 open or 3.3 V = 1 СЗ NSL2 NSL1 NSL0 VTUNE CPOUT VREGVCO 10 pF 100 kΩ 100 kΩ 100 kΩ pull up pull up pull up 30 pF V_{CC(DIV)} (3.3 V) lock: 2.5 V OUT-OF-LOCK no lock: 0 V LCKDET WINDOW DETECTOR 2.7 V 24 GND3(BUF) **PROCESSING BLOCK** All information provided in this document is subject to legal disclaimers 100 kΩ pull down GND1(REF) 8 23 BUF2_P V_{CC(BUF)} Rev. 1 — 1 pF $Z_0 = 50 \Omega$ RBUF_N 50 Ω R_{BUF}P 50 Ω 10 nF IN(REF)_P 9 $Z_0 = 50 \Omega$ VCO 50 Ω 22 BUF1_P 50 Ω VTUNE **CPOUT** PFD CP 28 March 2013 ± 50 Ω LOAD DC block 51 Ω DC block **DIVIDER** IN(REF)_N 21 BUF2_N REFERENCE SOURCE NSL0 NSL1 NSL2 not used input terminated with 24 Ω same impedance not used output GND2(REF) 11 20 BUF1_N terminated with same impedance V_{CC(REF)} 12 GND2(BUF) 3.3 V -16 GND1(BUF) V_{CC(BUF)} GND(DIV) V_{CC(DIV)} © NXP B.V. 2013. All rights reserved 001aal731

Application diagram with single ended source for IN(REF) and single ended load, loop filter is type 3 Fig 8.

TFF11092HN

16. Package outline

HVQFN24: plastic thermal enhanced very thin quad flat package; no leads; 24 terminals; body $4 \times 4 \times 0.85$ mm

SOT616-1

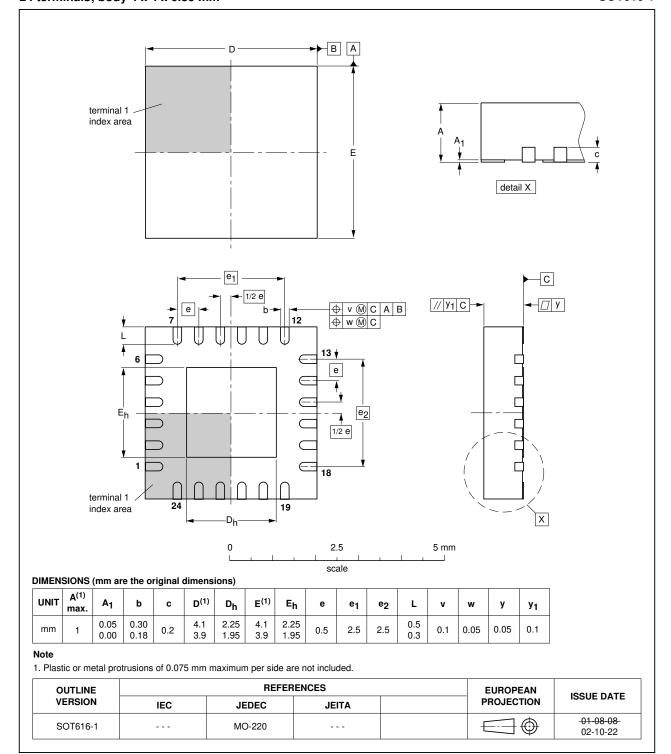


Fig 9. Package outline SOT616-1 (HVQFN24)

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17. Abbreviations

Table 13. Abbreviations

Acronym	Description
CMOS	Complementary Metal Oxide Semiconductor
CP	Charge Pump
K _u band	K-under band
LSB	Least Significant Bit
MSB	Most Significant Bit
PFD	Phase Frequency Detector
PLL	Phase-Locked Loop
VCO	Voltage Controlled Oscillator
VSAT	Very Small Aperture Terminal

18. Revision history

Table 14. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
TFF11092HN v.1	20130328	Product data sheet	-	-

Low phase noise LO generator for VSAT applications

19. Legal information

19.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
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Product data sheet

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