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1. Output A 8. Vcc 7. Output B 2. Input A (-) 6. Input B (-)

General Description

Utilizing the circuit designs perfected for recently introduced Quad Operational Amplifiers, these dual operational amplifiers have several distinct advantages over standard operational amplifier types in single supply applications. They can operate at supply voltages as low as 3.0 Volts or as high as 32 Volts with quiescent currents about one fifth of those associated with the LM741 (on a pet amplifier basis). The common mode input range includes the negative supply, thereby eliminating the necessity for external biasing components in many applications. The TS358 is equivalent to one half of TS324, and output voltage range also includes the negative supply voltage.

5. Input B (+)

Features

- Short circuit protected outputs •
- True differential input stage •
- Single supply operation: 3V to 32V •
- Low input bias currents •
- Internally compensated •
- Common mode range extends to negative supply •
- Single and split supply operation
- Similar performance to the popular MC1558

Ordering Information

Part No.	Package	Packing
TS358CD C3G	DIP-8	50pcs / Tube
TS358CS RLG	SOP-8	2,500pcs / 13" Reel

"G" denotes for Halogen free products

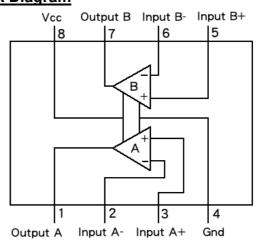
Absolute Maximum Rating

Parameter	Symbol	Limit	Unit
Supply Voltage	V _{CC}	+32 or ±16	V
Differential Input Voltage (Split Power Supplies)	V _{IDR}	32	V
Input Common Mode Voltage Range (note 1)	V _{ICR}	-0.3 to 32	V
Input Forward Current (note 2)	I _{IF}	50	mA
Output Short Circuit Duration	tsc	Continuous	
Operating Junction Temperature Range	TJ	0 ~ +70	°C
Storage Temperature Range	T _{STG}	-65 ~ +150	°C

Note 1: For supply, voltages less than 32V for the TS358 the absolute maximum input voltage is equal to the supply voltage.

Note 2: This input current will only exist when the voltage is negative at any of the input leads. Normal output states will reestablish when the input voltage returns to a voltage greater than -0.3V.

Block Diagram





Electrical Characteristics	$(V_{CC} = 5V, Ta = 25^{\circ}C; unless otherwise specified.)$
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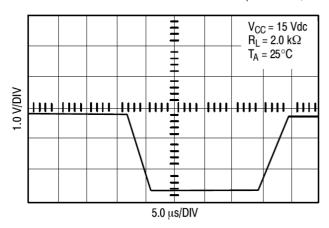
Characteristics	Symbol	Min	Тур	Мах	Unit
Input Offset Voltage					
$V_{CC}\text{=}$ 5.0V to 30V, $V_{IC}\text{=}$ 0V to Vcc -1.7 V, Vo= 1.4V, $R_{S}\text{=}$ 0 Ω	Vio		2.0	7.0	mV
$T_{LOW} \le Ta \le T_{HIGH}$				9.0	
Average Temperature Coefficient of Input Offset Voltage	ΔΙίο/ΔΤ		7.0		uV/°C
Input Offset Current	lio		5.0	50	54
$T_{LOW} \le Ta \le T_{HIGH}$	110			150	nA
Average Temperature Coefficient of input Offset Current	ΔΙίο/ΔΤ		10		pA/°C
Input Bias Current			-45	-250	n A
$T_{LOW} \le Ta \le T_{HIGH}$	I _{IB}		-50	-500	nA
Input Common-Mode Voltage Range)					
V _{CC} = 30 V (Note1)	V _{ICR}	0		28.3	V
$V_{CC} = 30 \text{ V}, T_{LOW} \le Ta \le T_{HIGH}$		0		28	
Differential Input Voltage Range	V _{IDR}			V _{CC}	V
Large Signal Open-Loop Voltage Gain					
R_L = 2.0K, V_{CC} =15V, For Large V_O Swing,	A _{VOL}	25	100		V/mV
$T_{LOW} \le Ta \le T_{HIGH}$		15			
Channel Separation			-120		dB
1.0 KHz to 20KHz			-120		uр
Common Mode Rejection Ratio	CMRR	65	70		dB
$R_{S} \le 10 \ k\Omega$	Civititi	00	70		uр
Power Supply Rejection Ratio	PSRR	65	100		dB
Output Voltage Range, $RL = 2K\Omega$	V _{OR}	0		3.3	V
Output Voltage High Limit					
V_{CC} = 30 V, R_L = 2 k Ω	V _{OH}	26			V
V_{CC} = 30 V, R_L = 10 k Ω		27	28		
Output Voltage Low Limit	V _{OL}		5.0	20	mV
$V_{CC} = 5.0 \text{ V}, \text{ R}_{\text{L}} = 10 \text{ k}\Omega$	VOL		5.0	20	111V
Output Source Current V _{ID} =+1.0V,V _{CC} =15V	I _{O+}	20	40		mA
Output Sink Current					
$V_{ID} = -1.0 V, V_{CC} = 15 V$	I _{O-}	10	20		mA
$V_{ID} = -1.0 \text{ V}, V_O = 200 \text{ mV}$		12	50		uA
Output Short Circuit to Ground (Note 2)	I _{OS}		40	60	mA
Power Supply Current,					
$V_{CC}=30~VV_O=0~V,~R_L=\infty$	I _{CC}		1.5	3.0	mA
$V_{CC}=5.0~V,~V_O=0~V,~R_L=\infty$			0.7	1.2	

Notes :

1. The input common mode voltage or either input signal voltage should not be allowed to go negative by more than

0.3 V. The upper end of the common mode voltage range is V_{CC} 17V, but either or both inputs can go to +32V.
2. Short circuits from the output to V_{CC} can cause excessive heating and eventual destruction. Destructive dissipation can recruit from simultaneous shorts on all amplifiers.







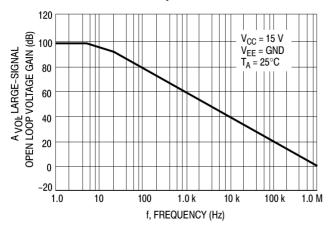


Figure 3. Open Loop Frequency

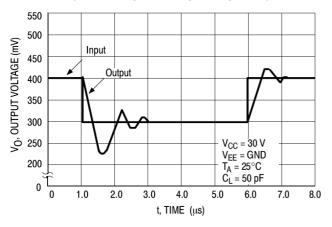


Figure 5. Small-Signal Voltage Follower Pulse **Response** (Noninverting)

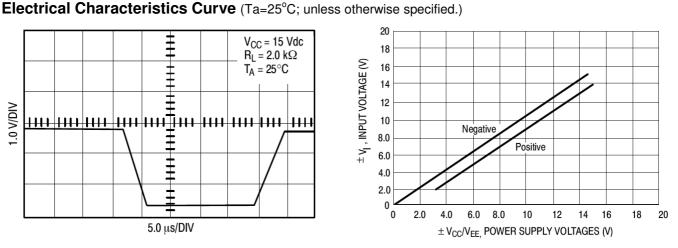


Figure 2. Input Voltage Range

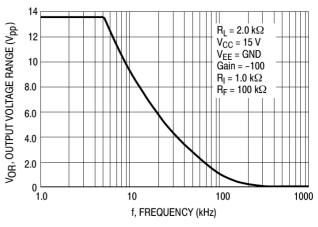


Figure 4. Large Signal Frequency Response

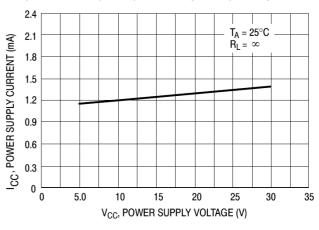


Figure 6. Power Supply Current vs. Supply Voltage

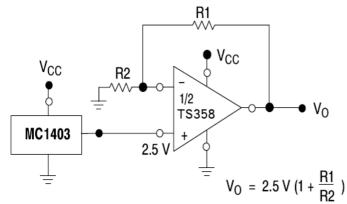
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Application Description

The TS358 made using two internally compensated, two-stage operational amplifiers. The first stage performs not only the first stage gain function but also performs the level shifting and transconductance reduction functions. By reducing the transconductance, a smaller compensation capacitor (only 5.0pF) can be employed, thus saving chip area. Another feature of this input stage is that the input common mode range can include the negative supply or ground, in single supply operation, without saturating either the input devices or the differential to single-ended converter. The second stage consists of a standard current source load amplifier stage.

Each amplifier is biased from an internal-voltage regulator, and which has a low temperature coefficient thus giving each amplifier good temperature characteristics as well as excellent power supply rejection.



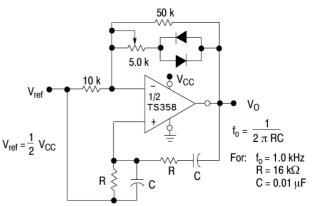


Figure 7. Voltage Reference

Figure 8. Wien Bridge Oscillator

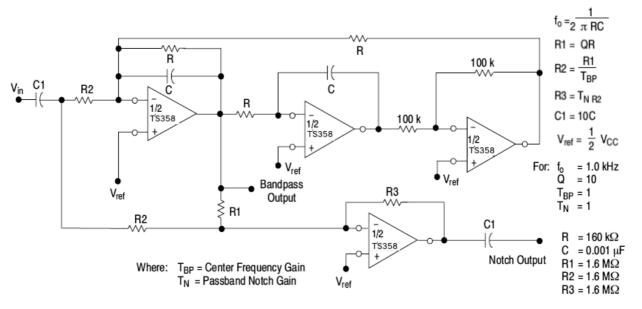
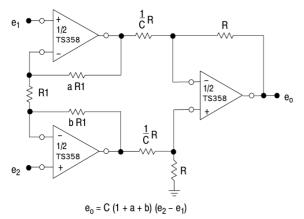


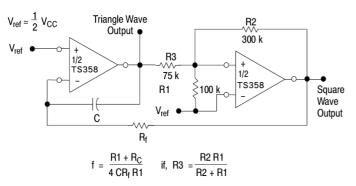
Figure 9. Bi-Quad Filter



Application Description (Continue)









С С

 \leq R2

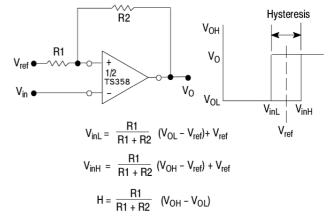
V_{in} R1

≥ R3

V_{ref}

V_{CC}

1/2 TS358





Given: fo = center frequency A(f_o) = gain at center frequency

Choose value fo, C

Then: R3 =
$$\frac{Q}{\pi f_0 C}$$

R1 = $\frac{R3}{2 A(f_0)}$

$$R2 = \frac{R1 R3}{4Q^2 R1 - R3}$$

For less than 10% error from operational amplifier. $\frac{Q_0 f_0}{BW} < 0.1$

Where fo and BW are expressed in Hz.

٧₀

CO CO = 10 C

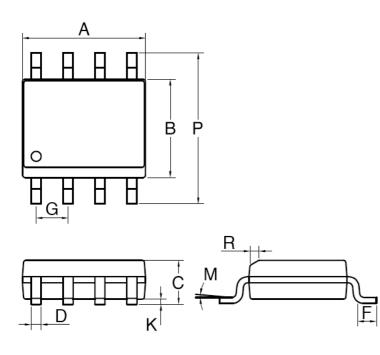
 $V_{ref} = \frac{1}{2} V_{CC}$

If source impedance varies, filter may be preceded with voltage follower buffer to stabilize filter parameters.

Figure 13. Multiple Feedback Bandpass Filter

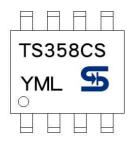


SOP-8 Mechanical Drawing



SOP-8 DIMENSION					
DIM	MILLIMETERS		INCHES		
	MIN	MAX	MIN	MAX.	
Α	4.80	5.00	0.189	0.196	
В	3.80	4.00	0.150	0.157	
С	1.35	1.75	0.054	0.068	
D	0.35	0.49	0.014	0.019	
F	0.40	1.25	0.016	0.049	
G	1.27	1.27BSC 0.05BSC		BSC	
K	0.10	0.25	0.004	0.009	
М	0º	7º	0º	7º	
Р	5.80	6.20	0.229	0.244	
R	0.25	0.50	0.010	0.019	

Marking Diagram



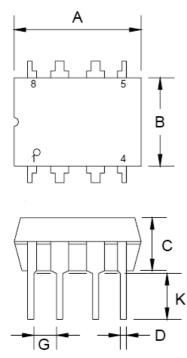
Υ	= `	Year	Code

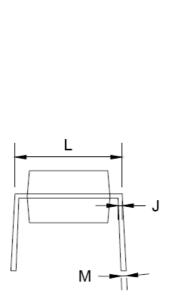
M = Month Code for Halogen Free Product

	O =Jan	P =Feb	Q =Mar	R =Apr
	S =May	T =Jun	U =Jul	V =Aug
	W =Sep	X =Oct	Y =Nov	Z =Dec
L	= Lot Code			



DIP-8 Mechanical Drawing





DIP-8 DIMENSION					
	MILLIMETERS		INCHES		
DIM	MIN	MAX	MIN	MAX	
Α	9.07	9.32	0.357	0.367	
В	6.22	6.48	0.245	0.255	
С	3.18	4.45	0.125	0.135	
D	0.35	0.55	0.019	0.020	
G	2.54 (typ)		0.10 (typ)		
J	0.29	0.31	0.011	0.012	
Κ	3.25	3.35	0.128	0.132	
L	7.75	8.00	0.305	0.315	
М	-	10°	-	10°	

Marking Diagram



Y = Year CodeM = Month Code	e for Haloge	n Free Proc	luct
O =Jan	P =Feb	Q =Mar	R =Apr
S =May	T =Jun	U =Jul	V =Aug
W =Sep	X =Oct	Y =Nov	Z =Dec
L = Lot Code			



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