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Low Dropout Linear Voltage Regulator

TLS850F0TAV50

TLS850F0TAV50

Linear Voltage Regulator

Data Sheet

Rev. 1.0, 2015-07-24

Automotive Power



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Low Dropout Linear Voltage Regulator

TLS850F0TAV50





1 Overview

Features

- Wide Input Voltage Range from 3.0 V to 40 V
- Fixed Output Voltage 5 V
- Output Voltage Precision ≤ ±2 %
- Output Current Capability up to 500 mA
- Ultra Low Current Consumption typ. 40 μA
- Very Low Dropout Voltage typ. 70 mV@100 mA
- Stable with Ceramic Output Capacitor of 1 μF
- · Delayed Reset at Power-On: 16.5 ms
- Adjustable Reset Threshold down to 2.50V
- Watchdog with fixed timing and current dependent deactivation: 96 ms, Activated at $I_{\rm Q}$ > 5.5 mA
- Enable, Undervoltage Reset, Overtemperature Shutdown
- · Output Current Limitation
- Wide Temperature Range
- Green Product (RoHS compliant)
- AEC Qualified

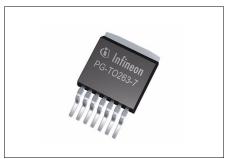


Figure 1 PG-TO263-7



Overview

Functional Description

The TLS850F0TAV50 is a high performance very low dropout linear voltage regulator for 5 V supply in a PG-TO263-7 package.

With an input voltage range of 3 V to 40 V and very low quiescent of only 40 μ A, these regulators are perfectly suitable for automotive or any other supply systems connected to the battery permanently. The TLS850F0TAV50 provides an output voltage accuracy of 2 % and a maximum output current up to 500 mA.

The new loop concept combines fast regulation and very good stability while requiring only one small ceramic capacitor of 1 μ F at the output. At currents below 100 mA the device will have a very low typical dropout voltage of only 70 mV. The operating range starts already at input voltages of only 3 V (extended operating range). This makes the TLS850F0TAV50 also suitable to supply automotive systems that need to operate during cranking condition.

The device can be switched on and off by the Enable feature as described in Chapter 5.5.

The output voltage is supervised by the Reset feature, including Undervoltage Reset, delayed Reset at Power-On and an adjustable lower Reset Threshold, more details can be found in **Chapter 5.7**.

In addition, a Watchdog circuit with fixed timing is integrated to monitor the microcontroller's operation.

Internal protection features like output current limitation and overtemperature shutdown are implemented to protect the device against immediate damage due to failures like output short circuit to GND, over-current and over-temperatures.

Choosing External Components

An input capacitor $C_{\rm I}$ is recommended to compensate line influences. The output capacitor $C_{\rm Q}$ is necessary for the stability of the regulating circuit. TLS850F0TAV50 is designed to be also stable with low ESR ceramic capacitors.

Туре	Package	Marking
TLS850F0TAV50	PG-TO263-7	850F0V50

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Block Diagram

2 Block Diagram

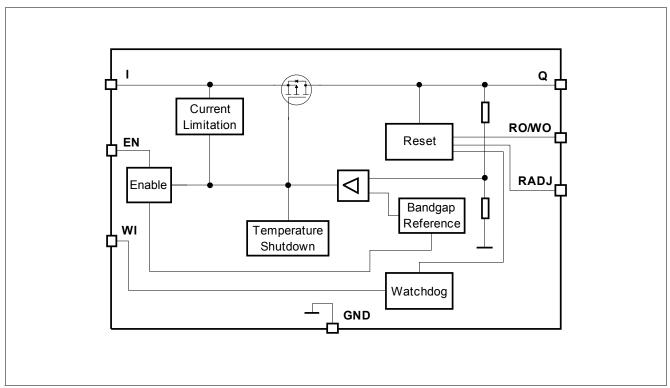


Figure 2 Block Diagram TLS850F0TAV50



Pin Configuration

3 Pin Configuration

3.1 Pin Assignment TLS850F0TAV50

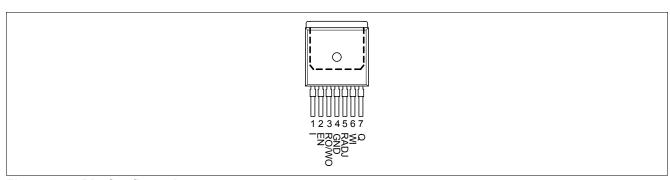


Figure 3 Pin Configuration

3.2 Pin Definitions and Functions TLS850F0TAV50

Pin	Symbol	Function
1	I	Input It is recommended to place a small ceramic capacitor (e.g. 100 nF) to GND, close to the IC terminals, in order to compensate line influences. See also Chapter 6.2.1
2	EN	Enable (integrated pull-down resistor) Enable the IC with high level input signal; Disable the IC with low level input signal;
3	RO/WO	Reset Output / Watchdog Output (intergrated pull-up resistor to Q) Open collector output; Leave open if the reset and watchdog function are not needed
4	GND	Ground
5	RADJ	Reset Threshold Adjustment Connect to GND to use standard value; Connect an external voltage divider to adjust reset threshold
6	WI	Watchdog Input (integrated pull-down resistor) Serve Watchdog with trigger input signal (usable for microcontroller monitoring)
7	Q	Output Voltage Connect output capacitor $C_{\rm Q}$ to GND close to the IC's terminals, respecting the values specified for its capacitance and ESR in "Functional Range" on Page 8
Heat Slug	-	Heat Slug Connect to heatsink area; Connect to GND



General Product Characteristics

4 General Product Characteristics

4.1 Absolute Maximum Ratings

Table 1 Absolute Maximum Ratings¹⁾

 T_i = -40 °C to +150 °C; all voltages with respect to ground (unless otherwise specified)

Parameter	Symbol		Value	S	Unit	Note / Test Condition	Number
		Min.	Тур.	Max.			
Input I, Enable EN							
Voltage	$V_{\rm I},V_{\rm EN}$	-0.3	_	45	V	_	P_4.1.1
Output Q, Reset/Watchdog Output RO	/WO						
Voltage	$V_{\rm Q},V_{\rm RO/WO}$	-0.3	_	7	V	_	P_4.1.3
Watchdog Input WI, Reset Threshold	Adjustment	RADJ		-	-	-	'
Voltage	V_{WI} , V_{RADJ}	-0.3	_	7	٧	_	P_4.1.5
Temperatures							
Junction Temperature	$T_{\rm j}$	-40	_	150	°C	_	P_4.1.7
Storage Temperature	T_{stg}	-55	_	150	°C	_	P_4.1.8
ESD Absorption							
ESD Susceptibility to GND	V_{ESD}	-2	_	2	kV	²⁾ HBM	P_4.1.9
ESD Susceptibility to GND	V_{ESD}	-500	_	500	٧	3) CDM	P_4.1.10
ESD Susceptibility Pin 1, 7 (corner pins) to GND	$V_{\mathrm{ESD1,7}}$	-750	-	750	V	3) CDM	P_4.1.12

¹⁾ Not subject to production test, specified by design.

Note:

- 1. Stresses above the ones listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
- 2. Integrated protection functions are designed to prevent IC destruction under fault conditions described in the data sheet. Fault conditions are considered as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.

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²⁾ ESD susceptibility, HBM according to ANSI/ESDA/JEDEC JS001 (1.5 k Ω , 100 pF)

³⁾ ESD susceptibility, Charged Device Model "CDM" according JEDEC JESD22-C101



General Product Characteristics

4.2 Functional Range

Table 2 Functional Range

 T_i = -40 °C to +150 °C; all voltages with respect to ground (unless otherwise specified)

Parameter	Symbol	V	alues		Unit	Note /	Number
		Min.	Тур.	Max.		Test Condition	
Input Voltage Range	V_1	$V_{\rm Q,nom}$ + $V_{\rm dr}$	_	40	V	1) _	P_4.2.1
Extended Input Voltage Range	$V_{I,ext}$	3.0	_	40	V	2) _	P_4.2.3
Enable Voltage Range	V_{EN}	0	_	40	V	_	P_4.2.5
Output Capacitor's	C_{Q}	1	_	_	μF	3)4) _	P_4.2.6
Requirements for Stability							
ESR	$ESR(C_{Q})$	_	_	100	Ω	3) _	P_4.2.7
Junction Temperature	T_{i}	-40	_	150	°C	_	P_4.2.9

¹⁾ Output current is limited internaly and depends on the input voltage, see Electrical Characteristics for more details.

Note: Within the functional or operating range, the IC operates as described in the circuit description. The electrical characteristics are specified within the conditions given in the Electrical Characteristics table.

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²⁾ When $V_{\rm l}$ is between $V_{\rm l,ext,min}$ and $V_{\rm Q,nom}$ + $V_{\rm dr}$, $V_{\rm Q}$ = $V_{\rm l}$ - $V_{\rm dr}$. When $V_{\rm l}$ is below $V_{\rm l,ext,min}$, $V_{\rm Q}$ can drop down to 0 V.

³⁾ Not subject to production test, specified by design.

⁴⁾ The minimum output capacitance requirement is applicable for a worst case capacitance tolerance of 30%



General Product Characteristics

4.3 Thermal Resistance

Note: This thermal data was generated in accordance with JEDEC JESD51 standards. For more information, go to www.jedec.org.

Table 3 Thermal Resistance

Parameter	Symbol		Value	s	Unit	Note / Test Condition	Number
		Min.	Тур.	Max.			
Package Version PG-TO26	63-7	-1		"		•	
Junction to Case	R_{thJC}	_	3	_	K/W	1) _	P_4.3.6
Junction to Ambient	R_{thJA}	_	21	_	K/W	¹⁾²⁾ 2s2p board	P_4.3.7
Junction to Ambient	R_{thJA}	_	75	_	K/W	1)3) 1s0p board, footprint only	P_4.3.8
Junction to Ambient	R_{thJA}	-	42	-	K/W	1)3) 1s0p board, 300 mm² heatsink area on PCB	P_4.3.9
Junction to Ambient	R_{thJA}	_	34	-	K/W	1)3) 1s0p board, 600 mm ² heatsink area on PCB	P_4.3.10

¹⁾ Not subject to production test, specified by design

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²⁾ Specified R_{thJA} value is according to Jedec JESD51-2,-5,-7 at natural convection on FR4 2s2p board; The Product (Chip+Package) was simulated on a 76.2 x 114.3 x 1.5 mm³ board with 2 inner copper layers (2 x 70μm Cu, 2 x 35μm Cu). Where applicable a thermal via array under the exposed pad contacted the first inner copper layer.

³⁾ Specified R_{thJA} value is according to JEDEC JESD 51-3 at natural convection on FR4 1s0p board; The Product (Chip+Package) was simulated on a 76.2 × 114.3 × 1.5 mm³ board with 1 copper layer (1 x 70 μ m Cu).



5 Block Description and Electrical Characteristics

5.1 Voltage Regulation

The output voltage V_Q is divided by a resistor network. This fractional voltage is compared to an internal voltage reference and the pass transistor is driven accordingly.

The control loop stability depends on the output capacitor $C_{\rm Q}$, the load current, the chip temperature and the internal circuit design. To ensure stable operation, the output capacitor's capacitance and its equivalent series resistor (ESR) requirements given in "Functional Range" on Page 8 have to be maintained. For details, also see the typical performance graph "Output Capacitor Series Resistor ESR(CQ) versus Output Current IQ" on Page 13. As the output capacitor also has to buffer load steps, it should be sized according to the application's needs.

An input capacitor C_1 is recommended to compensate line influences. In order to block influences like pulses and HF distortion at input side, an additional reverse polarity protection diode and a combination of several capacitors for filtering should be used. Connect the capacitors close to the component's terminals.

In order to prevent overshoots during start-up, a smooth ramp up function is implemented. This ensures almost no output voltage overshoots during start-up, mostly independent from load and output capacitance.

Whenever the load current exceeds the specified limit, e.g. in case of a short circuit, the output current is limited and the output voltage decreases.

The overtemperature shutdown circuit prevents the IC from immediate destruction under fault conditions (e.g. output continuously short-circuit) by switching off the power stage. After the chip has cooled down, the regulator restarts. This leads to an oscillatory behavior of the output voltage until the fault is removed. However, junction temperatures above 150 °C are outside the maximum ratings and therefore significantly reduce the IC's lifetime.

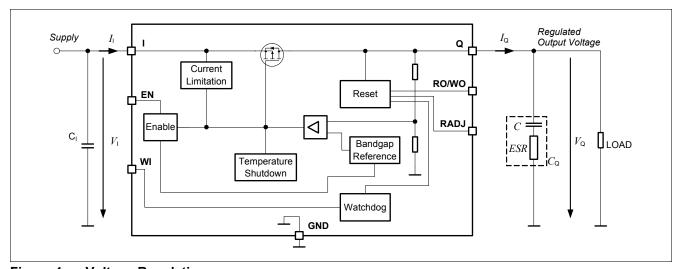


Figure 4 Voltage Regulation

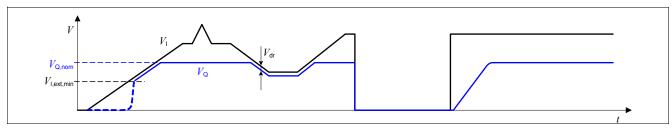


Figure 5 Output Voltage vs. Input Voltage



Table 4 Electrical Characteristics Voltage Regulator 5 V version

 $T_{\rm j}$ = -40 °C to +150 °C, $V_{\rm l}$ = 13.5 V, all voltages with respect to ground (unless otherwise specified) Typical values are given at $T_{\rm j}$ = 25 °C

Parameter	Symbol	Values			Unit	Note / Test Condition	Number
		Min.	Тур.	Max.			
Output Voltage Precision	V_{Q}	4.9	5.0	5.1	V	$0.05 \text{ mA} < I_{\text{Q}} < 500 \text{ mA}$ $5.95 \text{ V} < V_{\text{I}} < 28 \text{ V}$	P_5.1.3
Output Voltage Precision	V_{Q}	4.9	5.0	5.1	V	$0.05 \text{ mA} < I_{\text{Q}} < 200 \text{ mA}$ $5.44 \text{ V} < V_{\text{I}} < 40 \text{ V}$	P_5.1.4
Output Voltage Start-up slew rate	$\mathrm{d}V_{\mathrm{Q}}/\mathrm{dt}$	3.0	7.5	18	V/ms	$V_I > 18 \text{ V/ms}$ $C_Q = 1 \mu\text{F}$ $0.5 \text{ V} < V_Q < 4.5 \text{ V}$	P_5.1.7
Output Current Limitation	$I_{Q,max}$	501	650	1100	mA	0 V < V _Q < 4.8 V	P_5.1.9
Load Regulation steady-state	$\Delta V_{ m Q,load}$	-20	-1.5	5	mV	$I_{\rm Q}$ = 0.05 mA to 500 mA $V_{\rm I}$ = 6 V	P_5.1.11
Line Regulation steady-state	$\Delta V_{ m Q,line}$	-20	0	20	mV	$V_{\rm I}$ = 8 V to 32 V $I_{\rm Q}$ = 5 mA	P_5.1.13
Dropout Voltage $V_{dr} = V_1 - V_Q$	V_{dr}	-	175	425	mV	¹⁾ $I_{\rm Q}$ = 250 mA	P_5.1.16
Dropout Voltage $V_{dr} = V_1 - V_Q$	V_{dr}	-	70	170	mV	$I_{\rm Q}$ = 100 mA	P_5.1.17
Power Supply Ripple Rejection	PSRR	_	59	_	dB	$V_{\text{ripple}}^{(2)} = 100 \text{ Hz}$ Hz $V_{\text{ripple}}^{(2)} = 0.5 \text{ Vpp}$	P_5.1.18
Overtemperature Shutdown Threshold	$T_{ m j,sd}$	151	-	200	°C	$^{2)}$ $T_{\rm j}$ increasing	P_5.1.19
Overtemperature Shutdown Threshold Hysteresis	$T_{ m j,sdh}$	_	15	_	K	$^{2)}$ $T_{\rm j}$ decreasing	P_5.1.20

¹⁾ Measured when the output voltage $V_{\rm Q}$ has dropped 100 mV from the nominal value obtained at $V_{\rm I}$ = 13.5V

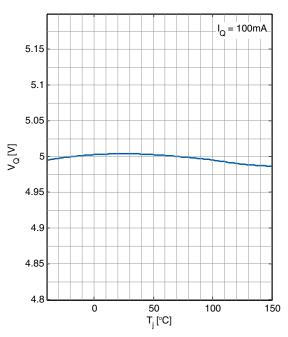
²⁾ Not subject to production test, specified by design



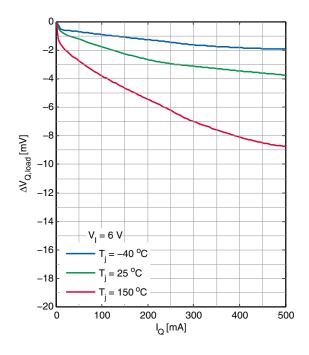
5.2 Typical Performance Characteristics Voltage Regulator

Typical Performance Characteristics

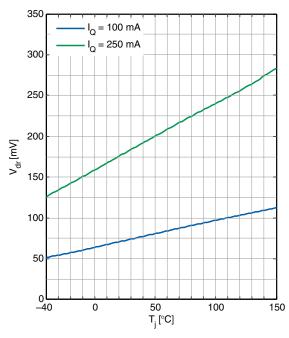
Output Voltage V_{Q} versus Junction Temperature T_{i}



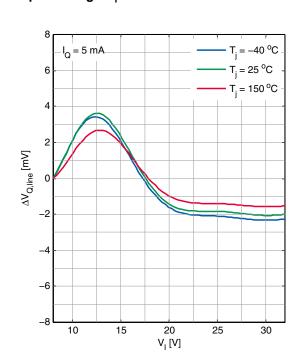
Load Regulation $\Delta V_{\rm Q,load}$ versus Output Current Change $I_{\rm Q}$



Dropout Voltage $V_{ m dr}$ versus Junction Temperature $T_{ m j}$

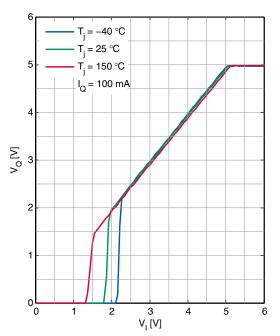


Line Regulation $\Delta V_{
m Q,line}$ versus Input Voltage $V_{
m I}$

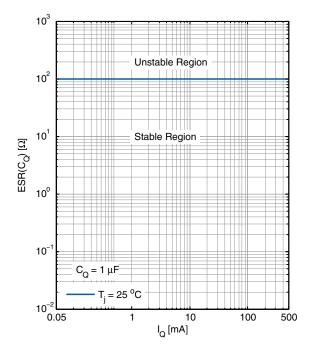




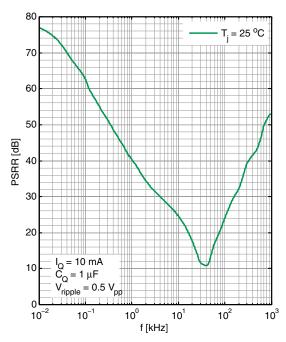
Output Voltage $V_{\rm Q}$ versus Input Voltage $V_{\rm I}$



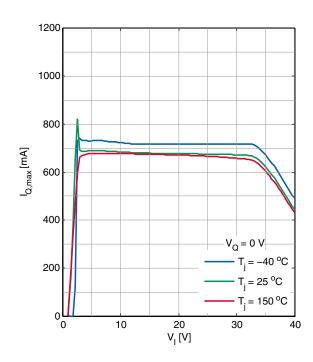
Output Capacitor Series Resistor $ESR(C_{\rm Q})$ versus Output Current $I_{\rm Q}$



Power Supply Ripple Rejection PSRR versus ripple frequency f

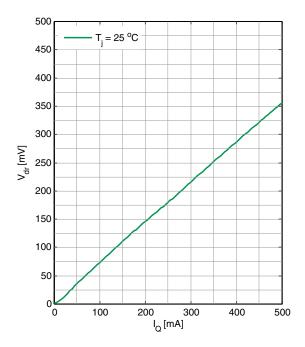


Maximum Output Current I_{Q} versus Input Voltage V_{I}





Dropout Voltage V_{dr} versus Output Current I_{Q}



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5.3 **Current Consumption**

Table 5 **Electrical Characteristics Current Consumption**

 $T_{\rm j}$ = -40 °C to +150 °C, $V_{\rm l}$ = 13.5 V (unless otherwise specified) Typical values are given at $T_{\rm j}$ = 25 °C Conditions of other pins: WI = GND

Parameter	Symbol		Values			Note / Test Condition	Number
		Min.	Тур.	Max.			
Current Consumption $I_q = I_1$	$I_{q,off}$	_	1.3	5	μΑ	$V_{\rm EN}$ = 0 V; $T_{\rm j}$ < 105 °C	P_5.3.1
Current Consumption $I_{q} = I_{l}$	$I_{q,off}$	_	_	8	μΑ	$V_{\rm EN}$ = 0.4 V; $T_{\rm j}$ < 125 °C	P_5.3.3
Current Consumption $I_{q} = I_{l} - I_{Q}$	I_{q}	_	40	52	μΑ	$I_{\rm Q}$ = 0.05 mA $T_{\rm j}$ = 25 °C	P_5.3.4
Current Consumption $I_q = I_1 - I_Q$	I_{q}	_	62	77	μΑ	$I_{\rm Q}$ = 0.05 mA $T_{\rm j}$ < 125 °C	P_5.3.7
Current Consumption $I_q = I_l - I_Q$	I_{q}	_	62	82	μΑ	$I_{\rm Q} = 500 \text{ mA}$ $I_{\rm j} < 125 ^{\circ}\text{C}$	P_5.3.11

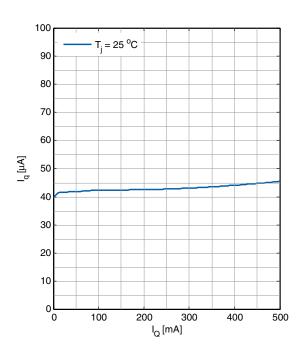
¹⁾ Not subject to production test, specified by design



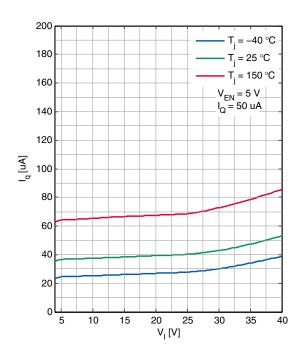
5.4 Typical Performance Characteristics Current Consumption

Typical Performance Characteristics

Current Consumption $I_{\rm q}$ versus Output Current $I_{\rm Q}$



Current Consumption $I_{\rm q}$ versus Input Voltage $V_{\rm I}$





5.5 Enable

The TLS850F0TAV50 can be switched on and off by the Enable feature: Connect a HIGH level as specified below (e.g. the battery voltage) to pin EN to enable the device; connect a LOW level as specified below (e.g. GND) to shut it down. The enable has a built in hysteresis to avoid toggling between ON/OFF state, if signals with slow slopes are applied to the EN input.

Table 6 Electrical Characteristics Enable

 $T_{\rm j}$ = -40 °C to +150 °C, $V_{\rm l}$ = 13.5 V, all voltages with respect to ground (unless otherwise specified) Typical values are given at $T_{\rm j}$ = 25 °C

Parameter	Symbol	Values			Unit	Note / Test Condition	Number
		Min.	Тур.	Max.			
High Level Input Voltage	$V_{EN,H}$	2	_	_	V	V_{Q} settled	P_5.5.1
Low Level Input Voltage	$V_{EN,L}$	-	_	8.0	V	V _Q ≤ 0.1 V	P_5.5.2
Enable Threshold Hysteresis	$V_{EN,Hy}$	100	_	_	mV	-	P_5.5.3
High Level Input Current	$I_{EN,H}$	_	_	3.5	μΑ	V _{EN} = 3.3 V	P_5.5.4
High Level Input Current	$I_{EN,H}$	_	_	22	μA	V _{EN} ≤ 18 V	P_5.5.6
Enable internal pull-down resistor	R_{EN}	0.95	1.5	2.6	ΜΩ	_	P_5.5.7

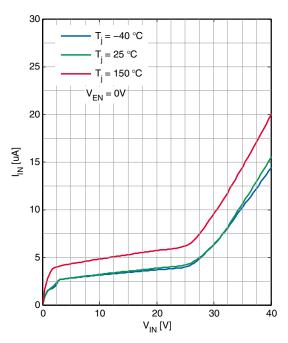
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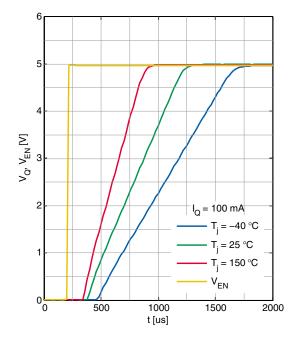
5.6 Typical Performance Characteristics Enable

Typical Performance Characteristics

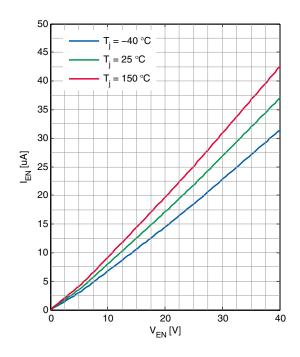
Input Current $I_{\rm IN}$ versus Input Voltage $V_{\rm IN}$ (condition: $V_{\rm EN}$ = 0 V)



Output Voltage $V_{\rm Q}$ versus time (EN switched ON)



Enabled Input Current I_{EN} versus Enabled Input Voltage V_{EN}





5.7 Reset

The TLS850F0TAV50's output voltage is supervised by the Reset feature, including Undervoltage Reset, delayed Reset at Power-On and an adjustable Reset Threshold.

The Undervoltage Reset function sets the pin RO/WO to LOW, in case $V_{\rm Q}$ is falling for any reason below the Reset Threshold $V_{\rm RT,low}$.

When the regulator is powered on, the pin RO/WO is held at LOW for the duration of the Power-On Reset Delay Time t_{rd} .

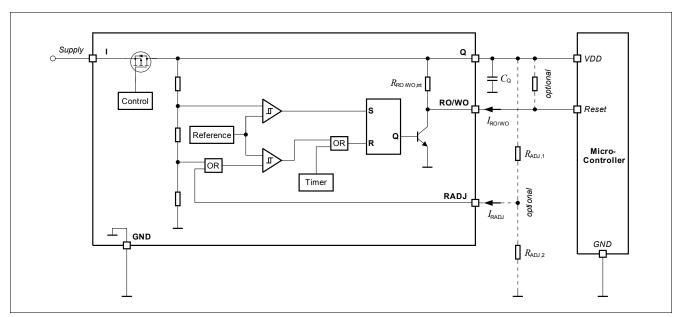


Figure 6 Block Diagram Reset Circuit

Reset Delay Time

The Reset Delay Time t_{rd} is fix defined according to **Table 7**.

Table 7 Reset DelayTime

Reset delay timing	$t_{\sf rd}$
fix	16.5 ms

Power-On Reset Delay Time

The power-on reset delay time is defined by the parameter t_{rd} and allows a microcontroller and oscillator to start up. This delay time is the time period from exceeding the upper reset switching threshold $V_{RT,high}$ until the reset is released by switching the reset output "RO/WO" from "LOW" to "HIGH".

Undervoltage Reset Delay Time

Unlike the power-on reset delay time, the undervoltage reset delay time is defined by the parameter t_{rd} and considers an output undervoltage event where the output voltage V_{O} trigger the $V_{RT,low}$ threshold.

Reset Blanking Time

The reset blanking time $t_{\text{rr,blank}}$ avoids that short undervoltage spikes trigger an unwanted reset "low" signal.



Reset Reaction Time

In case the output voltage of the regulator drops below the output undervoltage lower reset threshold $V_{\rm RT,low}$, the reset output "RO/WO" is set to low, after the delay of the internal reset reaction time $t_{\rm rr,int}$. The reset blanking time $t_{\rm rr,blank}$ is part of the reset reaction time $t_{\rm rr,int}$.

Reset Output "RO/WO"

The reset output "RO/WO" is an open collector output with an integrated pull-up resistor. In case a lower-ohmic "RO/WO" signal is desired, an external pull-up resistor can be connected to the output "Q". Since the maximum "RO/WO" sink current is limited, the minimum value of the optional external resistor " $R_{\text{RO/WO,ext}}$ " is given in Table "Reset Output / Watchdog Output RO/WO" on Page 22.

Reset Output "RO/WO" Low for VQ ≥ 1 V

In case of an undervoltage reset condition reset output "RO/WO" is held "low" for $V_Q \ge 1$ V, even if the input voltage V_I is 0 V. This is achieved by supplying the reset circuit from the output capacitor.

Reset Adjust Function

The undervoltage reset switching threshold can be adjusted according to the application's needs by connecting an external voltage divider (R_{ADJ1} , R_{ADJ2}) at pin "RADJ". For selecting the default threshold connect pin "RADJ" to GND. The reset adjustment range for the TLS850F0TAV50 is given in **Reset Threshold Adjustment Range**.

When dimensioning the voltage divider, take into consideration that there will be an additional current constantly flowing through the resistors.

With a voltage divider connected, the reset switching threshold $V_{\rm RT,new}$ is calculated as follows (neglecting the Reset Adjust Pin Current $I_{\rm RADJ}$):

$$V_{\text{RT,lo,new}} = V_{\text{RADJ,th}} \times (R_{\text{ADJ},1} + R_{\text{ADJ},2}) / R_{\text{ADJ},2}$$
(1)

with

- ullet $V_{
 m RT,lo,new}$: Desired undervoltage reset switching threshold.
- R_{AD,L1}, R_{AD,L2}: Resistors of the external voltage divider, see Figure 6.
- V_{RADJ,th}: Reset adjust switching threshold given in Reset Adjustment Switching Threshold.

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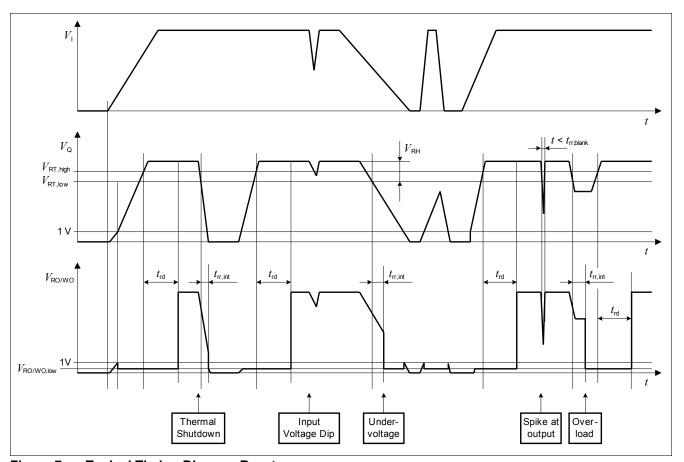


Figure 7 Typical Timing Diagram Reset

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Table 8 Electrical Characteristics Reset

 $T_{\rm j}$ = -40 °C to +150 °C, $V_{\rm l}$ = 13.5 V, all voltages with respect to ground (unless otherwise specified) Typical values are given at $T_{\rm i}$ = 25 °C

Parameter	Symbol	Values			Unit	Note / Test Condition	Number
		Min.	Тур.	Max.			
Output Undervoltage Reset 5V Ve	rsion only				"	ı	II.
Output Undervoltage Reset Upper Switching Threshold	$V_{RT,high}$	4.6	4.7	4.8	V	V_{Q} increasing	P_5.7.1
Output Undervoltage Reset Lower Switching Threshold - Default	$V_{RT,low}$	4.5	4.6	4.7	V	$V_{\rm Q}$ decreasing RADJ = GND	P_5.7.2
Output Undervoltage Reset Switching Hysteresis	$V_{RT,hy}$	60	100	-	mV	RADJ connected to GND	P_5.7.3
Output Undervoltage Reset Headroom $V_{\rm Q}$ - $V_{\rm RT}$	V_{RH}	200	400	_	mV	RADJ = GND	P_5.7.4
Reset Threshold Adjustment							
Reset Adjustment Switching Threshold	$V_{RADJ,th}$	1.15	1.20	1.25	V	_	P_5.7.9
Reset Threshold Adjustment Range	$V_{\mathrm{RT,range}}$	2.5	-	4.4	V	for $V_{Q,nom} = 5 \text{ V}$	P_5.7.10
Reset Output / Watchdog Output		1				· · · · · · · · · · · · · · · · · · ·	l
Reset Output Watchdog Output Low Voltage	$V_{\mathrm{RO/WO,low}}$	_	0.2	0.4	V	1 V $\leq V_{Q} \leq V_{RT}$; $R_{RO/WO} \geq 5.1 \text{ k}\Omega$	P_5.7.16
Reset Output Watchdog Output Internal Pull-Up Resistor	$R_{RO/WO,int}$	13	20	36	kΩ	internally connected to Q	P_5.7.17
Reset Output Watchdog Output External Pull-up Resistor to $V_{\rm Q}$	$R_{RO/WO,ext}$	5.1	_	_	kΩ	$1 \ V \le V_{Q} \le V_{RT}; \\ V_{RO/WO} \le 0.4 \ V$	P_5.7.18
Reset Delay Timing							
Reset Delay Time	$t_{\sf rd}$	13.2	16.5	19.8	ms	Fixed Timing	P_5.7.39
Reset blanking time	t _{rr,blank}	_	7	-	μs	¹⁾ for $V_{\text{Q,nom}}$ = 5 V	P_5.7.46
Internal Reset Reaction Time	$t_{\rm rr,int}$	_	10	33	μs	for $V_{\text{Q,nom}}$ = 5 V	P_5.7.36

¹⁾ Not subject to production test, specified by design.

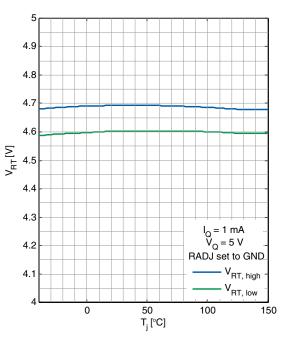
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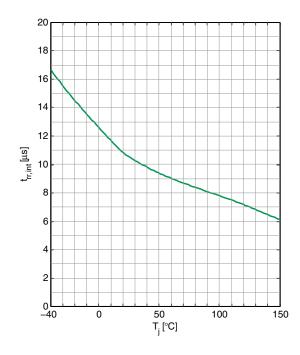
5.8 Typical Performance Characteristics Reset

Typical Performance Characteristics

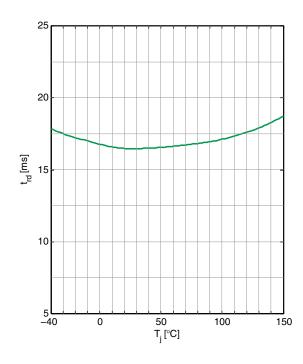
Undervoltage Reset Threshold V_{RT} versus Junction Temperature T_{I}



Internal Reset Reaction Time $t_{\rm rr,int}$ versus Junction Temperature $T_{\rm i}$



Power On Reset Delay Time $t_{\rm rd}$ versus Junction Temperature $T_{\rm j}$





5.9 Standard Watchdog

The TLS850F0TAV50 features a load dependent watchdog function. The watchdog function monitors a microcontroller, including time base failures. In case of a missing falling edge within a certain pulse repetition time, the watchdog output "RO/WO" is set to "low".

The watchdog uses an internal oscillator as timebase. The effective trigger window is derived from the watchdog timebase.

The watchdog output signal is provided by a combined Reset Output / Watchdog Output "RO/WO" pin.

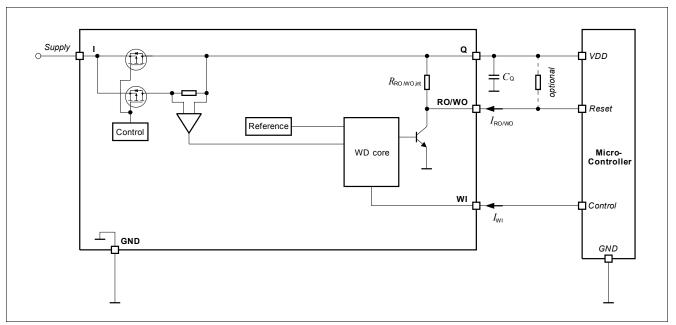


Figure 8 Block Diagram Watchdog Circuit

Watchdog Timing

Figure 9 shows the state diagram of the watchdog (WD) and the mode selection. After power-on, the reset output signal at the "RO/WO" pin (microcontroller reset) is kept LOW for the reset delay time $t_{\rm rd}$. With the LOW to HIGH transition of the signal at "RO/WO" the device starts the watchdog ignore time $t_{\rm WLi}$. Next, the WD starts the watchdog trigger time (time frame within a trigger at WI must occur).

From now on, the timing of the signal on WI from the microcontroller must fit to the WD-trigger time $t_{WI,tr}$. A ReTrigger of the WD-trigger time is done with a HIGH-to-LOW transient at the WI-pin within the active $t_{WI,tr}$.

Watchdog Output "RO/WO"

The watchdog output "RO/WO" is an open collector output with an integrated pull-up resistor. In case a lower-ohmic "RO/WO" signal is desired, an external pull-up resistor can be connected to the output "Q". Since the maximum "RO/WO" sink current is limited, the minimum value of the optional external resistor " $R_{\text{RO/WO,ext}}$ " is given in **Table "Reset Output / Watchdog Output RO/WO" on Page 22**. A HIGH to LOW transition of the watchdog trigger signal on pin WI is taken as a trigger. A watchdog signal is generated ("RO/WO" goes LOW), if there is no trigger pulse during the Watchdog trigger time.

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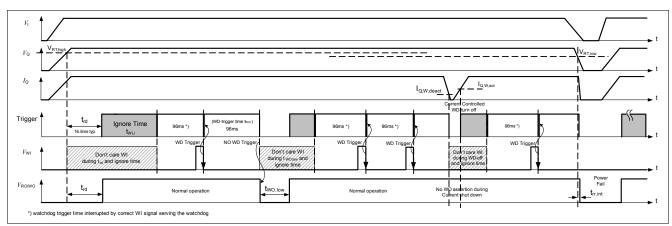


Figure 9 Typical Watchdog Timing Diagram, Watchdog and Reset Modes

Watchdog Input "WI"

The watchdog is triggered by a falling edge at the watchdog input pin "WI". The amplitude and slope of this signal has to comply with the specification (Table "Watchdog Input WI" on Page 26). For details regarding test pulses, see Figure 10 "Test Pulses Watchdog Input WI" on Page 25.

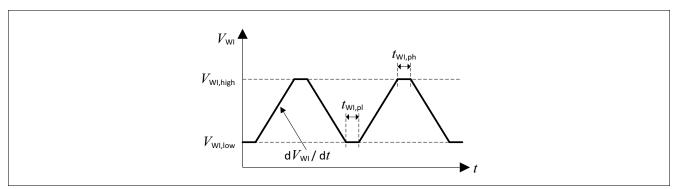


Figure 10 Test Pulses Watchdog Input WI

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