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## Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832

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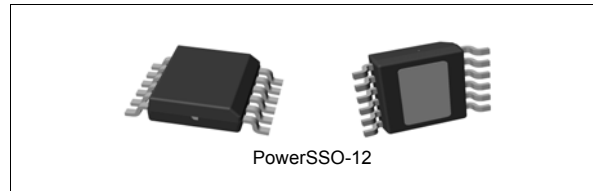


### Features

Max supply voltage	$V_{CC}$	41V
Operating voltage range	$V_{CC}$	4.5 to 36V
Max On-State resistance (per ch.)	$R_{ON}$	50 m $\Omega$
Current limitation (typ)	$I_{LIMH}$	19 A
Off state supply current	$I_S$	2 $\mu$ A <sup>(1)</sup>

(1) Typical value with all loads connected.

- Main
  - Inrush current active management by power limitation
  - Very low stand-by current
  - 3.0V CMOS compatible input
  - Optimized electromagnetic emission
  - Very low electromagnetic susceptibility
  - In compliance with the 2002/95/EC European directive
- Diagnostic functions
  - Open drain status output
  - On state open load detection
  - Off state open load detection
  - Thermal shutdown indication
- Protections
  - Undervoltage shut-down
  - Overvoltage clamp
  - Output stuck to  $V_{CC}$  detection
  - Load current limitation
  - Self limiting of fast thermal transients
  - Protection against loss of ground and loss of  $V_{CC}$
  - Thermal shut down
  - Reverse battery protection (see [Figure 27](#))
  - Electrostatic discharge protection



### Application

- All types of resistive, inductive and capacitive loads

### Description

The VN5050J-E is a monolithic device made using STMicroelectronics VIPower technology. It is intended for driving resistive or inductive loads with one side connected to ground. Active  $V_{CC}$  pin voltage clamp protects the device against low energy spikes (see ISO7637 transient compatibility table). The device detects open load condition both in on and off state, when STAT\_DIS is left open or driven low. Output shorted to  $V_{CC}$  is detected in the off state.

When STAT\_DIS is driven high, the STATUS pin is in a high impedance condition. Output current limitation protects the device in overload condition. In case of long duration overload, the device limits the dissipated power to safe level up to thermal shut-down intervention. Thermal shut-down with automatic restart allows the device to recover normal operation as soon as fault condition disappears.

**Table 1. Device summary**

Package	Order codes	
	Tube	Tape & Reel
PowerSSO-12	VN5050J-E	VN5050JTR-E

# Contents

- 1      Block diagram and pin description ..... 5**
  
- 2      Electrical specifications ..... 7**
  - 2.1 Absolute maximum ratings ..... 7
  - 2.2 Thermal data ..... 8
  - 2.3 Electrical characteristics ..... 9
  - 2.4 Electrical characteristics curves ..... 16
  
- 3      Application information ..... 20**
  - 3.1 GND protection network against reverse battery ..... 20
    - 3.1.1 Solution 1: resistor in the ground line (RGND only) ..... 20
    - 3.1.2 Solution 2: diode (DGND) in the ground line ..... 21
  - 3.2 Load dump protection ..... 21
  - 3.3 MCU I/Os protection ..... 21
  - 3.4 Open load detection in Off state ..... 21
  - 3.5 Maximum demagnetization energy (VCC = 13.5V) ..... 23
  
- 4      Package and PCB thermal data ..... 24**
  - 4.1 PowerSSO-12™ thermal data ..... 24
  
- 5      Package and packing information ..... 27**
  - 5.1 ECOPACK® packages ..... 27
  - 5.2 Package mechanical data ..... 27
  - 5.3 Packing information ..... 29
  
- 6      Revision history ..... 30**

## List of tables

Table 2.	Pin function . . . . .	5
Table 3.	Suggested connections for unused and N.C. pins . . . . .	6
Table 4.	Absolute maximum ratings . . . . .	7
Table 5.	Thermal data . . . . .	8
Table 6.	Power section . . . . .	9
Table 7.	Switching ( $V_{CC} = 13V$ ; $T_j = 25^{\circ}C$ ) . . . . .	9
Table 8.	Status pin ( $V_{SD}=0V$ ) . . . . .	10
Table 9.	Protections . . . . .	10
Table 10.	Openload detection . . . . .	11
Table 11.	Logic input . . . . .	11
Table 12.	Truth table . . . . .	12
Table 13.	Electrical transient requirements . . . . .	14
Table 14.	Thermal parameter . . . . .	26
Table 15.	PowerSSO-12™ mechanical data . . . . .	28
Table 16.	Document revision history . . . . .	30

## List of figures

Figure 1.	Block diagram . . . . .	5
Figure 2.	Configuration diagram (top view) . . . . .	6
Figure 3.	Current and voltage conventions . . . . .	7
Figure 4.	Status timings . . . . .	12
Figure 5.	Switching characteristics . . . . .	13
Figure 6.	Output voltage drop limitation . . . . .	13
Figure 7.	Waveforms . . . . .	15
Figure 8.	Off state output current . . . . .	16
Figure 9.	High level input current . . . . .	16
Figure 10.	Input clamp voltage . . . . .	16
Figure 11.	Input high level . . . . .	16
Figure 12.	Input low level . . . . .	16
Figure 13.	Input hysteresis voltage . . . . .	16
Figure 14.	Status low output voltage . . . . .	17
Figure 15.	Status leakage current . . . . .	17
Figure 16.	On state resistance vs $V_{CC}$ . . . . .	17
Figure 17.	Status clamp voltage . . . . .	17
Figure 18.	Openload On state detection threshold . . . . .	17
Figure 19.	Openload Off state voltage detection threshold . . . . .	18
Figure 20.	$I_{LIM}$ vs $T_{case}$ . . . . .	18
Figure 21.	Turn-On voltage slope . . . . .	18
Figure 22.	Undervoltage shutdown . . . . .	18
Figure 23.	Turn-Off voltage slope . . . . .	18
Figure 24.	STAT_DIS clamp voltage . . . . .	18
Figure 25.	High level STAT_DIS voltage . . . . .	19
Figure 26.	Low level STAT_DIS voltage . . . . .	19
Figure 27.	Application schematic . . . . .	20
Figure 28.	Open load detection in Off state . . . . .	22
Figure 29.	Maximum turn Off current versus inductance . . . . .	23
Figure 30.	PowerSSO-12™ PC Board . . . . .	24
Figure 31.	$R_{thj-amb}$ Vs. PCB copper area in open box free air condition . . . . .	24
Figure 32.	PowerSSO-12™ thermal impedance junction ambient single pulse . . . . .	25
Figure 33.	Thermal fitting model of a single channel HSD in PowerSSO-12™ . . . . .	25
Figure 34.	PowerSSO-12™ package dimensions . . . . .	27
Figure 35.	PowerSSO-12™ tube shipment (no suffix) . . . . .	29
Figure 36.	PowerSSO-12™ tape and reel shipment (suffix "TR") . . . . .	29

# 1 Block diagram and pin description

Figure 1. Block diagram

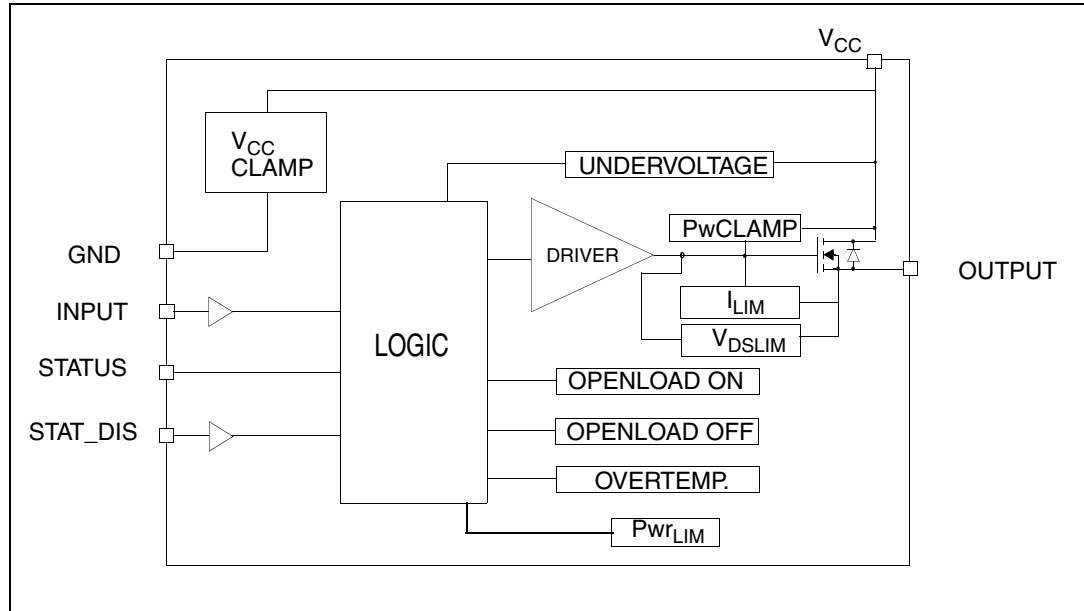
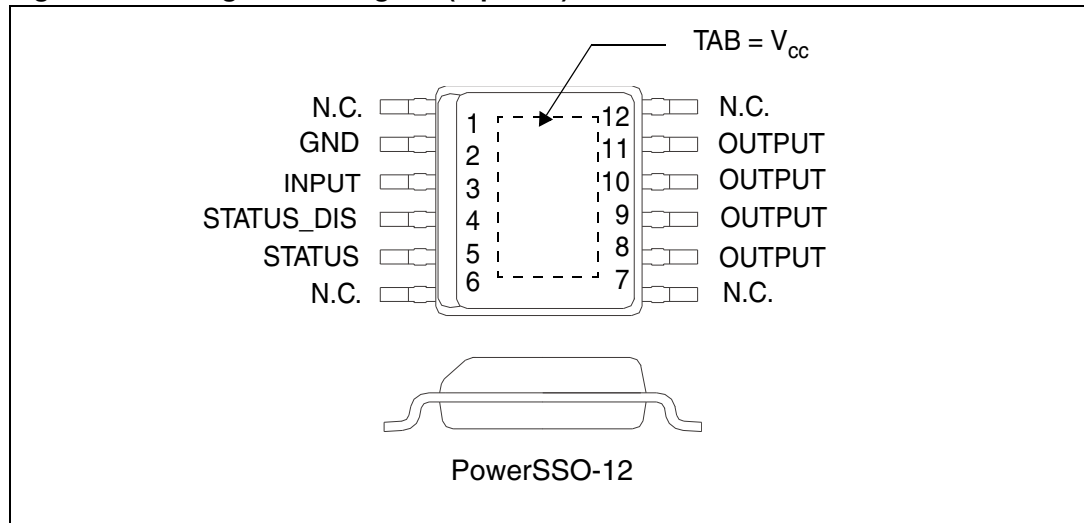


Table 2. Pin function

Name	Function
V <sub>CC</sub>	Battery connection.
OUTPUT	Power output.
GND	Ground connection. Must be reverse battery protected by an external diode/resistor network.
INPUT	Voltage controlled input pin with hysteresis, CMOS compatible. Controls output switch state.
STATUS	Open drain digital diagnostic pin.
STAT_DIS	Active high CMOS compatible pin, to disable the STATUS pin.

Figure 2. Configuration diagram (top view)



*Note:* The above pin configuration reflects the changes notified with PCN-APG-BOD/07/2886. The new pinout is backward compatible with existing PCB layouts where pins #1 and #6 are connected to Vcc and/or pins #7 and 12 are connected to OUTPUT. For new PCB designs, these pins should be left unconnected.

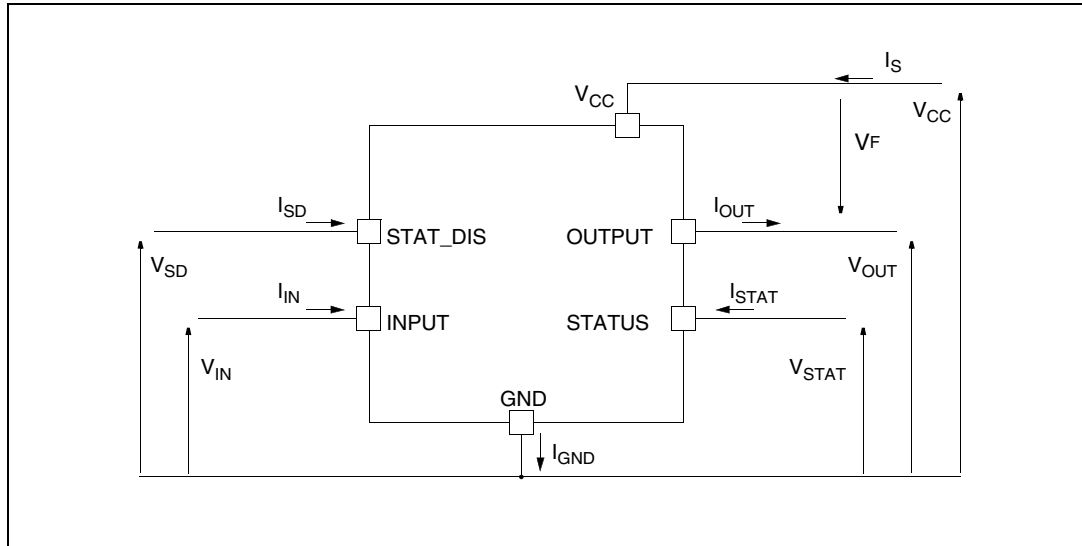
Table 3. Suggested connections for unused and N.C. pins

Connection / Pin	STATUS	N.C.	OUTPUT	INPUT	STAT_DIS
Floating	X	X	X	X	X
To ground	N.R. <sup>(1)</sup>	X	N.R.	Through 10KΩ resistor	Through 10KΩ resistor

(1) Not recommended.

## 2 Electrical specifications

Figure 3. Current and voltage conventions



Note:  $V_F = V_{OUT} - V_{CC}$  during reverse battery condition.

### 2.1 Absolute maximum ratings

Stressing the device above the rating listed in the “Absolute maximum ratings” table may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is not implied. Exposure to the conditions in table below for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE Program and other relevant quality document.

Table 4. Absolute maximum ratings

Symbol	Parameter	Value	Unit
$V_{CC}$	DC supply voltage	41	V
$-V_{CC}$	Reverse DC supply voltage	0.3	V
$-I_{GND}$	DC reverse ground pin current	200	mA
$I_{OUT}$	DC output current	Internally limited	A
$-I_{OUT}$	Reverse DC output current	12	A
$I_{IN}$	DC input current	+10 / -1	mA
$I_{STAT}$	DC status current	+10 / -1	mA
$I_{STAT\_DIS}$	DC status disable current	+10 / -1	mA
$E_{MAX}$	Maximum switching energy ( $L=3\text{mH}$ ; $R_L=0\Omega$ ; $V_{bat}=13.5\text{V}$ ; $T_{jstart}=150^\circ\text{C}$ ; $I_{OUT} = I_{limL}(Typ.)$ )	104	mJ



**Table 4. Absolute maximum ratings (continued)**

Symbol	Parameter	Value	Unit
$V_{ESD}$	Electrostatic discharge (Human Body Model: R=1.5K $\Omega$ ; C=100pF)		
$V_{ESD}$	Charge device model (CDM-AEC-Q100-011)	750	V
$T_j$	Junction operating temperature	-40 to 150	$^{\circ}\text{C}$
$T_{stg}$	Storage temperature	- 55 to 150	$^{\circ}\text{C}$

## 2.2 Thermal data

**Table 5. Thermal data**

Symbol	Parameter	Value	Unit
$R_{thj-case}$	Thermal resistance junction-case (Max.) (with one channel ON)	2.7	$^{\circ}\text{C}/\text{W}$
$R_{thj-amb}$	Thermal resistance junction-ambient (Max.)	See <a href="#">Figure 31</a>	$^{\circ}\text{C}/\text{W}$

## 2.3 Electrical characteristics

Values specified in this section are for  $8V < V_{CC} < 36V$ ;  $-40^{\circ}C < T_j < 150^{\circ}C$ , unless otherwise specified.

**Table 6. Power section**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{CC}$	Operating supply voltage		4.5	13	36	V
$V_{USD}$	Undervoltage shutdown			3.5	4.5	V
$V_{USDhyst}$	Undervoltage shut-down hysteresis			0.5		V
$R_{ON}$	On state resistance <sup>(2)</sup>	$I_{OUT}=1A$ ; $T_j=25^{\circ}C$			50	m $\Omega$
		$I_{OUT}=1A$ ; $T_j=150^{\circ}C$			100	m $\Omega$
		$I_{OUT}=1A$ ; $V_{CC}=5V$ ; $T_j=25^{\circ}C$			65	m $\Omega$
$V_{clamp}$	Clamp voltage	$I_S=20mA$	41	46	52	V
$I_S$	Supply current	Off State; $V_{CC}=13V$ ; $T_j=25^{\circ}C$ ; $V_{IN}=V_{OUT}=V_{SENSE}=V_{CSD}=0V$		2 <sup>(1)</sup>	5 <sup>(1)</sup>	$\mu A$
		On State; $V_{CC}=13V$ ; $V_{IN}=5V$ ; $I_{OUT}=0A$		1.9	3.5	mA
$I_{L(off1)}$	Off state output current <sup>(2)</sup>	$V_{IN}=V_{OUT}=0V$ ; $V_{CC}=13V$ ; $T_j=25^{\circ}C$	0	0.01	3	$\mu A$
		$V_{IN}=V_{OUT}=0V$ ; $V_{CC}=13V$ ; $T_j=125^{\circ}C$	0		5	
$I_{L(off2)}$	Off state output current <sup>(2)</sup>	$V_{IN}=0V$ ; $V_{OUT}=4V$	-75		0	
$V_F$	Output - $V_{CC}$ diode voltage <sup>(2)</sup>	$-I_{OUT}=2A$ ; $T_j=150^{\circ}C$			0.7	V

(1) PowerMOS leakage included.

(2) For each channel.

**Table 7. Switching ( $V_{CC} = 13V$ ;  $T_j = 25^{\circ}C$ )**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn- On delay time	$R_L = 6.5\Omega$ (see <a href="#">Figure 5</a> )		20		$\mu s$
$t_{d(off)}$	Turn- Off delay time	$R_L = 6.5\Omega$ (see <a href="#">Figure 5</a> )		35		$\mu s$
$dV_{OUT}/dt_{(on)}$	Turn- On voltage slope	$R_L = 6.5\Omega$		See <a href="#">Figure 21</a>		V/ $\mu s$
$dV_{OUT}/dt_{(off)}$	Turn- Off voltage slope	$R_L = 6.5\Omega$		See <a href="#">Figure 23</a>		V/ $\mu s$
$W_{ON}$	Switching energy losses during $t_{won}$	$R_L = 6.5\Omega$ (see <a href="#">Figure 5</a> )		0.2		mJ
$W_{OFF}$	Switching energy losses during $t_{woff}$	$R_L = 6.5\Omega$ (see <a href="#">Figure 5</a> )		0.2		mJ

**Table 8. Status pin ( $V_{SD}=0V$ )**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{STAT}$	Status low output voltage	$I_{STAT}=1.6\text{ mA}$ , $V_{SD}=0V$			0.5	V
$I_{LSTAT}$	Status leakage current	Normal Operation or $V_{SD}=5V$ , $V_{STAT}=5V$			10	$\mu A$
$C_{STAT}$	Status pin input capacitance	Normal Operation or $V_{SD}=5V$ , $V_{STAT}=5V$			100	pF
$V_{SCL}$	Status clamp voltage	$I_{STAT}=1\text{ mA}$ $I_{STAT}=-1\text{ mA}$	5.5	-0.7	7	V V

**Table 9. Protections (1)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{limH}$	DC short circuit current	$V_{CC}=13V$ $5V < V_{CC} < 36V$	13.5	19	26.5 26.5	A A
$I_{limL}$	Short circuit current during thermal cycling	$V_{CC}=13V$ $T_R < T_j < T_{TSD}$		7		A
$T_{TSD}$	Shutdown temperature		150	175	200	$^{\circ}C$
$T_R$	Reset temperature		$T_{RS} + 1$	$T_{RS} + 5$		$^{\circ}C$
$T_{RS}$	Thermal reset of STATUS		135			$^{\circ}C$
$T_{HYST}$	Thermal hysteresis ( $T_{TSD}-T_R$ )			7		$^{\circ}C$
$t_{SDL}$	Status delay in overload conditions	$T_j > T_{TSD}$ (see <a href="#">Figure 4</a> )			20	$\mu s$
$V_{DEMAG}$	Turn-Off output voltage clamp	$I_{OUT}=2A$ ; $V_{IN}=0$ ; $L=6\text{ mH}$	$V_{CC}-41$	$V_{CC}-46$	$V_{CC}-52$	V
$V_{ON}$	Output voltage drop limitation	$I_{OUT}=0.1A$ ; $T_j=-40^{\circ}C...+150^{\circ}C$ (see <a href="#">Figure 6</a> )		25		mV

(1) To ensure long term reliability under heavy overload or short circuit conditions, protection and related diagnostic signals must be used together with a proper software strategy. If the device is subjected to abnormal conditions, this software must limit the duration and number of activation cycles.

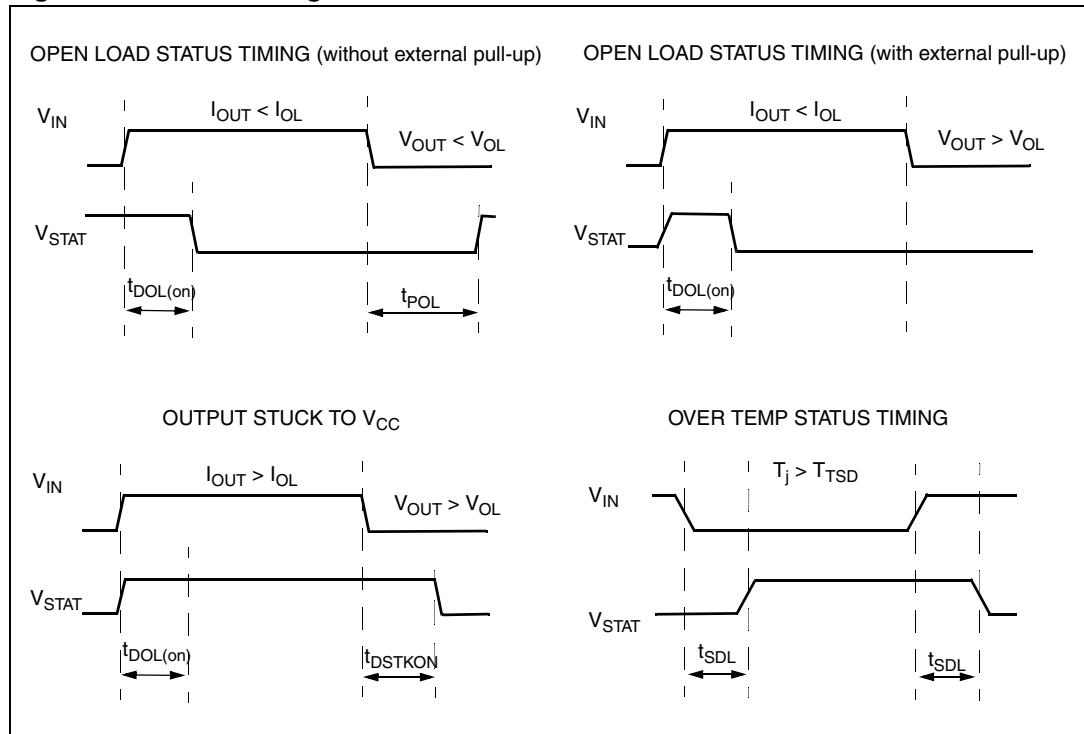
**Table 10. Openload detection**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{OL}$	Openload On State detection threshold	$V_{IN} = 5V, 8V < V_{CC} < 18V$	10	See <a href="#">Figure 18</a>	50	mA
$t_{DOL(on)}$	Openload On State detection delay	$I_{OUT} = 0A, V_{CC} = 13V$ (see <a href="#">Figure 4</a> )			200	$\mu s$
$t_{POL}$	Delay between INPUT falling edge and STATUS rising edge in Openload condition	$I_{OUT} = 0A$ (see <a href="#">Figure 4</a> )	200	500	1000	$\mu s$
$V_{OL}$	Openload Off State voltage detection threshold	$V_{IN} = 0V, 8V < V_{CC} < 16V$	2	See <a href="#">Figure 19</a>	4	V
$t_{DSTKON}$	Output short circuit to $V_{CC}$ detection delay at turn Off	See <a href="#">Figure 4</a>	180		$t_{POL}$	$\mu s$

**Table 11. Logic input**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{IL}$	Input low level				0.9	V
$I_{IL}$	Low level Input current	$V_{IN} = 0.9V$	1			$\mu A$
$V_{IH}$	Input high level		2.1			V
$I_{IH}$	High level Input current	$V_{IN} = 2.1V$			10	$\mu A$
$V_{I(hyst)}$	Input hysteresis voltage		0.25			V
$V_{ICL}$	Input clamp voltage	$I_{IN} = 1mA$ $I_{IN} = -1mA$	5.5	-0.7	7	V V
$V_{SDL}$	STAT_DIS low level voltage				0.9	V
$I_{SDL}$	Low level STAT_DIS current	$V_{SD} = 0.9V$	1			$\mu A$
$V_{SDH}$	STAT_DIS high level voltage		2.1			V
$I_{SDH}$	High level STAT_DIS current	$V_{SD} = 2.1V$			10	$\mu A$
$V_{SD(hyst)}$	STAT_DIS hysteresis voltage		0.25			V
$V_{SDCL}$	STAT_DIS clamp voltage	$I_{SD} = 1mA$ $I_{SD} = -1mA$	5.5	-0.7	7	V V

**Figure 4. Status timings**



**Table 12. Truth table**

Conditions	INPUT	OUTPUT	STATUS ( $V_{SD}=0V$ ) <sup>(1)</sup>
Normal operation	L	L	H
	H	H	H
Current limitation	L	L	H
	H	X	H
Overtemperature	L	L	H
	H	L	L
Undervoltage	L	L	X
	H	L	X
Output voltage $> V_{OL}$	L	H	Lv
	H	H	H
Output current $< I_{OL}$	L	L	H <sup>(2)</sup>
	H	H	L

(1) If the  $V_{SD}$  is high, the STATUS pin is in a high impedance.

(2) The STATUS pin becomes high with a delay equal to  $t_{POL}$  after INPUT falling edge.

Figure 5. Switching characteristics

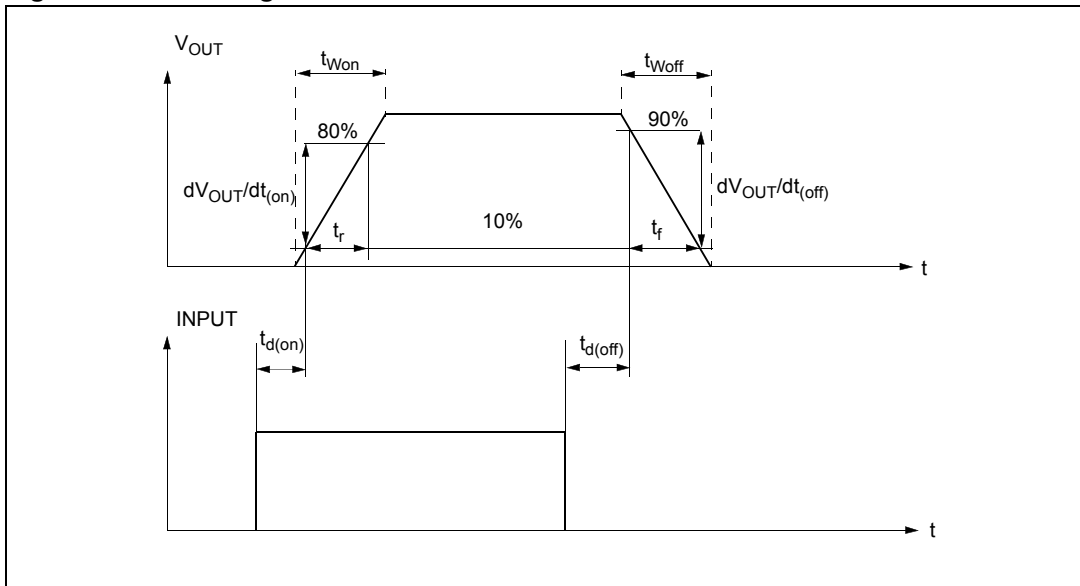
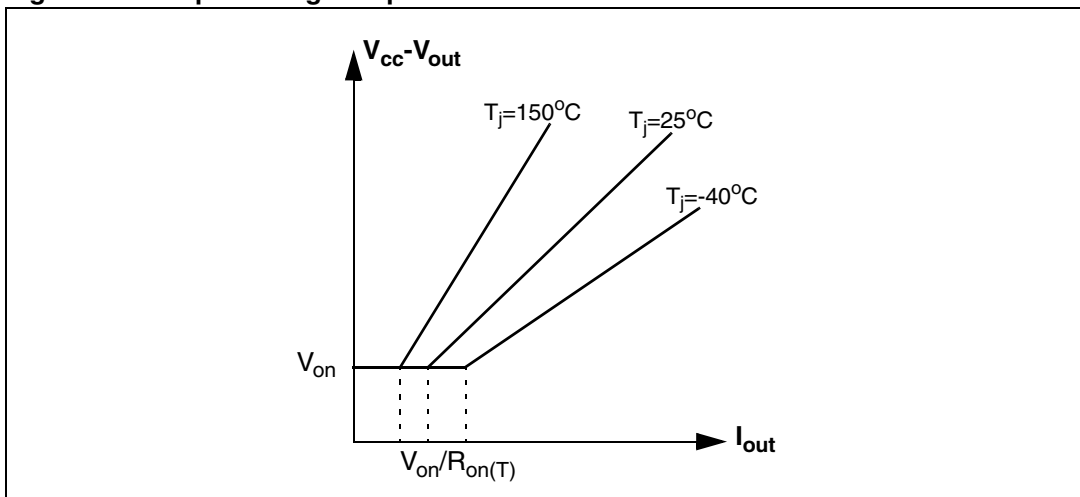


Figure 6. Output voltage drop limitation



**Table 13. Electrical transient requirements**

ISO 7637-2: 2004(E) Test pulse	Test levels		Number of pulses or test times	Burst cycle/pulse repetition time		Delays and Impedance
	III	IV				
1	-75V	-100V	5000 pulses	0.5 s	5 s	2 ms, 10 Ω
2a	+37V	+50V	5000 pulses	0.2 s	5 s	50 μs, 2 Ω
3a	-100V	-150V	1h	90 ms	100 ms	0.1 μs, 50 Ω
3b	+75V	+100V	1h	90 ms	100 ms	0.1 μs, 50 Ω
4	-6V	-7V	1 pulse			100 ms, 0.01Ω
5b <sup>(2)</sup>	+65V	+87V	1 pulse			400 ms, 2 Ω

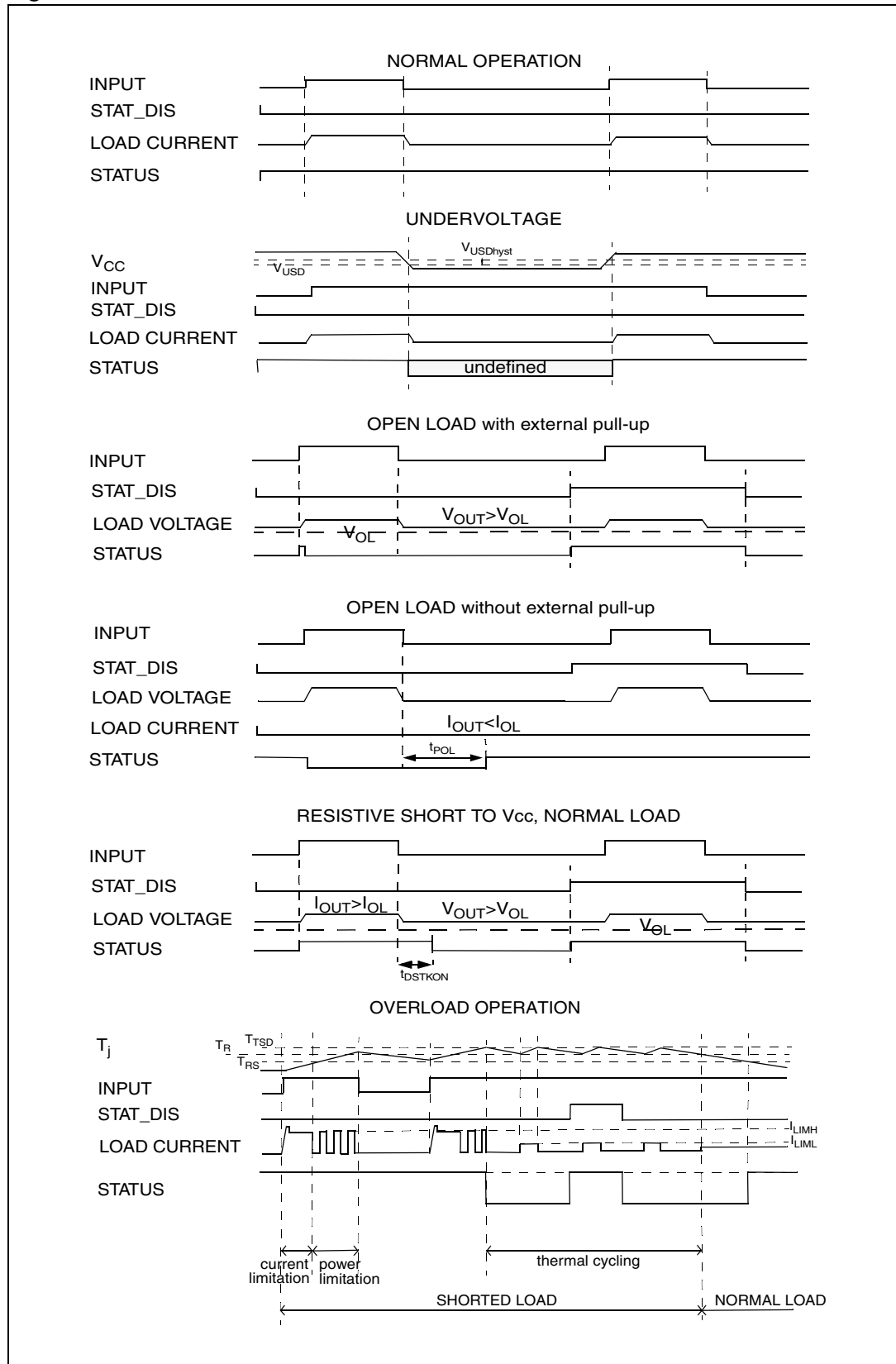
ISO 7637-2: 2004(E) Test pulse	Test level results <sup>(1)</sup>	
	III	IV
1	C	C
2a	C	C
3a	C	C
3b	C	C
4	C	C
5b <sup>(2)</sup>	C	C

(1) The above test levels must be considered referred to Vcc = 13.5V except for pulse 5b.

(2) Valid in case of external load dump clamp: 40V maximum referred to ground.

Class	Contents
C	All functions of the device are performed as designed after exposure to disturbance.
E	One or more functions of the device are not performed as designed after exposure to disturbance and cannot be returned to proper operation without replacing the device.

Figure 7. Waveforms





## 2.4 Electrical characteristics curves

Figure 8. Off state output current

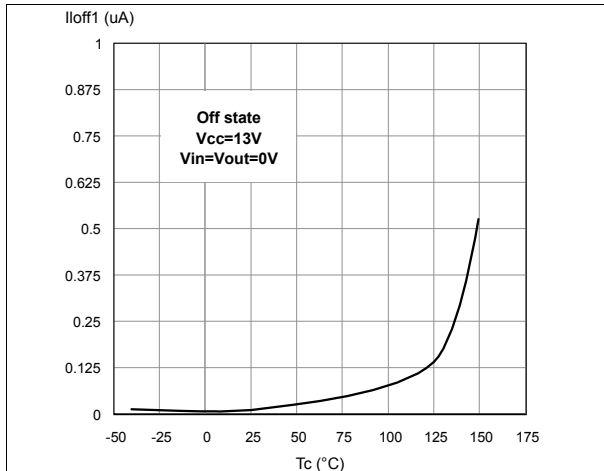


Figure 9. High level input current

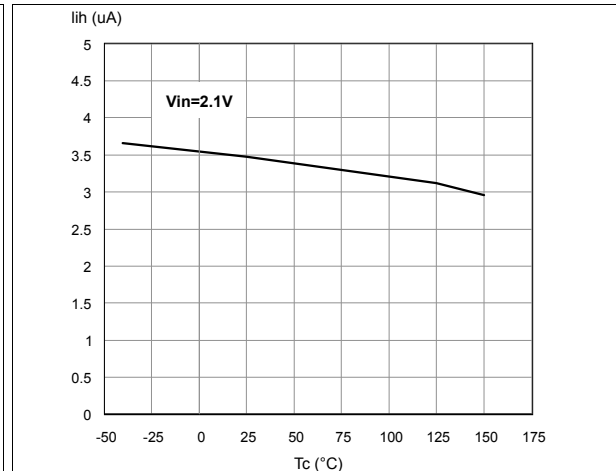


Figure 10. Input clamp voltage

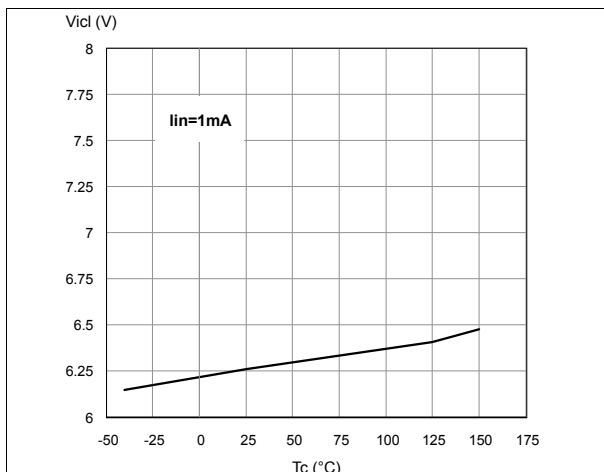


Figure 11. Input high level

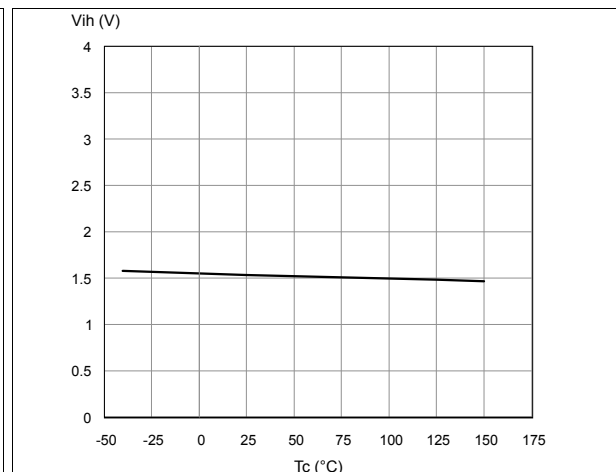


Figure 12. Input low level

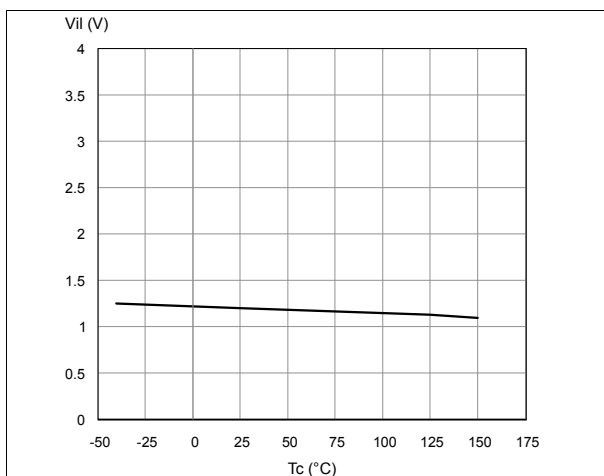


Figure 13. Input hysteresis voltage

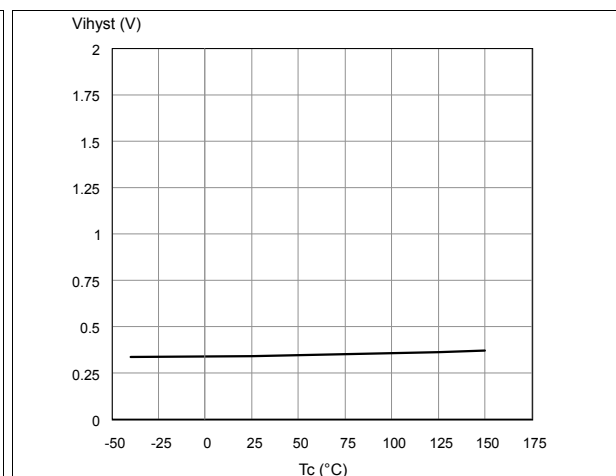
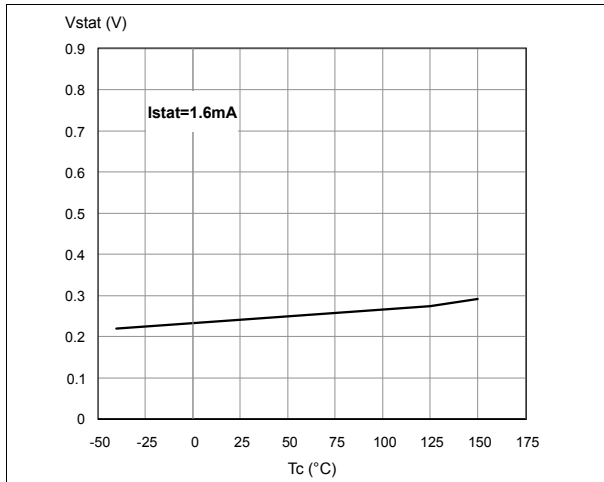


Figure 14. Status low output voltage



On state resistance vs T<sub>case</sub>

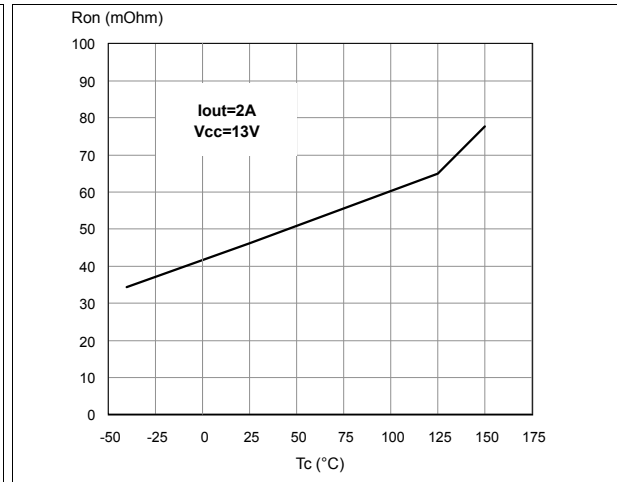


Figure 15. Status leakage current

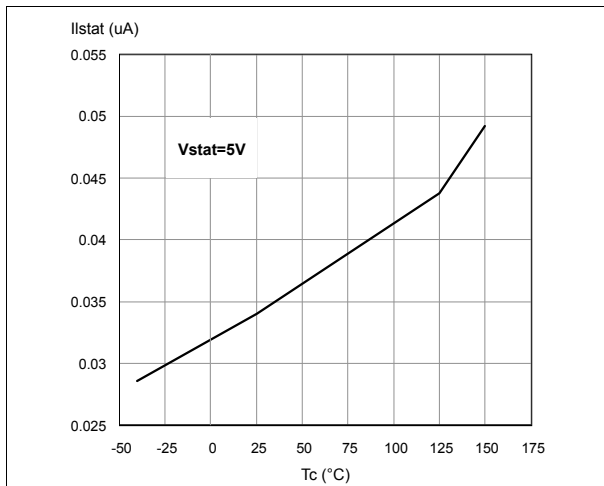


Figure 16. On state resistance vs V<sub>CC</sub>

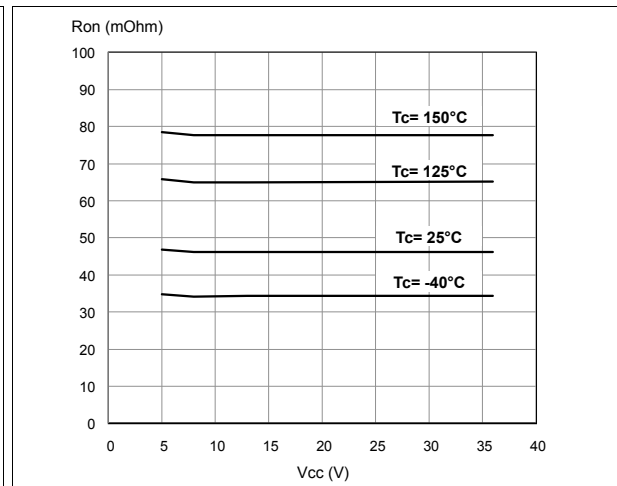


Figure 17. Status clamp voltage

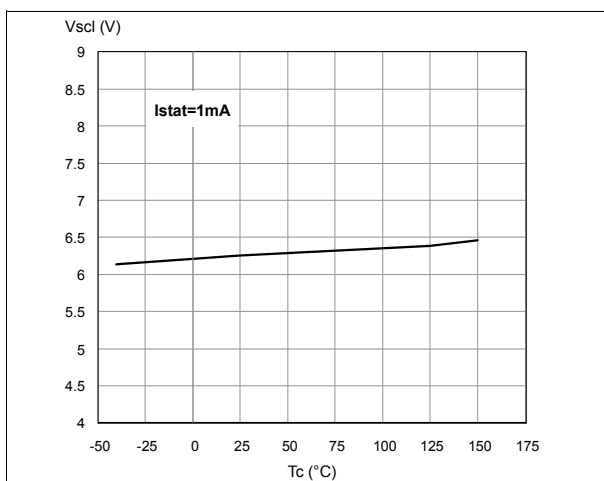


Figure 18. Openload On state detection threshold

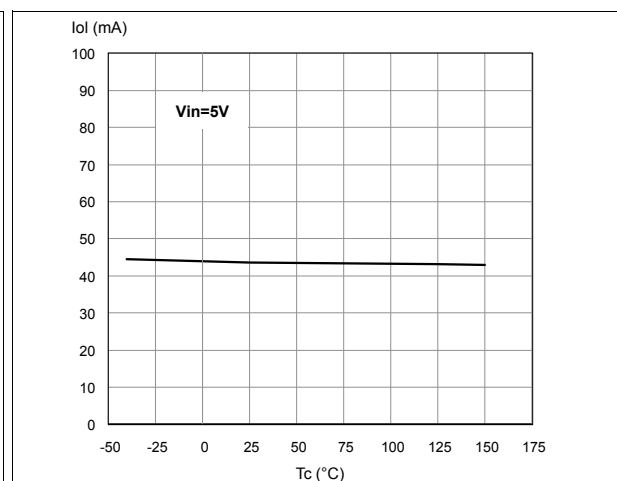


Figure 19. Openload Off state voltage detection threshold

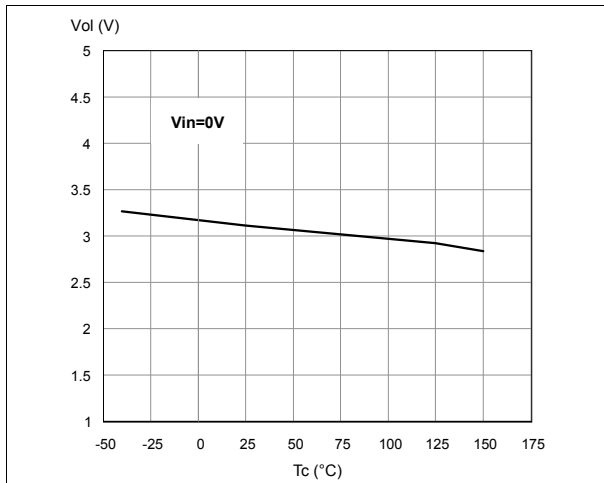


Figure 20. I<sub>LIM</sub> vs T<sub>case</sub>

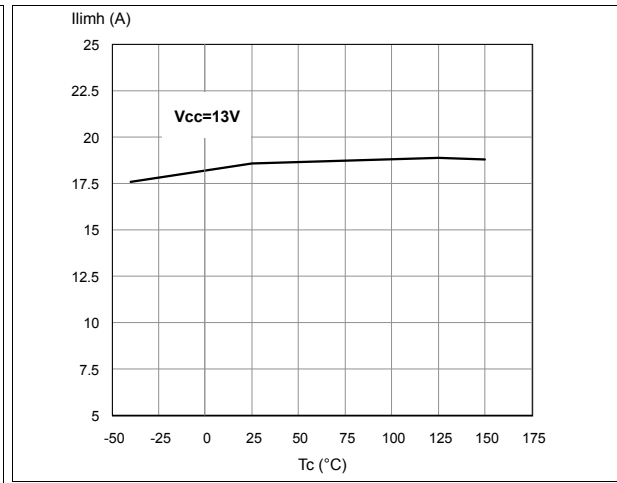


Figure 21. Turn-On voltage slope

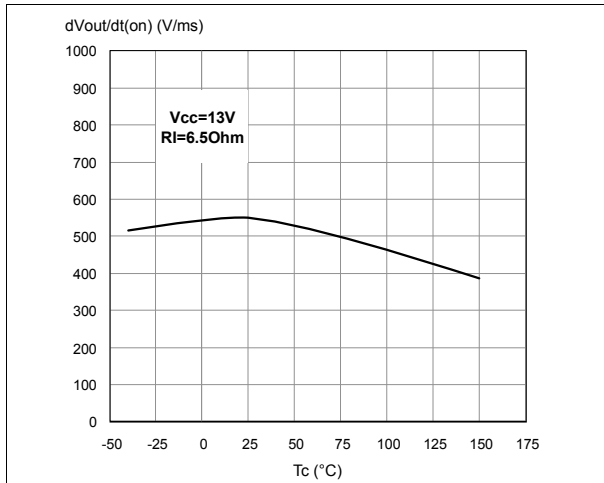


Figure 22. Undervoltage shutdown

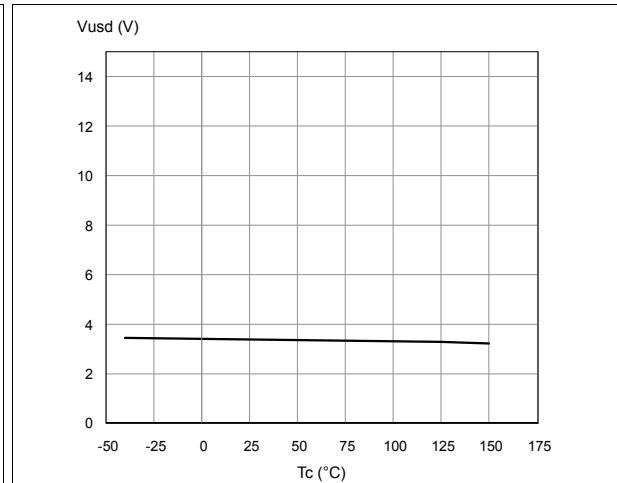


Figure 23. Turn-Off voltage slope

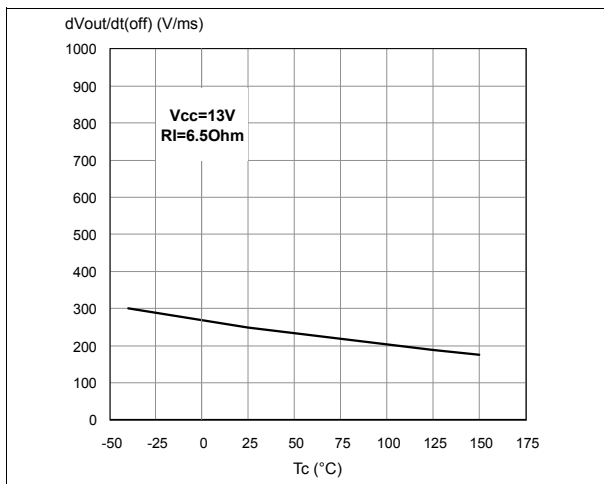


Figure 24. STAT\_DIS clamp voltage

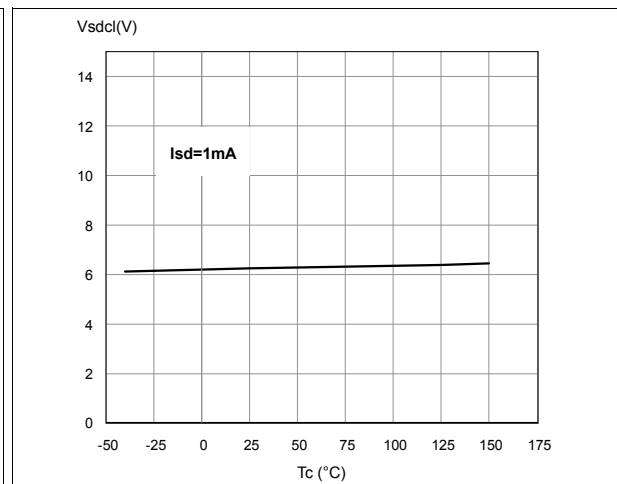


Figure 25. High level STAT\_DIS voltage

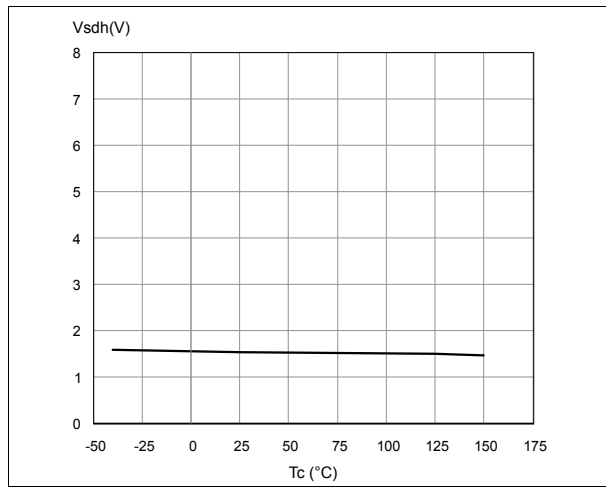
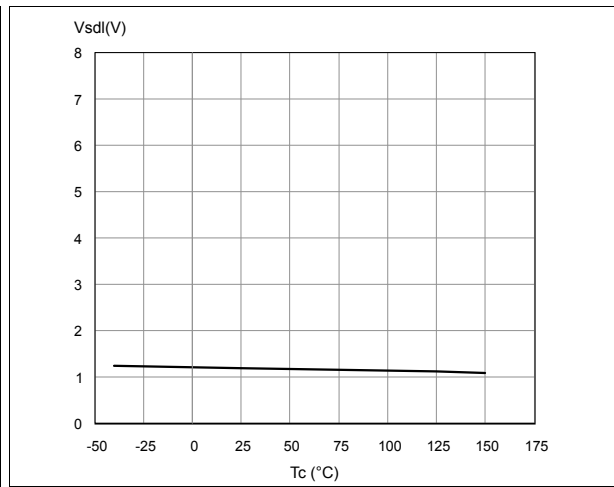
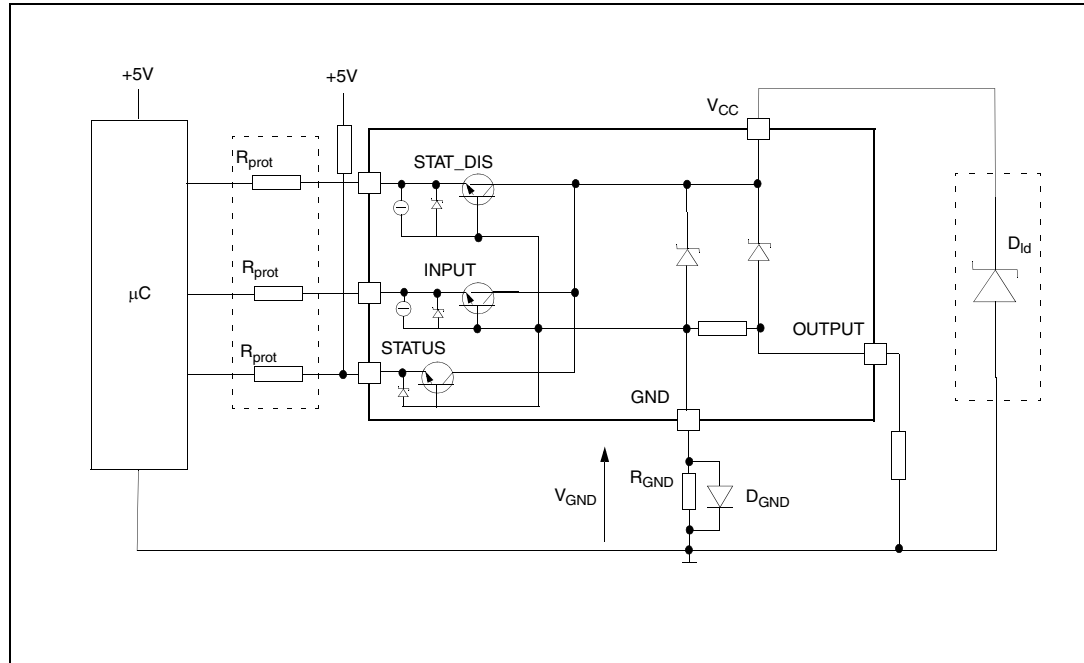


Figure 26. Low level STAT\_DIS voltage



### 3 Application information

Figure 27. Application schematic



#### 3.1 GND protection network against reverse battery

##### 3.1.1 Solution 1: resistor in the ground line (R<sub>GND</sub> only)

This can be used with any type of load.

The following is an indication on how to dimension the R<sub>GND</sub> resistor.

1.  $R_{GND} \leq 600\text{mV} / (I_{S(on)max})$ .
2.  $R_{GND} \geq (-V_{CC}) / (-I_{GND})$

where -I<sub>GND</sub> is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device datasheet.

Power Dissipation in R<sub>GND</sub> (when V<sub>CC</sub><0: during reverse battery situations) is:

$$P_D = (-V_{CC})^2 / R_{GND}$$

This resistor can be shared amongst several different HSDs. Please note that the value of this resistor should be calculated with formula (1) where I<sub>S(on)max</sub> becomes the sum of the maximum on-state currents of the different devices.

Please note that if the microprocessor ground is not shared by the device ground then the R<sub>GND</sub> will produce a shift (I<sub>S(on)max</sub> \* R<sub>GND</sub>) in the input thresholds and the status output values. This shift will vary depending on how many devices are ON in the case of several high side drivers sharing the same R<sub>GND</sub>.

If the calculated power dissipation leads to a large resistor or several devices have to share the same resistor then ST suggests to utilize Solution 2 (see below).

### 3.1.2 Solution 2: diode ( $D_{GND}$ ) in the ground line

A resistor ( $R_{GND}=1k\Omega$ ) should be inserted in parallel to  $D_{GND}$  if the device drives an inductive load.

This small signal diode can be safely shared amongst several different HSDs. Also in this case, the presence of the ground network will produce a shift ( $\approx 600mV$ ) in the input threshold and in the status output values if the microprocessor ground is not common to the device ground. This shift will not vary if more than one HSD shares the same diode/resistor network.

## 3.2 Load dump protection

$D_{ld}$  is necessary (Voltage Transient Suppressor) if the load dump peak voltage exceeds the  $V_{CC}$  max DC rating. The same applies if the device is subject to transients on the  $V_{CC}$  line that are greater than the ones shown in the ISO 7637-2: 2004(E) table.

## 3.3 MCU I/Os protection

If a ground protection network is used and negative transient are present on the  $V_{CC}$  line, the control pins will be pulled negative. ST suggests to insert a resistor ( $R_{prot}$ ) in line to prevent the  $\mu C$  I/Os pins to latch-up.

The value of these resistors is a compromise between the leakage current of  $\mu C$  and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of  $\mu C$  I/Os.

$$-V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH} - V_{GND}) / I_{IHmax}$$

Calculation example:

For  $V_{CCpeak} = -100V$  and  $I_{latchup} \geq 20mA$ ;  $V_{OH\mu C} \geq 4.5V$

$$5k\Omega \leq R_{prot} \leq 180k\Omega$$

Recommended values:  $R_{prot} = 10k\Omega$ .

## 3.4 Open load detection in Off state

Off state open load detection requires an external pull-up resistor ( $R_{PU}$ ) connected between OUTPUT pin and a positive supply voltage ( $V_{PU}$ ) like the +5V line used to supply the microprocessor.

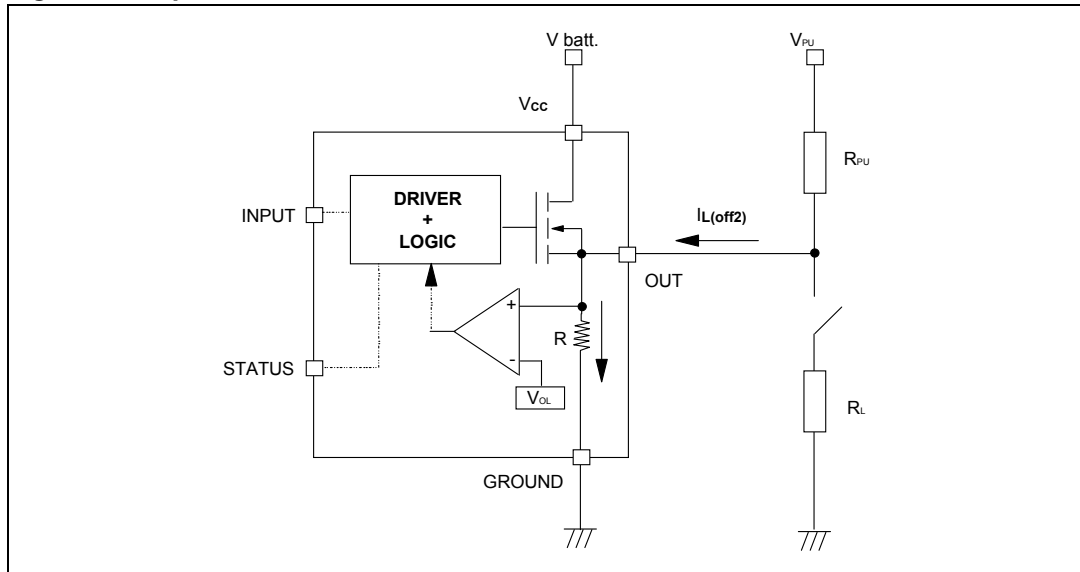
The external resistor has to be selected according to the following requirements:

- no false open load indication when load is connected: in this case we have to avoid  $V_{OUT}$  to be higher than  $V_{OLmin}$ ; this results in the following condition  $V_{OUT} = (V_{PU} / (R_L + R_{PU})) R_L < V_{OLmin}$ .
- no misdetection when load is disconnected: in this case the  $V_{OUT}$  has to be higher than  $V_{OLmax}$ ; this results in the following condition  $R_{PU} < (V_{PU} - V_{OLmax}) / I_{L(off2)}$ .

Because  $I_{s(OFF)}$  may significantly increase if  $V_{out}$  is pulled high (up to several mA), the pull-up resistor  $R_{PU}$  should be connected to a supply that is switched OFF when the module is in standby.

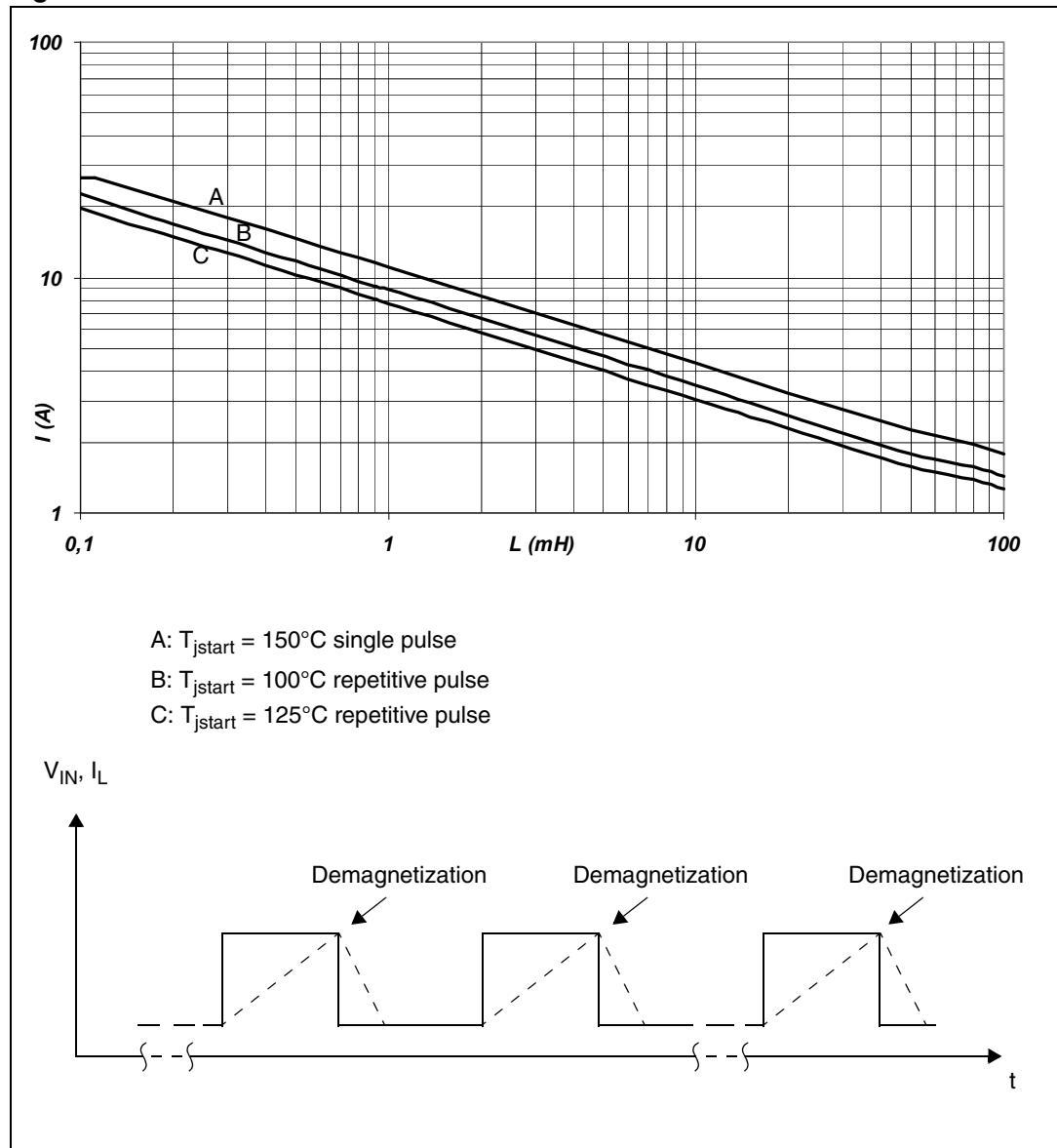
The values of  $V_{OLmin}$ ,  $V_{OLmax}$  and  $I_{L(off2)}$  are available in the Electrical characteristics section.

**Figure 28. Open load detection in Off state**



### 3.5 Maximum demagnetization energy ( $V_{CC} = 13.5V$ )

Figure 29. Maximum turn Off current versus inductance



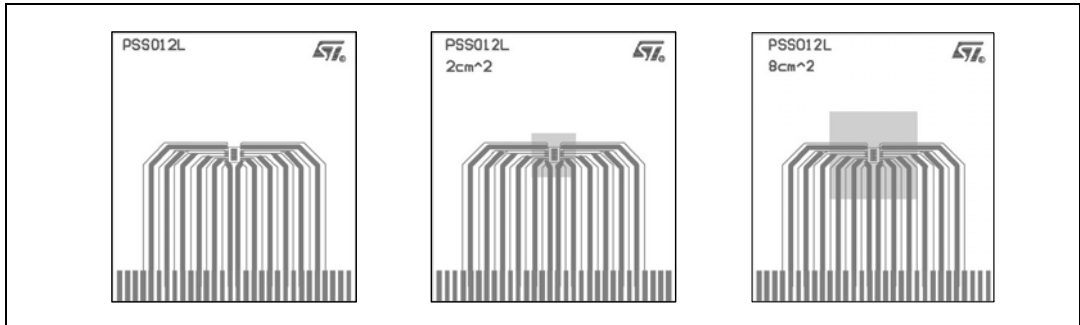
Note: Values are generated with  $R_L = 0 \Omega$ . In case of repetitive pulses,  $T_{jstart}$  (at beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves A and B.



## 4 Package and PCB thermal data

### 4.1 PowerSSO-12™ thermal data

Figure 30. PowerSSO-12™ PC Board



Note: Layout condition of  $R_{th}$  and  $Z_{th}$  measurements (PCB: Double layer, Thermal Vias, FR4 area= 77mm x 86mm, PCB thickness=1.6mm, Cu thickness=70  $\mu$ m (front and back side), Copper areas: from minimum pad lay-out to 8cm<sup>2</sup>).

Figure 31.  $R_{thj-amb}$  Vs. PCB copper area in open box free air condition

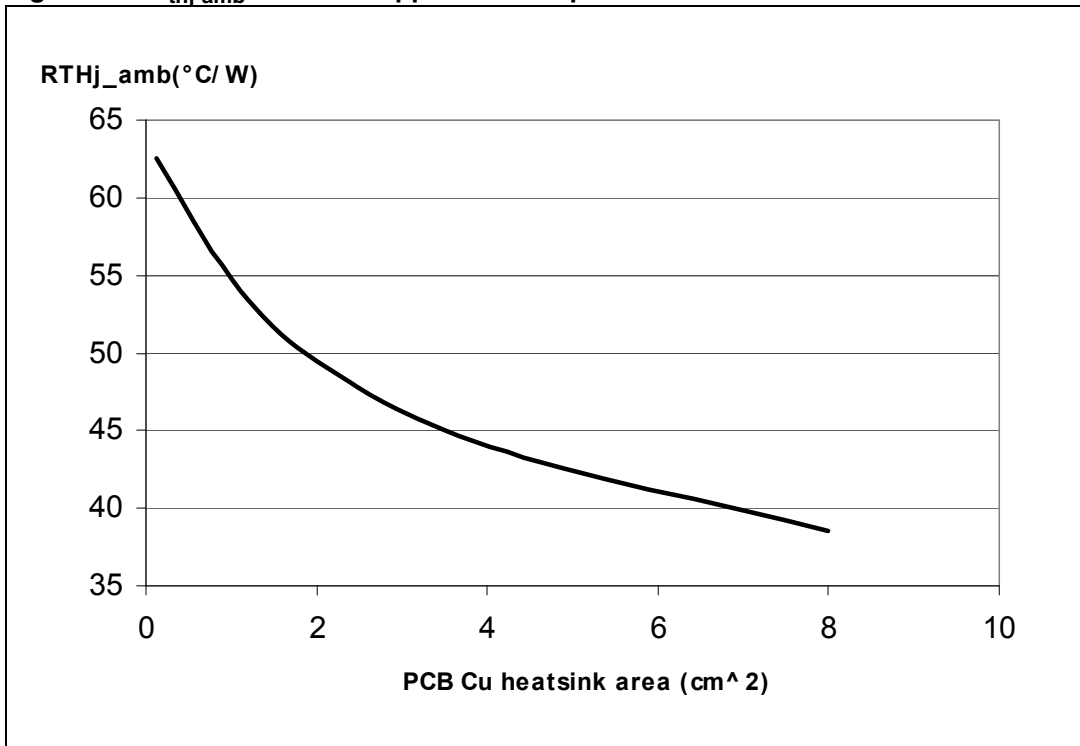
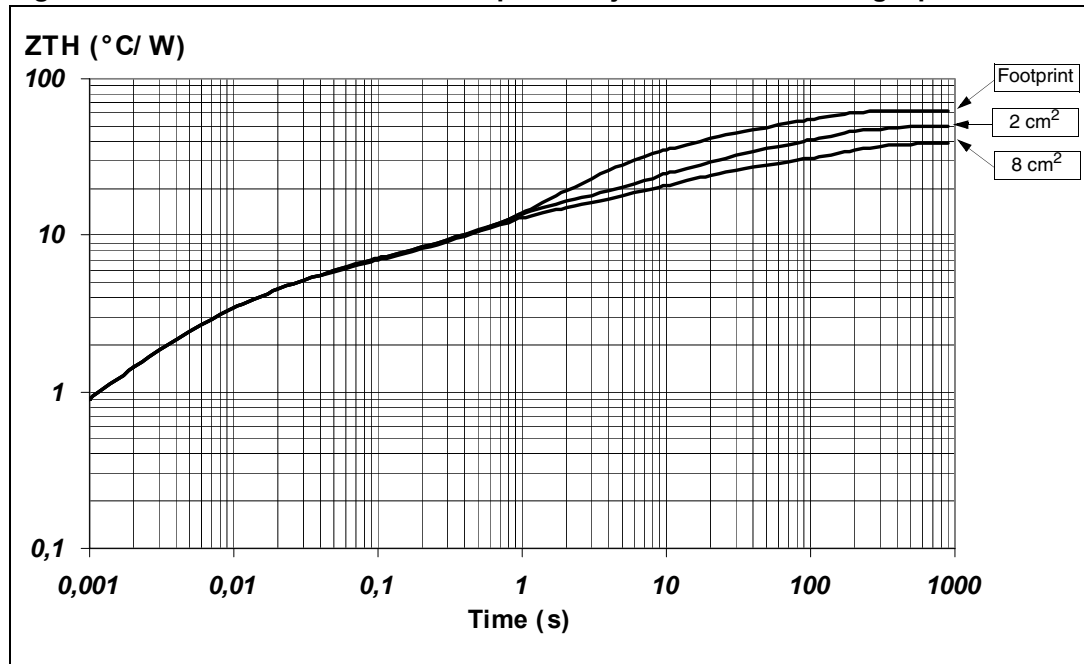


Figure 32. PowerSSO-12™ thermal impedance junction ambient single pulse

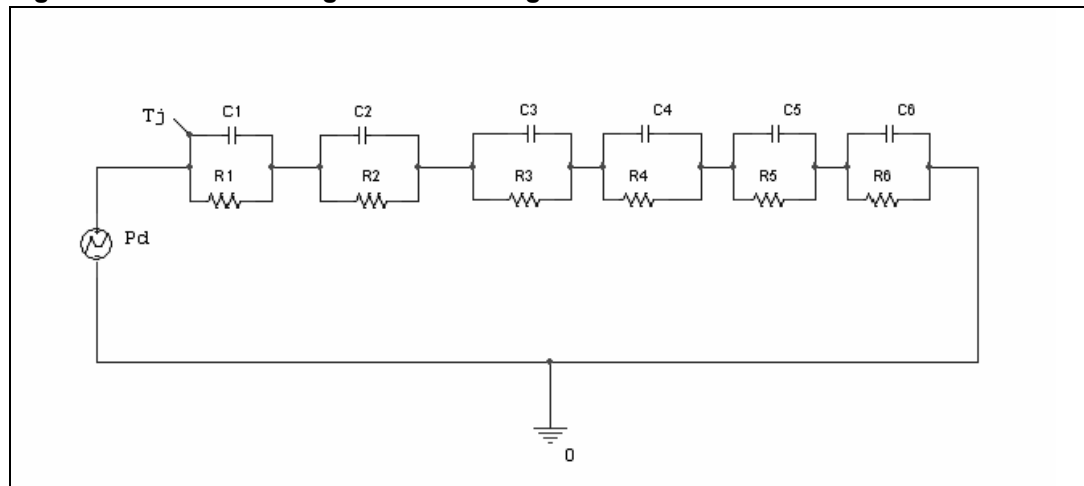


Equation 1: pulse calculation formula

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp}(1 - \delta)$$

where  $\delta = t_p/T$

Figure 33. Thermal fitting model of a single channel HSD in PowerSSO-12™ (a)



(a) The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.