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24-bit, 192kHz Stereo CODEC with Volume Control

DESCRIPTION

The WM8569 is a stereo audio codec ideal for DVD, PVR, LCD-TV and automotive applications. Independent ADC and DAC clocking permits separate record and playback sampling rates.

A stereo 24-bit multi-bit sigma delta DAC with oversampling and digital interpolation filters provides the output signal. Digital audio input word lengths from 16-24 bits and sampling rates from 8kHz to 192kHz are supported.

A stereo 24-bit multi-bit sigma delta ADC is used. Digital audio output word lengths from 16-24 bits and sampling rates from 32kHz to 96kHz are supported. The DAC and ADC support independent sampling rates.

The audio data interface supports I^2S , left justified, right justified and DSP digital audio formats.

The device is controlled via a 3-wire serial interface. The interface provides access to all features including volume controls, mutes, de-emphasis and power management facilities. A hardware control interface allows access to a limited feature set. The WM8569 is available in a 28-lead SSOP.

FEATURES

- Audio Performance
 - 103dB SNR ('A' weighted @ 48kHz) DAC
 - DAC Sampling Frequency: 8kHz 192kHz
 - 100dB SNR ('A' weighted @ 48kHz) ADC
 - ADC Sampling Frequency: 32kHz 96kHz
- Independent Sampling Rates for ADC and DAC
- 3-wire SPI Serial or Hardware Control Interface
- · Audio Mute and De-Emphasis Functions
- Programmable Audio Data Interface Modes
 - 16/20/24/32 bit Word Lengths
 - I²S, Left, Right Justified or DSP
- 2.7V to 5.5V Analogue Supply
- 2.7V to 3.6V Digital Supply
- 28-lead SSOP Package

APPLICATIONS

- DVD Recorder
- Personal Video Recorder
- PC Sound Card
- Automotive Audio
- LCD-TV

BLOCK DIAGRAM

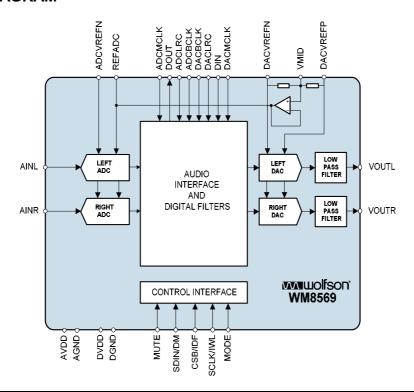


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PIN CONFIGURATION

DACMCLK [1 •	28	ADCMCLK
ADCBCLK [2	27	MODE
DACBCLK [3	26	AVDD
ADCLRC	4	25	AGND
DACLRC [5	24	NC
DVDD	6	23	VOUTR
DGND [7	22	VOUTL
DIN [8	21	AINL
NC [9	20	AINR
DOUT [10	19	VMID
CSB/IDF	11	18	DACVREFP
SCLK/IWL	12	17	DACVREFN
SDIN/DM	13	16	ADCVREFN
MUTE	14	15	REFADC

ORDERING INFORMATION

DEVICE	TEMPERATURE RANGE	PACKAGE	MOISTURE SENSITIVITY LEVEL	PEAK SOLDERING TEMPERATURE
WM8569SEDS/V	-25 to +85°C	28-lead SSOP (Pb-free)	MSL3	260°C
WM8569SEDS/RV	-25 to +85°C	28-lead SSOP (Pb-free, tape and reel)	MSL3	260°C

Note: Reel quantity = 2,000

PIN DESCRIPTION

PIN	NAME	TYPE	DESCRIPTION
1	DACMCLK	Digital Input	Master DAC clock: 256fs, 384fs, 512fs or 768fs
2	ADCBCLK	Digital Input	ADC audio interface bit clock
3	DACBLCK	Digital Input/Output	DAC audio interface bit clock
4	ADCLRC	Digital Input/Output	ADC left/right word clock
5	DACLRC	Digital Input/Output	DAC left/right word clock
6	DVDD	Supply	Digital positive supply
7	DGND	Supply	Digital ground
8	DIN	Digital input	DAC data input
9	NC	No Connect	No Connect
10	DOUT	Digital output	ADC data output
11	CSB/IDF	Digital input	Software Mode: Serial control interface latch signal
			Hardware Mode: Input audio data format
12	SCLK/IWL	Digital input	Software Mode: Serial control interface clock
			Hardware Mode: Audio data input word length
13	SDIN/DM	Digital input	Software Mode: Serial control interface data
			Hardware Mode: De-emphasis selection
14	MUTE	Digital input/output	DAC Zero Flag output or DAC mute input
15	REFADC	Analogue output	ADC reference buffer decoupling pin; 10uF external decoupling
16	ADCVREFN	Supply	ADC negative reference supply
17	DACVREFN	Supply	DAC negative reference supply
18	DACVREFP	Supply	DAC positive reference supply
19	VMID	Analogue output	Midrail divider decoupling pin; 10uF external decoupling
20	AINR	Analogue input	ADC right input
21	AINL	Analogue input	ADC left input
22	VOUTL	Analogue output	DAC left output
23	VOUTR	Analogue output	DAC right output
24	NC	No Connect	No Connect
25	AGND	Supply	Analogue negative supply and substrate connection
26	AVDD	Supply	Analogue positive supply
27	MODE	Digital Input	Control format selection:
			0 = Software mode
			1 = Hardware mode
28	ADCMCLK	Digital Input	Master ADC clock: 256fs, 384fs, 512fs or 768fs

 $\textbf{Note} \hbox{: Digital input pins have Schmitt trigger input buffers}.$



ABSOLUTE MAXIMUM RATINGS

Absolute Maximum Ratings are stress ratings only. Permanent damage to the device may be caused by continuously operating at or beyond these limits. Device functional operating limits and guaranteed performance specifications are given under Electrical Characteristics at the test conditions specified.



ESD Sensitive Device. This device is manufactured on a CMOS process. It is therefore generically susceptible to damage from excessive static voltages. Proper ESD precautions must be taken during handling and storage of this device.

Wolfson tests its package types according to IPC/JEDEC J-STD-020B for Moisture Sensitivity to determine acceptable storage conditions prior to surface mount assembly. These levels are:

MSL1 = unlimited floor life at <30°C / 85% Relative Humidity. Not normally stored in moisture barrier bag.

MSL2 = out of bag storage for 1 year at <30°C / 60% Relative Humidity. Supplied in moisture barrier bag.

MSL3 = out of bag storage for 168 hours at <30°C / 60% Relative Humidity. Supplied in moisture barrier bag.

The Moisture Sensitivity Level for each package type is specified in Ordering Information.

CONDITION	MIN	MAX
Digital supply voltage	-0.3V	+5V
Analogue supply voltage	-0.3V	+7V
Voltage range digital inputs ¹	DGND -0.3V	DVDD +0.3V
Voltage range analogue inputs ¹	AGND -0.3V	AVDD +0.3V
Master Clock Frequency		37MHz
Operating temperature range, T _A	-25°C	+85°C
Storage temperature after soldering	-65°C	+150°C

Notes:

1. Analogue and digital grounds must always be within 0.3V of each other.



RECOMMENDED OPERATING CONDITIONS

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Digital supply range	DVDD		2.7		3.6	V
Analogue supply range	AVDD, VREFP		2.7		5.5	V
Ground	AGND, VREFN, DGND			0		V
Difference DGND to AGND			-0.3	0	+0.3	V

Note: Digital supply DVDD must never be more than 0.3V greater than AVDD.

ELECTRICAL CHARACTERISTICS

Test Conditions

AVDD, VREFP = 5V, DVDD = 3.3V, AGND, VREFN = 0V, DGND = 0V, T_A = +25°C, fs = 48kHz, MCLK = 256fs.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DAC Performance (Load = 10kΩ	50pF)	1		•		
0dBFs Full scale output voltage				1.0 x		V _{rms}
				VREFP/5		
SNR (Note 1,2,4)		A-weighted,	95	103		dB
, , ,		@ fs = 48kHz				
SNR (Note 1,2,4)		A-weighted		102		dB
,		@ fs = 96kHz				
SNR (Note 1,2,4)		A-weighted		101		dB
,		@ fs = 192kHz				
SNR (Note 1,2,4)		A-weighted		99		dB
, , ,		@ fs = 48kHz, AVDD =				
		3.3V				
SNR (Note 1,2,4)		A-weighted		99		dB
		@ fs = 96kHz, AVDD =				
		3.3V				
Dynamic Range (Note 2,4)	DNR	A-weighted, -60dB full	90	103		dB
		scale input				
Total Harmonic Distortion (THD)		1kHz, 0dB.Fs		-90	-80	dB
Mute Attenuation		1kHz Input, 0dB gain		100		dB
DAC channel separation				100		dB
Power Supply Rejection Ratio	PSRR	1kHz 100mV _{p-p}		50		dB
		20Hz to 20kHz		45		dB
		100mV _{p-p}				
ADC Performance						
Input Signal Level (0dB)				2.0 x		V_{rms}
				REFADC/5		
Input resistance				20		kΩ
Input capacitance				10		pF
SNR (Note 1,2,4)		A-weighted, 0dB gain	80	100		dB
		@ fs = 48kHz				
SNR (Note 1,2,4)		A-weighted, 0dB gain		100		dB
		@ fs = 96kHz				
SNR (Note 1,2,4)		A-weighted, 0dB gain		93		dB
		@ fs = 48kHz, AVDD =				
		3.3V				
SNR (Note 1,2,4)		A-weighted, 0dB gain		93		dB
		@ fs = 96kHz, AVDD				
		= 3.3V				
Total Harmonic Distortion (THD)		1kHz, -1dBFs		-82	-72	dB
ADC Channel Separation		1kHz Input		90		dB



Test Conditions

AVDD, VREFP = 5V, DVDD = 3.3V, AGND, VREFN = 0V, DGND = 0V, T_A = $+25^{\circ}$ C, fs = 48kHz, MCLK = 256fs.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Mute Attenuation		1kHz Input, 0dB gain		90		dB
Power Supply Rejection Ratio	PSRR	1kHz 100mVpp		50		dB
		20Hz to 20kHz 100mVpp		45		dB
Digital Logic Levels (CMOS Lev	els)					
Input LOW level	V _{IL}				0.3 x DVDD	V
Input HIGH level	V _{IH}		0.7 x DVDD			V
Input leakage current				±0.2	±1	μΑ
Input capacitance				5		pF
Output LOW	V _{OL}	I _{OL} =1mA			0.1 x DVDD	V
Output HIGH	V _{OH}	I _{OH} = -1mA	0.9 x DVDD			V
Analogue Reference Levels						
Reference voltage	V_{VMID}		VREFP/2 – 50mV	VREFP/2	VREFP/2 + 50mV	V
Potential divider resistance	R _{VMID}	VREFP to VMID and VMID to VREFN		50		kΩ
Supply Current		•	•			
Analogue supply current		AVDD, VREFP = 5V		TBD		mA
Digital supply current		DVDD = 3.3V		TBD		mA

Notes:

- Ratio of output level with 1kHz full scale input, to the output level with all zeros into the digital input, measured 'A' weighted.
- All performance measurements done with 20kHz low pass filter, and where noted an A-weight filter. Failure to use such a filter will result in higher THD+N and lower SNR and Dynamic Range readings than are found in the Electrical Characteristics. The low pass filter removes out of band noise; although it is not audible it may affect dynamic specification values.
- 3. VMID decoupled with 10uF and 0.1uF capacitors (smaller values may result in reduced performance).

TERMINOLOGY

- 1. Signal-to-noise ratio (dB) SNR is a measure of the difference in level between the full scale output and the output with no signal applied. (No Auto-zero or Automute function is employed in achieving these results).
- 2. Dynamic range (dB) DNR is a measure of the difference between the highest and lowest portions of a signal. Normally a THD+N measurement at 60dB below full scale. The measured signal is then corrected by adding the 60dB to it. (e.g. THD+N @ -60dB= -32dB, DR= 92dB).
- 3. THD (dB) THD is a ratio, of the rms values, of Distortion/Signal.
- 4. Stop band attenuation (dB) Is the degree to which the frequency spectrum is attenuated (outside audio band).
- 5. Channel Separation (dB) Also known as Cross-Talk. This is a measure of the amount one channel is isolated from the other. Normally measured by sending a full scale signal down one channel and measuring the other.
- 6. Pass-Band Ripple Any variation of the frequency response in the pass-band region.

MASTER CLOCK TIMING

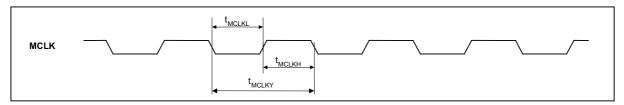


Figure 1 ADC and DAC Master Clock Timing Requirements

Test Conditions

AVDD, VREFP = 5V, DVDD = 3.3V, AGND, VREFN = 0V, AGND, DGND = 0V, T_A = +25°C, fs = 48kHz, DACMCLK and ADCMCLK = 256fs unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
System Clock Timing Information	on					
MCLK System clock pulse width high	t _{MCLKH}		11			ns
MCLK System clock pulse width low	t _{MCLKL}		11			ns
MCLK System clock cycle time	t _{MCLKY}		28			ns
MCLK Duty cycle			40:60		60:40	

Table 1 Master Clock Timing Requirements

DIGITAL AUDIO INTERFACE - MASTER MODE

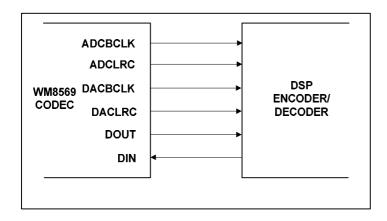


Figure 2 Audio Interface - Master Mode

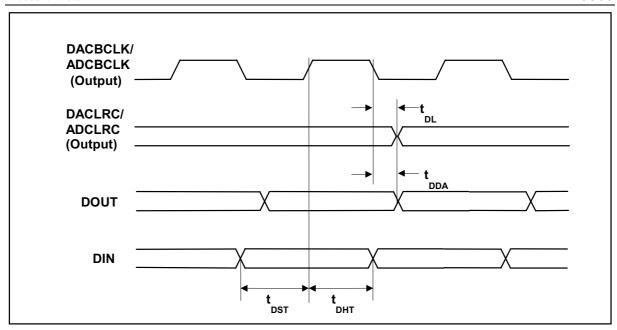


Figure 3 Digital Audio Data Timing – Master Mode

Test Conditions

AVDD, VREFP = 5V, DVDD = 3.3V, AGND, VREFN, DGND = 0V, T_A = +25°C, Master Mode, fs = 48kHz, MCLK = 256fs unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
Audio Data Input Timing Info	Audio Data Input Timing Information							
DACLRC/ADCLRC propagation delay from DACBCLK/ADCBCLK falling edge	t _{DL}		0		10	ns		
DOUT propagation delay from ADCBCLK falling edge	t _{DDA}		0		10	ns		
DIN setup time to DACBCLK rising edge	t _{DST}		10			ns		
DIN hold time from DACBCLK rising edge	t _{DHT}		10			ns		

Table 2 Digital Audio Data Timing – Master Mode

DIGITAL AUDIO INTERFACE – SLAVE MODE

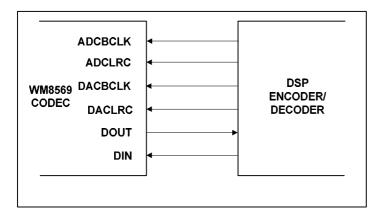


Figure 4 Audio Interface - Slave Mode

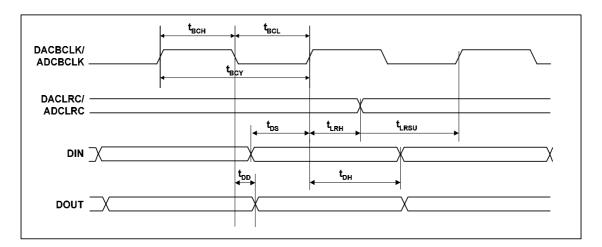


Figure 5 Digital Audio Data Timing - Slave Mode

Test Conditions

AVDD = 5V, DVDD = 3.3V, AGND = 0V, DGND = 0V, T_A = +25°C, Slave Mode, fs = 48kHz, DACMCLK and ADCMCLK = 256fs unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
Audio Data Input Timing Information								
ADCBCLK/DACBCLK cycle time	t _{BCY}		50			ns		
ADCBCLK/DACBCLK pulse width high	t _{BCH}		20			ns		
ADCBCLK/DACBCLK pulse width low	t _{BCL}		20			ns		
ADCLRC/DACLRC set-up time to ADCBCLK/DACBCLK rising edge	t _{LRSU}		10			ns		
ADCLRC/DACLRC hold time from ADCBCLK/DACBCLK rising edge	t _{LRH}		10			ns		
DIN set-up time to DACBCLK rising edge	t _{DS}		10			ns		
DIN hold time from DACBCLK rising edge	t _{DH}		10			ns		
DOUT propagation delay from ADCBCLK falling edge	t _{DD}		0		10	ns		

Table 3 Digital Audio Data Timing – Slave Mode

MPU INTERFACE TIMING

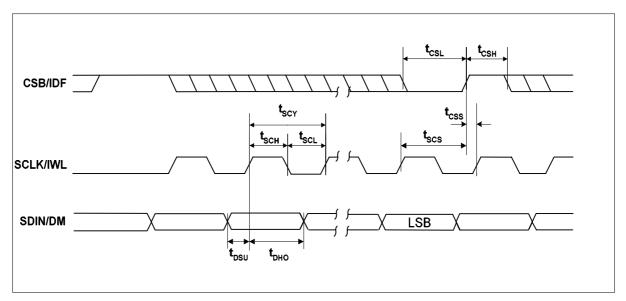


Figure 6 SPI Compatible Control Interface Input Timing

Test Conditions

AVDD = 5V, DVDD = 3.3V, AGND, DGND = 0V, $T_A = +25^{\circ}C$, fs = 48kHz, DACMCLK and ADCMCLK = 256fs unless otherwise stated

PARAMETER	SYMBOL	MIN	TYP	MAX	UNIT
SCLK/IWL rising edge to CSB/IDF rising edge	tscs	60			ns
SCLK/IWL pulse cycle time	t _{scy}	80			ns
SCLK/IWL pulse width low	t _{SCL}	30			ns
SCLK/IWL pulse width high	t _{scн}	30			ns
SDIN/DM to SCLK/IWL set-up time	t _{DSU}	20			ns
SCLK/IWL to SDIN/DM hold time	t _{DHO}	20			ns
CSB/IDF pulse width low	t _{CSL}	20			ns
CSB/IDF pulse width high	t _{CSH}	20			ns
CSB/IDF rising to SCLK/IWL rising	t _{CSS}	20			ns

Table 4 3-Wire SPI Compatible Control Interface Input Timing Information

INTERNAL POWER ON RESET CIRCUIT

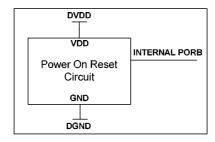


Figure 7 Internal Power on Reset Circuit Schematic

The WM8569 includes an internal Power-On-Reset Circuit, as shown in Figure 7, which is used reset the digital logic into a default state after power up. The POR circuit is powered from DVDD and monitors DVDD. It asserts PORB low if DVDD is below a minimum threshold.

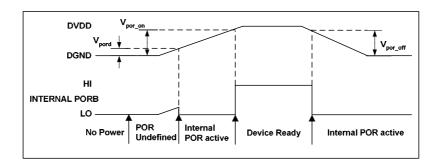


Figure 8 Typical Power-Up Sequence

Figure 8 shows a typical power-up sequence. When DVDD goes above the minimum threshold, Vpord, there is enough voltage for the circuit to guarantee PORB is asserted low and the chip is held in reset. In this condition, all writes to the control interface are ignored. When DVDD rises to Vpor_on, PORB is released high and all registers are in their default state and writes to the control interface may take place.

On power down, PORB is asserted low whenever DVDD drops below the minimum threshold $Vpor_off$.

SYMBOL	MIN	TYP	MAX	UNIT
V_{pord}	0.3	0.5	0.8	V
V_{por_on}	1.3	1.7	2.0	V
V_{por_off}	1.3	1.7	2.0	V

Table 5 Typical POR Operation (typical values, not tested)

DEVICE DESCRIPTION

INTRODUCTION

WM8569 is a complete 2-channel audio codec, including digital interpolation and decimation filters, multi-bit sigma delta stereo ADC, and switched capacitor multi-bit sigma delta DAC with volume control and output smoothing filter.

The device is implemented as a separate stereo DAC and a stereo ADC in a single package and controlled by a single 3-wire software or hardware interface.

The DAC has its own data input DIN, DAC word clock DACLRC, DAC bit clock DACBCLK and DAC master clock DACMCLK while the stereo ADC has its own data output DOUT, word clock ADCLRC, bit clock ADCBCLK and ADC master clock ADCMCLK. This allows the ADC and DAC to operate independently.

The Audio Interface may be configured to operate in either master or slave mode. In Slave mode DACLRC/ADCLRC and DACBCLK/ADCBCLK are all inputs. In Master mode DACLRC/ADCLRC and DACBCLK/ADCBCLK are all outputs.

The DAC has its own digital volume control that is adjustable in 0.5dB steps. A zero cross detect circuit is provided. The digital volume control detects a transition through the zero point before updating the volume. This minimises audible clicks and 'zipper' noise as the gain values change.

Control of internal functionality of the device is by 3-wire serial or pin programmable control interface.

Operation using master clocks of 128fs, 192fs, 256fs, 384fs, 512fs or 768fs is provided for the DAC, for operation of both the ADC and DAC master clocks of 256fs, 384fs, 512fs and 768fs is provided. In Slave mode, selection between clock rates is automatically controlled. In master mode, the sample rate is set by control bits DACRATE and ADCRATE. Audio sample rates (fs) from less than 8kHz up to 192kHz are allowed for the DAC and from less than 32kHz up to 96kHz for the ADC, provided the appropriate master clock is input.

The audio data interface supports right-justified, left-justified and I^2S interface formats along with a highly flexible DSP serial port interface.

AUDIO DATA SAMPLING RATES

In a typical digital audio system there is only one central clock source producing a reference clock to which all audio data processing is synchronised. This clock is often referred to as the audio system's Master Clock. The external master system clock can be applied directly through the ADC and DAC MCLK input pin(s) with no software configuration necessary. In a system where there are a number of possible sources for the reference clock it is recommended that the clock source with the lowest jitter be used to optimise the performance of the ADC and DAC.

The DAC master clock for WM8569 supports audio sampling rates from 128fs to 768fs, where fs is the audio sampling frequency (DACLRC) typically 32kHz, 44.1kHz, 48kHz, 96kHz or 192kHz. The ADC master clock for WM8569 supports audio sampling rates from 256fs to 768fs, where fs is the audio sampling frequency (ADCLRC) typically 32kHz, 44.1kHz, 48kHz or 96kHz. The master clock is used to operate the digital filters and the noise shaping circuits.

In Slave mode the WM8569 has a master clock detection circuit that automatically determines the relationship between the system clock frequency and the sampling rate (to within +/- 32 master clocks). If there is a greater than 32 clocks error the interface defaults to 768fs mode. The master clocks must be synchronised with LRC, although the WM8569 is tolerant of phase variations or jitter on this clock. Table 6 shows the typical master clock frequency inputs for the WM8569.

The signal processing for the WM8569 typically operates at an oversampling rate of 128fs for both ADC and DAC. The exception to this for the DAC is for operation with a 128/192fs system clock, e.g. for 192kHz operation, when the oversampling rate is 64fs. For ADC operation at 96kHz it is recommended that the user set the ADCOSR bit. This changes the ADC signal processing oversample rate to 64fs.



SAMPLING RATE		S	MHz)			
(DACLRC/	128fs	192fs	256fs	384fs	512fs	768fs
ADCLRC)	DAC only	DAC only				
32kHz	4.096	6.144	8.192	12.288	16.384	24.576
44.1kHz	5.6448	8.467	11.2896	16.9340	22.5792	33.8688
48kHz	6.144	9.216	12.288	18.432	24.576	36.864
96kHz	12.288	18.432	24.576	36.864	Unavailable	Unavailable
192kHz	24.576	36.864	Unavailable	Unavailable	Unavailable	Unavailable

Table 6 System Clock Frequencies Versus Sampling Rate

HARDWARE CONTROL MODES

When the MODE pin is held high, the following hardware modes of operation are available.

Note: When in hardware mode the ADC and DAC will only run in slave mode.

MUTE AND AUTOMUTE OPERATION

In both hardware and software modes, MUTE controls the selection of MUTE directly, and can be used to enable and disable the automute function. This pin becomes an output when left floating and indicates infinite ZERO detect (IZD) has been detected.

	DESCRIPTION
0	Normal Operation
1	Mute DAC channels
Floating	Enable IZD, MUTE becomes an output to indicate when IZD occurs.
	L=IZD detected, H=IZD not detected.

Table 7 Mute and Automute Control

Figure 9 shows the application and release of MUTE whilst a full amplitude sinusoid is being played at 48kHz sampling rate. When MUTE (lower trace) is asserted, the output (upper trace) begins to decay exponentially from the DC level of the last input sample. The output will decay towards V_{MID} with a time constant of approximately 64 input samples. If MUTE is applied to all channels for 1024 or more input samples the outputs will be connected directly to V_{MID} if IZD is set. When MUTE is deasserted, the output will restart immediately from the current input sample.

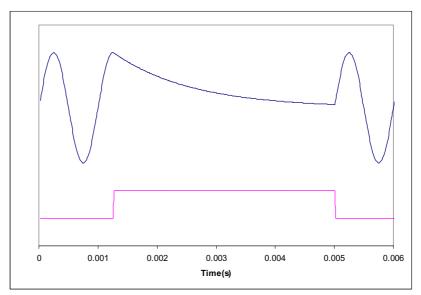


Figure 9 Application and Release of Soft Mute



The MUTE pin is an input to select mute or not mute. MUTE is active high; taking the pin high causes the filters to soft mute, ramping down the audio signal over a few milliseconds. Taking MUTE low again allows data into the filter.

The automute function detects a series of ZERO value audio samples of 1024 samples long being applied to both channels. After such an event, a latch is set whose output (AUTOMUTED) is wire OR'd through a $10k\Omega$ resistor to the MUTE pin. Thus if the MUTE pin is not being driven, the automute function will assert mute.

If MUTE is tied low, AUTOMUTED is overridden and will not mute unless the IZD register bit is set. If MUTE is driven from a bi-directional source, then both MUTE and automute functions are available. If MUTE is not driven, AUTOMUTED appears as a weak output ($10k\Omega$ source impedance) and can be used to drive external mute circuits. AUTOMUTED will be removed as soon as any channel receives a non-ZERO input.

A diagram showing how the various Mute modes interact is shown below.

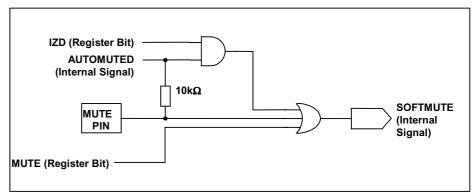


Figure 10 Control of MUTE Modes

INPUT FORMAT SELECTION

In hardware mode, CSB/IDF and SCLK/IWL become input controls for selection of input data format type and input data word length for both the ADC and DAC.

CSB/IDF	SCLK/IWL	INPUT DATA MODE
0	0	24-bit right justified
0	1	20-bit right justified
1	0	16-bit I ² S
1	1	24-bit I ² S

Table 8 Input Format Selection

Note:

In 24 bit I^2S mode, any width of 24 bits or less is supported provided that the left/right clocks (ADCLRC and DACLRC) are high for a minimum of 24 bit clocks (ADCBCLK and DACBCLK) and low for a minimum of 24 bit clocks.

DE-EMPHASIS CONTROL

In hardware mode, the SDIN/DM pin becomes an input control for selection of de-emphasis filtering to be applied.

SDIN/DM	DE-EMPHASIS
0	Off
1	On

Table 9 De-emphasis Control



DIGITAL AUDIO INTERFACE

MASTER AND SLAVE MODES

The audio interface operates in either Slave or Master mode, selectable using the DACMS and ADCMS control bits. In both Master and Slave modes DIN is always an input to the WM8569 and DOUT is always an output. The default is Slave mode.

In Slave mode, DACLRC, ADCLRC, DACBCLK and ADCBCLK are inputs to the WM8569 (Figure 11). DIN and DACLRC are sampled by the WM8569 on the rising edge of DACBCLK. ADC data is output on DOUT and changes on the falling edge of ADCBCLK.

By setting the control bit DACBCP the polarity of DACBCLK may be reversed so that DIN is sampled on the falling edge of DACBCLK. By setting the control bit ADCBCP, the polarity of ADCBCLK may be reversed so that DOUT changes on the rising edge of ADCBCLK.

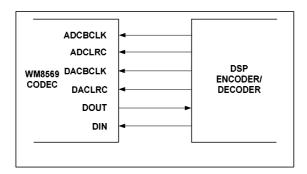


Figure 11 Slave Mode

In Master mode, DACLRC, ADCLRC, DACBCLK and ADCBCLK are outputs from the WM8569 (Figure 12). DACLRC, ADCLRC, DACBCLK and ADCBCLK are generated by the WM8569. DIN is sampled by the WM8569 on the rising edge of DACBCLK so the controller must output DAC data that changes on the falling edge of DACBCLK. ADC data is output on DOUT and changes on the falling edge of ADCBCLK.

By setting the control bit DACBCP the polarity of DACBCLK may be reversed so that DIN is sampled on the falling edge of DACBCLK. By setting the control bit ADCBCP, the polarity of ADCBCLK may be reversed so that DOUT changes on the rising edge of ADCBCLK.

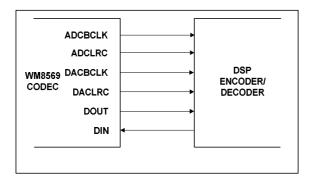


Figure 12 Master Mode

AUDIO INTERFACE FORMATS

Audio data is applied to the internal DAC filters or output from the ADC filters, via the Digital Audio Interface. 5 popular interface formats are supported:

- Left Justified mode
- · Right Justified mode
- I²S mode
- DSP Mode A
- DSP Mode B

All 5 formats send the MSB first and support word lengths of 16, 20, 24 and 32 bits, with the exception of 32 bit right justified mode, which is not supported.

In left justified, right justified and I²S modes, the digital audio interface receives DAC data on the DIN input and outputs ADC data on DOUT. Audio Data for each stereo channel is time multiplexed with DACLRC/ADCLRC indicating whether the left or right channel is present. DACLRC/ADCLRC is also used as a timing reference to indicate the beginning or end of the data words.

In left justified, right justified and I²S modes, the minimum number of BCLKs per LRC period is 2 times the selected word length. LRC must be high for a minimum of word length BCLKs and low for a minimum of word length BCLKs. Any mark to space ratio on LRC is acceptable provided the above requirements are met.

In DSP mode A or B, the DAC channels are time multiplexed onto DIN. LRC is used as a frame sync signal to identify the MSB of the first word. The minimum number of DACBCLKs per DACLRC period is 6 times the selected word length. Any mark to space ratio is acceptable on DACLRC provided the rising edge is correctly positioned. The ADC data may also be output in DSP mode A or B, with ADCLRC used as a frame sync to identify the MSB of the first word. The minimum number of ADCBCLKs per ADCLRC period is 2 times the selected word length if only the ADC is being operated.

LEFT JUSTIFIED MODE

In left justified mode, the MSB of DIN is sampled by the WM8569 on the first rising edge of DACBCLK following a DACLRC transition. The MSB of the ADC data is output on DOUT and changes on the same falling edge of ADCBCLK as ADCLRC and may be sampled on the rising edge of BCLK. LRC is high during the left samples and low during the right samples (Figure 13).

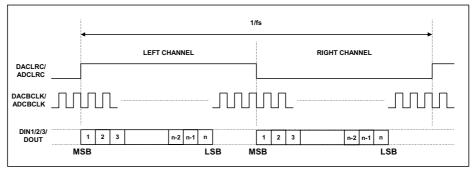


Figure 13 Left Justified Mode Timing Diagram



RIGHT JUSTIFIED MODE

In right justified mode, the LSB of DIN is sampled by the WM8569 on the rising edge of DACBCLK preceding a DACLRC transition. The LSB of the ADC data is output on DOUT and changes on the falling edge of ADCBCLK preceding a ADCLRC transition and may be sampled on the rising edge of ADCBCLK. LRC are high during the left samples and low during the right samples (Figure 14).

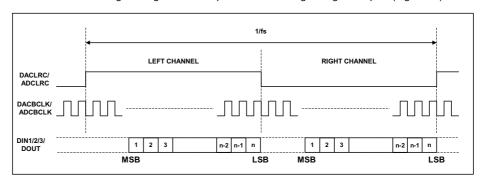


Figure 14 Right Justified Mode Timing Diagram

I2S MODE

In l^2S mode, the MSB of DIN is sampled by the WM8569 on the second rising edge of DACBCLK following a DACLRC transition. The MSB of the ADC data is output on DOUT and changes on the first falling edge of ADCBCLK following an ADCLRC transition and may be sampled on the rising edge of ADCBCLK. LRC are low during the left samples and high during the right samples.

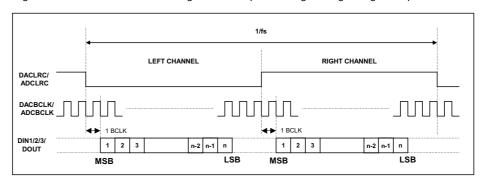


Figure 15 I²S Mode Timing Diagram

DSP MODE A

In DSP mode A, the MSB of DAC left channel data is sampled by the WM8569 on the second rising edge on DACBCLK following a DACLRC rising edge. DAC right channel follows DAC left channel (Figure 16).

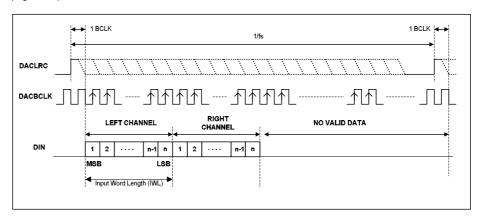


Figure 16 DSP Mode A Timing Diagram - DAC Data Input



The MSB of the left channel ADC data is output on DOUT and changes on the first falling edge of ADCBCLK following a low to high ADCLRC transition and may be sampled on the rising edge of ADCBCLK. The right channel ADC data is contiguous with the left channel data (Figure 17)

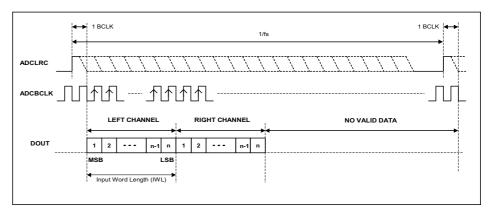


Figure 17 DSP Mode A Timing Diagram - ADC Data Output

DSP MODE B

In DSP mode B, the MSB of DAC left channel data is sampled by the WM8569 on the first DACBCLK rising edge following a DACLRC rising edge. (Figure 18).

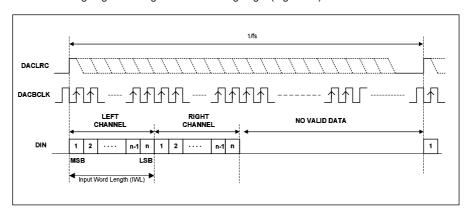


Figure 18 DSP Mode B Timing Diagram - DAC Data Input

The MSB of the left channel ADC data is output on DOUT and changes on the same falling edge of ADCBCLK as the low to high ADCLRC transition and may be sampled on the rising edge of ADCBCLK. The right channel ADC data is contiguous with the left channel data (Figure 19).

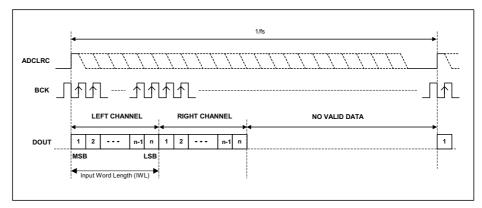


Figure 19 DSP Mode B Timing Diagram - ADC Data Output



POWERDOWN MODES

The WM8569 has powerdown control bits allowing specific parts of the WM8569 to be powered off when not being used. Control bit ADCPD powers off the ADC. The three stereo DACs each have a separate powerdown control bit, DACPD[2:0] allowing individual stereo DACs to be powered off when not in use. Setting ADCPD and DACPD[2:0] will powerdown everything except the references VMID and REFADC. These may be powered down by setting PDWN. Setting PDWN will override all other powerdown control bits. It is recommended that the ADC and DACs are powered down before setting PDWN.

ZERO DETECT

The WM8569 has a zero detect circuit for each DAC channel that detects when 1024 consecutive zero samples have been input. The MUTE pin output may be programmed to output the zero detect signal which may then be used to control external muting circuits. A '1' on MUTE indicates a zero detect. The zero detect may also be used to automatically enable DAC mute by setting IZD.

SOFTWARE CONTROL INTERFACE OPERATION

The WM8569 is controlled using a 3-wire serial interface in software mode or pin programmable in hardware mode.

The control mode is selected by the state of the MODE pin.

3-WIRE (SPI COMPATIBLE) SERIAL CONTROL MODE

SDIN/DM is used for the program data, SCLK/IWL is used to clock in the program data and CSB/IDF is used to latch the program data. SDIN/DM is sampled on the rising edge of SCLK/IWL. The 3-wire interface protocol is shown in Figure 20.

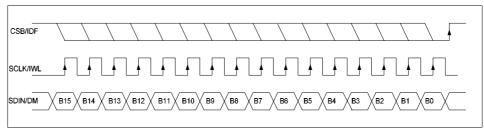


Figure 20 3-Wire SPI Compatible Interface

Notes:

- 1. B[15:9] are Control Address Bits
- 2. B[8:0] are Control Data Bits
- 3. CSB/IDF is edge sensitive the data is latched on the rising edge of CSB/IDF.

REGISTER MAP

The complete register map is shown below. The detailed description can be found in the relevant text of the device description. The WM8569 can be configured using the Control Interface. All unused bits should be set to '0'.

REGISTER	B15	B14	B13	B12	B11	B10	В9	B8	В7	В6	B5	B4	В3	B2	B1	В0	DEFAULT
R0(00h)	0	0	0	0	0	0	0	UPDATE	PDATE LDA[7:0]					011111111			
R1(01h)	0	0	0	0	0	0	1	UPDATE	PDATE RDA[7:0]					011111111			
R2(02h)	0	0	0	0	0	1	0		PL[8:5] IZD ATC 0 0 0				0	100100000			
R3(03h)	0	0	0	0	0	1	1	0	0	PHASE	DACIV	VL[5:4]	DACBCP	DACLRP	DACFI	MT[1:0]	000000000
R8(08h)	0	0	0	1	0	0	0	UPDATE	UPDATE MASTDA[7:0]					011111111			
R9(09h)	0	0	0	1	0	0	1	0	0	DEEMP	0	0	DMUTE	0	0	ZCD	000000000
R10(0Ah)	0	0	0	1	0	1	0	D/	ACRATE[8	:6]	DACMS	PWRDNALL	. 0	0	DACPD	ADCPD	010000000
R11(0Bh)	0	0	0	1	0	1	1	ADC OSR	ADCRATED 1 ADCMS 1 ADCIVITION ADCEMITED		MT[1:0]	001000000					
R12(0Ch)	0	0	0	1	1	0	0	0	0	MPD	ADCBCP	ADCLRP	ADCHP	AMUTE ALL	AMUTEL	AMUTER	000000000
R31(1Fh)	0	0	1	1	1	1	1		RESET					1	000000000		



CONTROL INTERFACE REGISTERS

ATTENUATOR CONTROL MODE

Setting the ATC register bit causes the left channel attenuation settings to be applied to both the left and right channel of the DAC from the next audio input sample. No update to the attenuation registers is required for ATC to take effect.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0000010	3	ATC	0	Attenuator Control Mode:
DAC Channel Control				0: Right channel use right attenuations
				1: Right channel use left attenuations

INFINITE ZERO DETECT ENABLE

Setting the IZD register bit will enable the internal infinite zero detect function:

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0000010	4	IZD	0	Infinite Zero Mute Enable
DAC Channel Control				0 : Disable inifinite zero mute
				1: Enable infinite zero mute

With IZD enabled, applying 1024 consecutive zero input samples each stereo channel will cause that stereo channel outputs to be muted to V_{MID} . Mute will be removed as soon as that stereo channel receives a non-zero input.

DAC OUTPUT CONTROL

The DAC output control word determines how the left and right input to the audio Interface are applied to the left and right DAC:

REGISTER ADDRESS	BIT	LABEL	DEFAULT	I	DESCRIPTION	ON
0000010 DAC Control	8:5	PL[3:0]	1001	PL[3:0]	Left Output	Right Output
				0000	Mute	Mute
				0001	Left	Mute
				0010	Right	Mute
				0011	(L+R)/2	Mute
				0100	Mute	Left
				0101	Left	Left
				0110	Right	Left
				0111	(L+R)/2	Left
				1000	Mute	Right
				1001	Left	Right
				1010	Right	Right
				1011	(L+R)/2	Right
				1100	Mute	(L+R)/2
				1101	Left	(L+R)/2
				1110	Right	(L+R)/2
				1111	(L+R)/2	(L+R)/2



DAC DIGITAL AUDIO INTERFACE CONTROL REGISTER

Interface format is selected via the DACFMT[1:0] register bits:

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0000011	1:0	DACFMT	00	Interface Format Select:
Interface Control		[1:0]		00 : Right justified mode
				01: Left justified mode
				10: I ² S mode
				11: DSP mode A orB

In left justified, right justified or I²S modes, the DACLRP register bit controls the polarity of DACLRC. If this bit is set high, the expected polarity of DACLRC will be the opposite of that shown Figure 13, Figure 14 and Figure 15. Note that if this feature is used as a means of swapping the left and right channels, a 1 sample phase difference will be introduced. In DSP modes, the DACLRP register bit is used to select between modes A and B.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0000011	2	DACLRP	0	In left/right/I ² S Modes:
Interface Control				DACLRC Polarity (normal)
				0 : Normal DACLRC polarity
				1: Inverted DACLRC polarity
				In DSP Mode:
				0 : Mode A
				1: Mode B

By default, DACLRC and DIN are sampled on the rising edge of DACBCLK and should ideally change on the falling edge. Data sources that change DACLRC and DIN on the rising edge of DACBCLK can be supported by setting the DACBCP register bit. Setting DACBCP to 1 inverts the polarity of DACBCLK to the inverse of that shown in Figure 13 to Figure 19.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0000011	3	DACBCP	0	DACBCLK Polarity (DSP Modes):
Interface Control				0: Normal BCLK polarity
				1: Inverted BCLK polarity

The DACIWL[1:0] bits are used to control the input word length.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0000011	5:4	DACIWL	00	Input Word Length:
Interface Control		[1:0]		00 : 16 bit data
				01: 20 bit data
				10: 24 bit data
				11: 32 bit data

Note: 32-bit right justified mode is not supported.

In all modes, the data is signed 2's complement. The digital filters always input 24-bit data. If the DAC is programmed to receive 16 or 20 bit data, the WM8569 pads the unused LSBs with zeros. If the DAC is programmed into 32 bit mode, the 8 LSBs are ignored.

Note: In 24 bit I²S mode, any width of 24 bits or less is supported provided that LRC is high for a minimum of 24 BCLKs and low for a minimum of 24 BCLKs.

A number of options are available to control how data from the Digital Audio Interface is applied to the DAC channels



DAC OUTPUT PHASE

The DAC Phase control word determines whether the output of the DAC is non-inverted or inverted

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0000011	6	PHASE	0	0 = non-inverted
DAC Phase				1 = inverted

DIGITAL ZERO CROSS-DETECT

The Digital volume control also incorporates a zero cross detect circuit which detects a transition through the zero point before updating the digital volume control with the new volume. This is enabled by control bit ZCD.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0001001	0	ZCD	0	DAC Digital Volume Zero Cross
DAC Control				Disable:
				0: Zero cross detect enabled
				1: Zero cross detect disabled

DAC MUTE MODES

The WM8569 has a mute for the DAC channels. Setting MUTE will apply a 'soft' mute to the input of the digital filters.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0001001	3	DMUTE	0	DAC Soft Mute Select
DAC Mute				0 = Not mute
				1 = Mute

Refer to Figure 9 for the plot of application and release of soft mute.

Note that all other means of muting the DAC channels: setting the PL[3:0] bits to 0, setting the PDWN bit or setting attenuation to 0 will cause much more abrupt muting of the output.

ADC MUTE MODES

Each ADC channel also has a mute control bit, which mutes the inputs to the ADC.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0001100	0	AMUTER	0	ADC Mute Select:
ADC Mute				0 : Normal operation
				1: mute ADC right
	1	AMUTEL	0	ADC Mute Select:
				0 : Normal operation
				1: mute ADC left
	2	AMUTEALL	0	ADC Mute Select:
				0 : Normal operation
				1: mute both ADC channels

DE-EMPHASIS MODE

Each stereo DAC channel has an individual de-emphasis control bit:

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0001001	6	DEEMPH	0	De-emphasis Channel Selection
DAC De-Emphahsis				Select:
Control				0 = Not de-emphasis
				1 = De-emphasis

