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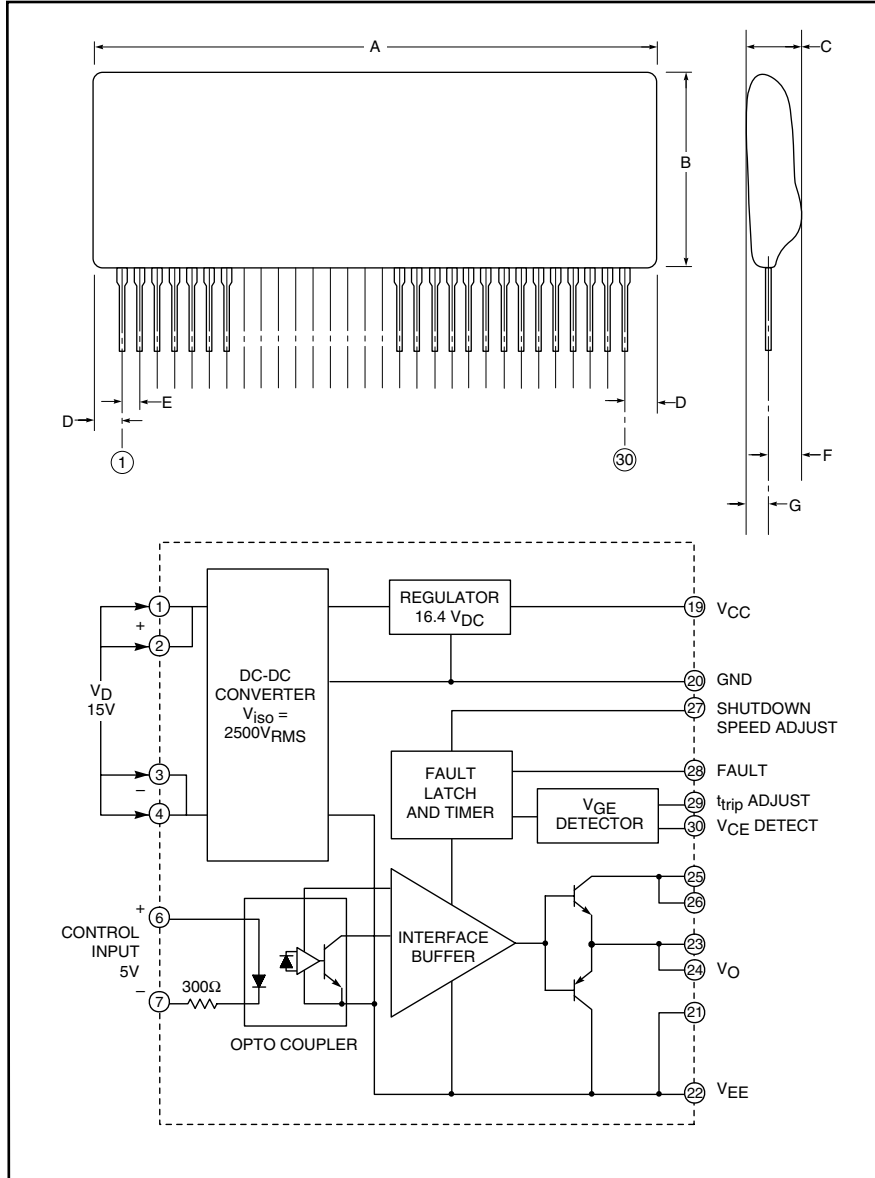
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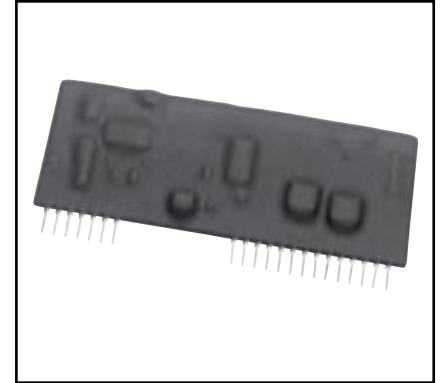
### Hybrid IC IGBT Gate Driver + DC/DC Converter



Outline Drawing and Circuit Diagram

Dimensions	Inches	Millimeters
A	3.27	83.0
B	1.3	33.0
C	0.61	15.5
D	0.2	5.0
E	0.1	2.54
F	0.45	11.5
G	0.18	4.5

Note: All dimensions listed are maximums except E.



#### Description:

VLA502-01 is a hybrid integrated circuit designed for driving IGBT modules. This device is a fully isolated gate drive circuit consisting of an optimally isolated gate drive amplifier and an isolated DC-to-DC converter. The gate driver provides an over-current protection function based on desaturation detection.

#### Features:

- Built-in Isolated DC-DC Converter for Gate Drive
- Propagation Delay Time: 0.4 μs (Typical)
- SIP Outline Allows More Space on Mounting Area
- Variable Fall Time on Short-Circuit Protection
- Electrical Isolation Voltage Between Input and Output (2500 V<sub>rms</sub> for 1 Minute)
- CMOS, TTL Compatible Input

#### Application:

To drive IGBT modules for inverter or AC servo systems applications

#### Recommended IGBT Modules:

600V and 1200V NFH Series IGBTs

**VLA502-01**  
**Hybrid IC IGBT Gate Driver +**  
**DC/DC Converter**

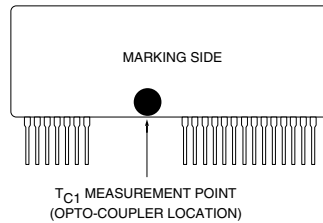
**Absolute Maximum Ratings,  $T_a = 25^\circ\text{C}$  unless otherwise specified**

Characteristics	Symbol	VLA502-01	Units
Supply Voltage, DC	$V_D$	-1 ~ 16.5	Volts
Input Signal Voltage (Applied between Pin 6 - 7, 50% Duty Cycle, Pulse Width 1ms)	$V_i$	-1 ~ 7	Volts
Output Voltage (When the Output Voltage is "H")	$V_O$	$V_{CC}$	Volts
Output Current	$I_{OHP}$	-12	Amperes
(Pulse Width 2 $\mu$ s)	$I_{OLP}$	12	Amperes
Isolation Voltage (Sine Wave Voltage 60HZ, for 1 Minute)	$V_{ISO}$	2500	$V_{rms}$
Case Temperature1 (Surface Temperature Opto-coupler Location)***	$T_{C1}$	85	$^\circ\text{C}$
Case Temperature2 (Surface Temperature Except Opto-coupler Location)	$T_{C2}$	100	$^\circ\text{C}$
Operating Temperature (No Condensation Allowable)	$T_{opr}$	-20 to 60	$^\circ\text{C}$
Storage Temperature (No Condensation Allowable)	$T_{stg}$	-25 to 100*	$^\circ\text{C}$
Fault Output Current (Applied Pin 28)	$I_{FO}$	20	mA
Input Voltage to Pin 30 (Applied Pin 30)	$V_{R30}$	50	Volts
Gate Drive Current	$I_{drive}$	210**	mA

\*Differs from temperature cycle condition.

\*\*Refer to  $I_{drive}$  VS.  $T_a$  CHARACTERISTICS (TYPICAL) graph. (Needs Derating)

\*\*\* $T_{C1}$  Measurement Point (opto-coupler location)



**Electrical and Mechanical Characteristics,  $T_a = 25^\circ\text{C}$  unless otherwise specified,  $V_D = 15\text{V}$ ,  $R_G = 2.2 \Omega$**

Characteristics	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Supply Voltage	$V_D$	Recommended Range	14.2	15	15.8	Volts
Pull-up Voltage on Input Side	$V_{IN}$	Recommended Range	4.75	5	5.25	Volts
"H" Input Current	$I_{IH}$	Recommended Range	9.5	10	14	mA
Switching Frequency	$f$	Recommended Range	—	—	60	kHz
Gate Resistance	$R_G$	Recommended Range	1	—	—	$\Omega$
"H" Input Current	$I_{IH}$	$V_{IN} = 5\text{V}$	—	10	—	mA
Gate Positive Supply Voltage	$V_{CC}$	—	15.2	—	17.5	Volts
Gate Negative Supply Voltage	$V_{EE}$	—	-6	—	-11.5	Volts
Gate Supply Efficiency	$E_{ta}$	Load Current = 210mA	60	75	—	%
$E_{ta} = (V_{CC} +  V_{EE} ) \times 0.21 / (15 \times I_D) \times 100$						
"H" Output Voltage	$V_{OH}$	10k $\Omega$ Connected Between Pin 23-20	14	15.3	16.5	Volts
"L" Output Voltage	$V_{OL}$	10k $\Omega$ Connected Between Pin 23-20	-5.5	—	-11	Volts
"L-H" Propagation Time	$t_{PLH}$	$I_{IH} = 10\text{mA}$	0.25	0.4	0.65	$\mu\text{s}$
"L-H" Rise Time	$t_r$	$I_{IH} = 10\text{mA}$	—	0.3	1	$\mu\text{s}$
"H-L" Propagation Time	$t_{PHL}$	$I_{IH} = 10\text{mA}$	0.25	0.4	0.65	$\mu\text{s}$
"H-L" Fall Time	$t_f$	$I_{IH} = 10\text{mA}$	—	0.3	1	$\mu\text{s}$



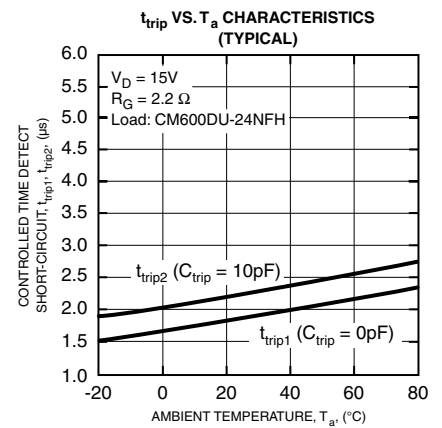
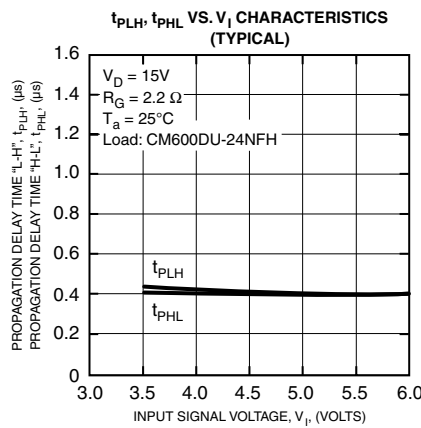
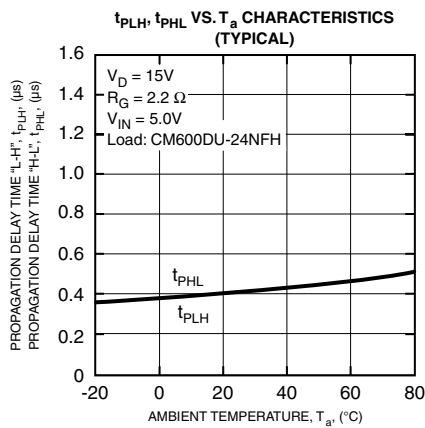
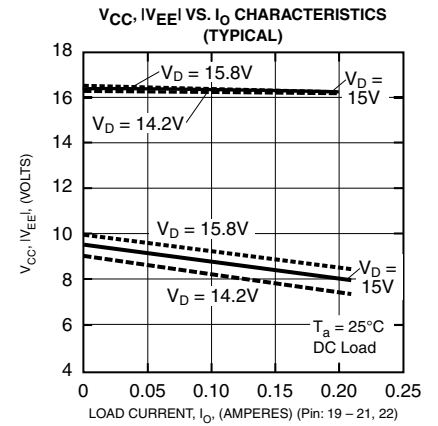
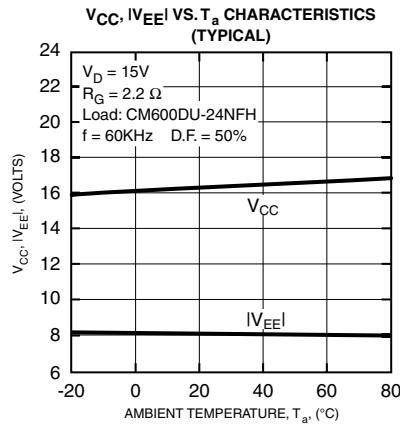
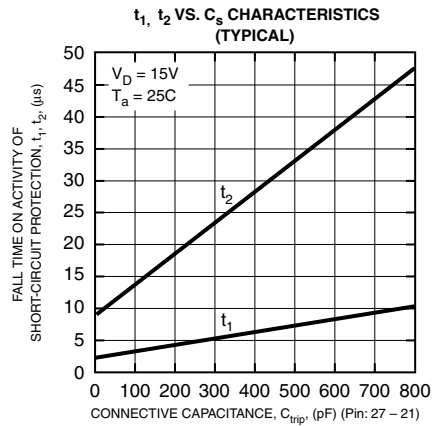
Powerex, Inc., 200 E. Hillis Street, Youngwood, Pennsylvania 15697-1800 (724) 925-7272

**VLA502-01**  
**Hybrid IC IGBT Gate Driver +**  
**DC/DC Converter**

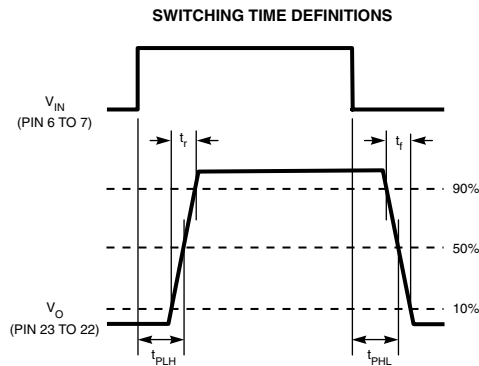
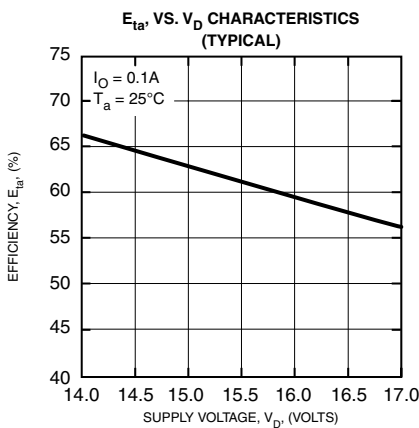
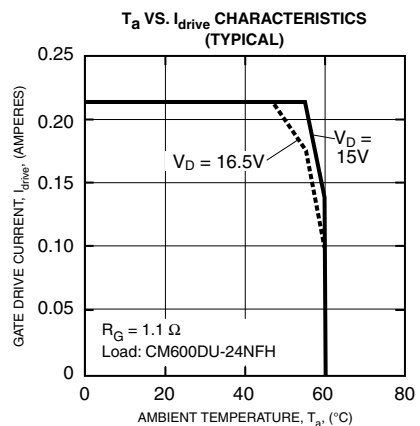
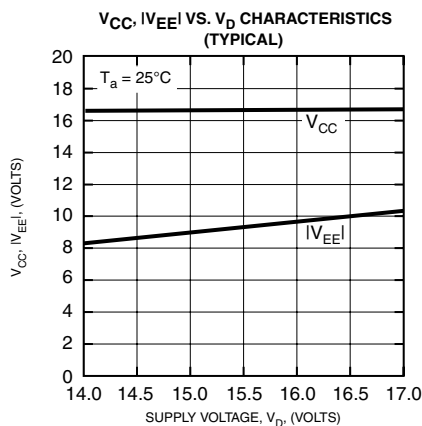
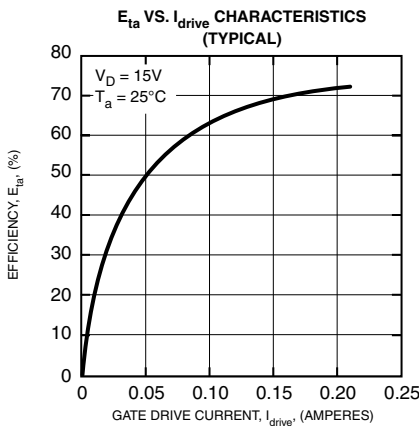
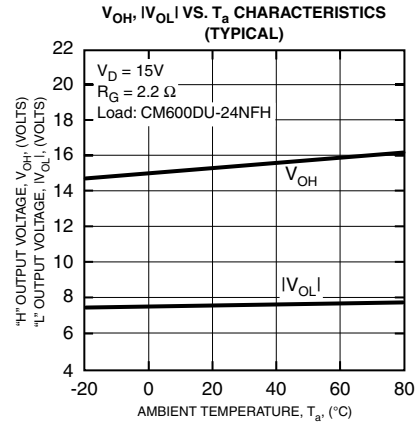
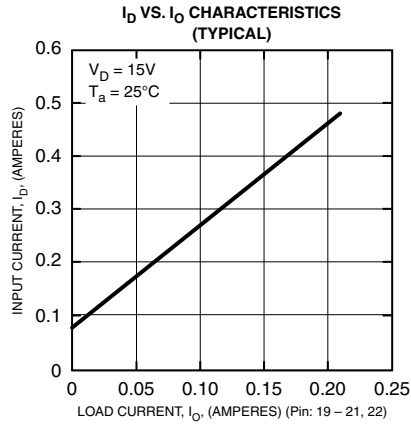
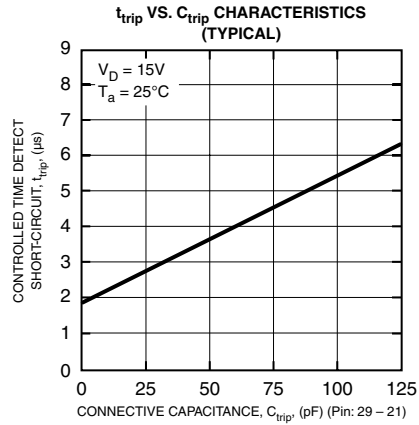
**Electrical and Mechanical Characteristics,  $T_a = 25^\circ\text{C}$  unless otherwise specified,  $V_D = 15\text{V}$ ,  $R_G = 2.2\ \Omega$**

Characteristics	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Timer	$t_{\text{timer}}$	Between Start and Cancel (Under Input Sign "L")	1	—	2	ms
Fault Output Current	$I_{\text{FO}}$	Applied Pin 28, $R = 4.7\text{k}\Omega$	—	5	—	mA
Controlled Time Detect Short-Circuit 1	$t_{\text{trip1}}$	Pin 30 : 15V and More, Pin 29 : Open	—	1.8	—	$\mu\text{s}$
Controlled Time Detect Short-Circuit 2*	$t_{\text{trip2}}$	Pin 30 : 15V and More, Pin 29-21, 22 : 10pF (Connective Capacitance)	—	2.2	—	$\mu\text{s}$
SC Detect Voltage	$V_{\text{SC}}$	Collector Voltage of Module	15	—	—	Volts

\*Length of wiring from  $C_{\text{trip}}$  to Pins 21, 22, and 29 must be less than 5cm.

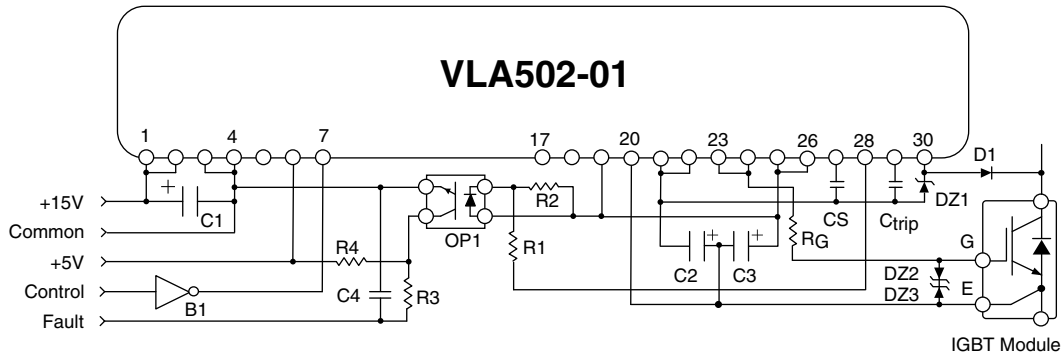


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**Application Circuit**



**Component Selection:**

Dsgn.	Typ. Value	Description
D1	0.5A	VCE detection diode – fast recovery, $V_{rrm} > V_{CES}$ of IGBT being used (Note 1)
DZ1	30V, 0.5W	Detect input pin surge voltage protection (Note 2)
DZ2, DZ3	18V, 1.0W	Gate surge voltage protection
C1	150 $\mu$ F, 35V	$V_D$ supply decoupling – Electrolytic, long life, low Impedance, 105°C (Note 3)
C2, C3	100-1000 $\mu$ F, 35V	DC/DC output filter - Electrolytic, long life, low Impedance, 105°C (Note 3,4)
C4	0.01 $\mu$ F	Fault feedback signal noise filter
CS	0-1000pF	Adjust soft shutdown – Multilayer ceramic or film (see application note)
Ctrip	0-200pF	Adjust trip time - Multilayer ceramic or film (see application note)
R1	4.7k $\Omega$ , 0.25W	Fault sink current limiting resistor
R2	3.3k $\Omega$ , 0.25W	Fault signal noise suppression resistor
R3	1k $\Omega$ , 0.25W	Fault feedback signal noise filter
R4	4.7k, 0.25W	Fault feedback signal pull-up
OP1	NEC PS2501	Opto-coupler for fault feedback signal isolation
B1	CMOS Buffer	74HC04 or similar – Must actively pull high to maintain noise immunity

**Notes:**

- (1) The VCE detection diode should have a blocking voltage rating equal to or greater than the  $V_{CES}$  of the IGBT being driven. Recovery time should be less than 200ns to prevent application of high voltage to Pin 30.
- (2) DZ1 is necessary to protect Pin 30 of the driver from voltage surges during the recovery of D1
- (3) Power supply input and output decoupling capacitors should be connected as close as possible to the pins of the gate driver.
- (4) DC-to-DC converter output decoupling capacitors must be sized to have appropriate ESR and ripple current capability for the IGBT being driven.

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### General Description

The VLA502-01 is a hybrid integrated circuit designed to provide gate drive for high power IGBT modules. This circuit has been optimized for use with Powerex NFH-Series IGBT modules. However, the output characteristics are compatible with most MOS gated power devices. The VLA502-01 features a compact single-in-line package design. The upright mounting design minimizes printed circuit board space to allow efficient and flexible layout. The VLA502-01 converts logic level control signals into fully isolated +15V/-8V gate drive with up to 12A of peak drive current. Isolated drive power is provided by a built in DC-to-DC converter and control signal isolation is provided by an integrated high speed opto-coupler. Short circuit protection is provided by means of desaturation detection.

### Short Circuit Protection

Figure 1 shows a block diagram of a typical desaturation detector. In this circuit, a high voltage fast recovery diode (D1) is connected to the IGBT's collector to monitor the collector to emitter voltage. When the IGBT is in the off state,  $V_{CE}$  is high and D1 is reverse biased. With D1 off the (+) input of the comparator is pulled up to the positive gate drive power supply ( $V+$ ) which is normally +15V. When the IGBT turns on, the comparators (+) input is pulled down by D1 to the IGBT's  $V_{CE(sat)}$ . The (-) input of the comparator is supplied with a fixed voltage ( $V_{trip}$ ). During a normal on-state condition the comparator's (+) input will be less than  $V_{trip}$  and its output will be low. During a normal off-state condition the comparator's (+) input will be larger than  $V_{trip}$  and its output will be high. If the IGBT turns on into a short circuit, the high current will cause the IGBT's collector-emitter voltage to rise above  $V_{trip}$  even though the gate of the IGBT is being driven on. This abnormal presence of high  $V_{CE}$  when the IGBT is supposed to be on is often called **desaturation**. Desaturation can be detected by a logical AND of the driver's input signal and the comparator output. When the output of the AND goes high a short circuit is indicated. The output of the AND can be used to command the IGBT to shut down in order to protect it from the short circuit. A delay ( $t_{trip}$ ) must be provided after the comparator output to allow for the normal turn on time of the IGBT. The  $t_{trip}$  delay is set so that the IGBT's  $V_{CE}$  has enough time to fall below  $V_{trip}$  during normal turn on switching. If  $t_{trip}$  is set too short, erroneous desaturation detection will occur. The maxi-

mum allowable  $t_{trip}$  delay is limited by the IGBT's short circuit withstanding capability. In typical applications using Powerex IGBT modules the recommended limit is 10us.

### Operation of the VLA502-01 Desaturation Detector

The Powerex VLA502-01 incorporates short circuit protection using desaturation detection as described above. A flow chart for the logical operation of the short-circuit protection is shown in Figure 2. When a desaturation is detected the hybrid gate driver performs a soft shutdown of the IGBT and starts a timed ( $t_{timer}$ ) 1.5ms lock out. The soft turn-off helps to limit the transient voltage that may be generated while interrupting the large short circuit current flowing in the IGBT. During the lock out the driver pulls Pin 28 low to indicate the fault status. Normal operation of the driver will resume after the lock-out time has expired and the control input signal returns to its off state.

### Adjustment of Trip time

The VLA502-01 has a default short-circuit detection time delay ( $t_{trip}$ ) of approximately 3 $\mu$ s. This will prevent erroneous detection of short-circuit conditions as long as the series gate resistance ( $R_G$ ) is near the minimum recommended value for the module being used. The 3 $\mu$ s delay is appropriate for most applications so adjustment will not be necessary. However, in some low frequency applications it may be desirable to use a larger series gate resistor to slow the switching of the IGBT, reduce noise, and limit turn-off transient

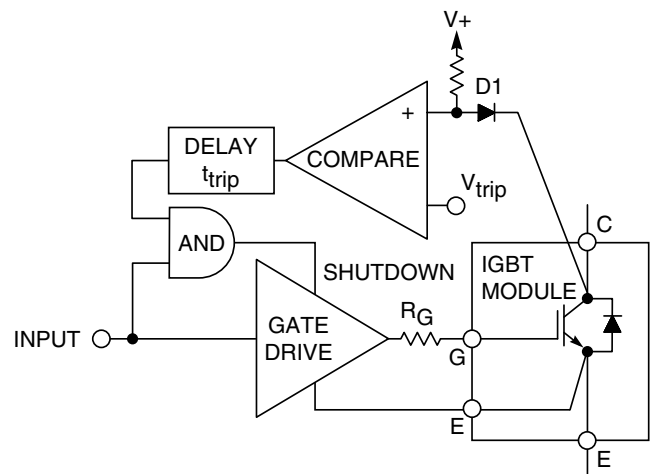
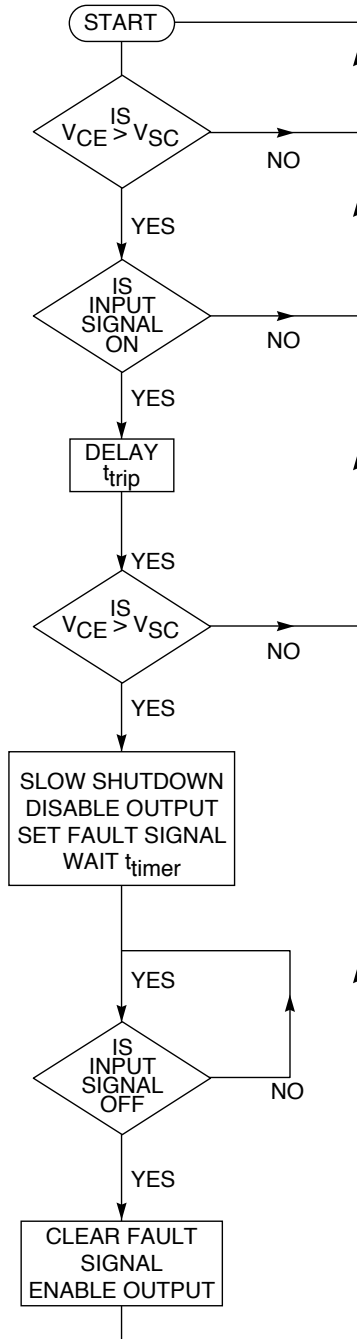


Figure 1. Desaturation Detector

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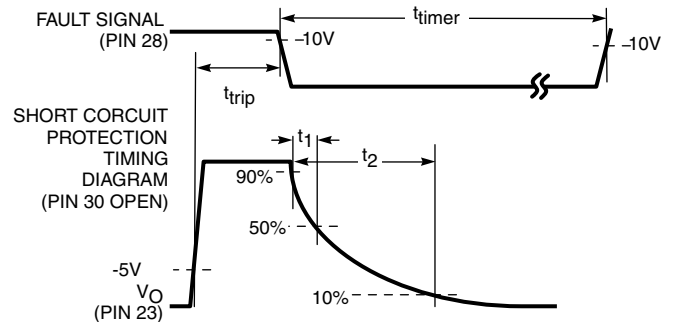


**Figure 2. VLA502-01 Desaturation Detector**

voltages. When  $R_G$  is increased, the switching delay time of the IGBT will also increase. If the delay becomes long enough so that the voltage on the detect Pin 30 is greater than  $V_{SC}$  at the end of the  $t_{trip}$  delay the driver will erroneously indicate that a short circuit has occurred. To avoid this condition the VLA502-01 has provisions for extending the  $t_{trip}$  delay by connecting a capacitor ( $C_{trip}$ ) between Pin 29 and  $V_{EE}$  (Pins 21 and 22). A curve showing the effect of adding  $C_{trip}$  on time is given in the characteristic data section of this data sheet. The waveform defining trip time ( $t_{trip}$ ) is shown in Figure 3. If  $t_{trip}$  is extended care must be exercised not to exceed the short-circuit withstanding capability of the IGBT module.

### Adjustment of Soft Shutdown Speed

As noted above the VLA502-01 provides a soft turn-off when a short circuit is detected in order to help limit the transient voltage surge that occurs when large short circuit currents are interrupted. The default shutdown speed will work for most applications so adjustment is usually not necessary. In this case  $C_S$  can be omitted. In some applications using large modules or parallel connected devices it may be helpful to make the shut-down even softer. This can be accomplished by connecting a capacitor ( $C_S$ ) at Pin 27. A curve showing the effect of  $C_S$  on short circuit fall time ( $t_1$ ,  $t_2$ ) is given in the characteristic data section of this data sheet. The waveform defining the fall time characteristics is shown in Figure 3.



**Figure 3. Adjustment of  $t_{trip}$  and Slow Shutdown Speed**