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VN5010AK-E

High side driver with analog current sense
for automotive applications

Features

Max supply voltage	V_{CC}	41 V
Operating voltage range	V_{CC}	4.5 to 36 V
Max on-state resistance	R_{ON}	10 mΩ
Current limitation (typ)	I_{LIMH}	65 A
Off-state supply current (typ)	I_S	2 µA

■ Main features
 – Inrush current active management by power limitation
 – Very low standby current
 – 3.0v CMOS compatible input
 – Optimized electromagnetic emission
 – Very low electromagnetic susceptibility
 – In compliance with the 2002/95/EC European directive

■ Diagnostic functions
 – Proportional load current sense
 – High current sense precision for wide range currents
 – Current sense disable
 – Thermal shutdown indication
 – Very low current sense leakage

■ Protections
 – Undervoltage shutdown
 – Overvoltage clamp
 – Load current limitation
 – Self limiting of fast thermal transients
 – Protection against loss of ground and loss of V_{CC}
 – Thermal shutdown

PowerSSO-24™

 – Reverse battery protection (see *Figure 26*)
 – Electrostatic discharge protection

Application

■ All types of resistive, inductive and capacitive loads

Description

The VN5010AK-E is a monolithic device made using STMicroelectronics VIPower M0-5 technology. It is intended for driving resistive or inductive loads with one side connected to ground. Active V_{CC} pin voltage clamp protects the device against low energy spikes (see ISO7637 transient compatibility table). This device integrates an analog current sense which delivers a current proportional to the load current (according to a known ratio) when CS_DIS is driven low or left open. When CS_DIS is driven high, the CURRENT SENSE pin is in a high impedance condition. Output current limitation protects the device in overload condition. In case of long overload duration, the device limits the dissipated power to safe level up to thermal shutdown intervention. Thermal shutdown with automatic restart allows the device to recover normal operation as soon as fault condition disappears.

Table 1. Device summary

Package	Order codes	
	Tube	Tape and reel
PowerSSO-24™	VN5010AK-E	VN5010AKTR-E

Contents

List of tables

List of figures

1   Block diagram and pin description

Figure 1.   Block diagram

Table 2.   Pin function

Name	Function
V_{CC}	Battery connection.
OUTPUT	Power output.
GND	Ground connection. Must be reverse battery protected by an external diode/resistor network.
INPUT	Voltage controlled input pin with hysteresis, CMOS compatible. Controls output switch state.
CURRENT SENSE	Analog current sense pin, delivers a current proportional to the load current.
CS_DIS	Active high CMOS compatible pin, to disable the current sense pin.

Figure 2.   Connection diagram (top view)
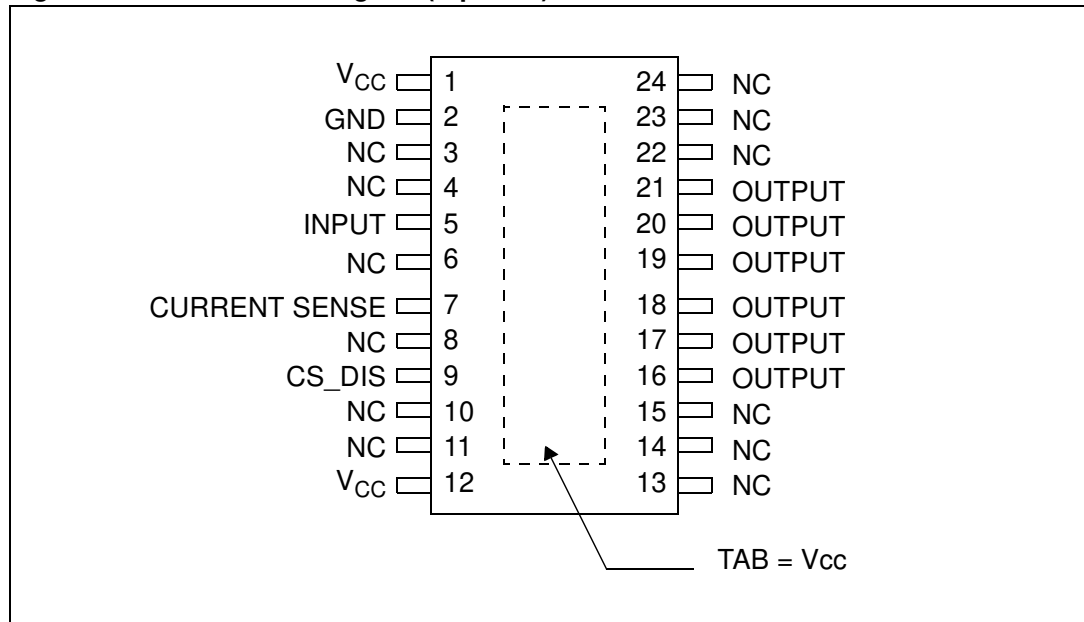
Table 3.   Suggested connections for unused and not connected pins

Connection/pin	Current sense	N.C.	Output	Input	CS_DIS
Floating	N.R.[1]	X	X	X	X
To ground	Through 1 kΩ resistor	X	N.R.	Through 10 kΩ resistor	Through 10 kΩ resistor

1. Not recommended.

2 Electrical specifications

Figure 3. Current and voltage conventions
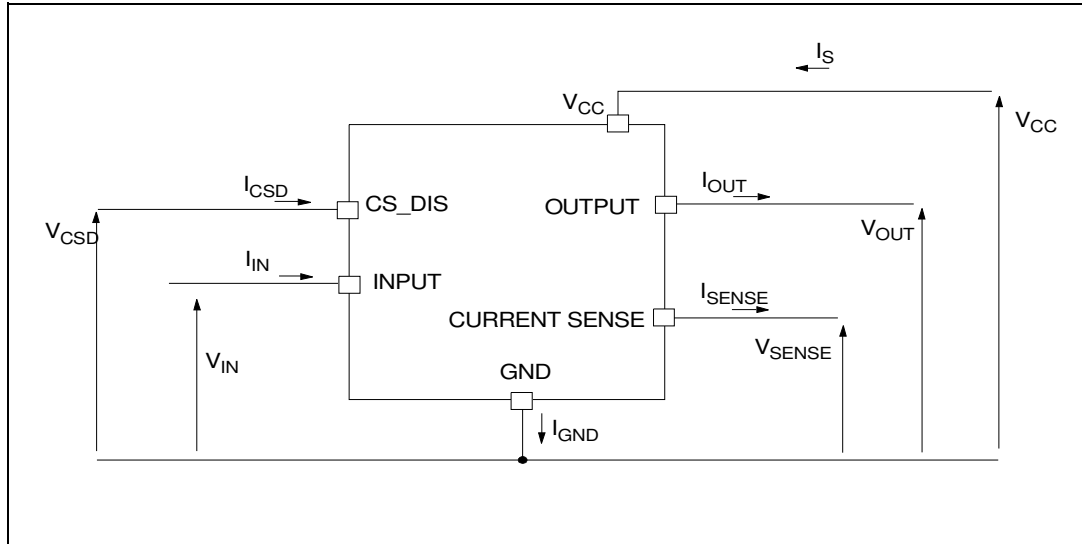
Note: $V_{Fn} = V_{OUT} - V_{CC}$ *during reverse battery condition.*

2.1 Absolute maximum ratings

Stressing the device above the ratings listed in the "Absolute maximum ratings" tables may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is not implied. Exposure to the conditions in this section for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE Program and other relevant quality documents.

Table 4. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{CC}	DC supply voltage	41	V
$-V_{CC}$	Reverse DC supply voltage	0.3	V
$-I_{GND}$	DC reverse ground pin current	200	mA
I_{OUT}	DC output current	Internally limited	A
$-I_{OUT}$	Reverse DC output current	30	A
I_{IN}	DC input current	-1 to 10	mA
I_{CSD}	DC current sense disable input current	-1 to 10	mA
$-I_{CSENSE}$	DC reverse CS pin current	200	mA

Table 4.   Absolute maximum ratings (continued)

Symbol	Parameter	Value	Unit
V_{CSENSE}	Current sense maximum voltage	V_{CC}-41 +V_{CC}	V V
E_{MAX}	Maximum switching energy (single pulse) (L=1.25mH; R_L=0Ω; V_{bat}=13.5V; T_{jstart}=150ºC; I_{OUT} = I_{limL}(typ.))	609	mJ
V_{ESD}	Electrostatic discharge (Human Body Model: R=1.5KΩ; C=100pF) – Input – Current Sense – CS_DIS – Output – V_{CC}	4000 2000 4000 5000 5000	V
V_{ESD}	Charge device model (CDM-AEC-Q100-011)	750	V
T_j	Junction operating temperature	-40 to 150	°C
T_{stg}	Storage temperature	-55 to 150	°C

Table 5.   Thermal data

Symbol	Parameter	Max value	Unit
$R_{thj-case}$	Thermal resistance junction case (max)	0.3	°C/W
$R_{thj-amb}$	Thermal resistance junction ambient (max)	See *Figure 29*	°C/W

2.2   Electrical characteristics

Values specified in this section are for 8 V< V_{CC}< 36 V; -40 °C< T_j< 150 °C, unless otherwise stated (for each channel).

Table 6.   Power section

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC}	Operating supply voltage		4.5	13	36	V
V_{USD}	Undervoltage shutdown			3.5	4.5	V
$V_{USDhyst}$	Undervoltage shutdown hysteresis			0.5		V
R_{ON}	On-state resistance	I_{OUT}= 6A; T_j= 25°C I_{OUT}= 6A; T_j= 150°C I_{OUT}=6A; V_{CC}=5V;T_j=25°C			10 20 13	mΩ mΩ mΩ
V_{clamp}	Clamp voltage	I_{CC}= 20 mA	41	46	52	V
I_S	Supply current	Off-state; V_{CC}= 13V; T_j= 25°C; V_{IN}=V_{OUT}=V_{SENSE}=V_{CSD}=0V On-state; V_{CC}=13V; V_{IN}=5V; I_{OUT}=0A		2[1] 1.5	5[1] 3	μA mA

Table 6. **Power section (continued)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{L(off)}$	Off-state output current	$V_{IN}=V_{OUT}=0V$; $V_{CC}= 13V$; $T_j= 25°C$ $V_{IN}=V_{OUT}=0V$; $V_{CC}= 13V$; $T_j= 125°C$	0 0	0.01	3 5	µA
V_F	Output - V_{CC} diode voltage	$-I_{OUT}= 10A$; $T_j= 150°C$			0.7	V

1. PowerMOS leakage included.

Table 7. **Switching (V_{CC}=13V)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn-on delay time	$R_L= 2.6Ω$ (see *Figure 8*)	-	35	-	µs
$t_{d(off)}$	Turn-off delay time	$R_L= 2.6Ω$ (see *Figure 8*)	-	65	-	µs
$(dV_{OUT}/dt)_{on}$	Turn-on voltage slope	$R_L= 2.6Ω$	-	See *Figure 20*	-	V/µs
$(dV_{OUT}/dt)_{off}$	Turn-off voltage slope	$R_L= 2.6Ω$	-	See *Figure 22*	-	V/µs
W_{ON}	Switching energy losses during t_{won}	$R_L= 2.6Ω$ (see *Figure 8*)	-	1.5	-	mJ
W_{OFF}	Switching energy losses during t_{woff}	$R_L= 2.6Ω$ (see *Figure 8*)	-	0.8	-	mJ

Table 8. **Logic input**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{IL}	Input low level voltage				0.9	V
I_{IL}	Low level input current	$V_{IN}= 0.9V$	1			µA
V_{IH}	Input high level voltage		2.1			V
I_{IH}	High level input current	$V_{IN}= 2.1V$			10	µA
$V_{I(hyst)}$	Input hysteresis voltage		0.25			V
V_{ICL}	Input clamp voltage	$I_{IN}= 1mA$ $I_{IN}= -1mA$	5.5	-0.7	7	V V
V_{CSDL}	CS_DIS low level voltage				0.9	V
I_{CSDL}	Low level CS_DIS current	$V_{CSD}= 0.9V$	1			µA
V_{CSDH}	CS_DIS high level voltage		2.1			V
I_{CSDH}	High level CS_DIS current	$V_{CSD}= 2.1V$			10	µA
$V_{CSD(hyst)}$	CS_DIS hysteresis voltage		0.25			V
V_{CSCL}	CS_DIS clamp voltage	$I_{CSD}= 1mA$ $I_{CSD}= -1mA$	5.5	-0.7	7	V V

Table 9. Protections and diagnostics[1]

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{limH}	Short circuit current	V_{CC}= 13V 5V<V_{CC}<36V	46	65	91 91	A A
I_{limL}	Short circuit current during thermal cycling	V_{CC}= 13V; T_R<T_j<T_{TSD}		24		A
T_{TSD}	Shutdown temperature		150	175	200	°C
T_R	Reset temperature		T_{RS}+1	T_{RS}+5		°C
T_{RS}	Thermal reset of STATUS		135			°C
T_{HYST}	Thermal hysteresis (T_{TSD}-T_R)			7		°C
V_{DEMAG}	Turn-off output voltage clamp	I_{OUT}=2A; V_{IN}=0; L=6mH	V_{CC}-41	V_{CC}-46	V_{CC}-52	V
V_{ON}	Output voltage drop limitation	I_{OUT}=0.5A (see *Figure 9*); T_j= -40°C...+150°C		25		mV

1. To ensure long term reliability under heavy overload or short circuit conditions, protection and related diagnostic signals must be used together with a proper software strategy. If the device is subjected to abnormal conditions, this software must limit the duration and number of activation cycles.

Table 10. Current sense (8V<V_{CC}<16V)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K_0	I_{OUT}/I_{SENSE}	I_{OUT}= 0.25A; V_{SENSE}=0.5V;V_{CSD}=0V; T_j= -40°C...150°C	2770	5490	8220	
K_1	I_{OUT}/I_{SENSE}	I_{OUT}= 6A; V_{SENSE}=0.5V; V_{CSD}=0V; T_j= -40°C...150°C I_{OUT}= 6A; V_{SENSE}=0.5V; V_{CSD}=0V; T_j= 25°C...150°C	3610 3930	4580 4580	5630 5230	
dK_1/K_1[1]	Current sense ratio drift	I_{OUT}= 6A; V_{SENSE}= 0.5V; V_{CSD}= 0V; T_J= -40 °C to 150 °C	-8		+8	%
K_2	I_{OUT}/I_{SENSE}	I_{OUT}= 10A; V_{SENSE}=4V; V_{CSD}=0V; T_j=-40°C...150°C I_{OUT}= 10A; V_{SENSE}=4V; V_{CSD}=0V; T_j=25°C...150°C	4000 4180	4570 4570	5220 4960	
dK_2/K_2[1]	Current sense ratio drift	I_{OUT}= 10A; V_{SENSE}= 4V; V_{CSD}=0V; T_J= -40 °C to 150 °C	-5		+5	%
K_3	I_{OUT}/I_{SENSE}	I_{OUT}= 25A; V_{SENSE}=4V; V_{CSD}=0V; T_j= -40°C...150°C I_{OUT}= 25A; V_{SENSE}=4V; V_{CSD}=0V; T_j= 25°C...150°C	4480 4500	4660 4660	4980 4820	

Table 10.   Current sense (8V<V_{CC}<16V) (continued)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
dK_3/K_3[(1)]	Current sense ratio drift	I_{OUT}= 25A; V_{SENSE}= 4V; V_{CSD}=0V; T_J= -40 °C to 150 °C	-3		+3	%
I_{SENSE0}	Analog sense leakage current	I_{OUT}= 0A; V_{SENSE}=0V; V_{CSD}= 5V; V_{IN}=0V; T_j= -40°C...150°C V_{CSD}= 0V; V_{IN}=5V; T_j= -40°C...150°C I_{OUT}= 2A; V_{SENSE}=0V; V_{CSD}=5V; V_{IN}=5V; T_j= -40°C...150°C	0 0 0		1 2 1	µA µA µA
I_{OL}	Openload on-state current detection threshold	V_{IN} = 5V, I_{SENSE}= 5 µA	10		45	mA
V_{SENSE}	Max analog sense output voltage	I_{OUT}=15A; V_{CSD}=0V;	5			V
V_{SENSEH}	Analog sense output voltage in over temperature condition	V_{CC}= 13V; R_{SENSE}= 3.9KΩ		9		V
I_{SENSEH}	Analog sense output current in over temperature condition	V_{CC}= 13V; V_{SENSE}= 5V		8		mA
$t_{DSENSE1H}$	Delay response time from falling edge of CS_DIS pin	V_{SENSE}<4V, 1.5A<Iout<25A I_{SENSE}= 90% of $I_{SENSE\ max}$ (see *Figure 4*)		50	100	µs
$t_{DSENSE1L}$	Delay response time from rising edge of CS_DIS pin	V_{SENSE}<4V, 1.5A<Iout<25A I_{SENSE}= 10% of $I_{SENSE\ max}$ (see *Figure 4*)		5	20	µs
$t_{DSENSE2H}$	Delay response time from rising edge of INPUT pin	V_{SENSE}<4V, 1.5A<Iout<25A I_{SENSE}= 90% of $I_{SENSE\ max}$ (see *Figure 4*)		270	500	µs
$\Delta t_{DSENSE2H}$	Delay response time between rising edge of output current and rising edge of current sense	V_{SENSE} < 4V, I_{SENSE} = 90% of $I_{SENSEMAX}$, I_{OUT} = 90% of I_{OUTMAX} I_{OUTMAX}=15A (see *Figure 5*)			310	□□µs
$t_{DSENSE2L}$	Delay response time from falling edge of INPUT pin	V_{SENSE}<4V, 1.5A<Iout<25A I_{SENSE}=10% of $I_{SENSE\ max}$ (see *Figure 4*)		100	250	µs

1. Parameter guaranteed by design; it is not tested.

Figure 4. Current sense delay characteristics
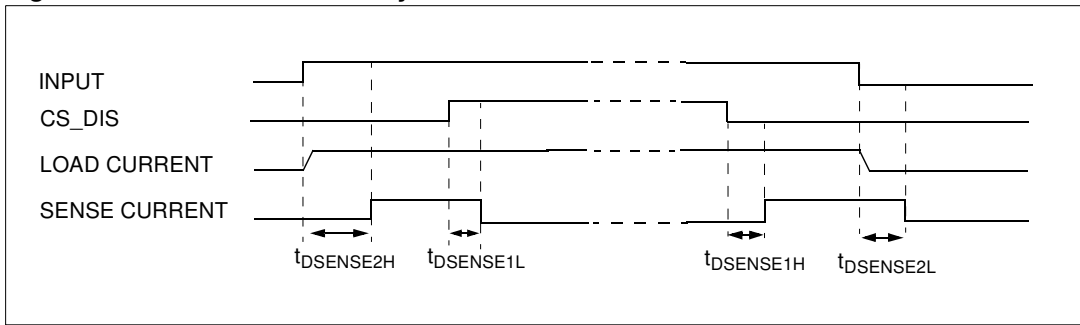
Figure 5. Delay response time between rising edge of output current and rising edge of Current Sense (CS enabled)
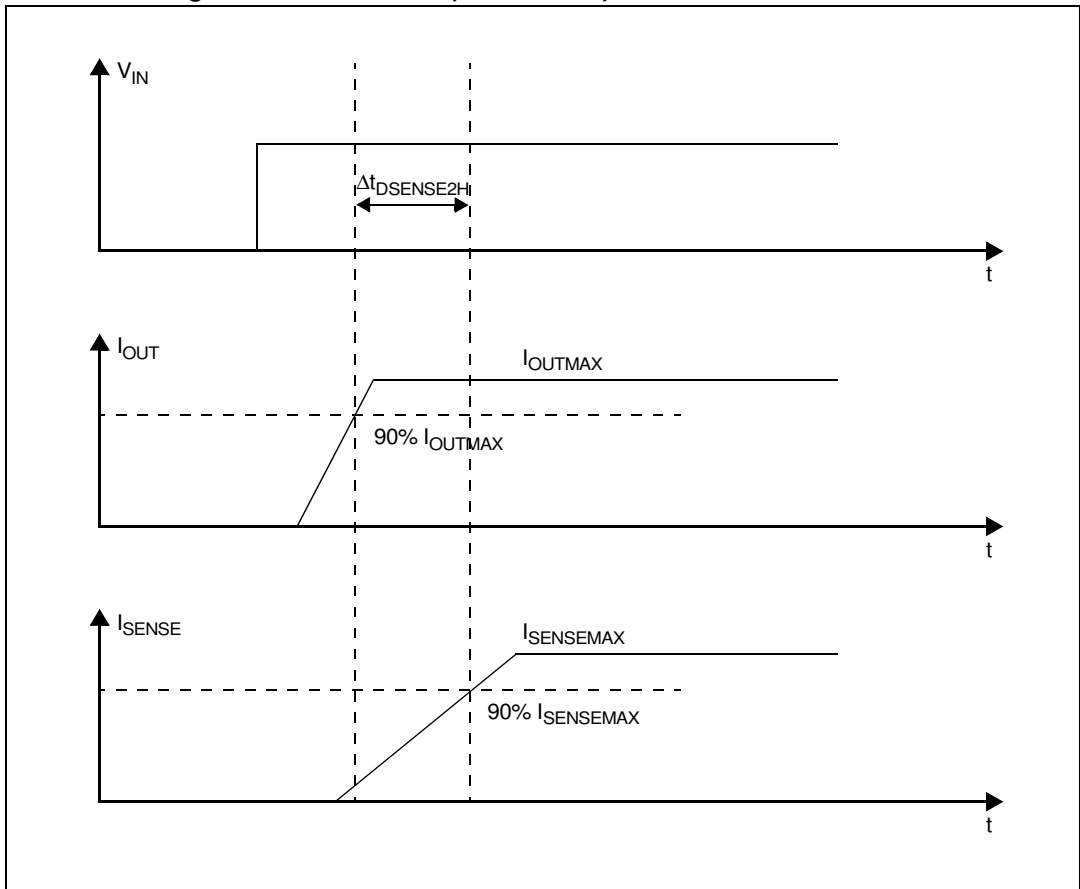
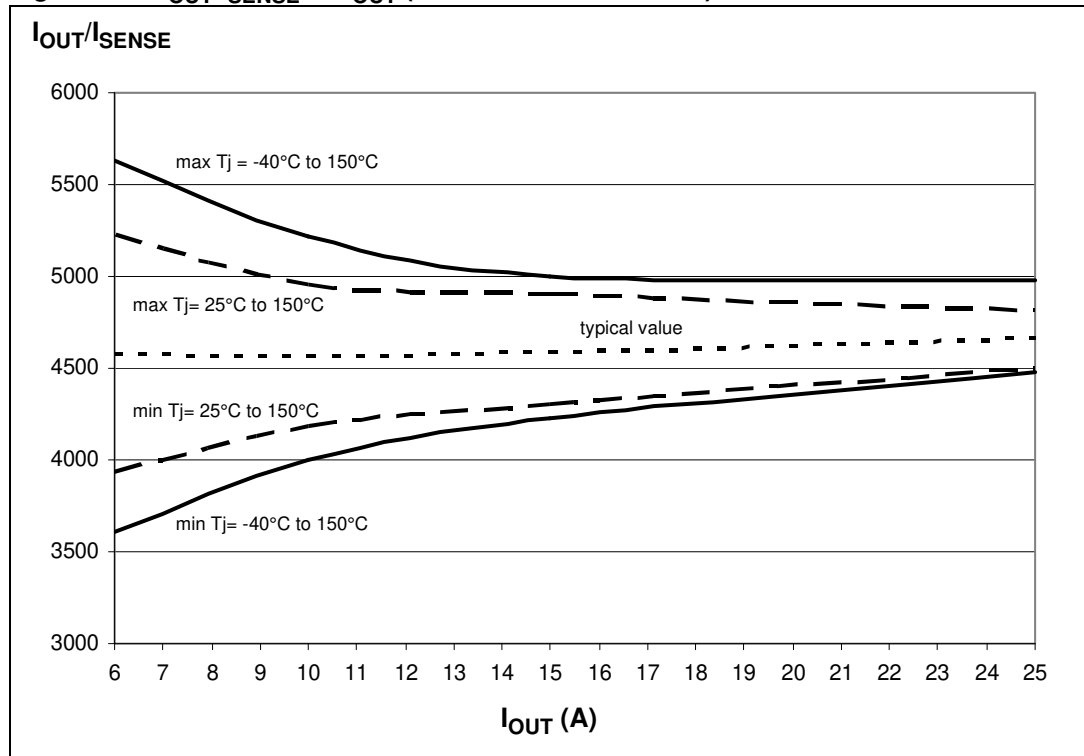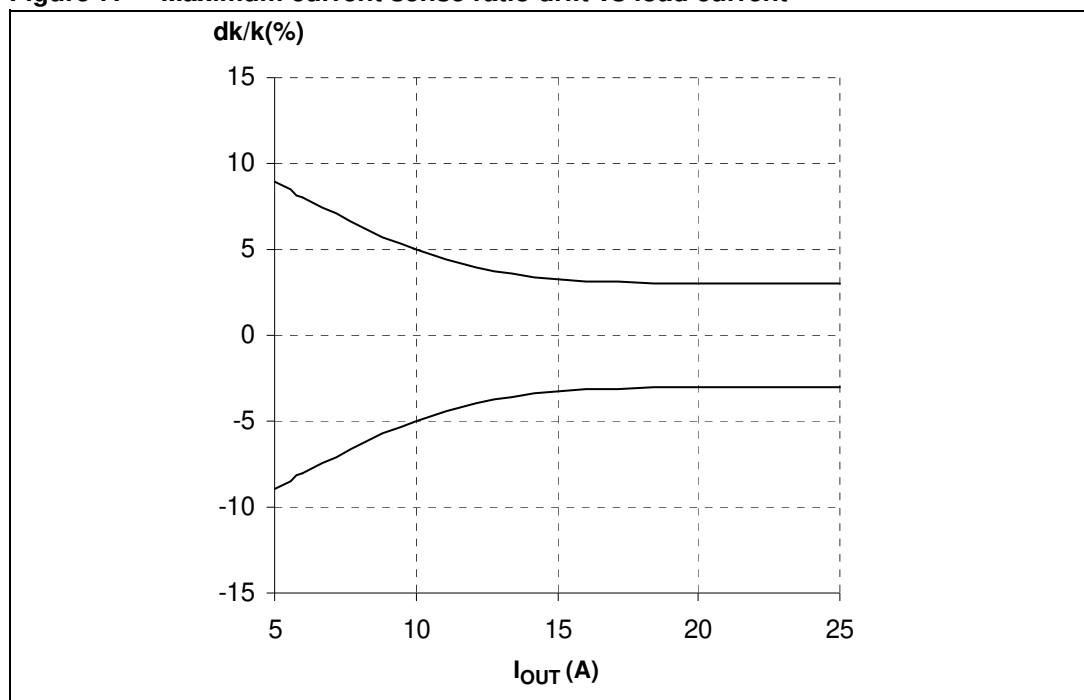**Figure 6.** I_{OUT}/I_{SENSE} vs I_{OUT} (see *Table 10* for details)

Figure 7. Maximum current sense ratio drift vs load current

Note: *Parameter guaranteed by design; it is not tested.*

Table 11. Truth table

Conditions	Input	Output	Sense (V$_{CSD}$=0V)[1]
Normal operation	L H	L H	0 Nominal
Overtemperature	L H	L L	0 V$_{SENSEH}$
Undervoltage	L H	L L	0 0
Short circuit to GND (R$_{sc}$ ≤ 10 mΩ)	L H H	L L L	0 0 if T$_j$ < T$_{TSD}$ V$_{SENSEH}$ if T$_j$ > T$_{TSD}$
Short circuit to V$_{CC}$	L H	H H	0 < Nominal
Negative output voltage clamp	L	L	0

1. If the V$_{CSD}$ is high, the SENSE output is at a high impedance, its potential depends on leakage currents and external circuit.

Figure 8. Switching characteristics
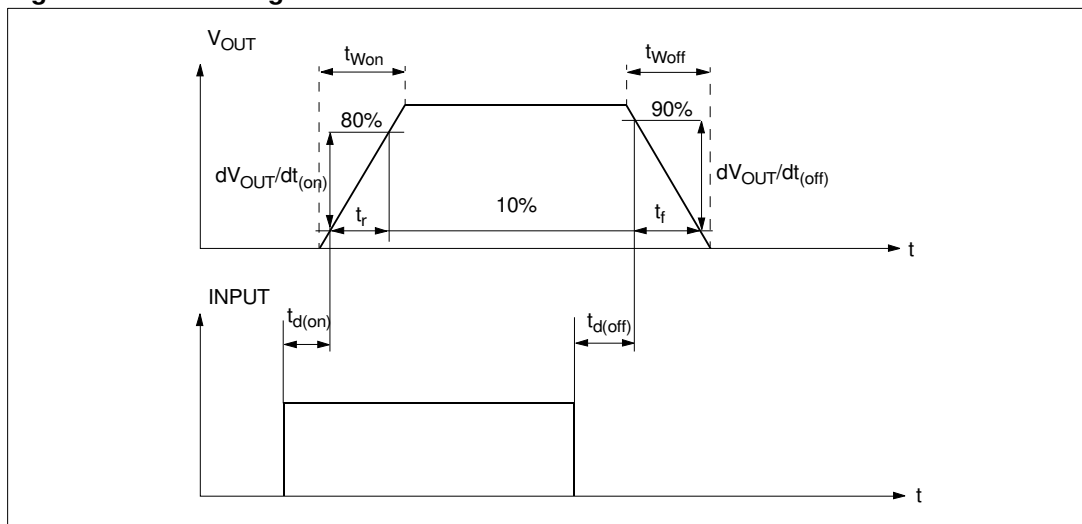
Figure 9. **Output voltage drop limitation**
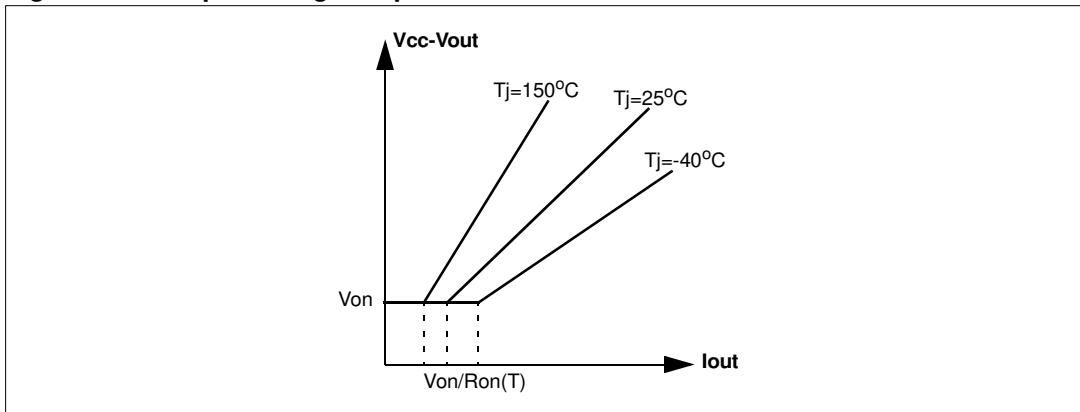
Table 12. **Electrical transient requirements (part 1/3)**

ISO 7637-2: 2004(E) Test pulse	Test levels [1]		Number of pulses or test times	Burst cycle/pulse repetition time		Delays and impedance
	III	IV				
1	-75V	-100V	5000 pulses	0.5 s	5 s	2 ms, 10 Ω
2a	+37V	+50V	5000 pulses	0.2 s	5 s	50 µs, 2 Ω
3a	-100V	-150V	1h	90 ms	100 ms	0.1 µs, 50 Ω
3b	+75V	+100V	1h	90 ms	100 ms	0.1 µs, 50 Ω
4	-6V	-7V	1 pulse			100 ms, 0.01Ω
5b [2]	+65V	+87V	1 pulse			400 ms, 2 Ω

Table 13. **Electrical transient requirements (part 2/3)**

ISO 7637-2: 2004(E) Test pulse	Test level results[1]	
	III	IV
1	C	C
2a	C	C
3a	C	C
3b	C	C
4	C	C
5b [2]	C	C

1. The above test levels must be considered referred to V_{CC} = 13.5V except for pulse 5b.

2. Valid in case of external load dump clamp: 40V maximum referred to ground.

Table 14. **Electrical transient requirements (part 3/3)**

Class	Contents
C	All functions of the device are performed as designed after exposure to disturbance.
E	One or more functions of the device are not performed as designed after exposure to disturbance and cannot be returned to proper operation without replacing the device.

Figure 10.   Waveforms
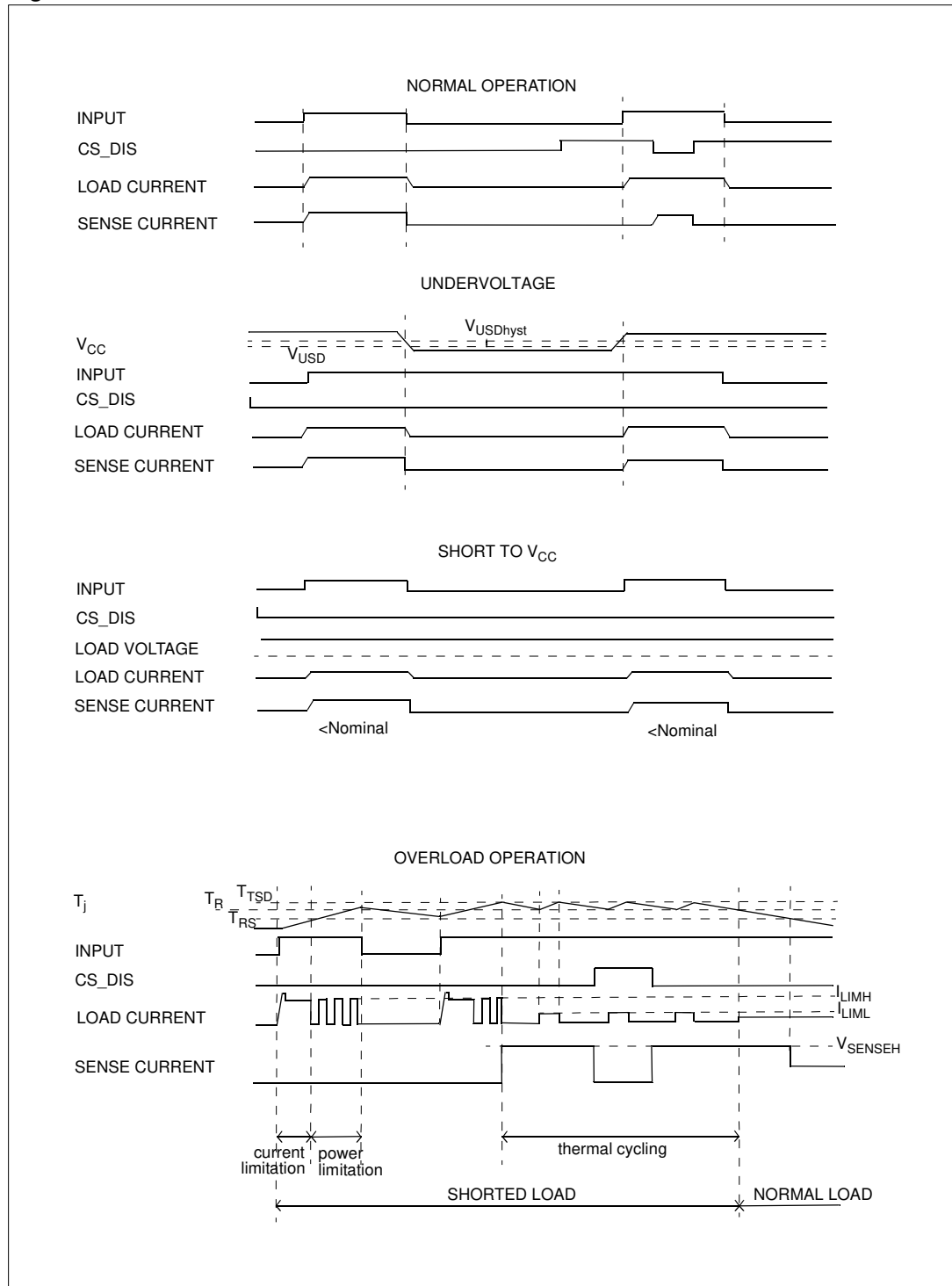
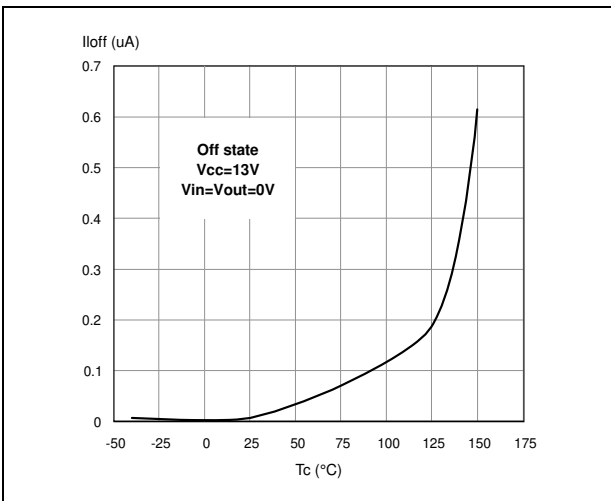## 2.3 Electrical characteristics curves

Figure 11. Off-state output current

Figure 12. High level input current
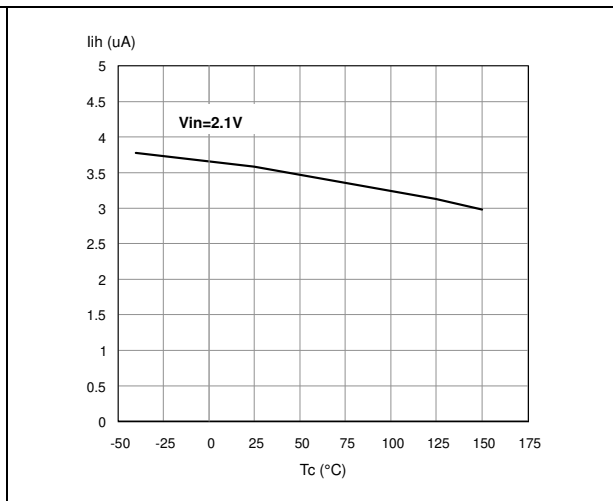
Figure 13. Input clamp voltage
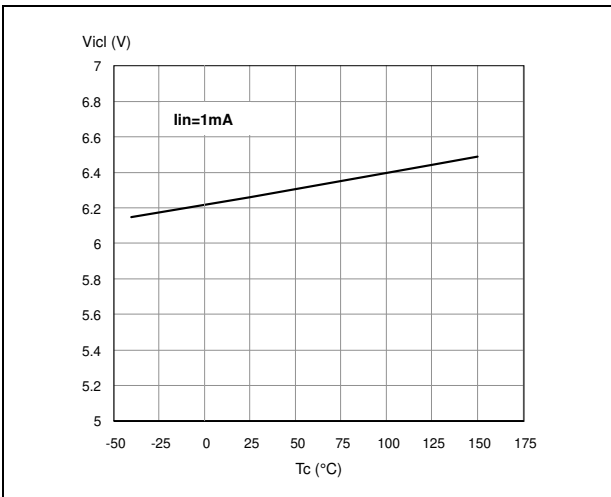
Figure 14. Input low level
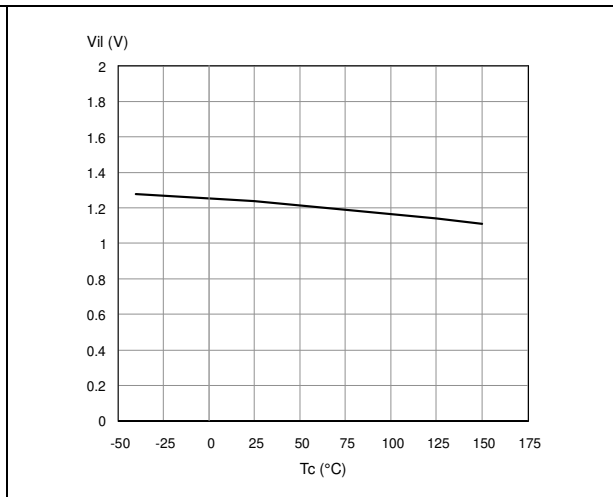
Figure 15. Input high level
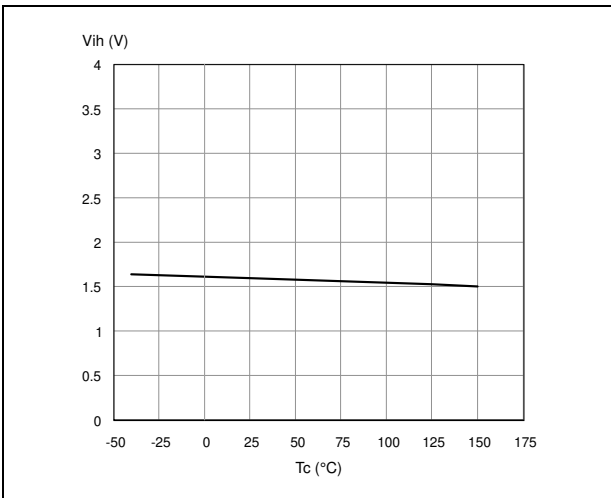
Figure 16. Input hysteresis voltage
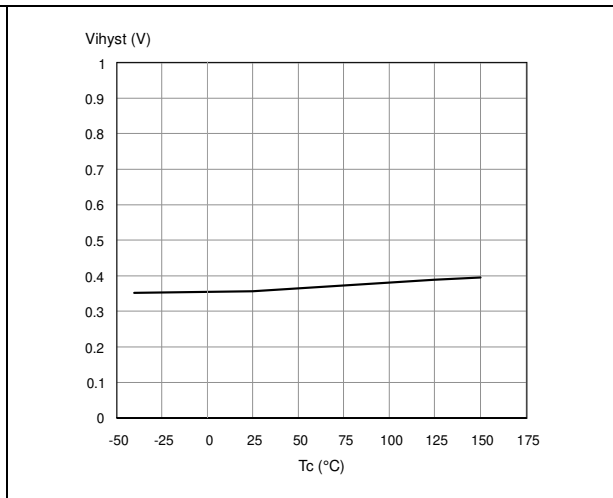
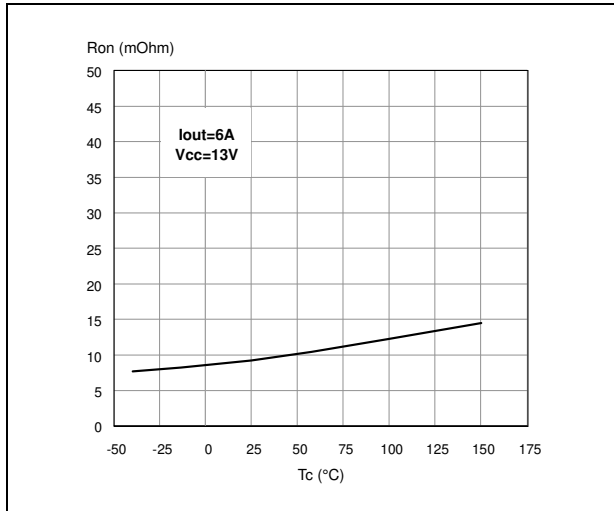**Figure 17. On-state resistance vs T$_{case}$**
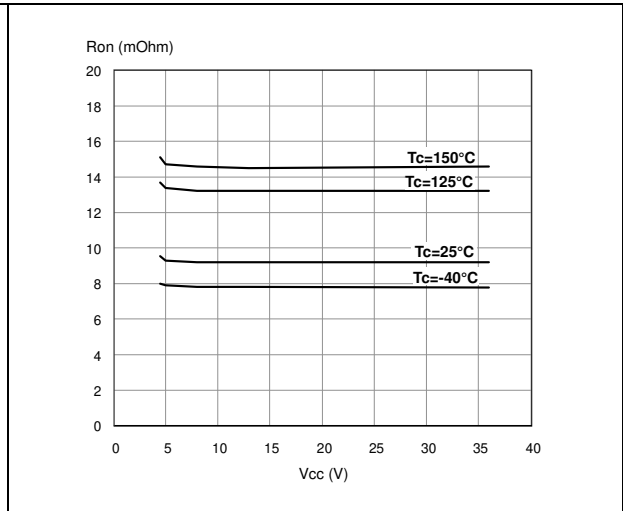

Figure 18. On-state resistance vs V$_{CC}$

Figure 19. Undervoltage shutdown
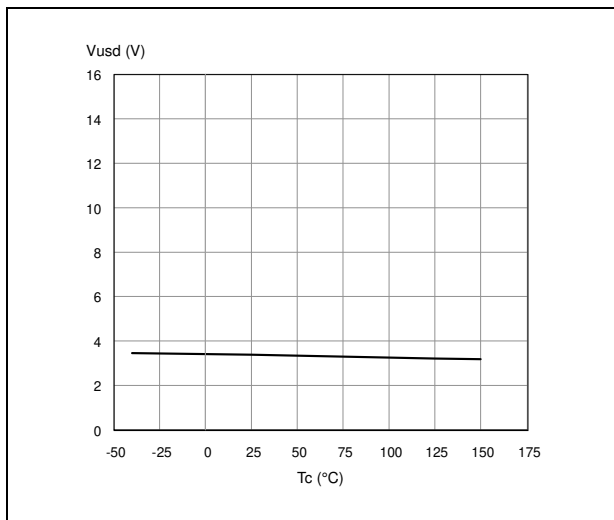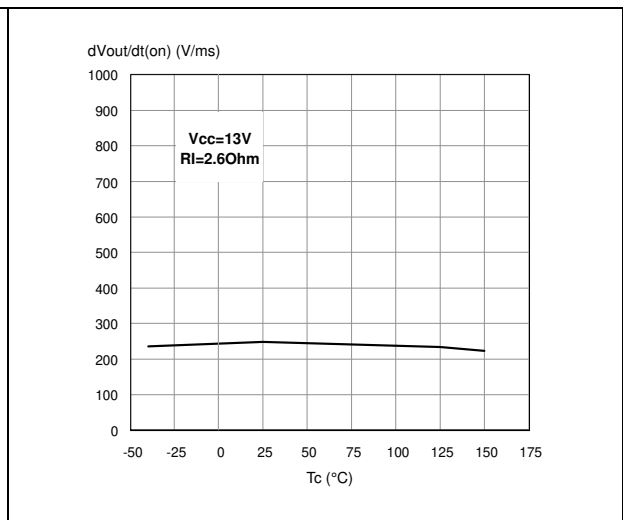

Figure 20. Turn-on voltage slope
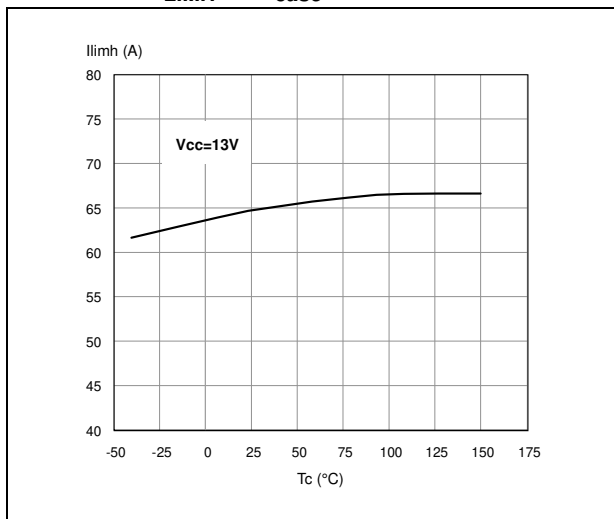

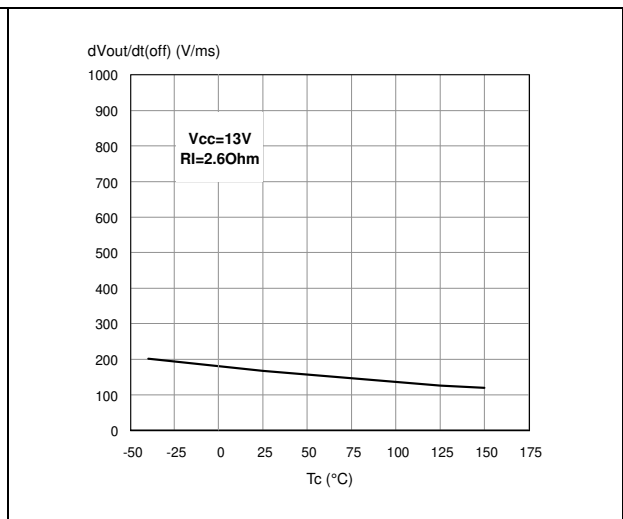**Figure 21. I$_{LIMH}$ vs T$_{case}$**

Figure 22. Turn-off voltage slope
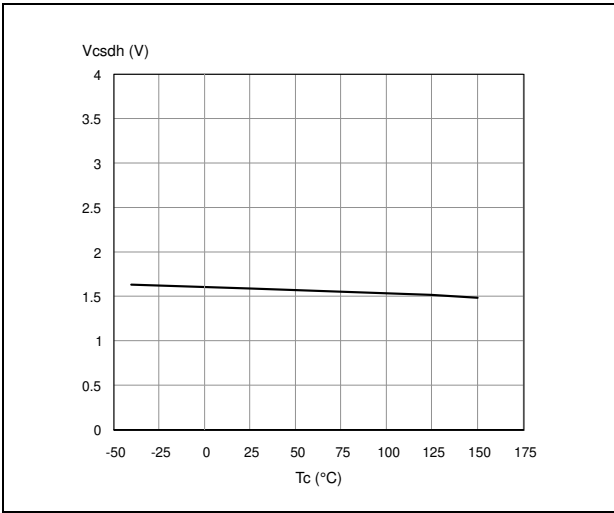
Figure 23. CS_DIS high level voltage

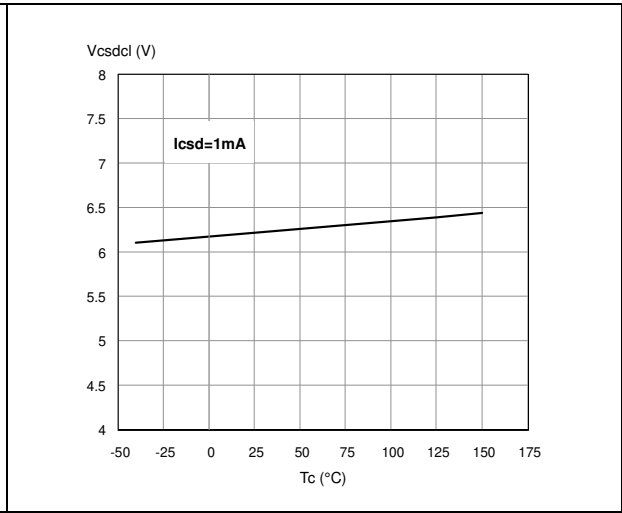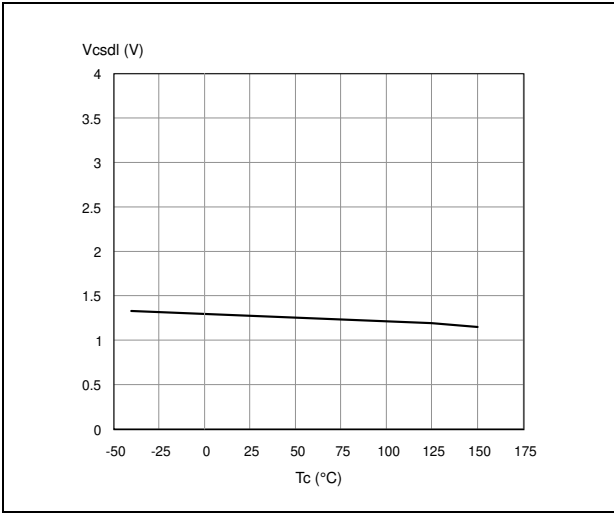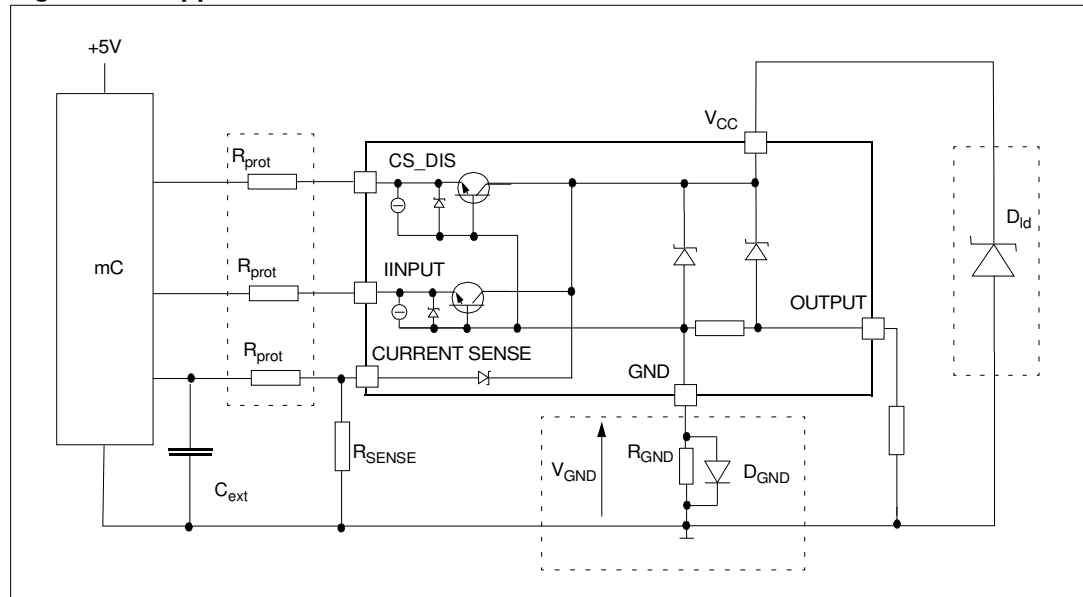**Figure 24. CS_DIS clamp voltage**

Figure 25. CS_DIS low level voltage

3 Application information

Figure 26. Application schematic

3.1 GND protection network against reverse battery

3.1.1 Solution 1: resistor in the ground line (R_{GND} only)

This can be used with any type of load.

The following is an indication on how to dimension the R_{GND} resistor.

1. $R_{GND} \leq 600$ mV / ($I_{S(on)max}$).
2. $R_{GND} \geq (-V_{CC}) / (-I_{GND})$

where $-I_{GND}$ is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device datasheet.

Power Dissipation in R_{GND} (when $V_{CC}<0$: during reverse battery situations) is:

$P_D = (-V_{CC})^2 / R_{GND}$

This resistor can be shared amongst several different HSDs. Please note that the value of this resistor should be calculated with formula (1) where $I_{S(on)max}$ becomes the sum of the maximum on-state currents of the different devices.

Please note that if the microprocessor ground is not shared by the device ground then the R_{GND} will produce a shift ($I_{S(on)max} * R_{GND}$) in the input thresholds and the status output values. This shift will vary depending on how many devices are ON in the case of several high side drivers sharing the same R_{GND}.

If the calculated power dissipation leads to a large resistor or several devices have to share the same resistor then ST suggests to utilize Solution 2 (see below).

3.1.2 Solution 2: diode (D$_{GND}$) in the ground line

A resistor (R$_{GND}$=1 kΩ) should be inserted in parallel to D$_{GND}$ if the device drives an inductive load.

This small signal diode can be safely shared amongst several different HSDs. Also in this case, the presence of the ground network will produce a shift ($\approx$600 mV) in the input threshold and in the status output values if the microprocessor ground is not common to the device ground. This shift will not vary if more than one HSD shares the same diode/resistor network.

3.2 Load dump protection

D$_{ld}$ is necessary (voltage transient suppressor) if the load dump peak voltage exceeds the V$_{CC}$ max DC rating. The same applies if the device is subject to transients on the V$_{CC}$ line that are greater than the ones shown in the ISO 7637-2: 2004(E) table.

3.3 Microcontroller I/Os protection

If a ground protection network is used and negative transient are present on the V$_{CC}$ line, the control pins will be pulled negative. ST suggests to insert a resistor (R$_{prot}$) in line to prevent the microcontroller I/Os pins to latch-up.

The value of these resistors is a compromise between the leakage current of microcontroller and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of microcontroller I/Os.

$-V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C}-V_{IH}-V_{GND}) / I_{IHmax}$

Calculation example:

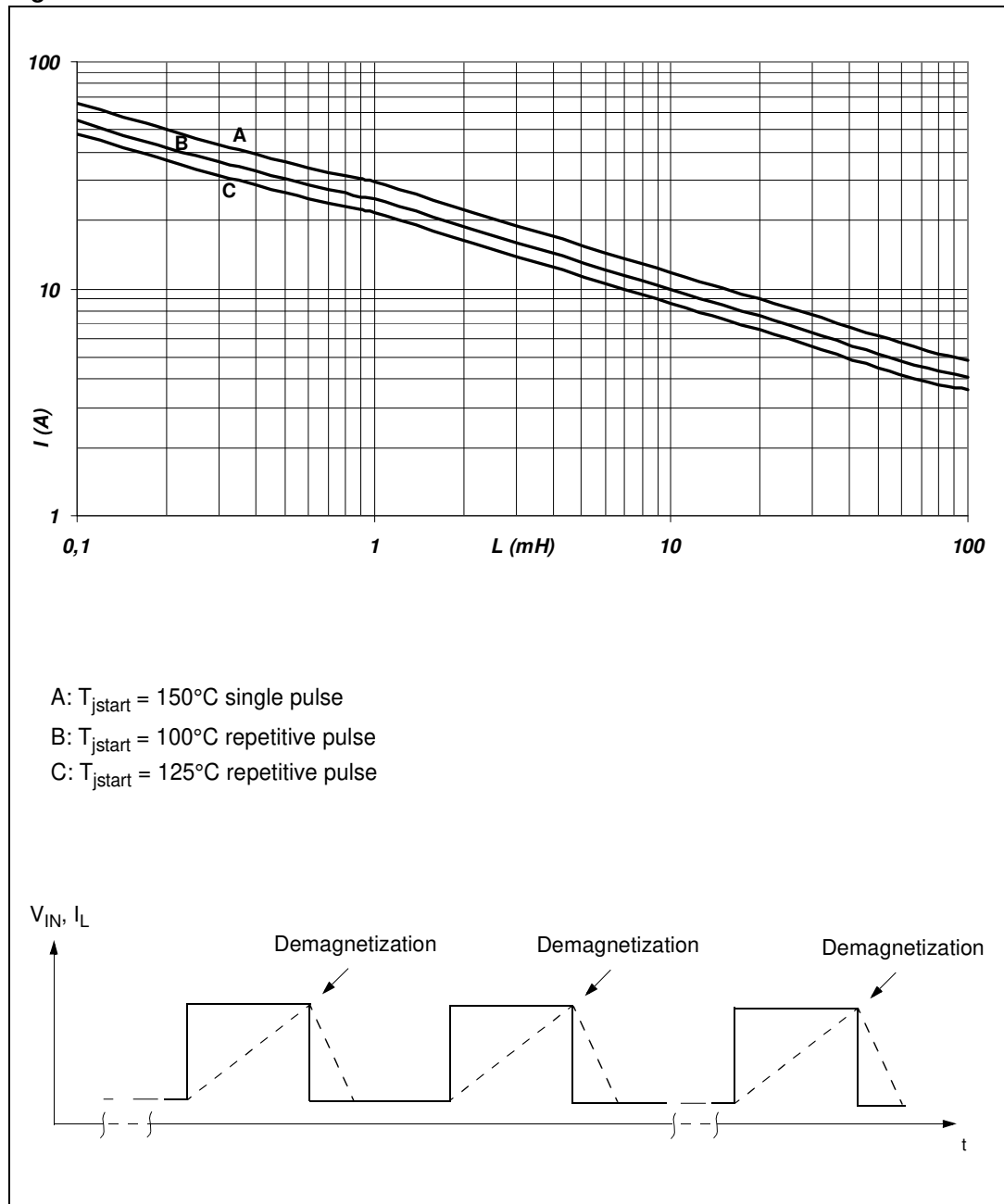For V$_{CCpeak}$= - 100 V and I$_{latchup}$ ≥ 20 mA; V$_{OH\mu C}$ ≥ 4.5 V

5 kΩ ≤ R$_{prot}$ ≤ 180 kΩ

Recommended values: R$_{prot}$ =10 kΩ, C$_{EXT}$=10 nF.

3.4 Maximum demagnetization energy (V_{CC}=13.5V)

Figure 27. Maximum turn-off current versus load inductance

A: T_{jstart} = 150°C single pulse

B: T_{jstart} = 100°C repetitive pulse
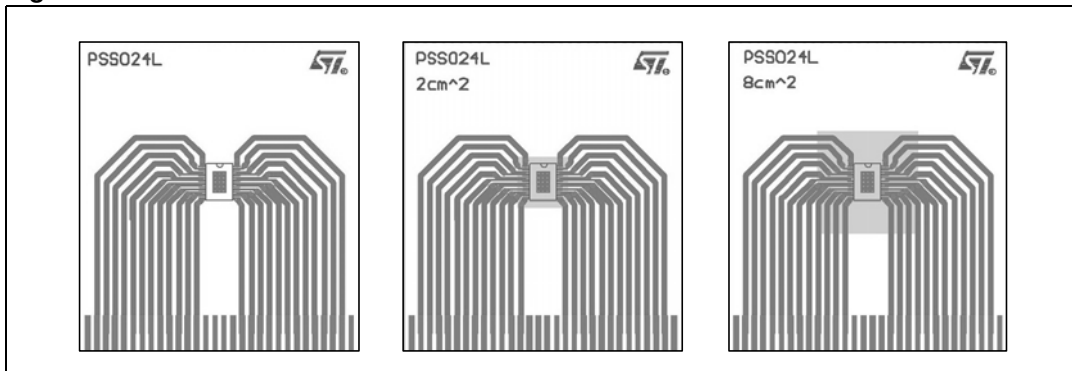
C: T_{jstart} = 125°C repetitive pulse

Note: *Values are generated with R_L=0 Ω.*

In case of repetitive pulses, T_{jstart} (at beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves B and C.

4 Package and PCB thermal data

4.1 PowerSSO-24™ thermal data

Figure 28. PowerSSO-24™ PC board

Note: *Layout condition of R_{th} and Z_{th} measurements (PCB: Double layer, Thermal Vias, FR4 area= 77 mm x 86 mm, PCB thickness=1.6mm, Cu thickness=70 µm (front and back side), Copper areas: from minimum pad lay-out to 8 cm^2).*

Figure 29. $R_{thj\text{-}amb}$ vs PCB copper area in open box free air condition
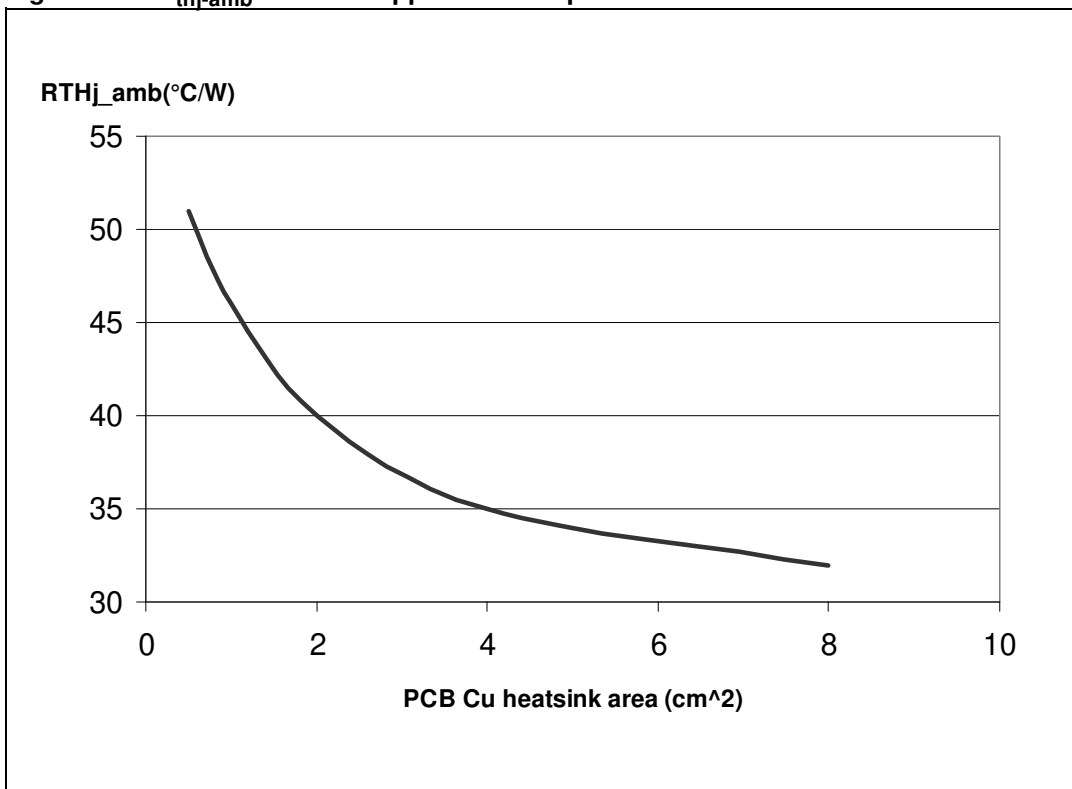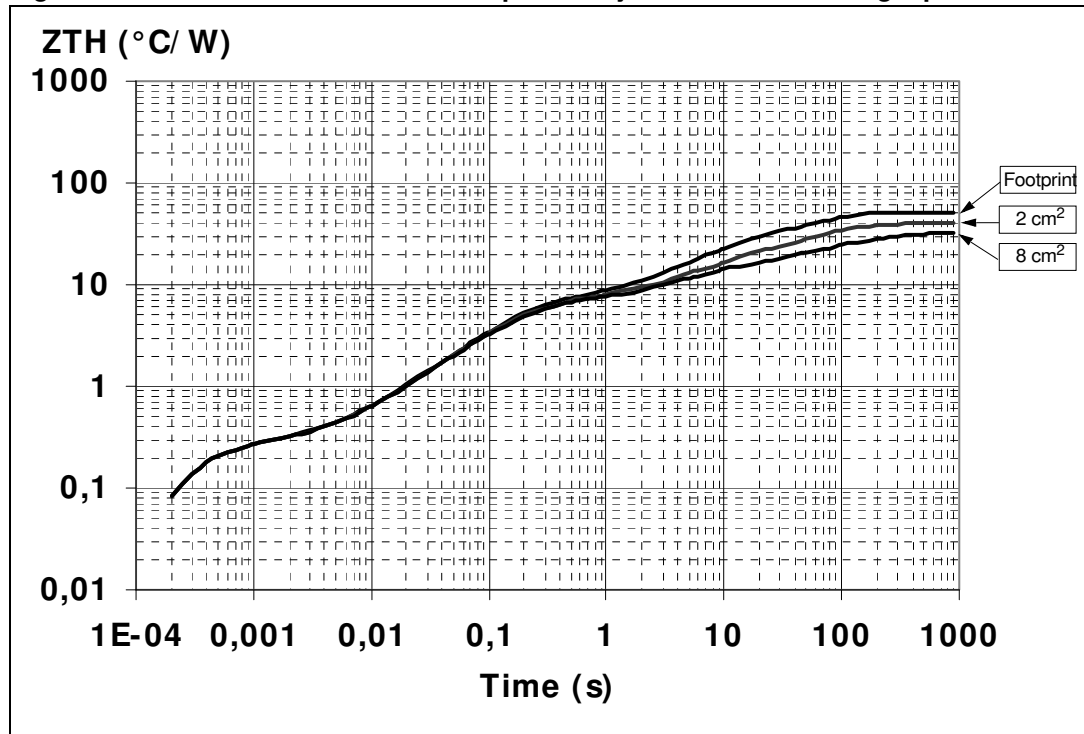
Figure 30. PowerSSO-24™ thermal impedance junction ambient single pulse
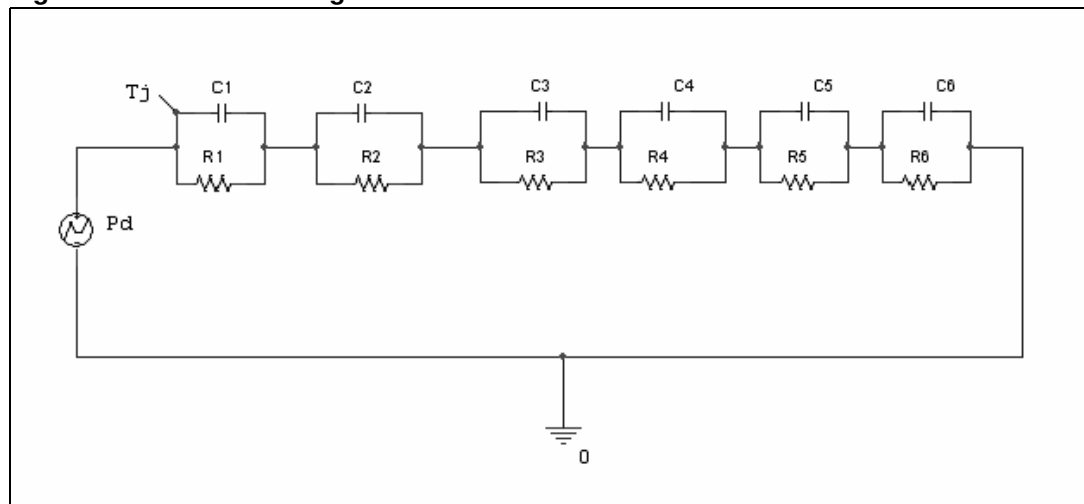

Equation 1: pulse calculation formula

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp}(1 - \delta)$$

where $\delta = t_P/T$

Figure 31. Thermal fitting model of a double channel HSD in PowerSSO-24™(a)

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a. The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.