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VND5025AK-E

Double channel high side driver with analog current sense for automotive applications

Features

Max transient supply voltage	V_{CC}	41V
Operating voltage range	V_{CC}	4.5 to 36V
Max on-state resistance (per ch.)	R_{ON}	25 m Ω
Current limitation (typ)	I_{LIMH}	41 A
Off-state supply current	I_S	2 μ A ⁽¹⁾

1. Typical value with all loads connected.

■ Main

- In-rush current active management by power limitation
- Very low standby current
- 3.0V CMOS compatible input
- Optimized electromagnetic emission
- Very low electromagnetic susceptibility
- In compliance with the 2002/95/EC European directive
- Package: ECOPACK®

■ Diagnostic functions Doc ID 12581

- Proportional load current sense
- High current sense precision for wide range currents
- Current sense disable
- Thermal shutdown indication
- Very low current sense leakage

■ Protection

- Undervoltage shut-down
- Overvoltage clamp
- Load current limitation
- Self-limiting of fast thermal transients
- Protection against loss of ground and loss of V_{CC}
- Thermal shutdown
- Reverse battery protection (see [Application schematic on page 21](#))



- Electrostatic discharge protection

Application

- All types of resistive, inductive and capacitive loads
- Suitable as LED driver

Description

The VND5025AK-E is a monolithic device made using STMicroelectronics VIPower M0-5 technology, intended for driving resistive or inductive loads with one side connected to ground, and suitable for driving LEDs. Active V_{CC} pin voltage clamp protects the device against low energy spikes (see ISO7637 transient compatibility table). This device integrates an analog current sense which delivers a current proportional to the load current (according to a known ratio) when CS_DIS is driven low or left open. When CS_DIS is driven high, the CURRENT SENSE pin is in a high impedance condition. Output current limitation protects the device in overload condition. In case of long overload duration, the device limits the dissipated power to safe level up to thermal shut-down intervention. Thermal shut-down with automatic restart allows the device to recover normal operation as soon as fault condition disappears.

Table 1. Device summary

Package	Order codes	
	Tube	Tape and reel
PowerSSO-24™	VND5025AK-E	VND5025AKTR-E

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1 Block diagram and pin description

Figure 1. Block diagram

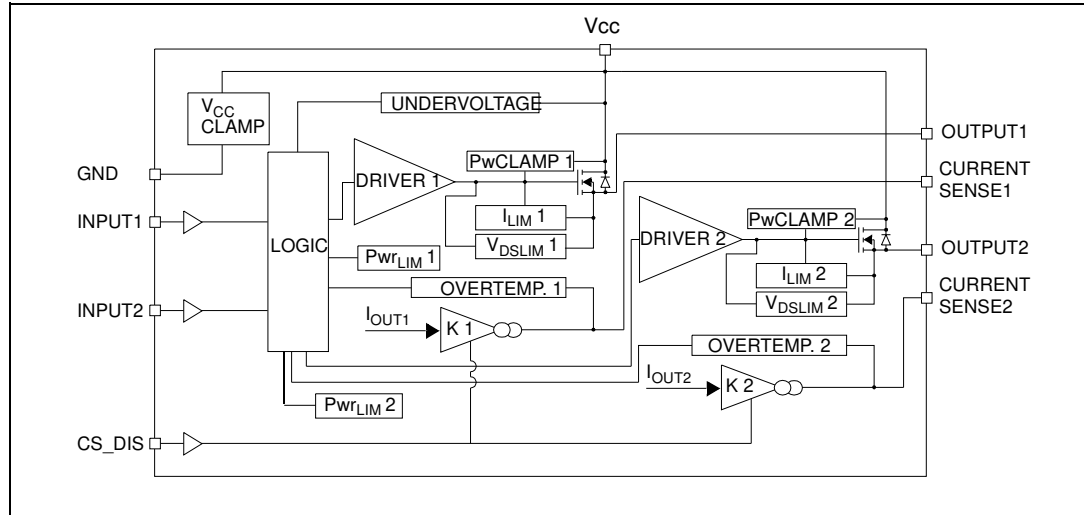


Table 2. Pin functions

Name	Function
V _{CC}	Battery connection.
OUTPUT _{1,2}	Power output.
GND	Ground connection; must be reverse battery protected by an external diode/resistor network.
INPUT _{1,2}	Voltage controlled input pin with hysteresis, CMOS compatible; controls output switch state.
CURRENT SENSE _{1,2}	Analog current sense pin; delivers a current proportional to the load current.
CS_DIS	Active high CMOS compatible pin to disable the current sense pin.

Figure 2. Configuration diagram (top view)

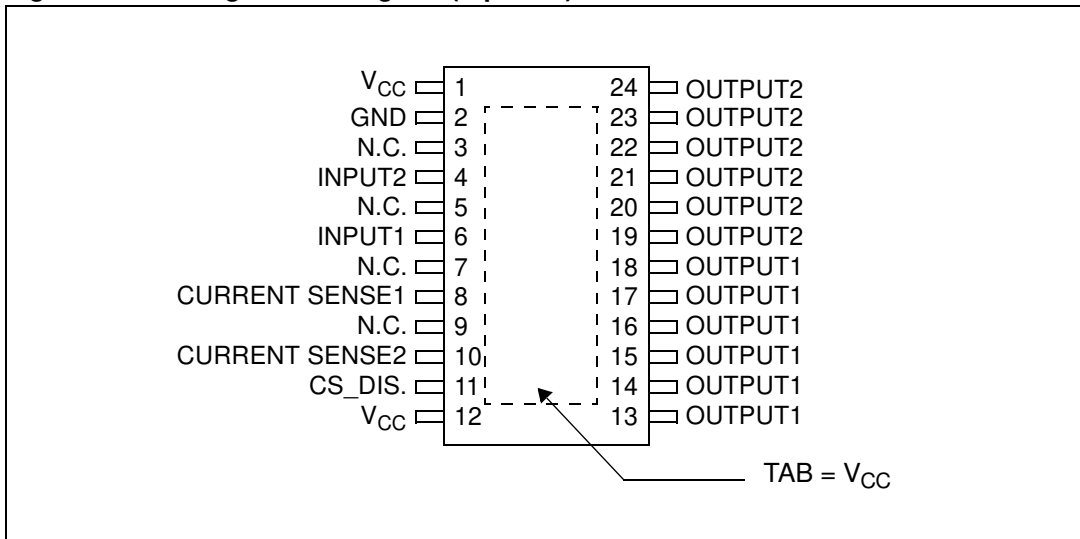


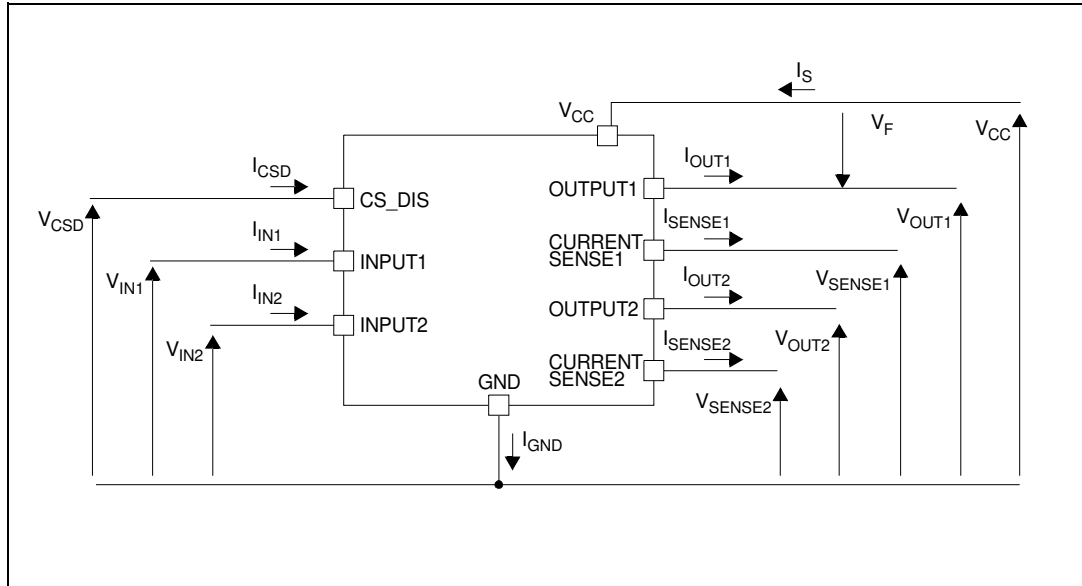
Table 3. Suggested connections for unused and not connected pins

Connection / pin	Current sense	N.C.	Output	Input	CS_DIS
Floating	N.R. ⁽¹⁾	X	X	X	X
To Ground	Through 1kΩ resistor	X	N.R.	Through 10kΩ resistor	Through 10kΩ resistor

1. Not recommended.

2 Electrical specification

Figure 3. Current and voltage conventions



Note: $V_{Fn} = V_{OUTn} - V_{CC}$ during reverse battery condition.

2.1 Absolute maximum ratings

Stressing the device above the rating listed in the “Absolute maximum ratings” table may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is not implied. Exposure to the conditions in table below for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE Program and other relevant quality document.

Table 4. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{CC}	DC supply voltage	41	V
$-V_{CC}$	Reverse DC supply voltage	0.3	
$-I_{GND}$	DC reverse ground pin current	200	mA
I_{OUT}	DC output current	Internally limited	A
$-I_{OUT}$	Reverse DC output current	24	
I_{IN}	DC input current	-1 to 10	mA
I_{CSD}	DC current sense disable input current		
$-I_{CSENSE}$	DC reverse CS pin current	200	
V_{CSENSE}	Current sense maximum voltage	$V_{CC} - 41$ to $+V_{CC}$	V

Table 4. Absolute maximum ratings (continued)

Symbol	Parameter	Value	Unit
E_{MAX}	Maximum switching energy (single pulse) ($L = 0.8\text{mH}$; $R_L = 0\Omega$; $V_{bat} = 13.5\text{V}$; $T_{jstart} = 150^\circ\text{C}$; $I_{OUT} = I_{limL}(\text{Typ.})$)	140	mJ
V_{ESD}	Electrostatic discharge (Human Body Model: $R = 1.5\text{k}\Omega$; $C = 100\text{pF}$)		
	- Input	4000	V
	- Current sense	2000	V
	- CS_DIS	4000	V
	- Output	5000	V
	- V_{CC}	5000	V
V_{ESD}	Charge device model (CDM-AEC-Q100-011)	750	V
T_j	Junction operating temperature	-40 to 150	°C
T_{stg}	Storage temperature	-55 to 150	

2.2 Thermal data

Table 5. Thermal data

Symbol	Parameter	Max Value	Unit
$R_{thj-case}$	Thermal resistance junction-case (MAX) (with one channel ON)	1.35	°C/W
$R_{thj-amb}$	Thermal resistance junction-ambient (MAX)	See Figure 29	

2.3 Electrical characteristics

$8V < V_{CC} < 36V$; $-40^{\circ}C < T_j < 150^{\circ}C$, unless otherwise specified.

Table 6. Power section

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC}	Operating supply voltage		4.5	13	36	V
V_{USD}	Undervoltage shutdown			3.5	4.5	
$V_{USDhyst}$	Undervoltage shut-down hysteresis			0.5		
R_{ON}	On-state resistance ⁽¹⁾	$I_{OUT} = 3A$; $T_j = 25^{\circ}C$			25	m Ω
		$I_{OUT} = 3A$; $T_j = 150^{\circ}C$			50	
		$I_{OUT} = 3A$; $V_{CC} = 5V$; $T_j = 25^{\circ}C$			35	
V_{clamp}	Clamp voltage	$I_S = 20\text{ mA}$	41	46	52	V
I_S	Supply current	Off-state; $V_{CC} = 13V$; $T_j = 25^{\circ}C$; $V_{IN} = V_{OUT} = V_{SENSE} = V_{CSD} = 0V$		2 ⁽²⁾	5 ⁽²⁾	μA
		On-state; $V_{CC} = 13V$; $V_{IN} = 5V$; $I_{OUT} = 0A$		3	6	mA
$I_{L(off)}$	Off-state output current ⁽¹⁾	$V_{IN} = V_{OUT} = 0V$; $V_{CC} = 13V$; $T_j = 25^{\circ}C$	0	0.01	3	μA
		$V_{IN} = V_{OUT} = 0V$; $V_{CC} = 13V$; $T_j = 125^{\circ}C$	0		5	
V_F	Output - V_{CC} diode voltage ⁽¹⁾	$-I_{OUT} = 4A$; $T_j = 150^{\circ}C$			0.7	V

1. For each channel.

2. PowerMOS leakage included.

Table 7. Switching ($V_{CC} = 13V$; $T_j = 25^{\circ}C$)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn-on delay time	$R_L = 4.3\Omega$ (see Figure 6)		35		μs
$t_{d(off)}$	Turn-off delay time			50		
$(dV_{OUT}/dt)_{on}$	Turn-on voltage slope	$R_L = 4.3\Omega$		See Figure 21		V/ μs
$(dV_{OUT}/dt)_{off}$	Turn-off voltage slope			See Figure 22		
W_{ON}	Switching energy losses during t_{WON}	$R_L = 4.3\Omega$ (see Figure 6)		0.45		mJ
W_{OFF}	Switching energy losses during t_{WOFF}			0.35		

Table 8. Logic input

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{IL}	Input low level voltage				0.9	V
I_{IL}	Low level input current	$V_{IN} = 0.9V$	1			μA
V_{IH}	Input high level voltage		2.1			V
I_{IH}	High level input current	$V_{IN} = 2.1V$			10	μA
$V_{I(hyst)}$	Input hysteresis voltage		0.25			V
V_{ICL}	Input clamp voltage	$I_{IN} = 1mA$	5.5		7	
		$I_{IN} = -1mA$		-0.7		
V_{CSDL}	CS_DIS low level voltage				0.9	μA
I_{CSDL}	Low level CS_DIS current	$V_{CSD} = 0.9V$	1			
V_{CSDH}	CS_DIS high level voltage		2.1			V
I_{CSDH}	High level CS_DIS current	$V_{CSD} = 2.1V$			10	μA
$V_{CSD(hyst)}$	CS_DIS hysteresis voltage		0.25			V
V_{CSCL}	CS_DIS clamp voltage	$I_{CSD} = 1mA$	5.5		7	
		$I_{CSD} = -1mA$		-0.7		

Table 9. Protection and diagnostics⁽¹⁾

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{LIMH}	DC short circuit current	$V_{CC} = 13V$	29	41	57	A
		$5V < V_{CC} < 36V$				
I_{LIML}	Short circuit current during thermal cycling	$V_{CC} = 13V$; $T_R < T_j < T_{TSD}$		16		
T_{TSD}	Shutdown temperature		150	175	200	°C
T_R	Reset temperature		$T_{RS} + 1$	$T_{RS} + 5$		
T_{RS}	Thermal reset of STATUS		135			
T_{HYST}	Thermal hysteresis ($T_{TSD} - T_R$)			7		
V_{DEMAG}	Turn-off output voltage clamp	$I_{OUT} = 2A$; $V_{IN} = 0$; $L = 6mH$	$V_{CC} - 41$	$V_{CC} - 46$	$V_{CC} - 52$	V
V_{ON}	Output voltage drop limitation	$I_{OUT} = 0.20.1A$; $T_j = -40^\circ C$ to $+150^\circ C$ (see Figure 9)		25		mV

1. To ensure long term reliability under heavy overload or short circuit conditions, protection and related diagnostic signals must be used together with a proper software strategy. If the device is subjected to abnormal conditions, this software must limit the duration and number of activation cycles.

Table 10. Current sense (8V < V_{CC} < 16V)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K _{LED}	I _{OUT} /I _{SENSE}	I _{OUT} = 0.05A; V _{SENSE} = 0.5V; V _{CSD} = 0V; T _j = -40°C to 150°C	1450	3300	5180	
K ₀	I _{OUT} /I _{SENSE}	I _{OUT} = 0.5 A; V _{SENSE} = 0.5V; V _{CSD} = 0V; T _j = -40°C to 150°C	1720	3020	4360	
dK ₀ /K ₀ ⁽¹⁾	Current Sense ratio drift	I _{OUT} = 0.5A; V _{SENSE} = 0.5V; V _{CSD} = 0V; T _j = -40°C to 150°C	-12		+12	%
K ₁	I _{OUT} /I _{SENSE}	I _{OUT} = 2A; V _{SENSE} = 4V; V _{CSD} = 0V; T _j = -40°C to 150°C T _j = 25°C to 150°C	1940 2230	2810 2810	3740 3390	
dK ₁ /K ₁ ⁽¹⁾	Current Sense ratio drift	I _{OUT} = 2A; V _{SENSE} = 4V; V _{CSD} = 0V; T _j = -40°C to 150°C	-10		+10	%
K ₂	I _{OUT} /I _{SENSE}	I _{OUT} = 3A; V _{SENSE} = 4V; V _{CSD} = 0V; T _j = -40°C to 150°C T _j = 25°C to 150°C	2250 2400	2790 2790	3450 3180	
dK ₂ /K ₂ ⁽¹⁾	Current Sense ratio drift	I _{OUT} = 3A; V _{SENSE} = 4V; V _{CSD} = 0V; T _j = -40°C to 150°C	-7		+7	%
K ₃	I _{OUT} /I _{SENSE}	I _{OUT} = 10A; V _{SENSE} = 4V; V _{CSD} = 0V; T _j = -40°C to 150°C T _j = 25°C to 150°C	2610 2650	2760 2760	2970 2870	
dK ₃ /K ₃ ⁽¹⁾	Current Sense ratio drift	I _{OUT} = 10A; V _{SENSE} = 4V; V _{CSD} = 0V; T _j = -40°C to 150°C	-4		+4	%
I _{SENSE0}	Analog Sense leakage current	I _{OUT} = 0A; V _{SENSE} = 0V; V _{CSD} = 5V; V _{IN} = 0V; T _j = -40°C to 150°C V _{CSD} = 0V; V _{IN} = 5V; T _j = -40°C to 150°C I _{OUT} = 2A; V _{SENSE} = 0V; V _{CSD} = 5V; V _{IN} = 5V; T _j = -40°C to 150°C	0 0 0		1 2 1	μA μA μA
I _{OL}	Openload on-state current detection threshold	V _{IN} = 5V, I _{SENSE} = 5 μA	5		30	mA
V _{SENSE}	Max analog Sense output voltage	I _{OUT} = 3 A; V _{CSD} = 0V	5			V
V _{SENSEH}	Analog Sense output voltage in overtemperature condition	V _{CC} = 13V; R _{SENSE} = 3.9kΩ		9		

Table 10. Current sense (8V < V_{CC} < 16V) (continued)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I _{SENSEH}	Analog Sense output current in overtemperature condition	V _{CC} = 13V; V _{SENSE} = 5V		8		mA
t _{DSENSE1H}	Delay response time from falling edge of CS_DIS pin	V _{SENSE} < 4V, 0.5 < I _{OUT} < 10A I _{SENSE} = 90% of I _{SENSEMAX} (see Figure 4)		50	100	μs
t _{DSENSE1L}	Delay response time from rising edge of CS_DIS pin	V _{SENSE} < 4V, 0.5 < I _{OUT} < 10A I _{SENSE} = 10% of I _{SENSEMAX} (see Figure 4)		5	20	
t _{DSENSE2H}	Delay response time from rising edge of INPUT pin	V _{SENSE} < 4V, 0.5 < I _{OUT} < 10A I _{SENSE} = 90% of I _{SENSEMAX} (see Figure 4)		70	300	
Δt _{DSENSE2H}	Delay response time between rising edge of output current and rising edge of current sense	V _{SENSE} < 4V, I _{SENSE} = 90% of I _{SENSEMAX} , I _{OUT} = 90% of I _{OUTMAX} , I _{OUTMAX} = 3A (see Figure 5)			200	
t _{DSENSE2L}	Delay response time from falling edge of INPUT pin	V _{SENSE} < 4V, 0.5 < I _{OUT} < 10A I _{SENSE} = 10% of I _{SENSEMAX} (see Figure 4)		100	250	

1. Parameter guaranteed by design; it is not tested.

Figure 4. Current sense delay characteristics

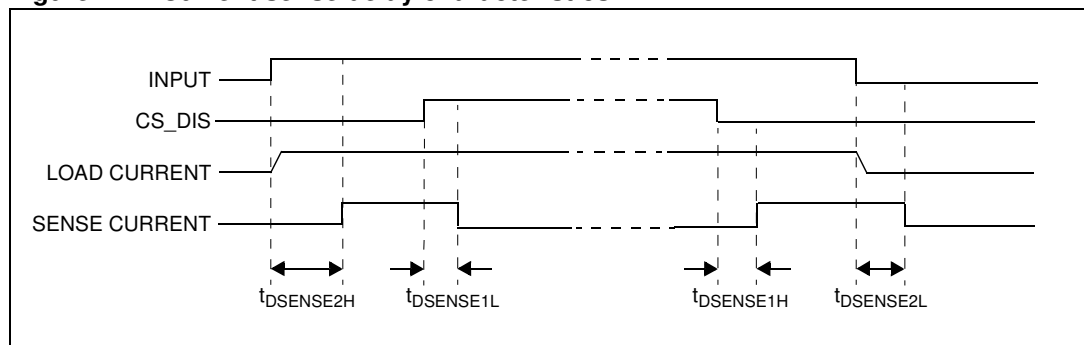


Figure 5. Delay response time between rising edge of output current and rising edge of Current Sense (CS enabled)

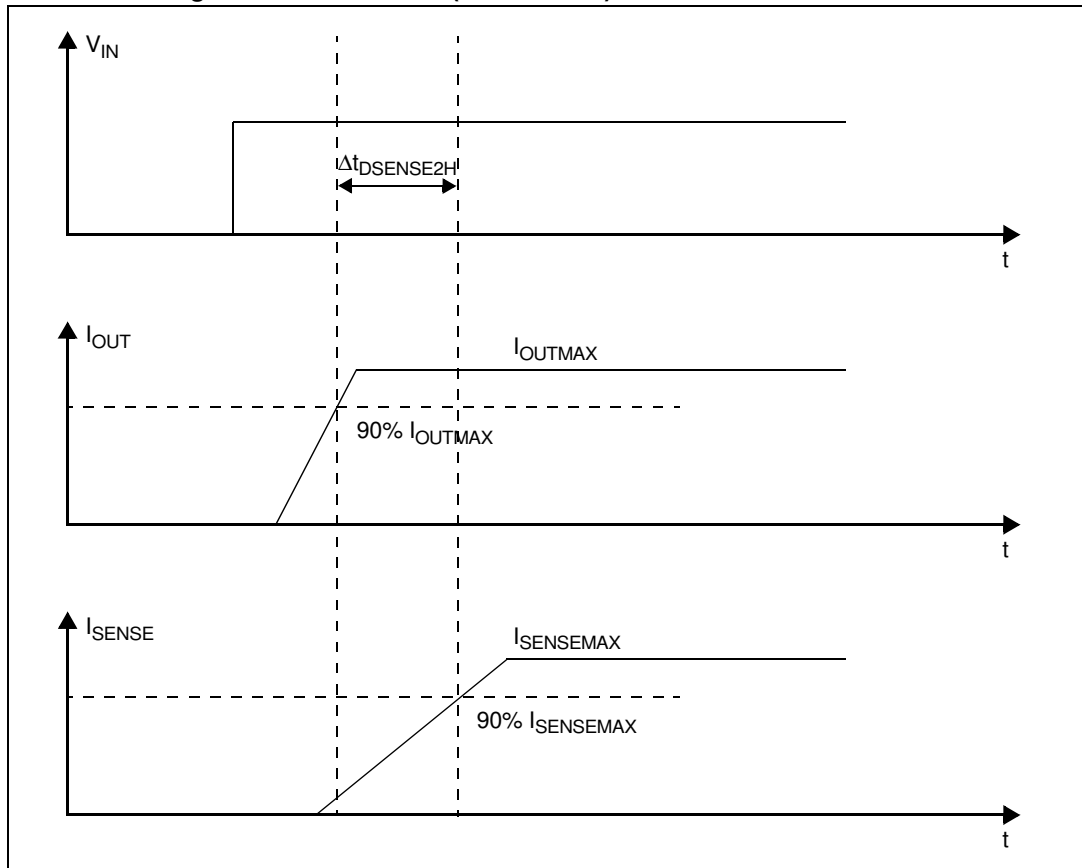


Figure 6. Switching characteristics

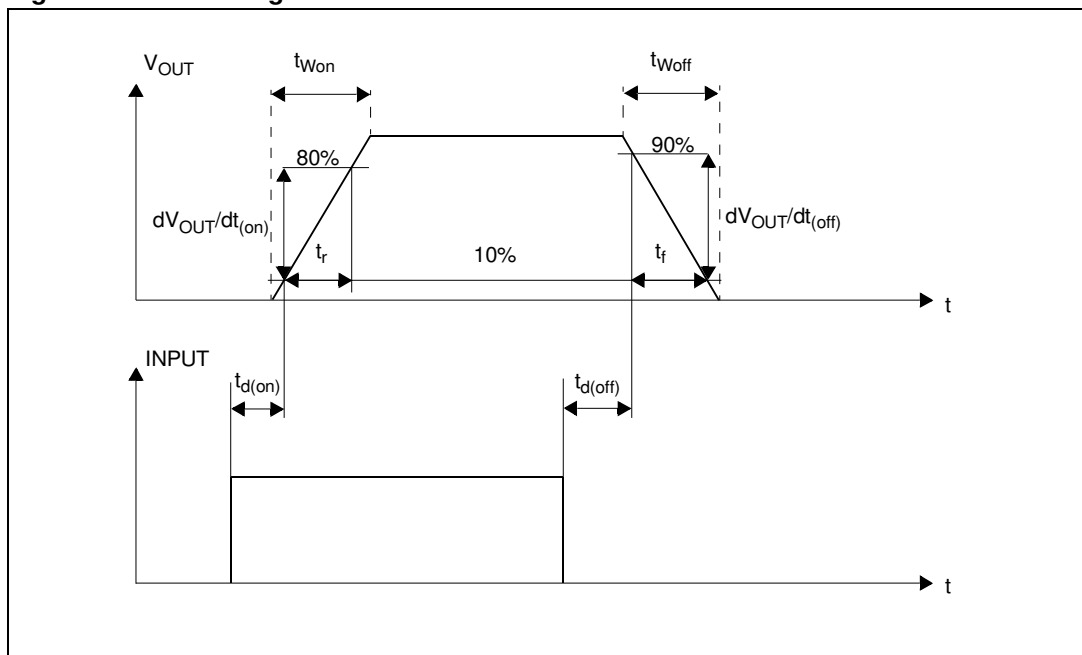
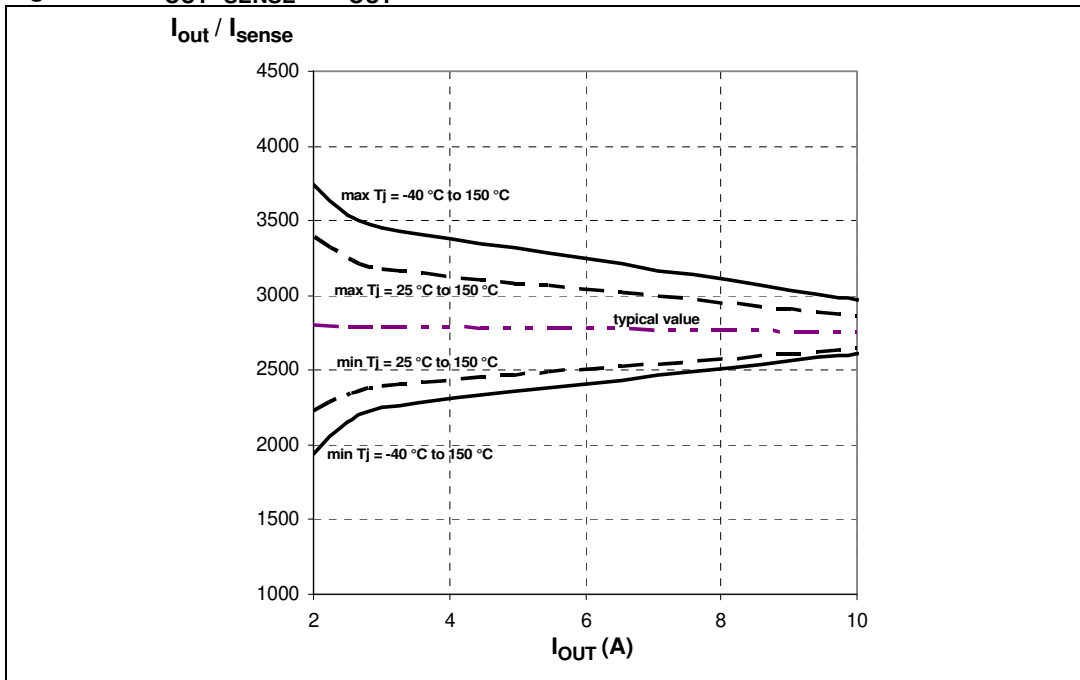
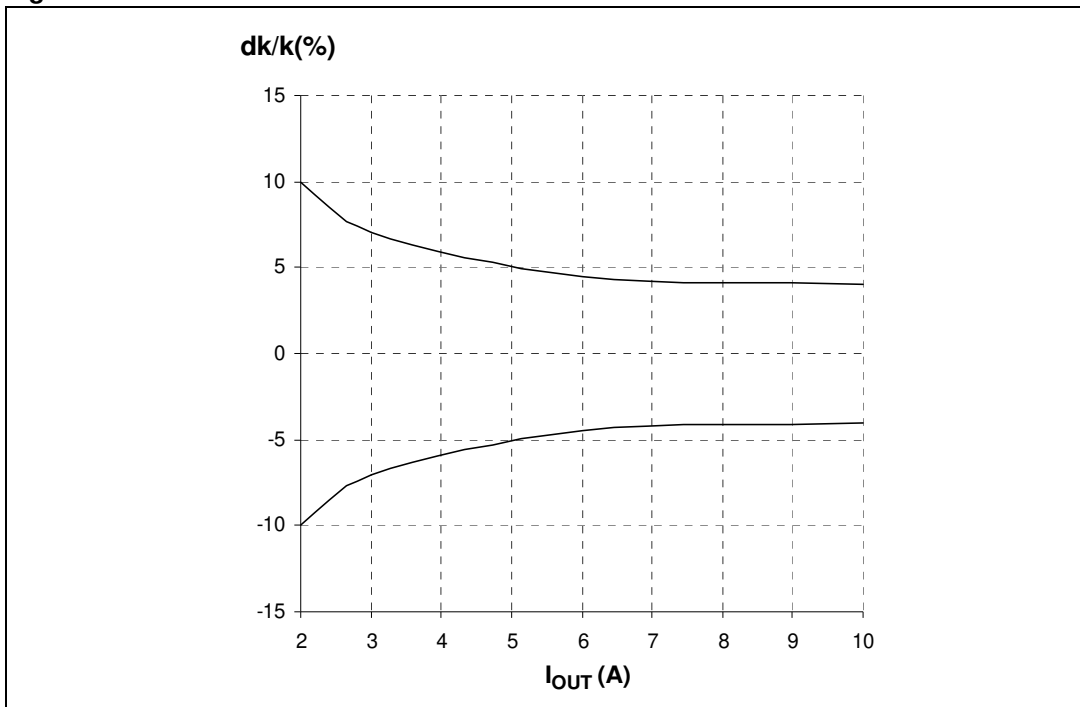


Figure 7. I_{OUT}/I_{SENSE} vs I_{OUT} (1)



1. See [Table 10](#) for details.

Figure 8. Maximum current sense ratio drift vs load current(1)



1. Parameter guaranteed by design; it is not tested.

Table 11. Truth table

Conditions	Input	Output	Sense ($V_{CSD} = 0V$) ⁽¹⁾
Normal operation	L	L	0
	H	H	Nominal
Over temperature	L	L	0
	H		V_{SENSEH}
Undervoltage	L	L	0
	H		
Short circuit to GND ($R_{SC} \leq 10m\Omega$)	L	L	0
	H		0 if $T_j < T_{TSD}$
			V_{SENSEH} if $T_j > T_{TSD}$
Short circuit to V_{CC}	L	H	0
	H		< Nominal
Negative output voltage clamp	L	L	0

1. If the V_{CSD} is high, the SENSE output is at a high impedance; its potential depends on leakage currents and external circuit.

Figure 9. Output voltage drop limitation

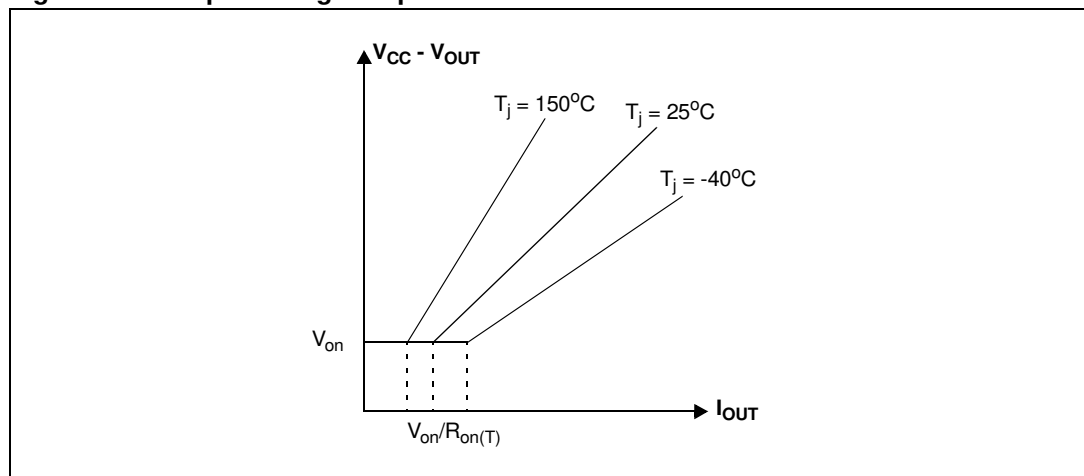


Table 12. Electrical transient requirements (part 1/3)

ISO 7637-2: 2004(E) Test pulse	Test levels ⁽¹⁾		Number of pulses or test times	Burst cycle/pulse repetition time		Delays and Impedance
	III	IV		Min.	Max.	
1	-75V	-100V	5000 pulses	0.5s	5s	2 ms, 10Ω
2a	+37V	+50V	5000 pulses	0.2s	5s	50μs, 2Ω
3a	-100V	-150V	1h	90ms	100ms	0.1μs, 50Ω
3b	+75V	+100V	1h	90ms	100ms	0.1μs, 50Ω
4	-6V	-7V	1 pulse			100ms, 0.01Ω
5b ⁽²⁾	+65V	+87V	1 pulse			400ms, 2Ω

1. The above test levels must be considered referred to $V_{CC} = 13.5V$ except for pulse 5b.
2. Valid in case of external load dump clamp: 40V maximum referred to ground.

Table 13. Electrical transient requirements (part 2/3)

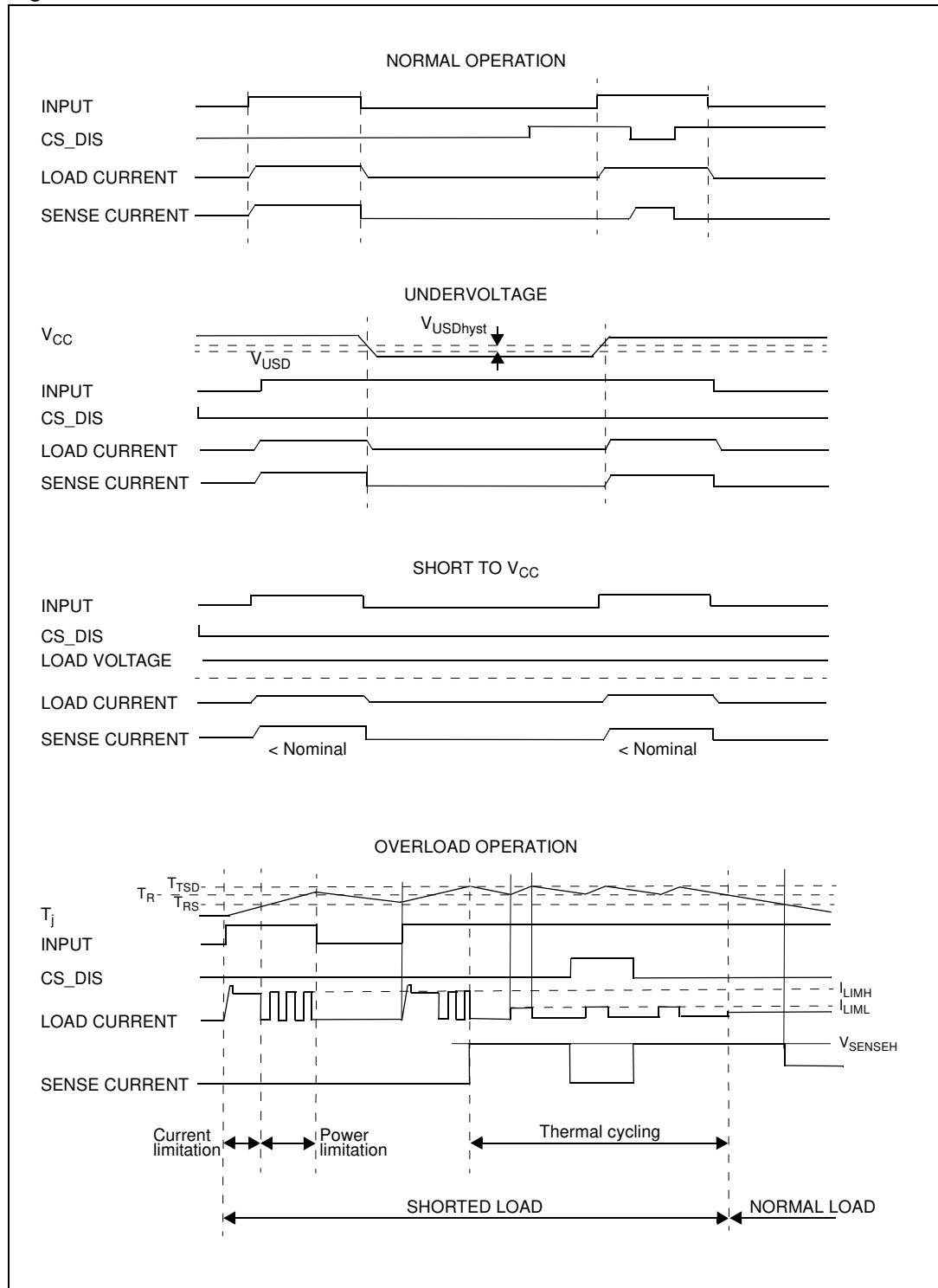
ISO 7637-2: 2004E Test pulse	Test level results	
	III	VI
1	C	C
2a	C	C
3a	C	C
3b	C	C
4	C	C
5b ⁽¹⁾	C	C

1. Valid in case of external load dump clamp: 40V maximum referred to ground.

Table 14. Electrical transient requirements (part 3/3)

Class	Contents
C	All functions of the device performed as designed after exposure to disturbance.
E	One or more functions of the device did not perform as designed after exposure to disturbance and cannot be returned to proper operation without replacing the device.

Figure 10. Waveforms



2.4 Electrical characteristics curves

Figure 11. Off-state output current

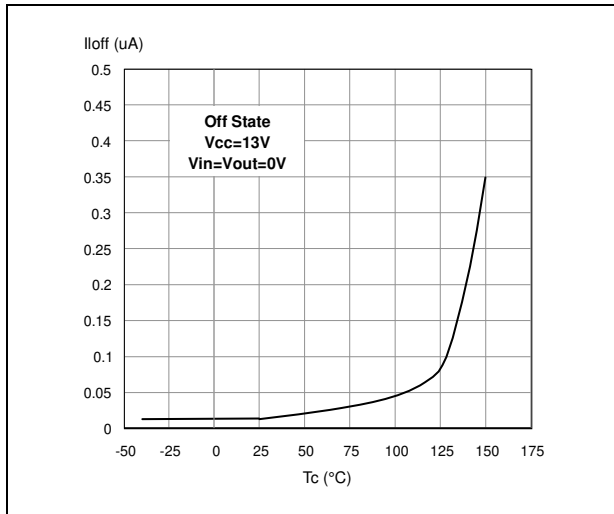


Figure 12. High level input current

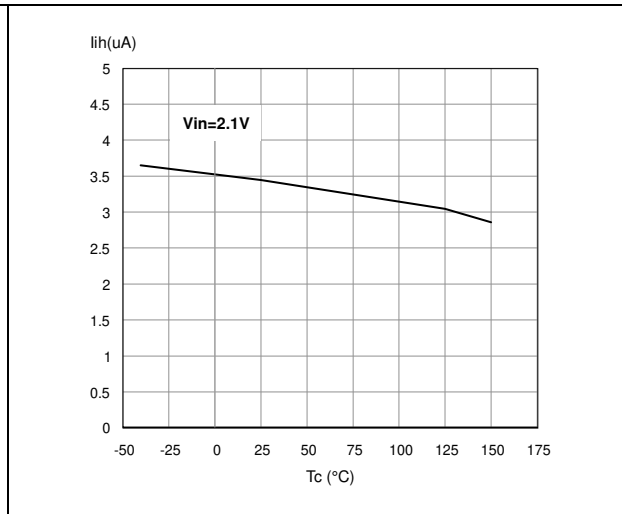


Figure 13. Input clamp voltage

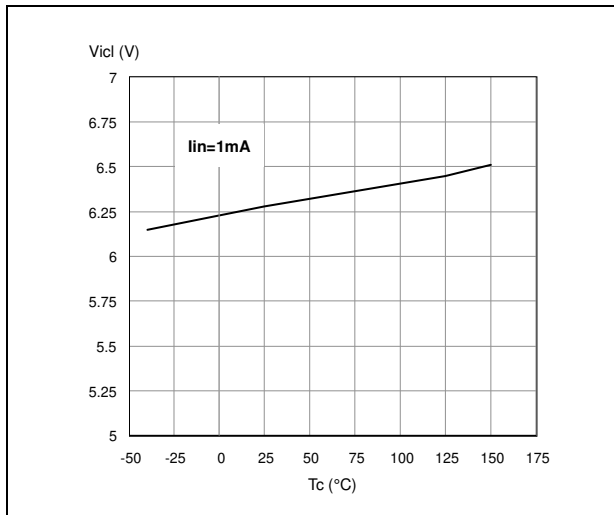


Figure 14. Input high level

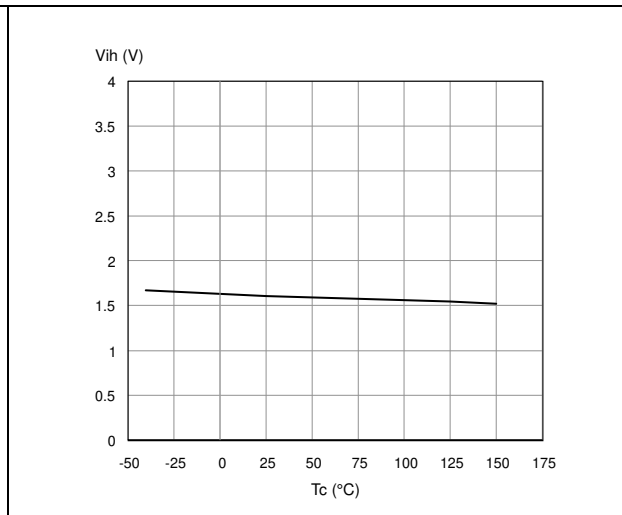


Figure 15. Input low level

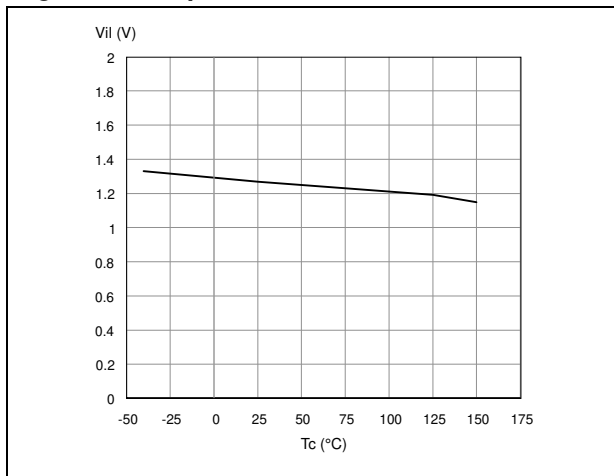


Figure 16. Input hysteresis voltage

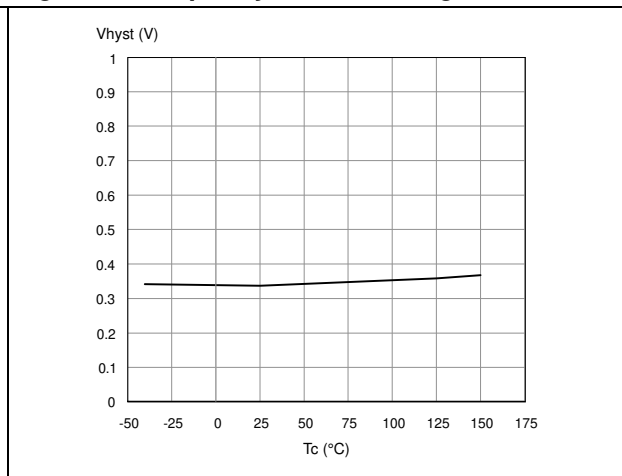


Figure 17. On-state resistance vs T_{case}

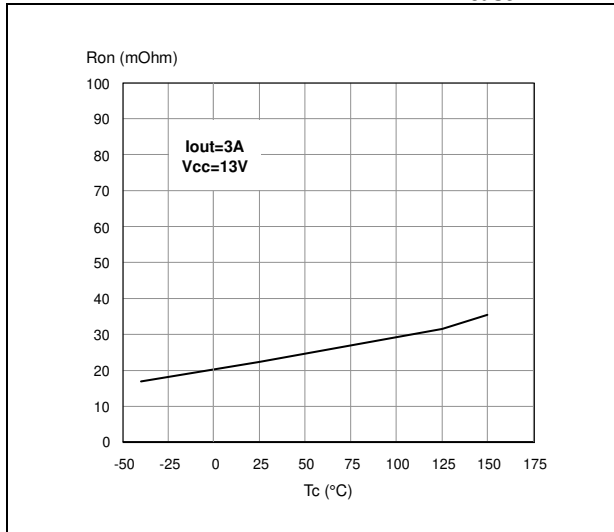


Figure 18. On-state resistance vs V_{CC}

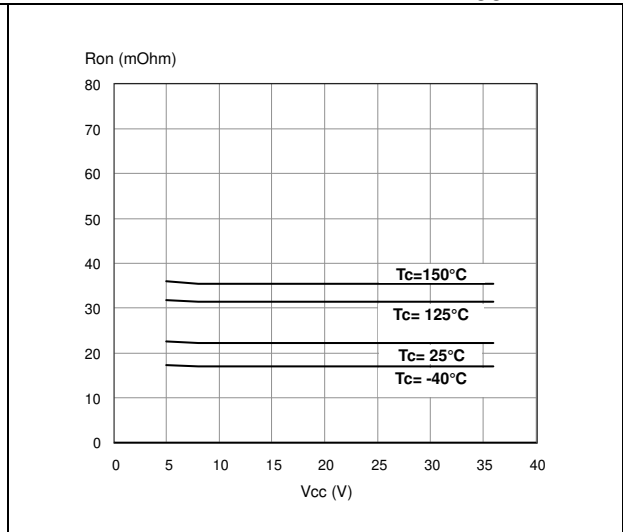


Figure 19. Undervoltage shutdown

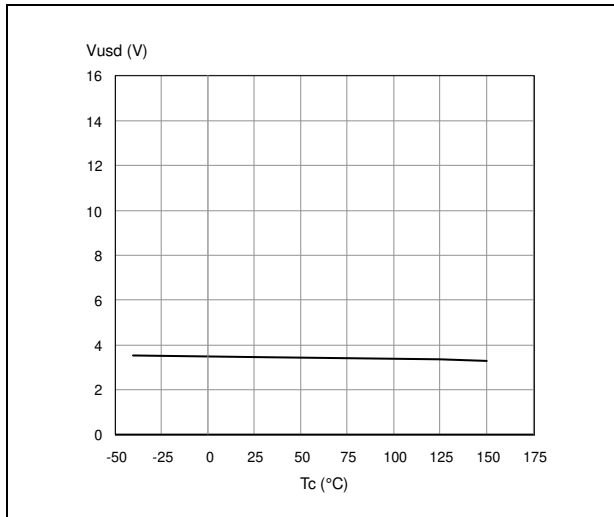


Figure 20. I_{LIMH} vs T_{case}

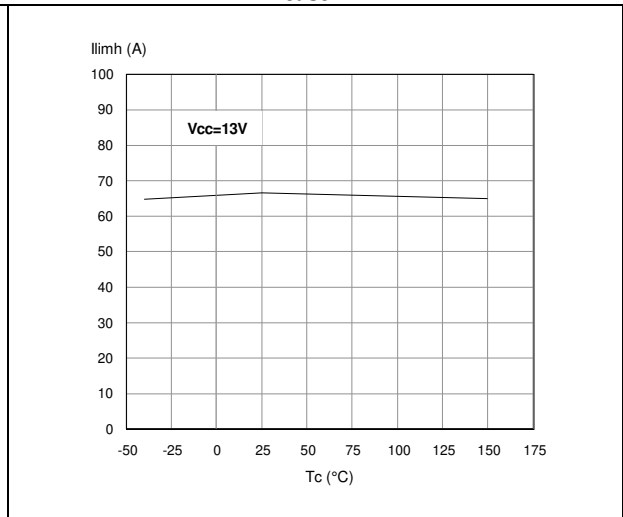


Figure 21. Turn-on voltage slope

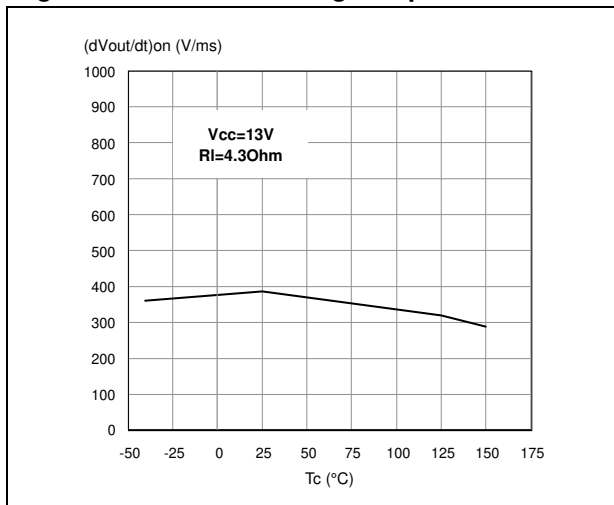


Figure 22. Turn-off voltage slope

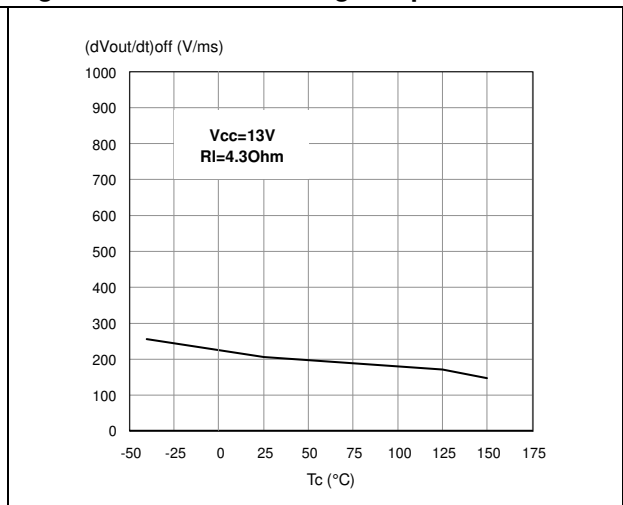


Figure 23. CS_DIS high level voltage

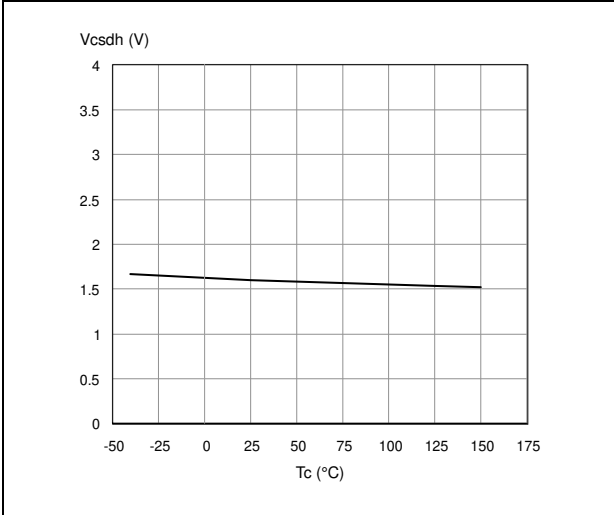


Figure 24. CS_DIS low level voltage

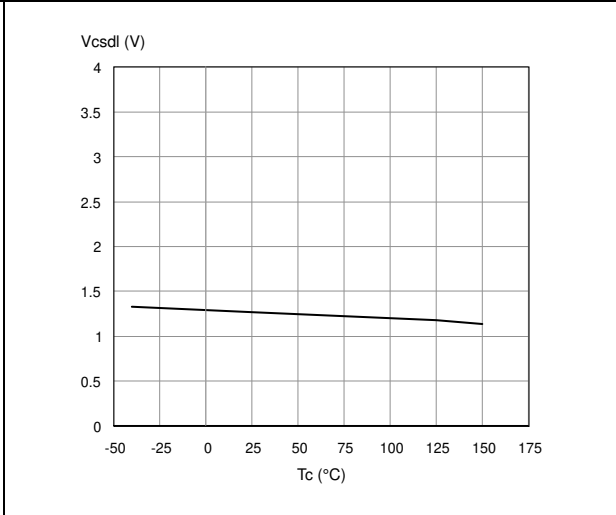
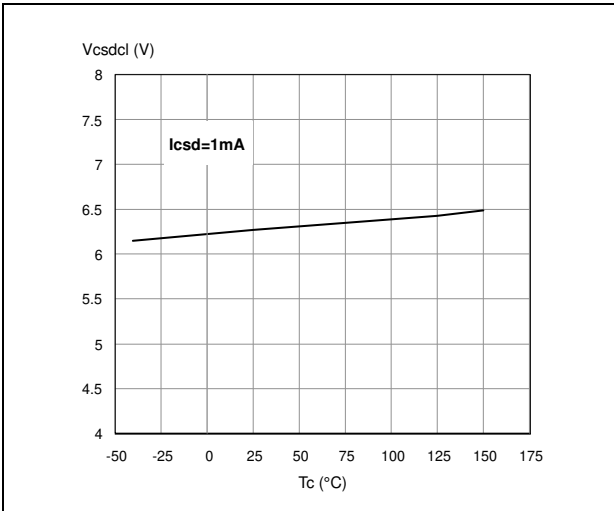
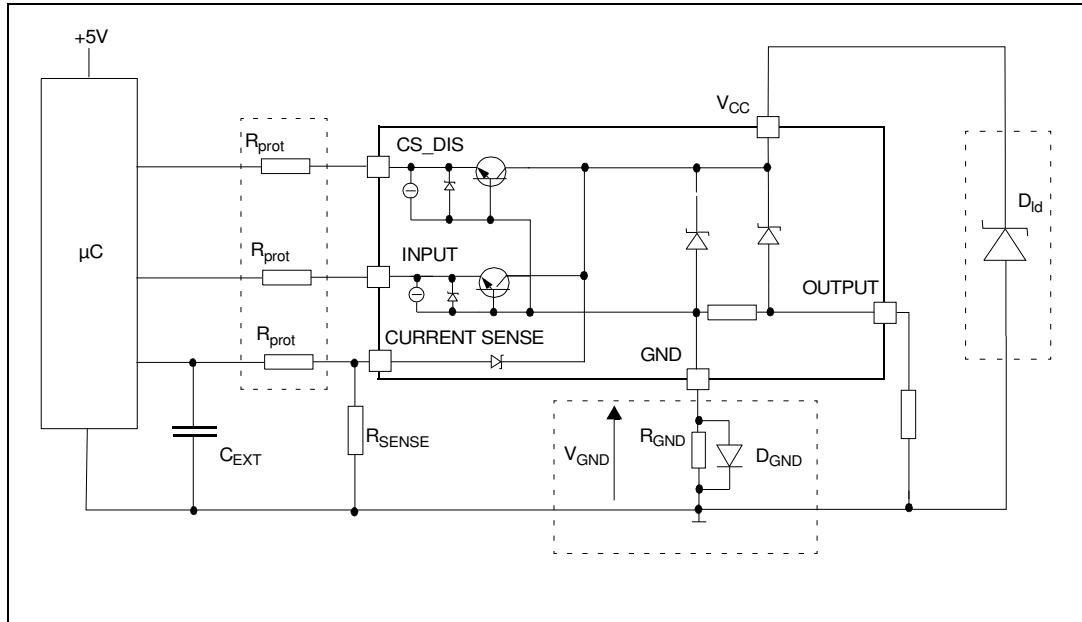


Figure 25. CS_DIS clamp voltage



3 Application information

Figure 26. Application schematic



Note: Channel 2 has the same internal circuit as channel 1.

3.1 GND protection network against reverse battery

This section provides two solutions for implementing a ground protection network against reverse battery.

3.1.1 Solution 1: resistor in the ground line (R_{GND} only)

This first solution can be used with any type of load.

The following formulas indicate how to dimension the R_{GND} resistor:

1. $R_{GND} \leq 600\text{mV} / (I_{S(ON)max})$
2. $R_{GND} \geq (-V_{CC}) / (-I_{GND})$

where $-I_{GND}$ is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device datasheet.

Power Dissipation in R_{GND} (when $V_{CC} < 0$ during reverse battery situations) is:

$$P_D = (-V_{CC})^2 / R_{GND}$$

This resistor can be shared among several different HSDs. Please note that the value of this resistor is calculated with formula (1), where $I_{S(ON)max}$ becomes the sum of the maximum on-state currents of the different devices.

Please note that if the microprocessor ground is not shared by the device ground, the R_{GND} produces a shift ($I_{S(ON)max} * R_{GND}$) in the input thresholds and the status output values. This shift varies depending on how many devices are ON in the case of several high-side drivers sharing the same R_{GND} .

If the calculated power dissipation leads to a large resistor or several devices have to share the same resistor, then ST suggests to utilize the following Solution 2.

3.1.2 Solution 2: diode (D_{GND}) in the ground line

If the device drives an inductive load, insert a resistor ($R_{GND} = 1k\Omega$) in parallel to D_{GND} .

This small signal diode can be safely shared among several different HSDs. Also in this case, the presence of the ground network produces a shift ($\approx 600mV$) in the input threshold and in the status output values if the microprocessor ground is not common to the device ground. This shift does not vary if more than one HSD shares the same diode/resistor network.

3.2 Load dump protection

D_{ld} is necessary (Voltage Transient Suppressor) if the load dump peak voltage exceeds the V_{CC} maximum DC rating. The same applies if the device is subject to transients on the V_{CC} line that are greater than the ones shown in the ISO 7637-2:2004E table.

3.3 MCU I/Os protection

If a ground protection network is used and negative transients are present on the V_{CC} line, the control pins are pulled negative. ST suggests to insert an in-line resistor (R_{prot}) to prevent the μC I/Os pins from latch-up.

The value of these resistors is a compromise between the leakage current of μC and the current required by the HSD I/Os (input levels compatibility) with the latch-up limit of μC I/Os.

$$-V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH} - V_{GND}) / I_{IHmax}$$

Calculation example:

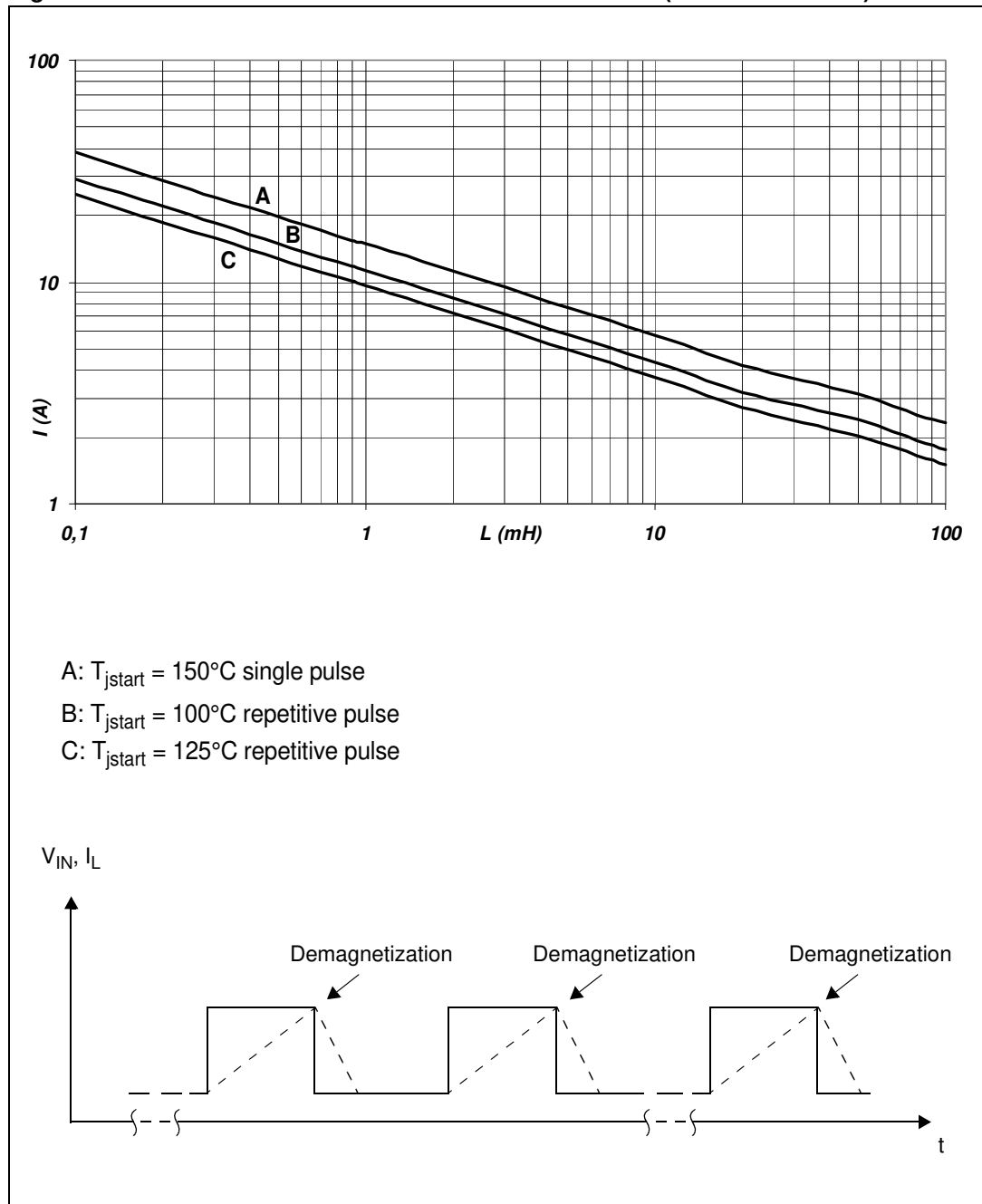
$$\text{For } V_{CCpeak} = -100V \text{ and } I_{latchup} \geq 20mA; V_{OH\mu C} \geq 4.5V$$

$$5k\Omega \leq R_{prot} \leq 180k\Omega$$

Recommended values: $R_{prot} = 10k\Omega$, $C_{EXT} = 10nF$.

3.4 Maximum demagnetization energy ($V_{CC} = 13.5V$)

Figure 27. Maximum turn-off current versus inductance (for each channel)

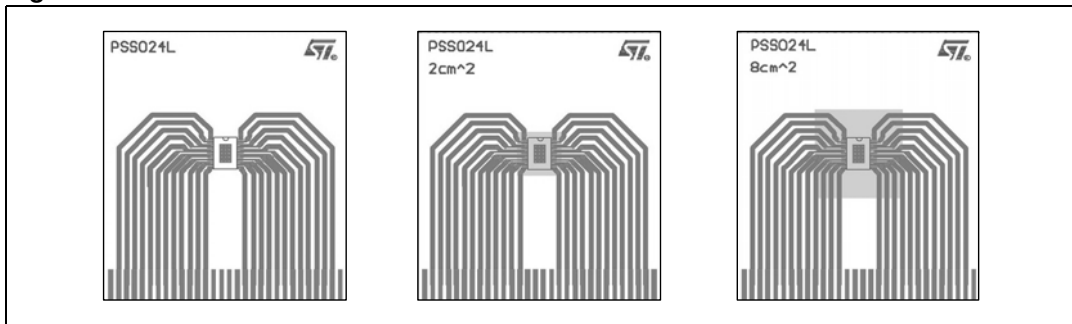


Note: Values are generated with $R_L = 0 \Omega$.
 In case of repetitive pulses, T_{jstart} (at beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves A and B.

4 Package and thermal data

4.1 PowerSSO-24™ thermal data

Figure 28. PowerSSO-24™ PC board



Note: Layout condition of R_{th} and Z_{th} measurements (PCB: Double layer, Thermal Vias, FR4 area = 77mm x 86mm, PCB thickness = 1.6mm, Cu thickness = 70µm (front and back side), Copper areas: from minimum pad layout to 8cm²).

Figure 29. $R_{thj-amb}$ vs PCB copper area in open box free air condition (one channel ON)

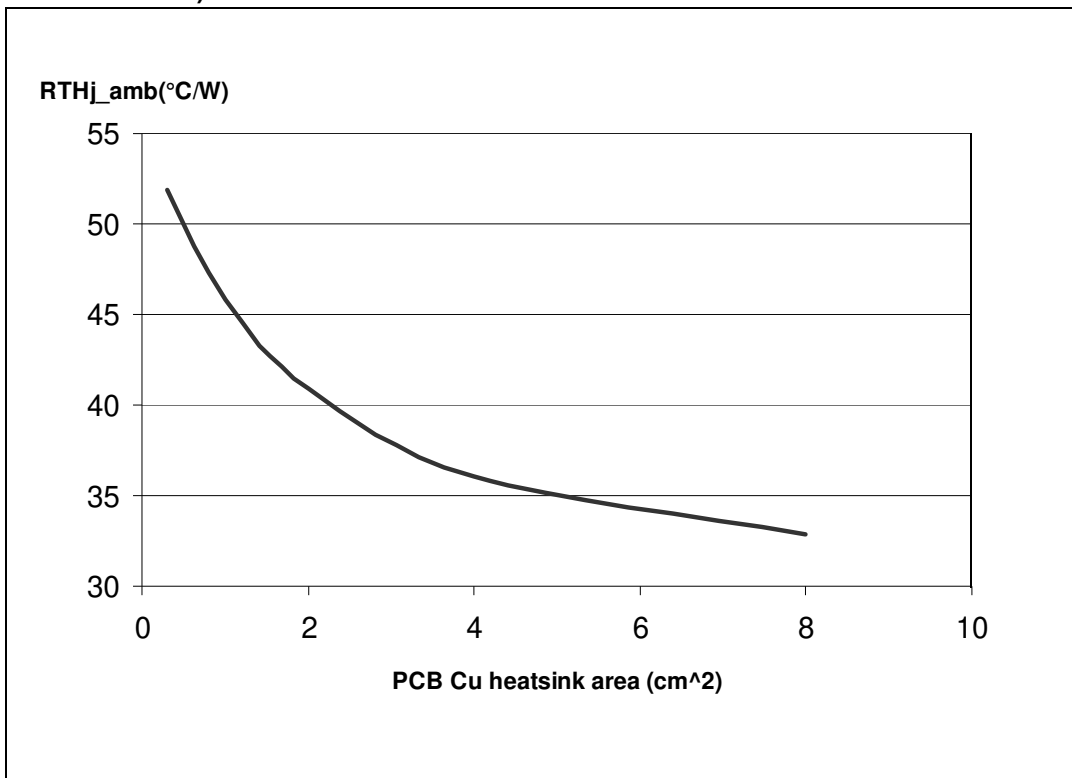
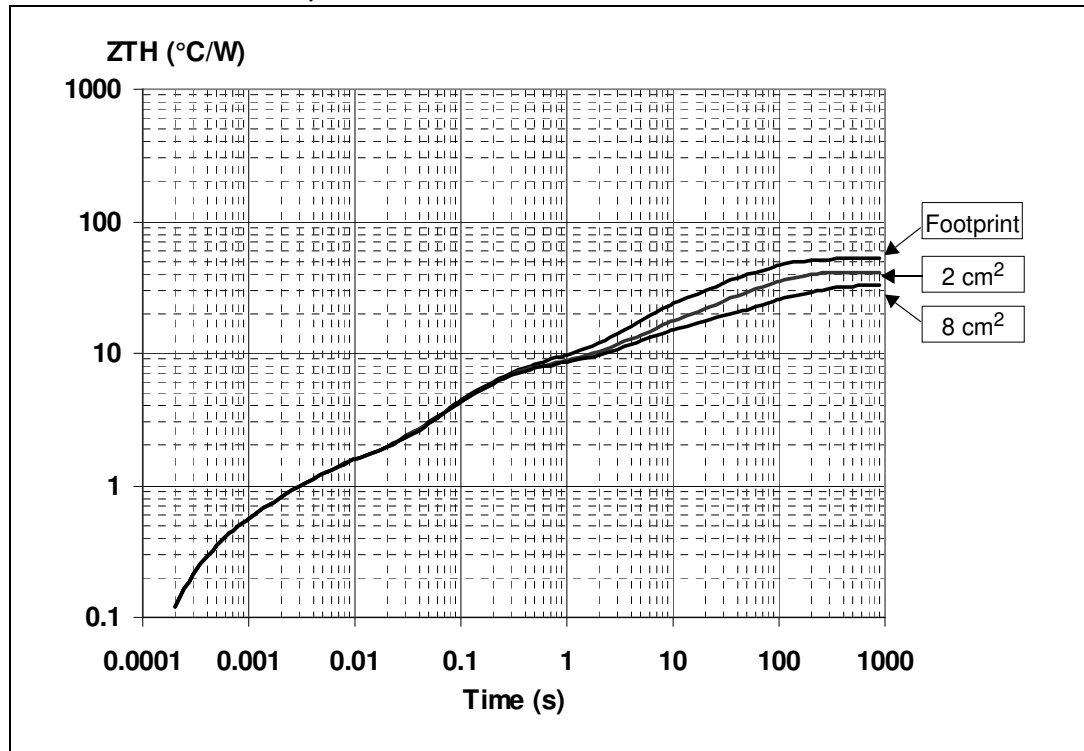


Figure 30. PowerSSO-24™ thermal impedance junction to ambient single pulse (one channel ON)

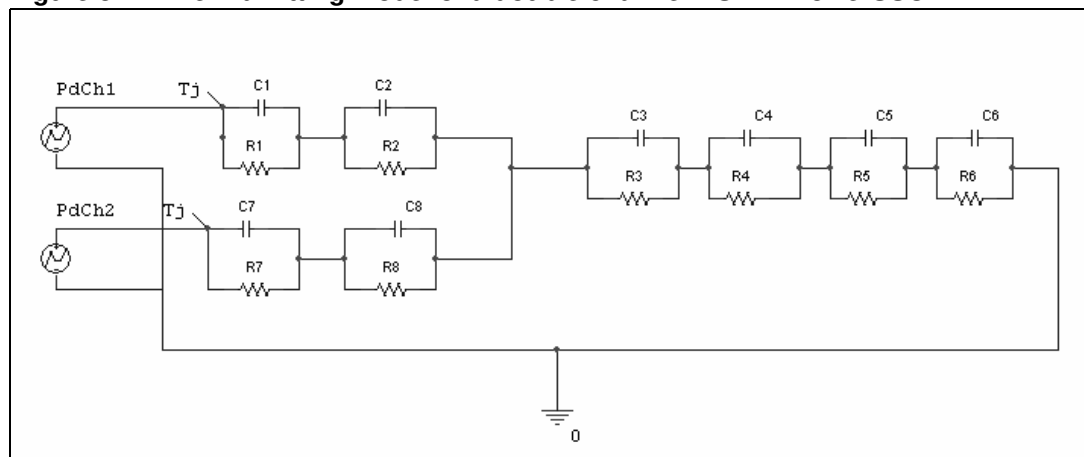


Equation 1: pulse calculation formula

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp}(1 - \delta)$$

where $\delta = t_p/T$

Figure 31. Thermal fitting model of a double channel HSD in PowerSSO-24™(1)



1. The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.