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VMMK-2403

2 to 4 GHz GaAs High Linearity LNA in Wafer Level Package



Data Sheet



Lead (Pb) Free
RoHS 6 fully
compliant

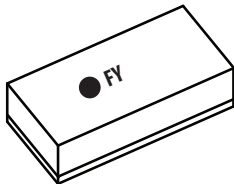


Description

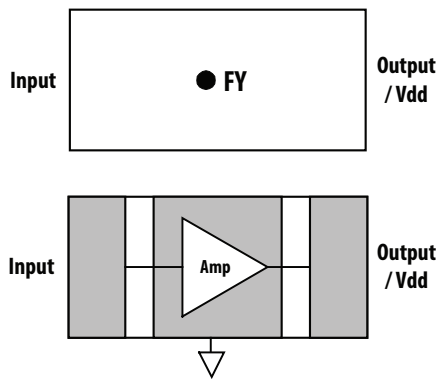
Avago's VMMK-2403 is an easy-to-use, high linearity low noise amplifier in a miniaturized wafer level package (WLP). The low noise and unconditionally stable performance makes this amplifier ideal as a primary or subsequent gain block of an RF receiver in applications from 2–4GHz. A 3V, 38mA power supply is required for optimal performance.

This amplifier is fabricated with enhancement E-pHEMT technology and industry leading revolutionary wafer level package. The GaAsCap wafer level sub-miniature leadless package is small and ultra thin yet can be handled and placed with standard 0402 pick and place assembly. This product is easy to use since it requires only positive DC voltages for bias and no matching coefficients are required for impedance matching to 50 Ω systems.

WLP 0402, 1mm x 0.5mm x 0.25 mm



Pin Connections (Top View)



Note:
"F" = Device Code
"Y" = Month Code

Features

- 1 x 0.5 mm Surface Mount Package
- Ultrathin (0.25mm)
- Unconditionally Stable
- 50Ohm Input and Output Match
- RoHS6 + Halogen Free

Specifications (Vdd = 3.0V, Idd = 37mA)

- Output IP3: 28dBm
- Small-Signal Gain: 16dB
- Noise Figure: 1.7dB

Applications

- 2.4 GHz, 3.5GHz WLAN and WiMax notebook computer, access point and mobile wireless applications
- 802.16 & 802.20 BWA systems
- Radar, radio and ECM systems



Attention: Observe precautions for handling electrostatic sensitive devices.
ESD Machine Model = 50V
ESD Human Body Model = 675V
Refer to Avago Application Note A004R:
Electrostatic Discharge, Damage and Control.

Table 1. Absolute Maximum Ratings^[1]

Sym	Parameters/Condition	Unit	Absolute Max
Vd	Supply Voltage (RF Output) ^[2]	V	5
Id	Device Current ^[2]	mA	60
P _{in,max}	CW RF Input Power (RF Input) ^[3]	dBm	+20
P _{diss}	Total Power Dissipation	mW	300
T _{ch}	Max channel temperature	°C	150
θ _{jc}	Thermal Resistance ^[4]	°C/W	160

Notes

1. Operation in excess of any of these conditions may result in permanent damage to this device.
2. Bias is assumed DC quiescent conditions
3. With the DC (typical bias) and RF applied to the device at board temperature T_b = 25°C
4. Thermal resistance is measured from junction to board using IR method

Table 2. DC and RF Specifications

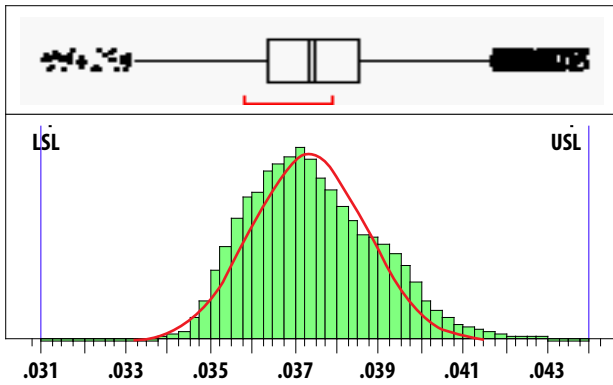
T_A = 25°C, Frequency = 3 GHz, V_d = 3V, Z_{in} = Z_{out} = 50Ω (unless otherwise specified)

Sym	Parameters/Condition	Unit	Minimum	Typ.	Maximum
Id	Device Current	mA	30	37	45
NF ^[1,2]	Noise Figure	dB	–	1.7	2.2
Ga ^[1,2]	Associated Gain	dB	14	16	17
OIP3 ^[1,2,3]	Output 3rd Order Intercept	dBm	+26	+27.8	–
P-1dB ^[1,2]	Output Power at 1dB Gain Compression	dBm		+16.5	–
IRL ^[1,2]	Input Return Loss	dB	–	-12	–
ORL ^[1,2]	Output Return Loss	dB	–	-12	–

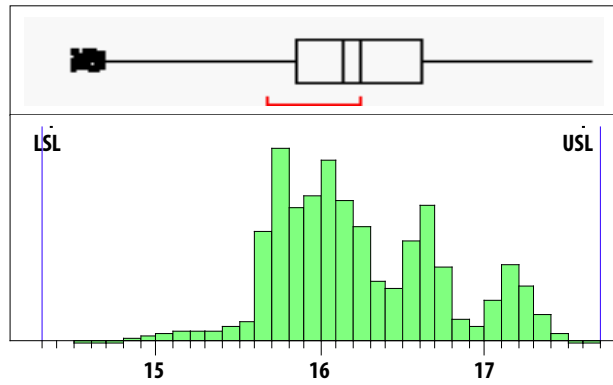
Notes:

1. Losses of test systems have been de-embedded from final data
2. Measure Data obtained from wafer-probing
3. OIP3 test condition: F1=3.0GHz, F2=3.01GHz, Pin=-20dBm

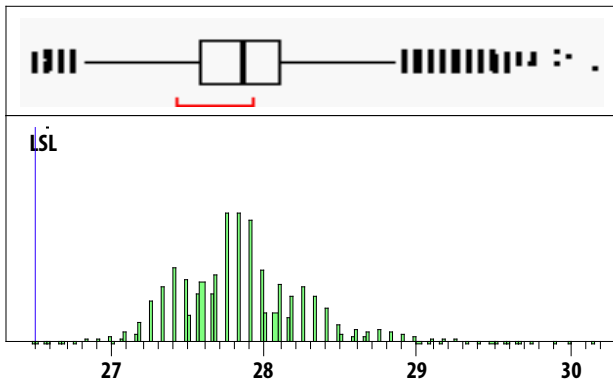
Product Consistency Distribution Charts at 3.0 GHz, Vd = 3V



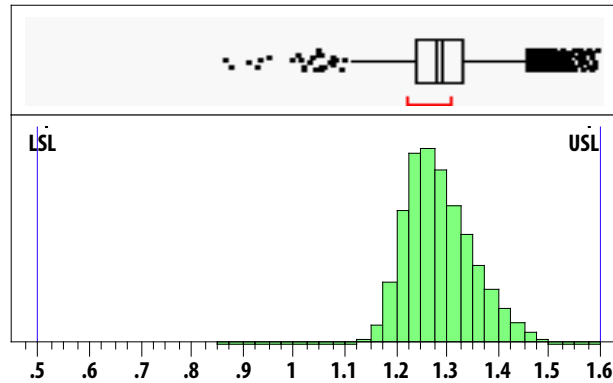
Id, LSL=0.030 A, Nominal=0.037 A, USL=0.045 A



Gain @ 3GHz, LSL=14 dB, Nominal=16 dB, USL=18 dB



OIP3 at 3GHz, LSL=26 dBm, Nominal=27.85 dBm



NF @ 3GHz, Nominal=1.69 dB, USL=2.20 dB

Note: Distribution data based on 500 part sample size from skew lots during initial characterization. Measurements were obtained using 300um G-S production wafer probe. Future wafers allocated to this product may have nominal values anywhere between the upper and lower limits.

VMMK-2403 Typical Performance

($T_A = 25^\circ\text{C}$, $V_{dd} = 3\text{V}$, $I_{dd} = 38\text{mA}$, $Z_{in} = Z_{out} = 50\ \Omega$ unless noted)

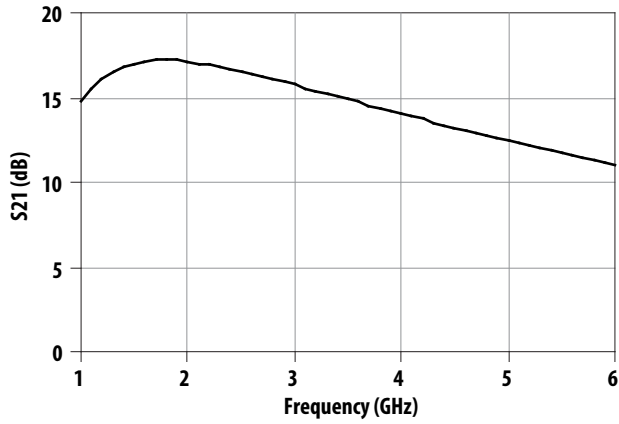


Figure 1. Small-signal Gain^[1]

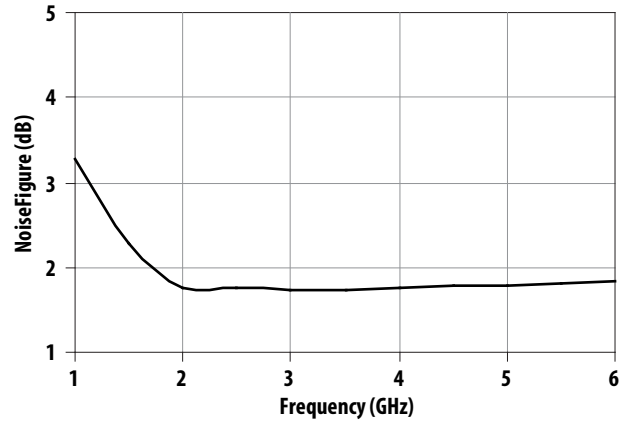


Figure 2. Noise Figure^[1]

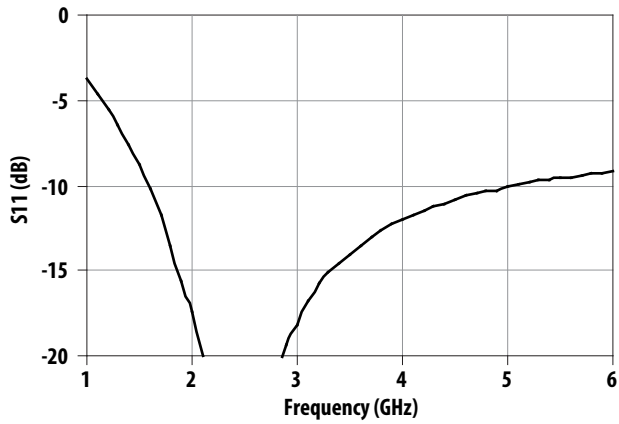


Figure 3. Input Return Loss^[1]

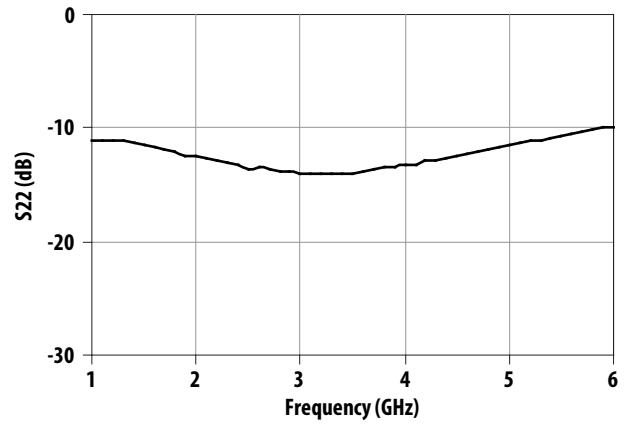


Figure 4. Output Return Loss^[1]

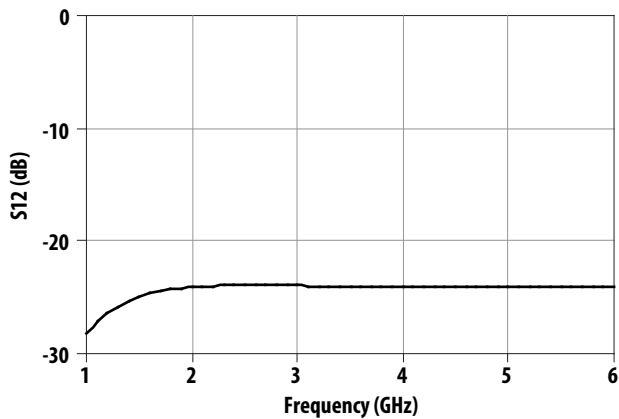


Figure 5. Isolation^[1]

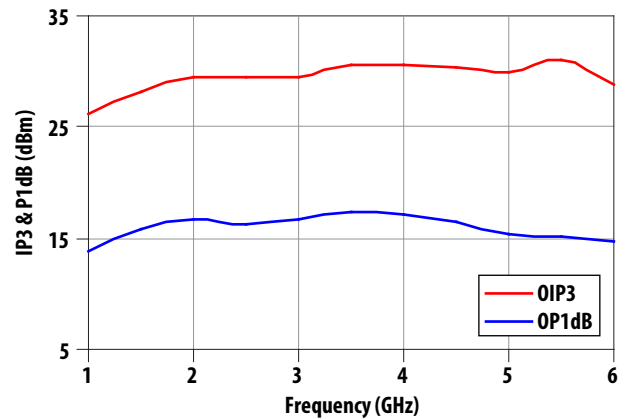


Figure 6. Output IP3 & Output P1dB^[1,2]

Notes:

1. Data taken on a G-S-G probe substrate fully de-embedded to the reference plane of the package
2. Output IP3 data taken at $P_{in} = -15\text{dBm}$

VMMK-2403 Typical Performance (continue)

($T_A = 25^\circ\text{C}$, $V_{dd} = 3\text{V}$, $I_{dd} = 38\text{mA}$, $Z_{in} = Z_{out} = 50\ \Omega$ unless noted)

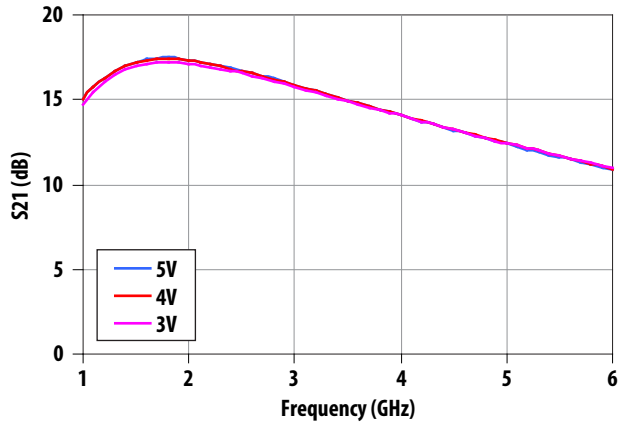


Figure 7. Gain over Vdd^[1]

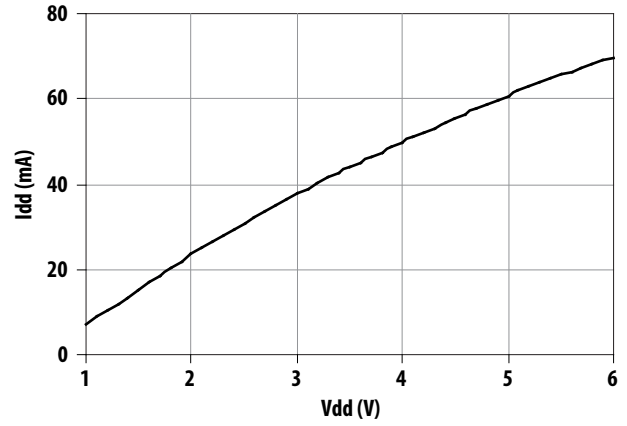


Figure 8. Total Current^[1]

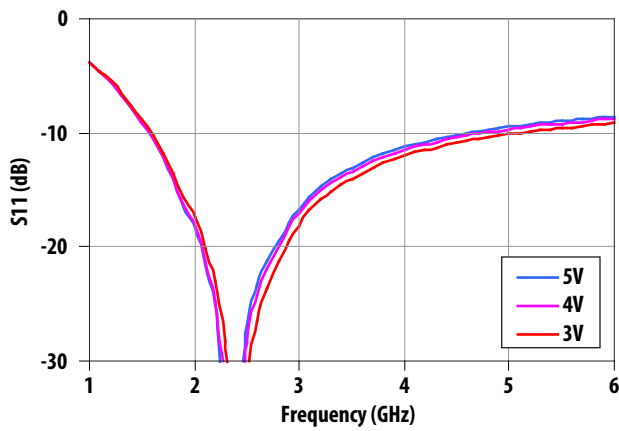


Figure 9. Input Return Loss over Vdd^[1]

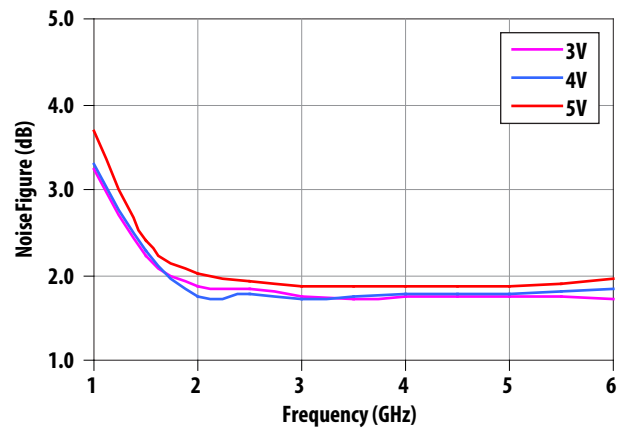


Figure 10. Noise Figure over Vdd^[1]

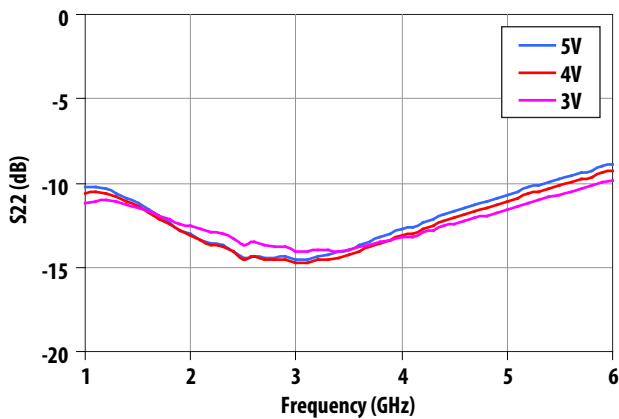


Figure 11. Output Return Loss Over Vdd^[1]

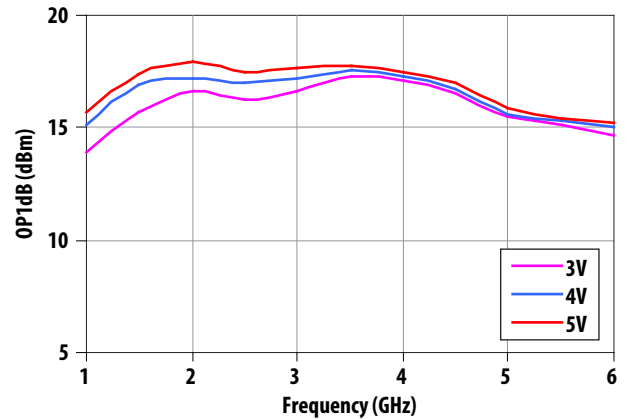


Figure 12. Output P1dB Over Vdd^[1]

Notes:

1. Data taken on a G-S-G probe substrate fully de-embedded to the reference plane of the package

VMMK-2403 Typical Performance (continue)

($T_A = 25^\circ\text{C}$, $V_{dd} = 3\text{V}$, $I_{dd} = 38\text{mA}$, $Z_{in} = Z_{out} = 50\ \Omega$ unless noted)

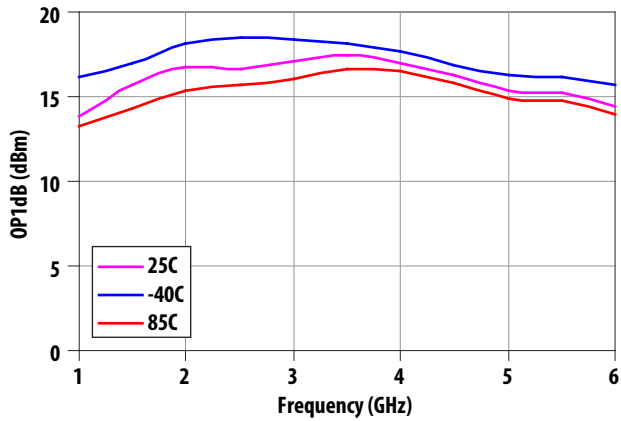


Figure 13. Output P-1dB over Temp^[3]

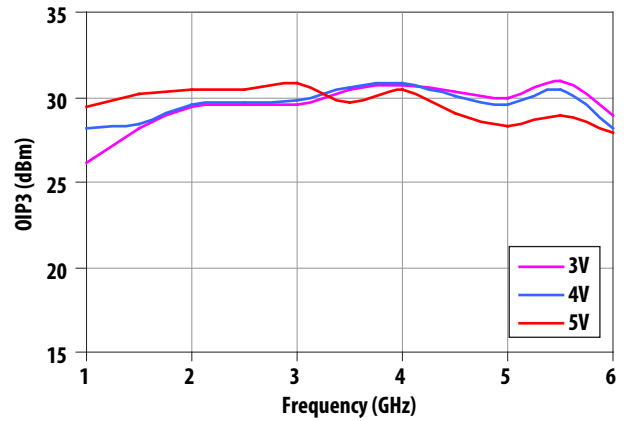


Figure 14. Output IP3 Over Vdd^[2,3]

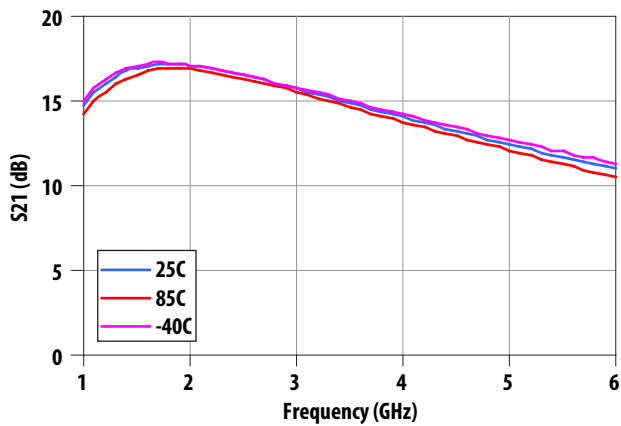


Figure 15. Gain over Temp^[3]

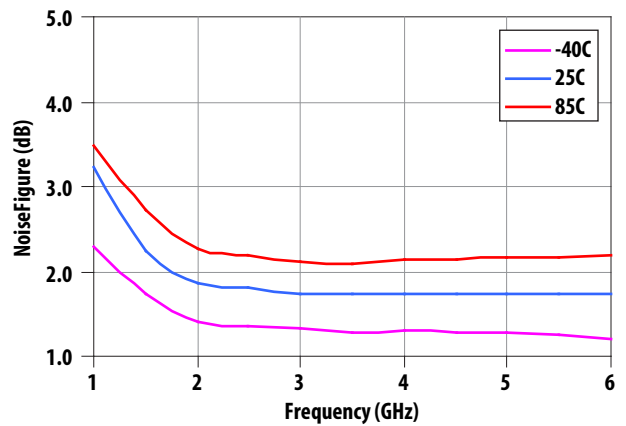


Figure 16 Noise Figure over Temp^[3]

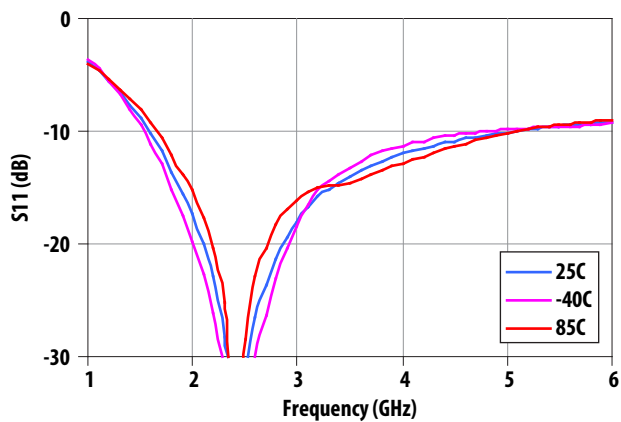


Figure 17. Input Return Loss Over Temp^[3]

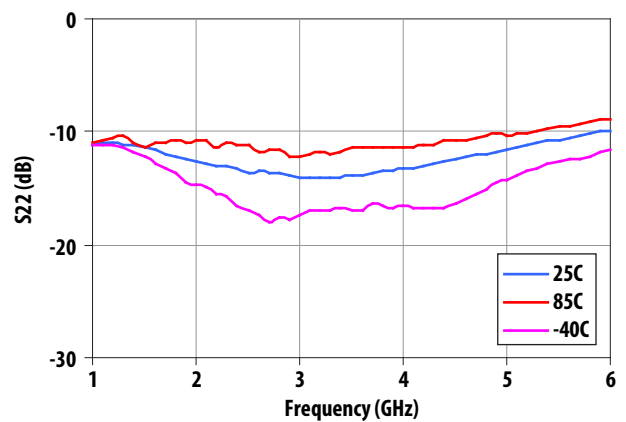


Figure 18. Output Return Loss Over Temp^[3]

Notes:

1. Data taken on a G-S-G probe substrate fully de-embedded to the reference plane of the package
2. Output IP3 data taken at $P_{in} = -15\text{dBm}$
3. Over temp data taken on a test fixture (Figure 20) without de-embedding

VMMK-2403 Typical S-parameters

$T_A = 25^\circ\text{C}$, $V_{dd} = 3\text{V}$, $I_{dd} = 38\text{mA}$, $Z_{in} = Z_{out} = 50\ \Omega$ unless noted)

Freq GHz	S11			S21			S12			S22		
	Mag	dB	Phase	Mag	dB	Phase	Mag	dB	Phase	Mag	dB	Phase
1	0.65	-3.79	113.52	5.47	14.76	-144.36	0.04	-28.13	87.94	0.28	-11.18	-10.64
1.5	0.36	-8.82	81.05	7.05	16.97	-178.05	0.05	-24.97	60.57	0.27	-11.46	-58.54
2.0	0.13	-17.40	55.74	7.16	17.09	155.06	0.06	-24.10	42.05	0.24	-12.53	-87.61
2.5	0.03	-31.80	-133.91	6.71	16.53	134.46	0.06	-23.90	30.73	0.21	-13.69	-100.64
3.0	0.12	-18.13	-151.55	6.12	15.74	117.90	0.06	-23.97	23.43	0.20	-14.03	-105.66
3.5	0.20	-14.05	-163.87	5.55	14.88	104.00	0.06	-24.04	19.19	0.20	-13.96	-107.15
4.0	0.25	-12.00	-174.32	5.04	14.05	91.92	0.06	-24.14	15.78	0.22	-13.22	-106.91
4.5	0.29	-10.89	175.98	4.58	13.22	81.08	0.06	-24.14	13.35	0.24	-12.42	-108.59
5.0	0.31	-10.13	167.30	4.19	12.44	71.14	0.06	-24.12	12.04	0.26	-11.58	-110.59
5.5	0.33	-9.59	159.43	3.84	11.69	61.94	0.06	-24.07	10.81	0.29	-10.73	-113.90
6.0	0.35	-9.16	151.41	3.54	10.97	53.04	0.06	-24.01	9.65	0.32	-9.87	-118.40
6.5	0.36	-8.91	143.69	3.26	10.27	44.57	0.06	-23.89	8.81	0.35	-9.03	-123.20
7.0	0.37	-8.69	136.41	3.02	9.59	36.31	0.07	-23.74	8.13	0.39	-8.27	-128.23
7.5	0.38	-8.50	129.27	2.79	8.91	28.19	0.07	-23.58	7.21	0.42	-7.52	-133.56
8.0	0.38	-8.36	122.37	2.58	8.25	20.21	0.07	-23.39	6.48	0.46	-6.81	-139.38
8.5	0.39	-8.21	115.19	2.39	7.56	12.37	0.07	-23.22	5.74	0.49	-6.17	-144.98
9.0	0.39	-8.12	108.46	2.20	6.85	4.43	0.07	-23.06	4.98	0.53	-5.53	-151.26
9.5	0.40	-7.98	101.84	2.02	6.11	-3.50	0.07	-22.83	4.19	0.56	-4.98	-158.01
10.0	0.41	-7.83	95.04	1.84	5.30	-11.35	0.07	-22.60	3.57	0.60	-4.39	-164.68
10.5	0.41	-7.72	88.39	1.66	4.40	-19.19	0.08	-22.34	2.72	0.64	-3.88	-172.30
11.0	0.42	-7.59	82.22	1.48	3.39	-26.92	0.08	-22.09	2.06	0.67	-3.42	179.91
11.5	0.43	-7.42	76.20	1.29	2.22	-34.30	0.08	-21.79	1.40	0.71	-2.99	171.62
12.0	0.43	-7.25	70.01	1.10	0.82	-41.13	0.08	-21.46	0.40	0.74	-2.67	162.75

VMMK-2403 Application and Usage

(Please always refer to the latest Application Note AN5378 in website)

Biasing and Operation

The VMMK-2403 is normally biased with a positive drain supply connected to the output pin through an external bias-tee and with bypass capacitors as shown in Figure 19. The recommended drain supply voltage is 3 V and the corresponding drain current is approximately 38mA. Aspects of the amplifier performance may be improved over a narrower bandwidth by application of additional conjugate, linearity, or low noise (τ_{opt}) matching.

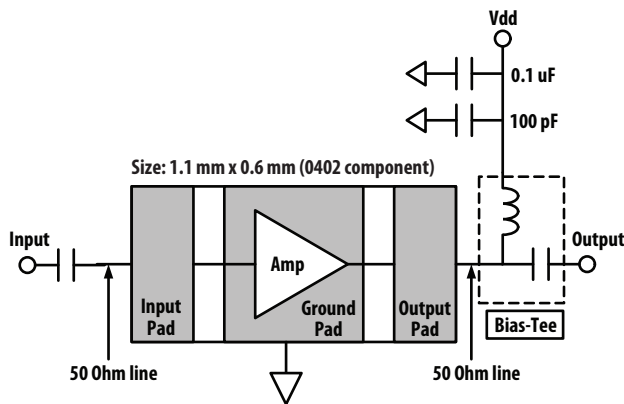


Figure 19. Usage of the VMMK-2403

Biasing the device at 3V results in slightly lower noise figure. In a typical application, the bias-tee can be constructed using lumped elements. The value of the output inductor can have a major effect on both low and high frequency operation. The demo board uses a 15 nH inductor which provides an optimum value for 2 to 4 GHz operation. If operation is desired at frequencies higher than 4 GHz then a smaller value such as 8.2 nH may be required to keep the self resonant frequency higher than the maximum desired frequency of operation.

Another approach for broadbanding the VMMK-2403 is to series two different value inductors with the smaller value inductor placed closest to the device and favoring the higher frequencies. The larger value inductor will then offer better low frequency performance by not loading the output of the device. The parallel combination of the 100pF and 0.1uF capacitors provide a low impedance in the band of operation and at lower frequencies and should be placed as close as possible to the inductor. The low frequency bypass provides good rejection of power supply noise and also provides a low impedance termination for third order low frequency mixing products that will be generated when multiple in-band signals are injected into any amplifier.

Refer the Absolute Maximum Ratings table for allowed DC and thermal conditions.

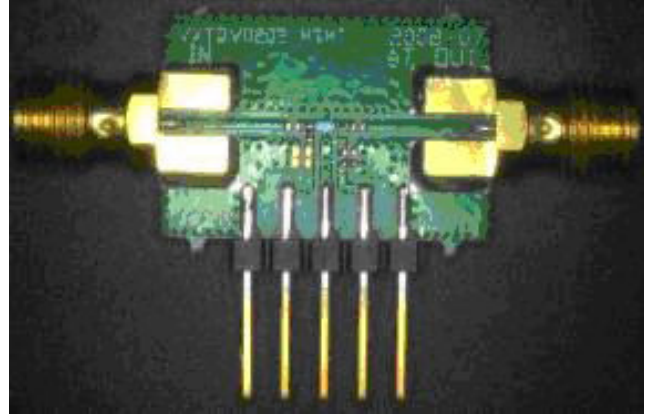


Figure 20. Evaluation/Test Board (available to qualified customer request)

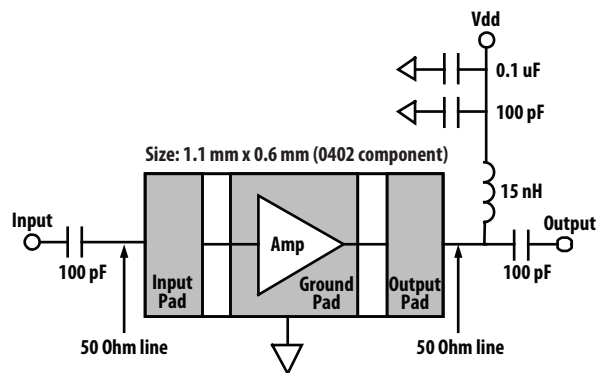


Figure 21. Example application of VMMK-2403 at 3.5GHz

S Parameter Measurements

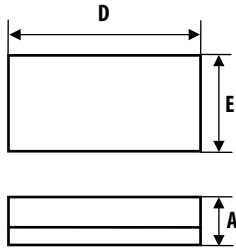
The S-parameters are measured on a .016 inch thick RO4003 printed circuit test board, using G-S-G (ground signal ground) probes. Coplanar waveguide is used to provide a smooth transition from the probes to the device under test. The presence of the ground plane on top of the test board results in excellent grounding at the device under test. A combination of SOLT (Short - Open - Load - Thru) and TRL (Thru - Reflect - Line) calibration techniques are used to correct for the effects of the test board, resulting in accurate device S-parameters. The reference plane for the S Parameters is at the edge of the package.

The product consistency distribution charts shown on page 2 represent data taken by the production wafer probe station using a 300um G-S wafer probe. The ground-signal probing that is used in production allows the device to be probed directly at the device with minimal common lead inductance to ground. Therefore there will be a slight difference in the nominal gain obtained at the test frequency using the 300um G-S wafer probe versus the 300um G-S-G printed circuit board substrate method.

Ordering Information

Part Number	Devices Per Container	Container
VMMK-2403-BLKG	100	Antistatic Bag
VMMK-2403-TR1G	5000	7" Reel

Package Dimension Outline

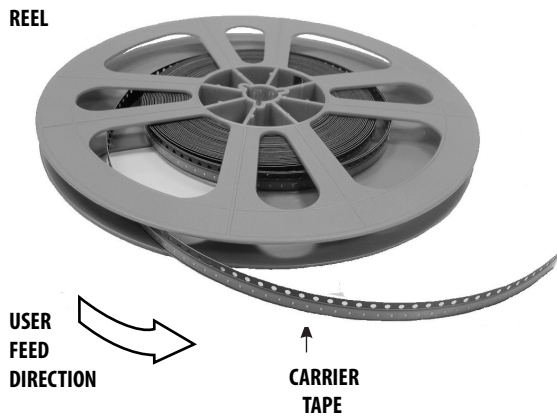


Die dimension:

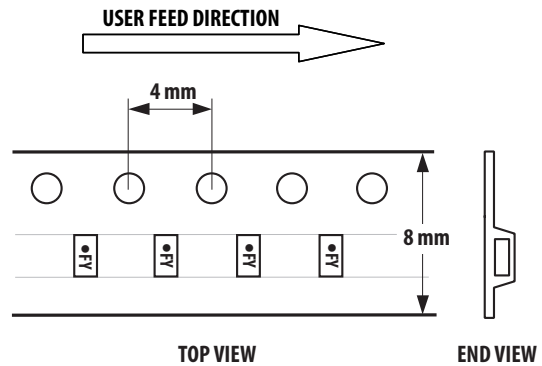
Dim	Range	Unit
D	1.004 - 1.085	mm
E	0.500 - 0.585	mm
A	0.225 - 0.275	mm

Note:
All dimensions are in mm

Reel Orientation

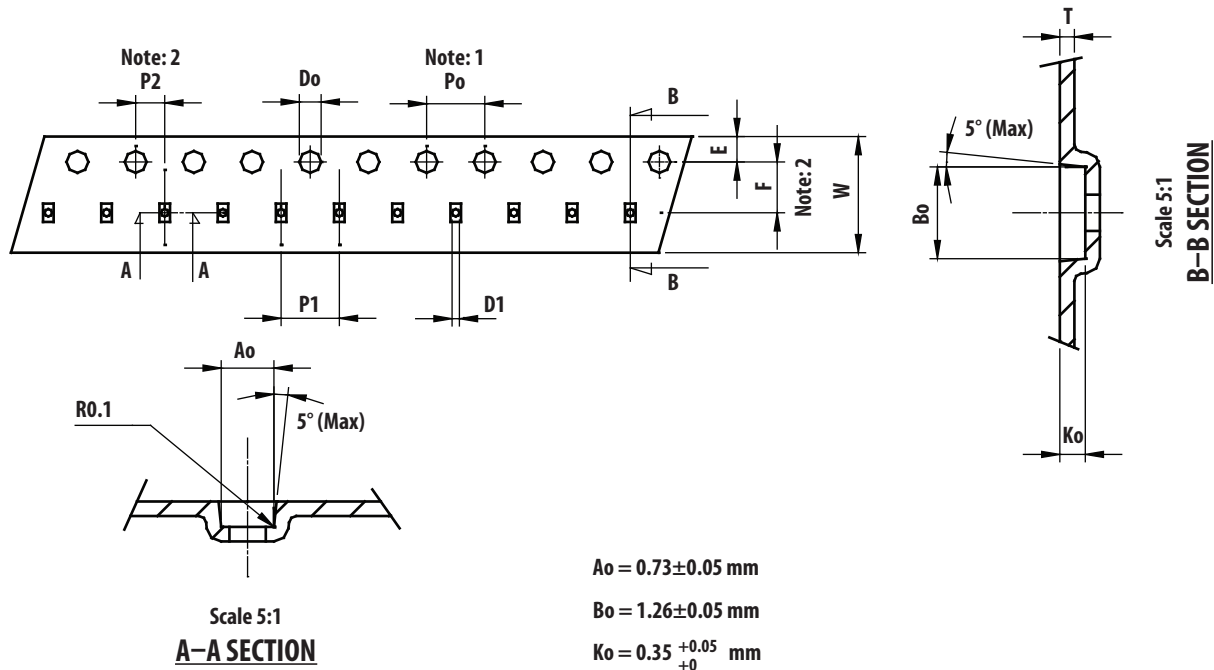


Device Orientation



Note:
"F" = Device Code
"Y" = Month Code

Tape Dimensions



Unit: mm

Symbol	Spec.
K1	-
Po	4.0±0.10
P1	4.0±0.10
P2	2.0±0.05
Do	1.55±0.05
D1	0.5±0.05
E	1.75±0.10
F	3.50±0.05
10Po	40.0±0.10
W	8.0±0.20
T	0.20±0.02

Notice:

- 10 Sprocket hole pitch cumulative tolerance is ±0.1mm.
- Pocket position relative to sprocket hole measured as true position of pocket not pocket hole.
- A_o & B_o measured on a plane 0.3mm above the bottom of the pocket to top surface of the carrier.
- K_o measured from a plane on the inside bottom of the pocket to the top surface of the carrier.
- Carrier camber shall be not than 1m per 100mm through a length of 250mm.

For product information and a complete list of distributors, please go to our web site: www.avagotech.com

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