Chipsmall Limited consists of a professional team with an average of over 10 year of expertise in the distribution of electronic components. Based in Hongkong, we have already established firm and mutual-benefit business relationships with customers from,Europe,America and south Asia,supplying obsolete and hard-to-find components to meet their specific needs.

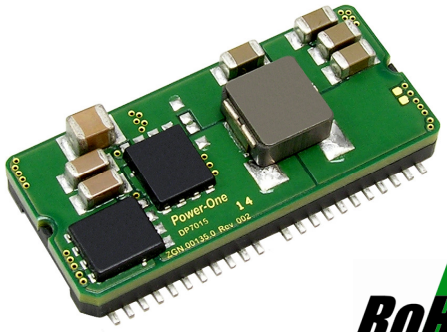
With the principle of “Quality Parts,Customers Priority,Honest Operation,and Considerate Service”,our business mainly focus on the distribution of electronic components. Line cards we deal with include Microchip,ALPS,ROHM,Xilinx,Pulse,ON,Everlight and Freescale. Main products comprise IC,Modules,Potentiometer,IC Socket,Relay,Connector.Our parts cover such applications as commercial,industrial, and automotives areas.

We are looking forward to setting up business relationship with you and hope to provide you with the best service and solution. Let us make a better world for our industry!

Contact us

Tel: +86-755-8981 8866 Fax: +86-755-8427 6832
Email & Skype: info@chipsmall.com Web: www.chipsmall.com
Address: A1208, Overseas Decoration Building, #122 Zhenhua RD., Futian, Shenzhen, China

power-one®
Changing the Shape of Power



RoHS Compliant

Applications

- Low voltage, high density systems with Intermediate Bus Architectures (IBA)
- Point-of-load regulators for high performance DSP, FPGA, ASIC, and microprocessor applications
- Desktops, servers, and portable computing
- Broadband, networking, optical, and communications systems

Benefits

- Integrates digital power conversion with intelligent power management
- Eliminates the need for external power management components
- Programmable via industry-standard I^2C communication bus (DPM required)
- Reduce the number of discrete parts within a power system.
- Reduces board space, system cost, complexity and time to market

Features

- Input voltage range: 8V–14V
- High continuous output current: 10A
- Wide digitally programmable output voltage range: 0.7V–5.5V
- Active patented current sharing
- Single-wire serial communication bus between dPOL and Digital Power Manager (DPM)
- Programmable dynamic output voltage positioning for better load transient response
- Overcurrent, overvoltage, undervoltage, and overtemperature protections with programmable thresholds and hiccup or latching modes
- Programmable fixed switching frequency: 500KHz or 1.0MHz
- Programmable switching frequency phasing
- Programmable turn-on and turn-off delays and slew rates.
- Auto Compensation
- In-System Loop Identification (SysID)through pseudo-random noise injection
- Power Good signal with programmable threshold and delay.
- Advanced fault management and propagation.
- Start up into pre-biased load.
- Real time voltage, current, and temperature measurements, monitoring, and reporting.
- Horizontal orientation SMT PCB attachment.
- Small footprint SMT package: 16x32mm.
- Extremely low profile of 7mm.
- Compatible with conventional pick-and-place equipment.
- Wide operating temperature range -40°C - 85°C
- UL 60950-1/CSA 22.2 No. 60950-1-07 Second Edition, IEC 60950-1: 2005, and EN 60950-1:2006 (pending)

Description

Power-One's DP7010 is an intelligent, fully programmable step-down point-of-load DC-DC converter integrating digital power conversion and intelligent power management. The dPOL is used in conjunction with DM73xx Series Digital Power Manager (DPM), and completely eliminates the need for external components for output voltage setting, sequencing, tracking, protection, monitoring, error amplifier compensation and reporting. All performance parameters of the DP7010 are programmable and managed through Digital Power Manager via the industry-standard I^2C communication bus and can be changed by a user at any time during product development and operation. Telemetry data is available in real time and can be accessed over the I²C bus.

Reference Documents:

- DM7300 Digital Power Manager Data Sheet
- DM7300 Digital Power Manager Programming Manual
- Power-One I2C GEN II Graphical User Interface
- DM00056-KIT USB to I^2C Adapter Kit. User Manual

1. Ordering Information

DP	70	10	G	–	zz
Product family: dPWER®	**Series:** Intelligent dPOL Converter	**Output Current:** 7A	**RoHS compliance:** **G** - RoHS compliant for all six substances	**Dash**	**Packaging Option**[1] **R100** - 100pcs T&R **R200** – 200pcs T&R **Q1** – 1pc sample for evaluation only

Example: **DP7010G-R200**: A 200-piece reel of RoHS compliant dPOL converters. Each dPOL converter is labeled DP7010G.

2. Absolute Maximum Ratings

Stresses in excess of the absolute maximum ratings may cause performance degradation, adversely affect long-term reliability, and cause permanent damage to the converter.

Parameter	Conditions/Description	Min	Max	Units
Inductor or Printed Circuit Board (PCB) Temperature	Input Voltage applied	-40	125	°C
Input Voltage	250ms Transient		15	VDC
Output Current	(See Output Current De-rating Curves)	-8	10	ADC

3. Environmental and Mechanical Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
Ambient Temperature Range		-40		85	°C
Storage Temperature (Ts)		-55		125	°C
Weight				8	grams
MTBF	Calculated Per Telcordia Technologies SR-332	6.24			MHrs
Peak Reflow Temperature	DP7010G		245	260	°C
Lead Plating	DP7010G	100% Matte Tin			
Moisture Sensitivity Level	DP7010G	3			

[1] Packaging option is used only for ordering and not included in the part number printed on the dPOL converter label.

4. Electrical Specifications

Specifications apply at the input voltage from 8V to 14V, output load from 0 to 10A, ambient temperature from -40°C to 85°C. Test conditions include an output filter with 2 x 330µF 20mΩ solid electrolytic plus 1 x 22µF X7R ceramic output capacitors, unless otherwise noted..

4.1 Input Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
Input voltage (V_{IN})		8		14	VDC
Input Current (at no load)	V_{IN}=14.0V, V_{OUT}=3.3V		50		mADC
Undervoltage Lockout	Ramping Up Ramping Down	5		7.5	VDC VDC
VLDO Input Current	Current drawn from the external low voltage supply at VLDO=8V		50		mADC

4.2 Output Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
Output Voltage Range (V_{OUT})		0.7		5.5	VDC
Output Voltage Setpoint Resolution		2.5mV (1 LSB)			
Output Voltage Setpoint Accuracy	2nd Vo Loop Enabled	±(0.6% + 5mV)			
Output Current (I_{OUT})	$V_{IN\ MIN}$ to $V_{IN\ MAX}$	-5.5[2]		7	ADC
Line Regulation	$V_{IN\ MIN}$ to $V_{IN\ MAX}$		±0.3		%V_{OUT}
Load Regulation	0 to $I_{OUT\ MAX}$		±0.2		%V_{OUT}
Dynamic Regulation Peak Deviation Settling Time	Slew rate 1A/µs, 50 -75% load step F_{SW}=500kHz to 10% of peak deviation See Output Load Transient Section		 50 60		 mV µs
Output Voltage Peak-to-Peak Ripple and Noise Scope BW=20MHz Full Load	V_{IN}=8.0V, V_{OUT}=0.7V V_{IN}=8.0V, V_{OUT}=2.5V V_{IN}=8.0V, V_{OUT}=5.5V V_{IN}=14V, V_{OUT}=0.7V V_{IN}=14V, V_{OUT}=2.5V V_{IN}=14V, V_{OUT}=5.5V		10 20 40 18 35 50		mV mV mV mV mV mV
Temperature Coefficient	V_{IN}=12V, I_{OUT}=0.5×$I_{OUT\ MAX}$		20		ppm/°C
Switching Frequency	Default		500		kHz
	Programmable to	500 / 1,000			
Duty Cycle Limit	Default Programmable, 1.56% steps	 3.125	90.5	 100	% %

[2] At negative (sink) output current (bus terminator mode) the efficiency of the DP7010 degrades resulting in increased internal power dissipation and switching noise. Therefore maximum allowable negative current under specific conditions is lower than the current determined from the de-rating curves shown in paragraph.

4.3 Protection Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
Output Overcurrent Protection					
Type	Default Programmable	Non-Latching, 130ms period Latching/Non-Latching			
Threshold	Default Programmable in 11 steps	36	132	132	%I_{OUT} %I_{OUT}
Threshold Accuracy		-20		+20	%$I_{OCP.SET}$
Output Overvoltage Protection					
Type	Default Programmable	Non-Latching, 130ms period Latching/Non-Latching			
Threshold	Default Programmable in 10% steps	110	130	130	%$V_{O.SET}$ %$V_{O.SET}$
Threshold Accuracy	Measured at $V_{O.SET}$=2.5V	-2		2	%$V_{OVP.SET}$
Delay	From instant when threshold is exceeded until the turn-off command is generated		6		µs
Turn Off Behavior[3]	Default Programmable to	Emergency Off Critical Off / Emergency Off			
Output Undervoltage Protection					
Type	Default Programmable	Non-Latching, 130ms period Latching/Non-Latching			
Threshold	Default Programmable in 5% steps	75	75	90	%$V_{O.SET}$ %$V_{O.SET}$
Threshold Accuracy	Measured at $V_{O.SET}$=2.5V	-2		2	%$V_{UVP.SET}$
Delay	From instant when threshold is exceeded until the turn-off command is generated		6		µs
Turn Off Behavior[3]	Default Programmable to	Sequenced Off Sequenced / Critical Off			
Overtemperature Protection					
Type	Default Programmable	Non-Latching, 130ms period Latching/Non-Latching			
Turn Off Threshold	Temperature is increasing		120		°C
Turn On Threshold	Temperature is decreasing after the module was shut down by OTP[4]		110		°C
Threshold Accuracy		-5		5	°C
Delay	From instant when threshold is exceeded until the turn-off command is generated		6		µs
Turn Off Behavior[3]	Default Programmable to	Sequenced Off Sequenced / Critical Off			
Tracking Protection (when Enabled)					

[3] Sequenced Off: The turn-off follows the turn-off delay and slew-rate settings; Critical Off: At turn-off both low and high switches are immediately disabled; Catastrophic Off: At turn-off the high side switch is disabled and the low side switch is enabled.
[4] OTP clears when Overtemp Warning (Status Register TW bit) turns off.

Type	Default Programmable	Disabled Latching/Non-Latching, 130ms			
Threshold	Enabled during output voltage ramping up			±250	mVDC
Threshold Accuracy		-50		50	mVDC
Delay	From instant when threshold is exceeded until the turn-off command is generated		6		µs
Overtemperature Warning					
Threshold	Always enabled, reported in Status register (TW bit)[5]		110		°C
Threshold Accuracy	From Nominal Set Point	-5		+5	°C
Hysteresis			1.7		°C
Power Good Signal (PG pin)					
Logic	V_{OUT} is inside the PG window V_{OUT} is outside the PG window	High Low			
Lower Threshold	Default Programmable in 5% steps	90	90	95	%$V_{O.SET}$ %$V_{O.SET}$
Upper Threshold	Default Programmable in 5% steps	105	110	110	%$V_{O.SET}$
Threshold Accuracy	Measured at $V_{O.SET}$=2.5V	-2		2	%$V_{O.SET}$
PG On Delay[6]	Default		0		ms
	Programmable at	0, 10, 50, 150			
PG Off Delay	Default	PG disabled when $V_{OUT} \le V_{UV}$ threshold			
	Programmable same as PG On Delay	PG disabled at turn-off command (Reset function)			

[5] Temp Warning error same sign and proportional with OTP error.
[6] From instant when threshold is exceeded until status of PG signal changes high

4.4 Feature Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
Current Share					
Type		Active, Single Line			
Maximum Number of Modules Connected in Parallel	I_{OUT} ≥ 20% $I_{OUT\ NOM}$	4			
Current Share Accuracy	I_{OUT} ≥ 20% $I_{OUT\ NOM}$			±20	%I_{OUT}
Interleave					
Interleave (Phase Shift)	Default Programmable in 22.5° steps	0	0	337.5	Degree degree
Sequencing[7]					
Turn ON Delay	Default Programmable in 1ms steps	0	0	255	ms ms
Turn OFF Delay	Default Programmable in 1ms steps	0	0	63	ms ms
Tracking					
Turn ON Slew Rate	Default Programmable in 8 steps	0.05	0.05	2.0[8]	V/ms V/ms
Turn OFF Slew Rate	Default Programmable in 8 steps	-0.05	-0.05	-2.0[8]	V/ms V/ms
Optimal Voltage Positioning					
Load Regulation	Default Programmable in 7 steps	0	0	2.45	mV/A mV/A
Feedback Loop Compensation					
Proportional (Kr)	Programmable	0.01		2	
Integral (Ti)	Programmable	1		100	µs
Differential (Td)	Programmable	1		100	µs
Differential Roll-Off (Tv)	Programmable	1		100	µs
Monitoring					
Voltage Monitoring Accuracy	12 Bit Resolution over 0.5…5.5V	-0.5		0.5	%
Current Monitoring Accuracy	20% $I_{OUT\ NOM}$ < I_{OUT} < $I_{OUT\ NOM}$	-20		+20	%I_{OUT}
Temperature Monitoring Accuracy	Junction temperature of dPOL controller	-5		+5	°C
Remote Voltage Sense (+VS and –VS pins)[9]					
Voltage Drop Compensation	Between +VS and VOUT			300	mV
Voltage Drop Compensation	Between -VS and PGND			100	mV

[7] Timing based on SD clock and subject to tolerances of SD.

[8] Achieving fast slew rates under specific line and load conditions may require feedback loop adjustment. See Rising and Falling Slew Rates..

[9] For remote sense, it is recommended to place a 0.01-0.1µF ceramic capacitor between +VS and –VS pins as close to the dPOL converter as possible.

4.5 Signal Specifications

Parameter	Conditions/Description	Min	Nom	Max	Units
VDD	Internal supply voltage	3.15	3.3	3.45	V
Logic In Max	Pull Up Logic max safe input			VDD+.4	
SYNC/DATA Line (SD pin)					
ViL_sd	LOW level input voltage	-0.5		0.3 x VDD	V
ViH_sd	HIGH level input voltage	0.75 x VDD		VDD + 0.5	V
Vhyst_sd	Hysteresis of input Schmitt trigger	0.25 x VDD		0.45 x VDD	V
VoL	LOW level sink current @ 0.5V	14		60	mA
Tr_sd	Maximum allowed rise time 10/90%VDD			300	ns
Cnode_sd	Added node capacitance		5	10	pF
Ipu_sd	Pull-up current source at Vsd=0V	0.3		1.0	mA
Freq_sd	Clock frequency of external SD line	475		525	kHz
Tsynq	Sync pulse duration	22		28	% of clock cycle
T0	Data=0 pulse duration	72		78	% of clock cycle
Inputs: ADDR0…ADDR4, EN, IM					
ViL_x	LOW level input voltage	-0.5		0.3 x VDD	V
ViH_x	HIGH level input voltage	0.7 x VDD		VDD+0.5	V
Vhyst_x	Hysteresis of input Schmitt trigger	0.1 x VDD		0.3 x VDD	V
RdnL_ADDR	External pull down resistance ADDRX forced low			10	kOhm
Power Good and OK Inputs/Outputs					
Iup_PG	Pull-up current source input forced low PG	25		110	µA
Iup_OK	Pull-up current source input forced low OK	175		725	µA
ViL_x	LOW level input voltage	-0.5		0.3 x VDD	V
ViH_x	HIGH level input voltage	0.7 x VDD		VDD+0.5	V
Vhyst_x	Hysteresis of input Schmitt trigger	0.1 x VDD		0.3 x VDD	V
IoL	LOW level sink current at 0.5V	4		20	mA
Current Share Bus (CS pin)					
Iup_CS	Pull-up current source at VCS = 0V	0.84		3.1	mA
ViL_CS	LOW level input voltage	-0.5		0.3 x VDD	V
ViH_CS	HIGH level input voltage	0.75 x VDD		VDD+0.5	V
Vhyst_CS	Hysteresis of input Schmitt trigger	0.25 x VDD		0.45 x VDD	V
IoL	LOW level sink current at 0.5V	14		60	mA
Tr_CS	Maximum allowed rise time 10/90% VDD			100	ns

5. Pin Assignments and Descriptions

Pin Name	Pin Number	Pin Type	Buffer Type	Pin Description	Notes
VIN	1	P		Not Used	Not connected internally
IM	2			Not Used	Leave floating
NC	3			Not Used	Leave floating
NC	4			Not Used	Leave floating
NC	5			Not Used	Leave floating
NC	6			Not Used	Leave floating
NC	7			Not Used	Leave floating
NC	8			Not Used	Leave floating
VREF	9		A	Not Used	Nominally 2.5V. Leave floating
EN	10			Not Used	Leave Floating
OK	11	I/O	PU	Fault/Status Condition	Connect to OK pin of the DPM and any other dPOLs of the same group.
SD	12	I/O	PU	Sync/Data Line	Connect to SD pin of DPM
PG	13	I/O	PU	Power Good	Pin state reflected in Status Register.
TRIM	14			Not Used	Leave floating
CS	15	I/O	PU	Current Share	Connect to CS pin of other dPOLs connected in parallel. Leave floating if not in sharing.
ADDR4	16	I	PU	dPOL Address Bit 4	Tie to PGND for 0 or leave floating for 1
ADDR3	17	I	PU	dPOL Address Bit 3	Tie to PGND for 0 or leave floating for 1
ADDR2	18	I	PU	dPOL Address Bit 2	Tie to PGND for 0 or leave floating for 1
ADDR1	19	I	PU	dPOL Address Bit 1	Tie to PGND for 0 or leave floating for 1
ADDR0	20	I	PU	dPOL Address Bit 0	Tie to PGND for 0 or leave floating for 1
-VS	21	I	PU	Negative Voltage Sense	Connect to the negative point close to the load or PGND
+VS	22	I	PU	Positive Voltage Sense	Connect to the positive point close to the load or VOUT
VOUT	23	P		Output Voltage	
PGND	24	P		Power Ground	
VIN	25	P		Input Voltage	

Legend: I=input, O=output, I/O=input/output, P=power, A=analog, PU=internal pull-up

6. Typical Performance Characteristics

6.1 Thermal De-rating Curves

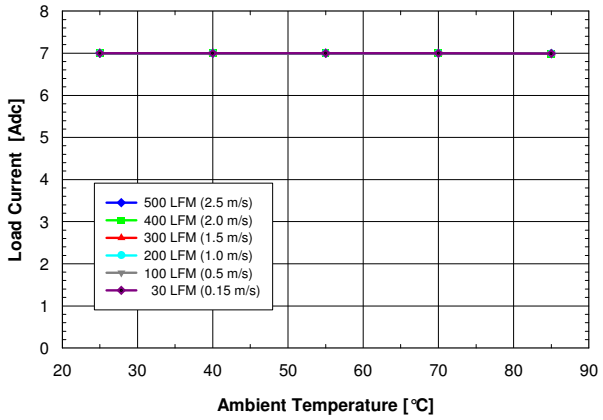


Figure 1. Available output current vs. ambient air temperature and airflow rates for converter DP7010 mounted horizontally with air flowing from input to output, MOSFET temperature ≤ 120 °C, Vin = 12 V, Vout = 5 V, and Fsw= 500KHz

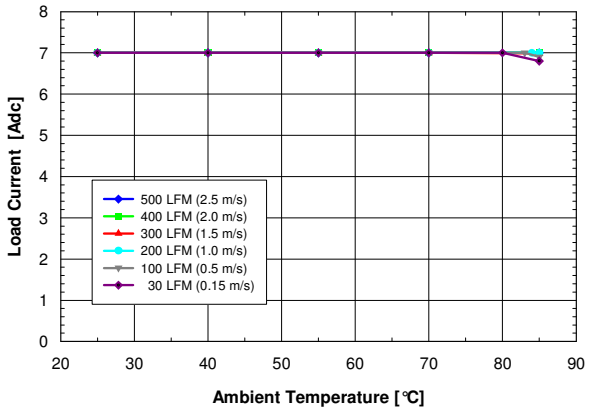


Figure 2. Available output current vs. ambient air temperature and airflow rates for converter DP7010 mounted horizontally with air flowing from input to output, MOSFET temperature ≤ 120 °C, Vin = 12 V, Vout = 5 V, and Fsw= 1MHzw

6.2 Efficiency Curves

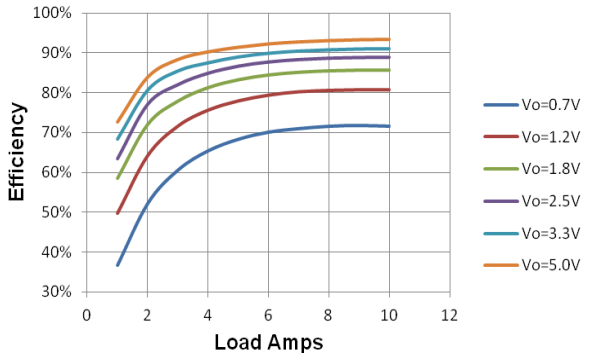


Figure 3. Efficiency vs. Load. Vin=12V, Fsw=500KHz

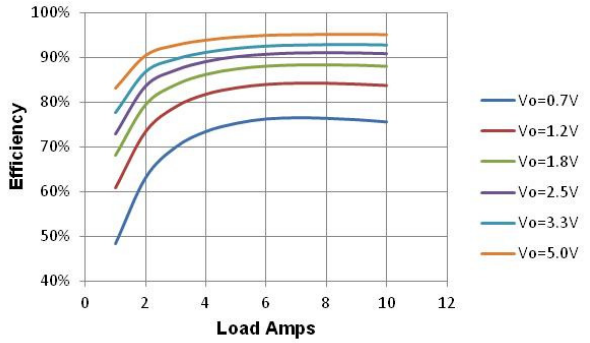


Figure 4 Efficiency vs. Load, Vin=8V, FSW=500KHz

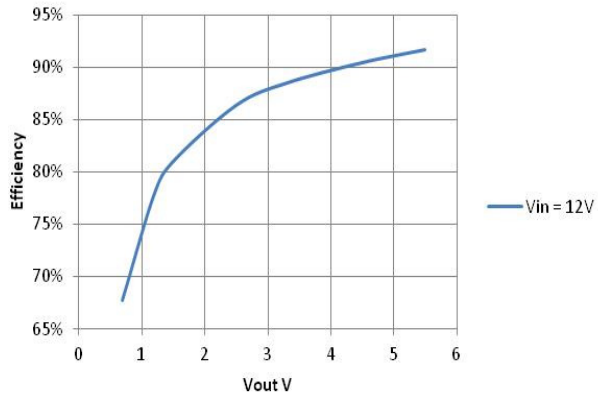


Figure 5. Efficiency vs. Output Voltage, Iout=7A, Fsw=500kHz

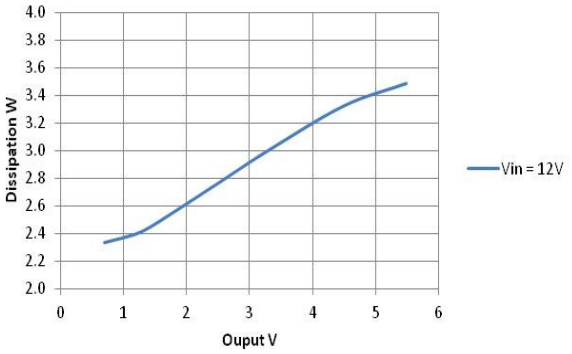


Figure 6. Dissipation vs Voltage. Iout=7A, Fsw=500kHz

7. Programmable Features

Performance parameters of DP7010 dPOL converters can be programmed via the industry standard I²C communication bus. Each parameter has a default value stored in the volatile memory registers detailed in Table 1.

Table 1. DP7010 Memory Registers

CONFIGURATION REGISTERS		
Name	**Register**	**Address**
PC1	Protection Configuration 1	0x00
PC2	Protection Configuration 2	0x01
PC3	Protection Configuration 3	0x02
TC	Tracking Configuration	0x03
INT	Interleave and Frequency Configuration	0x04
DON	Turn-On Delay	0x05
DOF	Turn-Off Delay	0x06
VLC	Voltage Loop Configuration	0x07
CLS	Current Limit Set-point	0x08
DCL	Duty Cycle Limit	0x09
PC4	Protection Configuration 4	0x0A
V1H	Output Voltage Setpoint 1 (Low Byte)	0x0B
V1L	Output Voltage Setpoint 1 (High Byte)	0x0C
V2H	Output Voltage Setpoint 2 (Low Byte)	0x0D
V2L	Output Voltage Setpoint 2 (High Byte)	0x0E
V3H	Output Voltage Setpoint 3 (Low Byte)	0x0F
V3L	Output Voltage Setpoint 3 (High Byte)	0x10
CP	Controller Proportional Coefficient	0x11
CI	Controller Integral Coefficient	0x12
CD	Controller Derivative Coefficient	0x13
B1	Controller Derivative Roll-Off Coefficient	0x14
STATUS REGISTERS		
Name	**Register**	**Address**
RUN	Run enable / status	0x15
ST	Status	0x16
MONITORING REGISTERS		
Name	**Register**	**Address**
VOH	Output Voltage High Byte (Monitoring)	0x17
VOL	Output Voltage Low Byte (Monitoring)	0x27
IO	Output Current (Monitoring)	0x18
TMP	Temperature (Monitoring)	0x19

Setup registers 00h through 14h are programmed at the system power-up. When the user programs new performance parameters, they are stored in the DPM, which overwrites the values in the registers with the new data. Upon removal of the input voltage, the default values are restored.

DP7010 converters can be programmed using the Graphical User Interface or directly via the I²C bus by using high and low level commands as described in the '"DPM Programming Manual".

DP7010 parameters can be reprogrammed at any time during the system operation and service except for the digital filter coefficients, the switching frequency and the duty cycle limit, that can only be changed when the dPOL output is turned off.

7.1 Output Voltage

The output voltage can be programmed in the GUI Output Configuration window shown in the Figure 7 or directly via the I²C bus by writing into the VOS register shown in Figure 8.

Note that the GUI shows the effect of setting PG, OV and UV limits as both values and graphical limit bars. Vertical hashed lines are error bars for the Overcurrent (OC) limit.

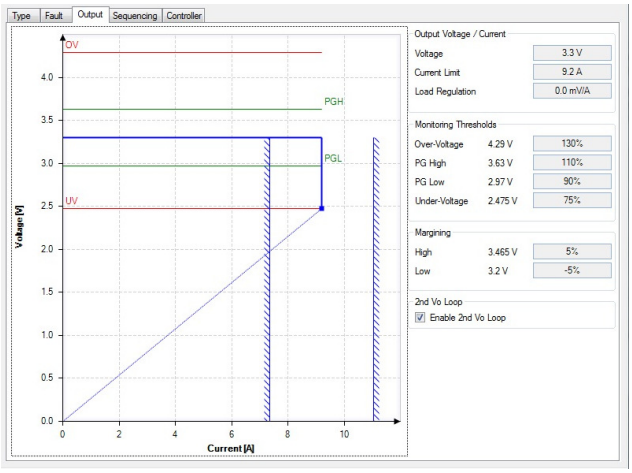


Figure 7. Output Configuration Window

7.1.1 Output Voltage Setpoint

The output voltage programming range is from 0.7 V to 5.5 V. The resolution is constant across the range and is 2.5 mV. A Total of 3 registers are provided: one should be used for the normal setpoint voltage;

the other two can be used to define a low/high margining voltage setpoint. Note that each register is 16bit wide and that the high byte needs always to be written / read first. The writing of the low byte triggers the refresh of the whole 16bit register (the high byte is written to a shadow register).

VOS: Output Voltage Set-Point
Address: 0x0B ... 0x10

	Coefficient	Addr	Bits	Default
V1H	First Vo Setpoint High Byte	0x0B	8	
V1L	First Vo Setpoint Low Byte	0x0C	8	
V2H	Second Vo Setpoint High Byte	0x0D	8	
V2L	Second Vo Setpoint Low Byte	0x0E	8	
V3H	Third Vo Setpoint High Byte	0x0F	8	
V3L	Third Vo Setpoint Low Byte	0x10	8	

Mapping:
- 12 bit data word, left aligned
- 1LSB = 2.5mV

Note:
- all registers are readable and writeable
- always write and read the high byte first

Figure 8. Output Voltage Setpoint Register VOS

Unlike other configuration registers, the dPOL controller's VOS registers are dynamic. Changes to VOS values can be made while the output is enabled over the I2C bus through register bypass commands and the dPOL will change its output immediately.

7.1.2 Output Voltage Margining

If the output voltage needs to be varied by a certain percentage, the margining function can be utilized. The margining can be programmed in the dPOL Configuration window or directly via the I^2C bus using high level commands as described in the "DM7300 Digital Power Manager Programming Manual".

In order to properly margin dPOLs that are connected in parallel, the dPOLs must be members of one of the Parallel Buses. Refer to the GUI System Configuration Window shown in Figure 47.

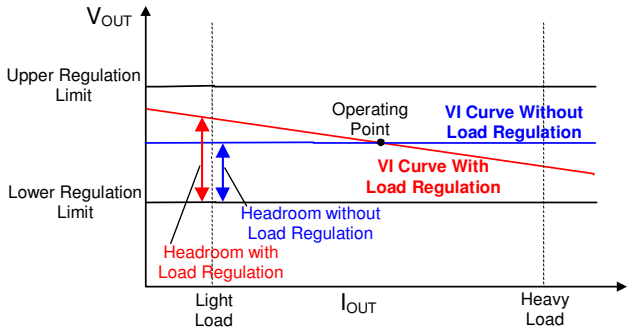


Figure 9. Optimal Voltage Positioning Concept

7.1.3 Load Regulation Control

Load Regulation provides for dynamic output voltage change proportional to load current. This feature helps to improve step load response by changing the VI characteristic slope at the point of regulation. This can be programmed in the GUI Output Configuration window shown in Figure 7 or directly via the I^2C bus by writing into the CLS register shown in Figure 25. Load Regulation can be set to one of eight values: 0, 0.74, 1.48, 2.22, 2.96, 3.71, 4.45, or 5.19 mv/A.

Figure 10 shows a DP7010 dPOL with 0 mv/A (load current) regulation. Alternating high and low output load currents causes large transients in Vout to appear with each change.

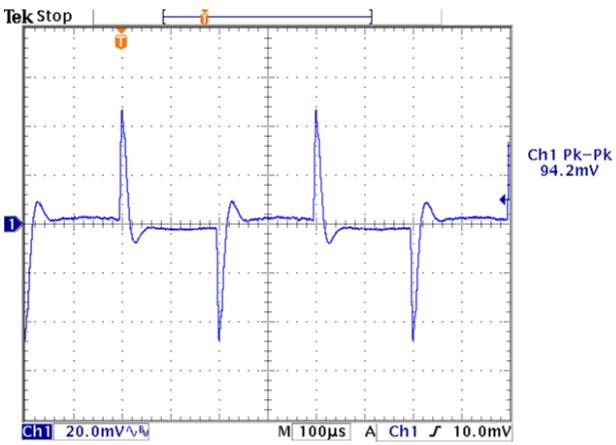


Figure 10 Transient Response with Regulation set to 0 mV/A.

As the Load Regulation parameter is increased, step offsets in output voltage begin to appear, as shown in Figure 11.

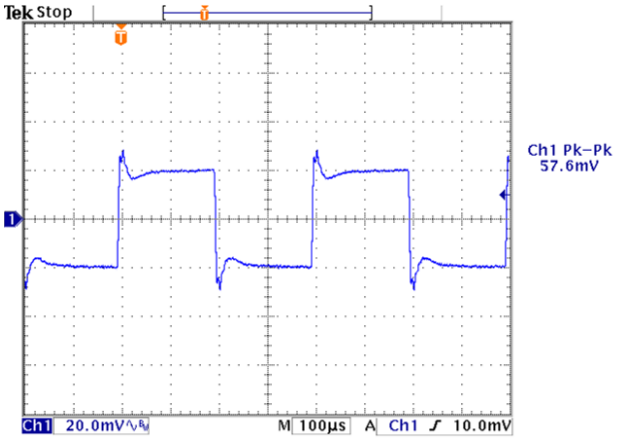


Figure 11 Transient response with non-zero Regulation.

The Load Regulation parameter is an important part of Current Sharing. It is used to set one dPOL as a "master", by assigning a lower mV/A load regulation than all other dPOLs which share the load as "slaves". The dPOL with the lowest Regulation parameter sets the effective overall regulation. (See Current Sharing elsewhere in this document.)

7.2 Sequencing and Tracking

Turn-on delay, turn-off delay, and rising and falling output voltage slew rates can be programmed in the dPOL Configure Sequencing window shown in Figure 12 or directly via the I2C bus by writing into the DON, DOF, and TC registers, respectively. The registers are shown in Figure 13, Figure 15, and Figure 16.

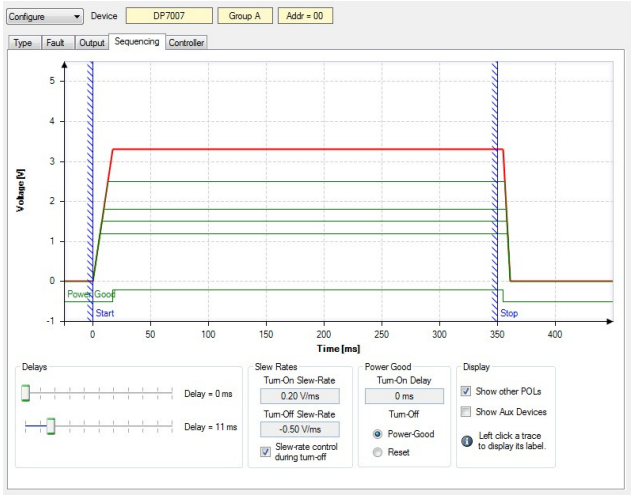


Figure 12. dPOL Configure Sequencing Window

7.2.1 Turn-On Delay

Turn-on delay is defined as an interval from the application of the Turn-On command until the output voltage starts ramping up.

DON: Turn-On Delay Configuration
Address: 0x05

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
DON7	**DON6**	**DON5**	**DON4**	**DON3**	**DON2**	**DON1**	**DON0**
Bit 7							Bit 0

Bit 7:0 **DON[7:0]**: Turn-On delay in ms
0x00 = 0ms (default)
0x01 = 1ms
…
0xFF = 255ms

Figure 13. Turn-On Delay Register DON

7.2.2 Turn-Off Delay

Turn-off delay is defined as an interval from the application of the Turn-Off command until the output voltage reaches zero (if the falling slew rate is programmed) or until both high side and low side switches are turned off (if the slew rate is not programmed). Therefore, for the slew rate controlled turn-off the ramp-down time is included in the turn-off delay as shown in Figure 14

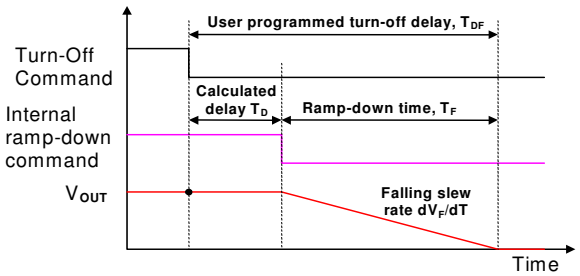


Figure 14. Relationship between Turn-Off Delay and Falling Slew Rate

As it can be seen from the figure, the internally calculated delay T_D is determined by the equation below.

$$T_D = T_{DF} - \frac{V_{OUT}}{dV_F/dT},$$

For proper operation T_D shall be greater than zero. The appropriate value of the turn-off delay needs to be programmed to satisfy the condition.

If the falling slew rate control is not utilized, the turn-off delay only determines an interval from the application of the Turn-Off command until both high

side and low side switches are turned off. In this case, the output voltage ramp-down process is determined by load parameters.

DOF: Turn-Off Delay Configuration
Address: 0x06

U	U	R/W-0	R/W-0	R/W-1	R/W-0	R/W-1	R/W-1
---	---	DOF5	DOF4	DOF3	DOF2	DOF1	DOF0
Bit 7							Bit 0

Bit 7:6 **Unimplemented:** read as '0'
Bit 5:0 **DOF[5:0]**: Turn-Off delay in ms
0x00 = 0ms
0x01 = 1ms
…
0x0B = 11ms (default)
…
0x3F = 63ms

Figure 15. Turn-Off Delay Register DOF

7.3 Turn-On Characteristics

Once delays are accounted for, turn-on and turn-off characteristics are simply a function of slew rates, which are selectable.

7.3.1 Rising and Falling Slew Rates

Output voltage ramp up (and down) control is accomplished by programming the rising and falling slew rates of the output voltage, supported in the GUI as shown in Figure 12, which is implemented by the DPM through writing data to the TC register, Figure 16.

To achieve programmed slew rates, the output voltage is being changed in 10mV steps where duration of each step determines the slew rate. For example, ramping up a 1.0V output with a slew rate of 0.5V/ms will require 100 steps duration of 20µs each.

Duration of each voltage step is calculated by dividing the master clock frequency generated by the DPM. Since all dPOLs in the system are synchronized to the master clock, the matching of voltage slew rates of different outputs is very accurate as it can be seen in Figure 17 and Figure 22.

During the turn on process, a dPOL not only delivers current required by the load (I_{LOAD}), but also charges the load capacitance. The charging current can be determined from the equation below:

$$I_{CHG} = C_{LOAD} \times dV_R / dt$$

Where, C_{LOAD} is load capacitance, dV_R/dt is rising voltage slew rate, and I_{CHG} is charging current.

TC: Tracking Configuration
Address: 0x03

U	R/W-0	R/W-0	R/W-1	R/W-1	R/W-1	R/W-0	R/W-0
---	R2	R1	R0	SC	F2	F1	F0
Bit 7							Bit 0

Bit 7 **Unimplemented**: read as '0'
Bit 6:4 **R[2:0]**: Vo rising slew rate
0 = 0.05 V/ms (default when in bus terminator mode)
1 = 0.1 V/ms (default)
2 = 0.2 V/ms
3 = 0.25 V/ms
4 = 0.5 V/ms
5 = 1.0 V/ms
6 = 2.0 V/ms
7 = Reserved
Bit 3 **SC**: Turn-off slew rate control
0 = disabled
1 = enabled (default)
Bit 2:0 **F[2:0]**: Vo falling slew rate
0 = -0.05 V/ms
1 = -0.1 V/ms
2 = -0.2 V/ms
3 = -0.25 V/ms (default when in bus terminator mode)
4 = -0.5 V/ms (default)
5 = -1.0 V/ms
6 = -2.0 V/ms
7 = Reserved

Figure 16. Tracking Configuration Register TC

When selecting the rising slew rate, a user needs to ensure that

$$I_{LOAD} + I_{CHG} < I_{OCP}$$

Where I_{OCP} is the overcurrent protection threshold of the dPOL. If the condition is not met, then the overcurrent protection will be triggered during the turn-on process. To avoid this, dV_R/dt and the overcurrent protection threshold should be programmed to meet the condition above.

7.3.2 Delay and Slew Rate Combination

The effect of setting slew rates and turn on/off delays is illustrated in the following sets of figures.

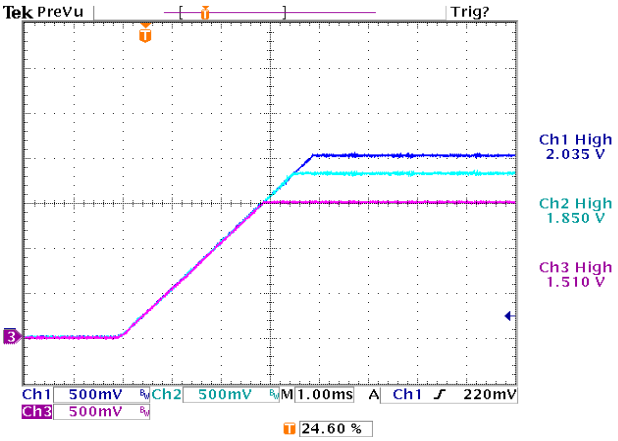


Figure 17. Tracking Turn-On. Rising Slew Rate is Programmed at 0.5V/ms for each output.

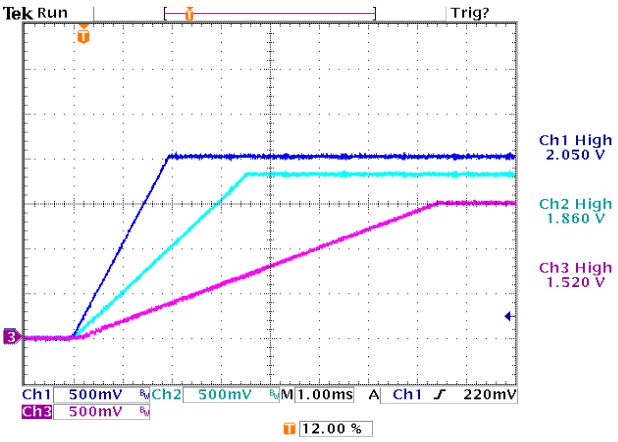


Figure 18. Turn-On with Different Rising Slew Rates. Rising Slew Rates are V1-1V/ms, V2-0.5V/ms, V3-0.2V/ms.

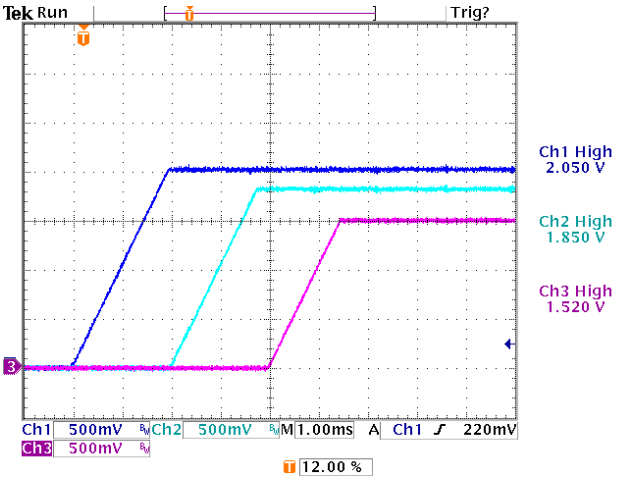


Figure 19. Sequenced Turn-On. Rising Slew Rate is Programmed at 1V/ms. V2 Delay is 2ms, V3 delay is 4ms.

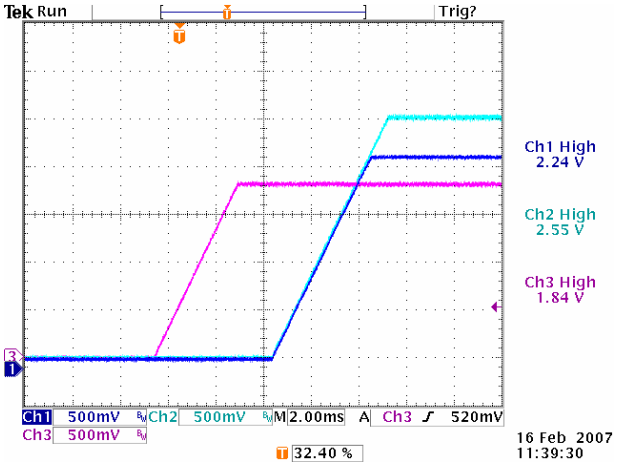


Figure 20 Two outputs delayed 5ms. All slew rates at 0.5V/ms.

7.3.3 Pre-Bias

In some applications, current leaking from a powered circuit to an unpowered bus, typically through ESD protection diodes, will accumulate charge on the unpowered bus filter capacitors. dPWER® controller in the DP7010 holds off turn on its output until the desired ramp up point crosses the pre-bias point, as seen in Figure 21.

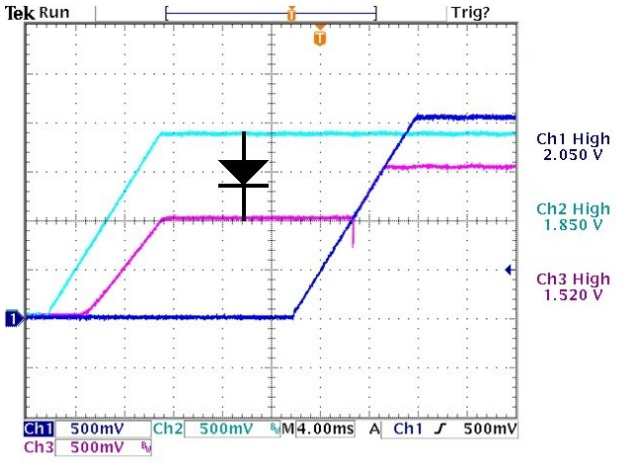


Figure 21. Turn On into Prebiased Load. V3 is Prebiased by V2 via a Diode.

This figure was captured with an actual system where a diode was added to pre-bias a 1.5V bus from a 1.85V bus in order to simulate the effect of current leakage through protection circuits of unpowered logic connected to powered logic outputs (a common source of pre-bias in power systems).

7.4 Turn-Off Characteristics

Turn of captures show that combining turn off delays and ramp rates. Note that while turnoff delays have a lower upper time limit as compared to turn on delays, all ramp down rates are available independently to turn on and off.

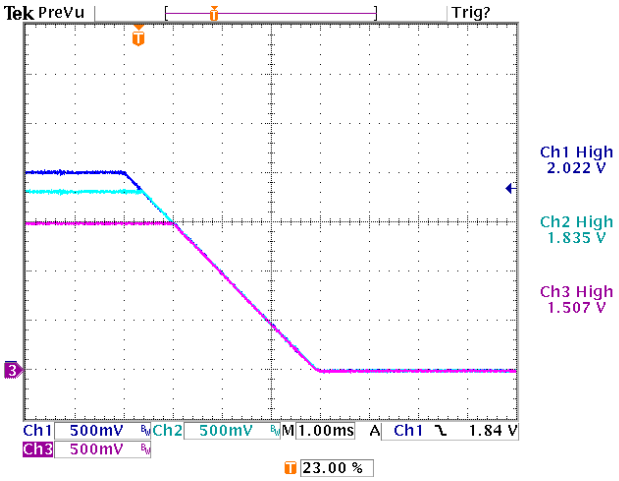


Figure 22. Tracking Turn-Off. Falling Slew Rate is Programmed at 0.5V/ms.

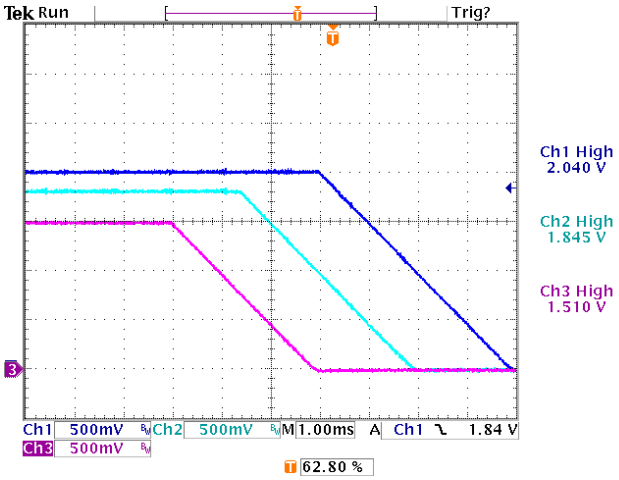


Figure 23. Turn-Off with Tracking and Sequencing. Falling Slew Rate is Programmed at 0.5V/ms.

7.5 Faults, Errors and Warnings

All dPOL series converters have a comprehensive set of programmable fault and error protection functions that can be classified into three groups based on their effect on system operation: warnings, faults, and errors. These are *warnings*, *errors* and *faults*. Warnings include Thermal (Overtemperature limit near) and Power Good (a warning in a negative sense.)

Faults in DP7xxx and DP8xxx series sPOLs include overcurrent protection, overvoltage, overtemperature and tracking failure detection. Errors include only undervoltage. Control of responses to Faults and Errors are distributed between different dPOL registers and are configurable in the GUI.

Thresholds of overcurrent, over- and undervoltage detection, and Power Good limits can be programmed in the GUI Output Configuration window (Figure 7) or directly via the I^2C bus by writing into the PC2 registers shown in Figure 24.

PC2: Protection Configuration Register 2 [1)]
Address: 0x01

U	U	R/W-0	R/W-0	R/W-1	R/W-0	R/W-0	R/W-0
---	---	PGHL	PGLL	OVPL1	OVPL0	UVPL1	UVPL0
Bit 7							Bit 0

Bit7:6 **Unimplemented**: read as '0'
Bit 5 **PGHL**: Power Good High Level
1 = 105% of Vo
0 = 110% of Vo (default)
Bit 4 **PGLL**: Power Good Low Level
1 = 95% of Vo
0 = 90% of Vo (default)
Bit 3:2 **OVPL**: Over Voltage Protection Level
00 = 110% of Vo
01 = 120% of Vo
10 = 130% of Vo (default)
11 = 130% of Vo
Bit 1:0 **UVPL**: Under Voltage Protection Level
00 = 75% of Vo (default)
01 = 80% of Vo
10 = 85% of Vo
11 = 90% of Vo

[1)] This register can only be written when PWM is not active (RUN[RUN] is '0')

Figure 24. Protection Configuration Register PC2

Note that the overvoltage and undervoltage protection thresholds and Power Good limits are defined as percentages of the output voltage. Therefore, the absolute levels of the thresholds change when the output voltage setpoint is changed either by output voltage adjustment or by margining.

Overcurrent limits are set either in the GUI dPOL Output configuration dialog or in the dPOL's CLS register as shown in Figure 25

Note that the CLS register includes bits which control the Regulation option settings. When writing into this register be careful to not change Regulation by accident.

CLS: Current Limit Setting
Address: 0x08

R/W-0	R/W-0	R/W-0	R/W-1	R/W-1	R/W-0	R/W-1	R/W-1
LR2	**LR1**	**LR0**	**TCE**	**CL3**	**CL2**	**CL1**	**CL0**
Bit 7							Bit 0

Bit 7:5 **LR[2:0]**: Load Regulation setting
0 = 0 V/A/Ω (default)
1 = 0.39 V/A/Ω
2 = 0.78 V/A/Ω
3 = 1.18 V/A/Ω
4 = 1.57 V/A/Ω
5 = 1.96 V/A/Ω
6 = 2.35 V/A/Ω
7 = 2.75 V/A/Ω

Bit 4 **TCE**: Temperature Compensation for Current Limitation Enable
0 = disabled
1 = enabled (default)

Bit 3:0 **CLS[3:0]**: Current Limit set-point when Vo Stationary or Falling
0x0 = 37%
0x1 = 47%
…
0xB = 140% (default)
values higher than 0xB are translated to 0xB (140%)

Figure 25. Current Limit Setpoint Register CLS

7.5.1 Warnings

This group includes Overtemperature Warning and Power Good Signal. Warnings do not turn off dPOLs but rather generate signals that can be transmitted to a host controller via the I^2C bus.

7.5.1.1 Overtemperature Warning

The Overtemperature Warning is generated when temperature of the controller exceeds 120°C. The Overtemperature Warning changes the TW bit of the status register ST. When the temperature falls below 117°C, the PT bit is cleared and the Overtemperature Warning is removed.

7.5.1.2 Power Good

Power Good (PG) is an open collector output that is pulled low, if the output voltage is outside of the Power Good window. The window is formed by the Power Good High threshold that is programmable at 105 or 110% of the output voltage and the Power Good Low threshold that can be programmed at 90 or 95% of the output voltage.

Power Good protection is only enabled after the output voltage reaches its steady state level. A programmable delay can be set between 0 and 150ms to delay the release of the PG pin after the voltage has reached the steady state level (see Figure 12). This allows using the PG pin to reset load circuits properly. The Power Good protection remains active during margining voltage transitions. The threshold will vary proportionally to the voltage change (see Figure 26).

The Power Good Warning pulls the PG pin low and changes the PG bit of the status register ST to 0. When the output voltage returns within the Power Good window, the PG pin is released high, the PG bit is cleared and the Power Good Warning is removed. The Power Good pin can also be pulled low by an external circuit to initiate the Power Good Warning.

At turn-off the PG pin can be programmed to either be pulled low immediately following the turn-off command, or then when the voltage actually starts to ramp down (Reset vs. Power Good functionality in Figure 12).

Note: To retrieve status information, Status Monitoring in the GUI DPM Configure Devices window should be enabled (refer to Digital Power Manager Data Sheet). The DPM will retrieve the status information from each dPOL on a continuous basis.

7.5.2 Faults

This group includes overcurrent, overtemperature, undervoltage, and tracking protections. Triggering any protection in this group will turn off the dPOL.

7.5.2.1 Overcurrent Protection

Overcurrent protection is active whenever the output voltage of the dPOL exceeds the prebias voltage (if any). When the output current reaches the OC threshold, the POL control chip asserts an OC fault. The dPOL sets the OC bit in the register ST to 0. Both high side and low side switches of the dPOL are turned off instantly (fast turn-off).

Current sensing is across the dPOLs choke. To compensate for copper winding T_C, compensation is added to keep the OC threshold approximately constant at temperatures above room temperature. Note that the temperature compensation can be disabled in the dPOL Configure Output window or directly via the I^2C by writing into the CLS register. However, it is recommended to keep the temperature compensation enabled.

7.5.2.2 Undervoltage Protection

The undervoltage protection is only active during steady state operation of the dPOL to prevent nuisance tripping. If the output voltage decreases below the UV threshold and there is no OC fault, the UV fault signal is generated, the dPOL turns off, and the UV bit in the register ST is changed to 0. The output voltage is ramped down according to sequencing and tracking settings (regular turn-off).

7.5.2.3 Overtemperature Protection

Overtemperature protection is active whenever the dPOL is powered up. If temperature of the controller exceeds 130°C, the OT fault is generated, dPOL turns off, and the OT bit in the register ST is changed to 0. The output voltage is ramped down according to sequencing and tracking settings (regular turn-off).

If non-latching OTP is programmed, the dPOL will restart as soon as the temperature of the controller decreases below the Overtemperature Warning threshold of 120°C.

7.5.2.4 Tracking Protection

Ramp up and down operations are under control by the dPOL. Tracking protection, however, is active only when the output voltage is ramping up. The purpose of the protection is to ensure that the voltage differential between multiple rails being tracked does not exceed 250mV. This protection eliminates the need for external clamping diodes between different voltage rails which are frequently recommended by ASIC manufacturers.

When the tracking protection is enabled, the dPOL continuously compares actual value of the output voltage to its programmed value as defined by the output voltage and its rising slew rate. If absolute value of the difference exceeds 250mV, the tracking fault signal is generated, the dPOL turns off, and the TR bit in the register ST is changed to 0. Both high side and low side switches of the dPOL are turned off instantly (fast turn-off).

The tracking protection can be disabled, if it contradicts requirements of a particular system (for example turning into high capacitive load where rising slew rate is not important). It can be disabled in the dPOL Configure Fault window or directly via the I^2C bus by writing into the PC1 register.

7.5.3 Faults and Margining

As noted earlier, UV and OV protection settings are a percentage of Vout. As Vout ramps between nominal, low or high margin values, UVP and OVP limits adjust accordingly.This is illustrated in Figure 26. The middle plot of Vo (Vout) level is the result of a Low Margining command. Note that Tracking is not re-enabled during changes to Vout from margining commands.

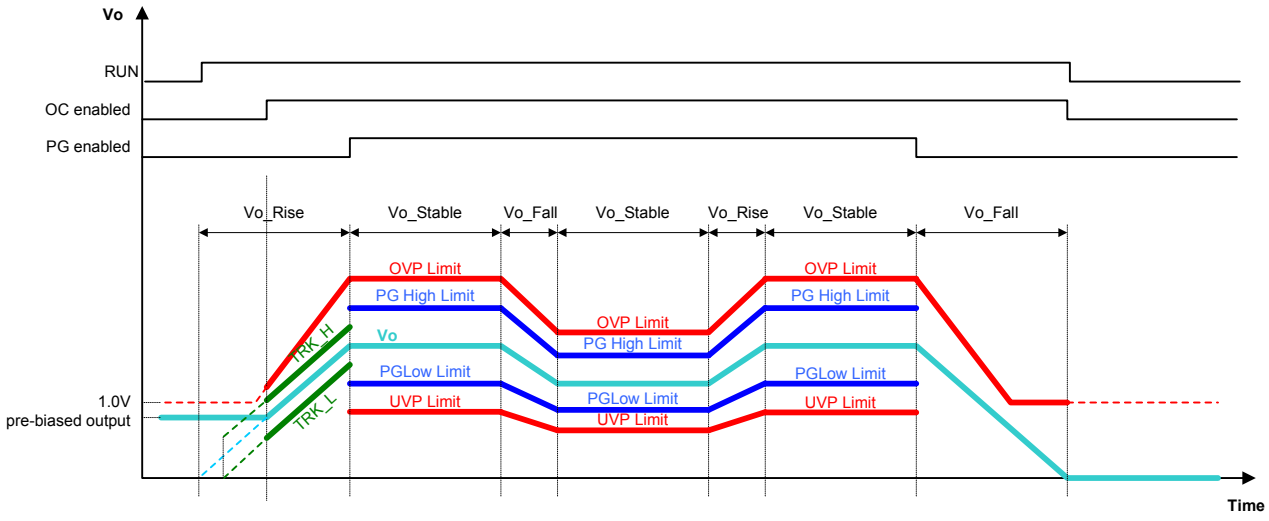


Figure 26. Protection Enable Conditions

7.5.4 Errors

This protection group only includes overvoltage protection.

7.5.4.1 Overvoltage Protection

The overvoltage protection is active whenever the output voltage of the dPOL exceeds the pre-bias voltage (if any). If the output voltage exceeds the overvoltage protection threshold, the overvoltage error signal is generated, the dPOL turns off, and the OV bit in the register ST is changed to 0. The high side switch is turned off instantly, and simultaneously the low side switch is turned on to ensure reliable protection of sensitive loads. The low side switch provides low impedance path to quickly dissipate energy stored in the output filter and achieve effective voltage limitation. The OV threshold can be programmed from 110% to 130% of the output voltage setpoint, but not lower than 0.5V. Also the OV threshold will always be at least 0.25V above the setpoint.

7.5.5 Fault and Error Latching

The user has the option of setting up any protection option as either latching/non-latching and propagating or non-propagating.

Propagation and Latching for each dPOL is set in the GUI (Figure 27 below) or directly via the I^2C by writing into the PC1 register shown in Figure 28.

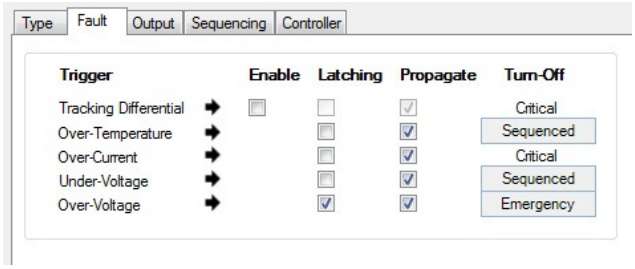


Figure 27. GUI dPOL Fault Propagation Option Window

If the non-latching protection is selected, a dPOL will attempt to restart every 130ms until the condition that triggered the protection is removed. When restarting, the output voltages follow tracking and sequencing settings.

If the latching type is selected, a dPOL will turn off and stay off. The dPOL can be turned on after 130ms, if the condition that caused the fault is removed and the respective bit in the ST register was cleared, or the Turn On command was recycled, or the input voltage was recycled.

PC1: Protection Configuration Register 1
Address: 0x00

R/W-0	R/W-1	R/W-0	R/W-0	R/W-0	R/W-0	R/W-1	R/W-1
TRE	**PVE**	**TRC**	**OTC**	**OCC**	**UVC**	**OVC**	**PVC**
Bit 7							Bit 0

Bit 7 **TRE**: Tracking fault enable
1 = enabled
0 = disabled

Bit 6 **PVE**: Phase voltage error enable
1 = enabled
0 = disabled

Bit 5 **TRC**: Tracking Fault Protection Configuration
1 = latching
0 = non-latching

Bit 4 **OTC**: Over Temperature Protection Configuration
1 = latching
0 = non-latching

Bit 3 **OCC**: Over Current Protection Configuration
1 = latching
0 = non- latching

Bit 2 **UVC**: Under Voltage Protection Configuration
1 = latching
0 = non- latching

Bit 1 **OVC**: Over Voltage Protection Configuration
1 = latching
0 = non- latching

Bit 0 **PVC**: Phase Voltage Protection Configuration
1 = latching
0 = non- latching

Figure 28. Protection Configuration Register PC1

7.5.6 Fault and Error Turn Off Control

In the GUI dPOL Fault dialog is a column of spin controls which set the Turn-Off style OT, UV and OV events. The choices are defined as:

Sequenced: Outputs shut down according to ramp down rate control settings.

Critical: Both high side and low side switches of the dPOL are turned off instantly

Emergency: The high side switch is turned off instantly, and simultaneously the low side switch is turned on to ensure reliable protection of sensitive loads

7.5.7 Fault and Error Status

Status of dPOL protection logic is stored in the dPOL's ST register shown in

When Status monitoring is enabled for a group, the DPM will read this register and make the information available for uses such as GUI Monitor display.

ST: Status register
Address: 0x16

R-1	R-0	R/W-1[1]	R/W-1[1]	R/W-1[1]	R/W-1[1]	R/W-1[1]	R/W-1[1]
TW	**PG**	**TR**	**OT**	**OC**	**UV**	**OV**	**PV**
Bit 7							Bit 0

Bit 7 **TW**: Temperature Warning
Bit 6 **PG**: Power Good Warning (high and low)
Bit 5 **TR**: Tracking Fault
Bit 4 **OT**: Over Temperature Fault
Bit 3 **OC**: Over Current Fault
Bit 2 **UV**: Under Voltage Fault
Bit 1 **OV**: Over Voltage Error
Bit 0 **PV**: Reserved

Note: an activated fault is encoded as '0'
1) Writing a '1' into a fault/error bit clears a latching fault/error

Figure 29. Protection Status Register ST

7.5.8 Fault and Error Propagation

The feature adds flexibility to the fault management scheme by giving users control over propagation of fault signals within and outside of the system. The propagation means that a fault in one dPOL can be programmed to turn off other dPOLs and devices in the system, even if they are not directly affected by the fault

7.5.8.1 Fault Propagation

When propagation is enabled, the faulty dPOL pulls its OK pin low. This signals to the DPM and any other dPOL connected to that signal, that the dPOL has a Fault or Error condition. A low OK line initiates turn-off of other dPOLs connected to the same OK line with the same turn-off behavior as the faulty dPOL. The turn-off type is encoded into the OK line when it transitions from high to low.

7.5.8.2 Grouping of dPOLs

dPWER® dPOLs can be arranged in groups of up to 4, 8, 16 or 32 dPOLs (depending upon the DPM model used). Membership in a group is set in the GUI in the **DPM** / **Configure** / **Devices** dialog, and implemented in hardware by connecting the OK pins of each dPOL in the group to the matching OK input on the DPM.

In order for a particular Fault or Error to propagate through the OK line, Propagation needs to be checked in the GUI dPOL **Configure** / **Fault** Management Window. shown in Figure 30

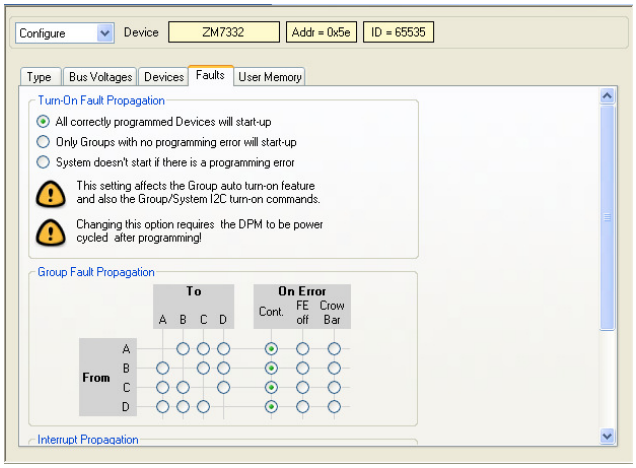


Figure 30. DPM Configure Faults Window

Note that the turn-off type of the fault as it propagates through the DPM will remain unchanged.

Propagation options for dPOLs can be read or set in the dPOL PC3 register shown in Figure 31

PC3: Protection Configuration Register 3
Address: 0x02

U	U	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
---	**---**	**TRP**	**OTP**	**OCP**	**UVP**	**OVP**	**PVP**
Bit 7							Bit 0

Bit 7:6 **Unimplemented**: Read as '0'
Bit 5 **TRP**: Tracking Protection Propagation
0 = disabled
1 = enabled
Bit 4 **OTP**: Over Temperature Protection Propagation
0 = disabled
1 = enabled
Bit 3 **OCP**: Over Current Protection Propagation
0 = disabled
1 = enabled
Bit 2 **UVP**: Under Voltage Protection Propagation
0 = disabled
1 = enabled
Bit 1 **OVP**: Over Voltage Protection Propagation
0 = disabled
1 = enabled
Bit 0 **PVP**: Reserved

Figure 31. Protection Configuration Register PC3

7.5.9 Front End and Crowbar

If an error is propagated to at least the Group level, the DPM can also be configured to generate commands to turn off a front end (a DC-DC converter generating the intermediate bus voltage) and to trigger an optional crowbar protection to accelerate removal of the IBV voltage.

7.5.10 Propagation Process

Understanding Fault and Error propagation is easier with the following examples.

The First example is of of non-propagation from a dPOL, as shown in Figure 32. An undervoltage error shuts down the Vo, but since propagation was not enabled, OK-A is not pulled down and Vo2 stays up.

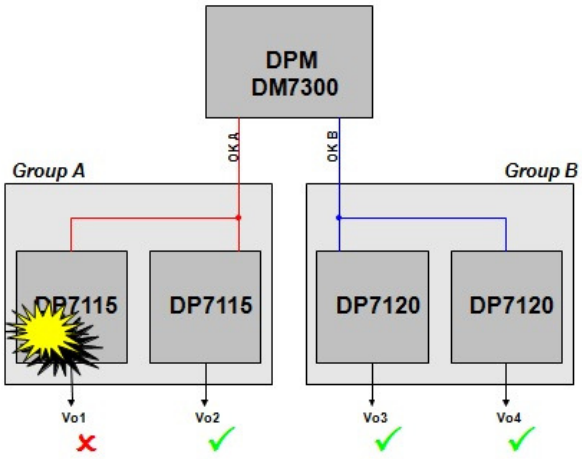


Figure 32. No Group Fault Propagation

Figure 33 shows a scope capture an actual system when undervoltage error detection is set to not propagate.

In this example, the dPOL connected to scope Ch 1 encounters the undervoltage fault after turn-on. Because fault propagation is not enabled for this dPOL, it alone turns off and generates the UV fault signal. Because a UV fault triggers the sequenced turn-off, the dPOL meets its turn-off delay and falling slew rate settings during the turn-off process as shown in the trace for Ch1. Since the UV fault is programmed to be non-latching, the dPOL will attempt to restart every 130 ms, repeating the process described above until the condition causing the undervoltage is removed. The 130ms hiccup interval is guaranteed regardless of the turn-off delay setting.

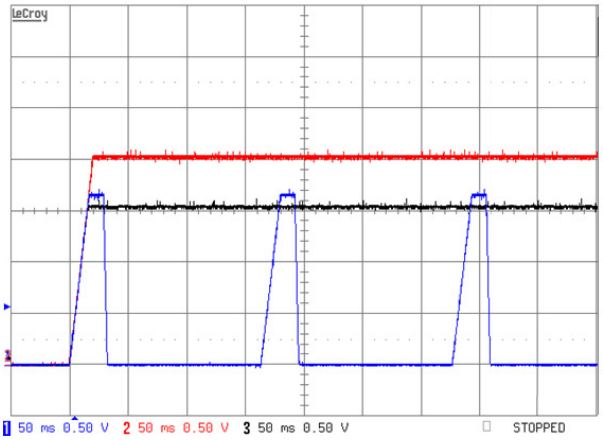


Figure 33. Turn-On into UVP on V3. The UV Fault Is Programmed To Be Non-Latching. Ch1 – Vo1, Ch2 – Vo2(Group A), Ch3 – Vo3 (Group B) Vo4 not shown.

The next example is intra-group propagation, the dPOL propagates its fault or error events. Here fault propagation between dPOLs is enabled.

In Figure 34 the dPOL powering output Vo1 again encounters an undervoltage error. It pulls its OK line low. Since the dPOL powering output Vo2 (Ch3 in the picture) belongs to the same group (A in this case), pulling down OK-A tells that dPOL to execute a regular turn-off.

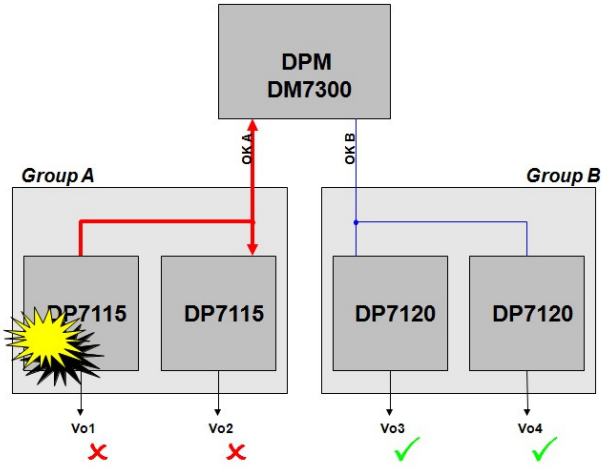


Figure 34 Intra Group Fault Propagation

Since both Vo1 and Vo2 have the same delay and slew rate settings they will continue to turn off and on synchronously every 130ms as shown in Figure 35 until the condition causing the undervoltage is removed.

Note that the dPOL powering the output Vo2 (Ch3) actually reaches its voltage set point before the error in Vo1 is detected.

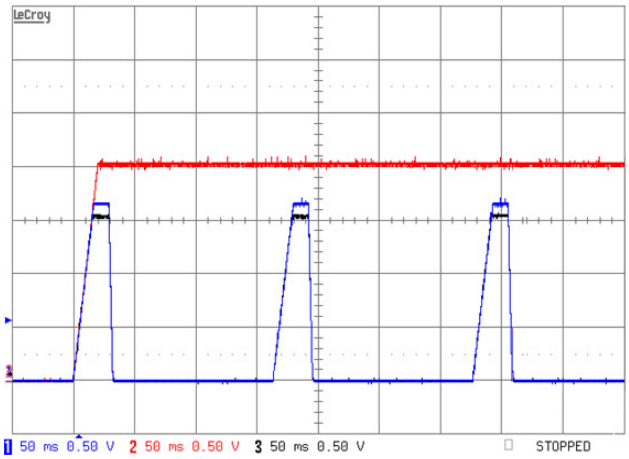


Figure 35. Turn-On into UVP on V3. The UV Fault Is Programmed To Be Non-Latching and Propagate From Group C to Group A. Ch1 – V3 (Group C), Ch2 – V2, Ch3 – V1 (Group A)

The turn-off type of a dPOL fault/error as propagated by the faulty dPOL via the OK line is propagated through the DPM to other dPOLs connected to other Groups (per configuration in) through its connection to their OK line or lines.

This behavior assures that all dPOLs configured to be affected through Group linkages will switch off with the same turn-off type.

7.5.11 Protection Summary

A summary of protection support, their parameters and features are shown in Table 2.

Table 2. Summary of Protection Parameters and Features

Code	Name	Type	When Active	Turn Off	Low Side Switch	Propagation	Disable
TW	Temperature Warning	Warning	Whenever V_{IN} is applied	No	N/A	Status Bit	No
PG	Power Good	Warning	During steady state	No	N/A	PG	No
TR	Tracking	Fault	During ramp up	Fast	Off	Critical	Yes
OT	Overtemperature	Fault	Whenever V_{IN} is applied	Regular	Off	Sequenced or Critical	No
OC	Overcurrent	Fault	When V_{OUT} exceeds prebias	Fast	Off	Critical	No
UV	Undervoltage	Fault	During steady state	Regular	Off	Sequenced or Critical	No
OV	Overvoltage	Error	When V_{OUT} exceeds prebias	Fast	On	Critical or Emergency	No

7.6 OK Coding of Faults and Errors

dPWER® dPOLs have an additional functionality added to the OK line signal. The OK line is used to propagate and receive information from other devices in the power system belonging to the same group as to the kind of turn-off procedure a device has initiated because of a fault.

Figure 36 shows the three types of OK encoding. The bubbles show when the SD and OK line logic levels are sampled by dPOL and the DPM logic.

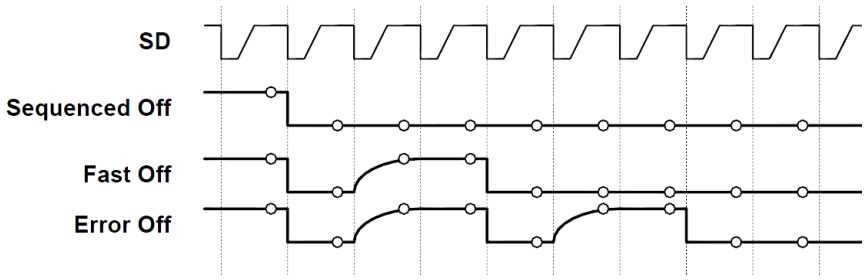


Figure 36. OK Severity Encoding Waveforms

Note that the OK line state changes are always executed by dPOLs at the negative edge of the SD line.

The chart shows shut down response types as the user can select the kind of response desired for each type of Fault or Error (within the limits of choice provided for each type of Fault or Error). All dPOL devices in the same Group are expected to trigger the same turn-off procedure in order to maintain overall tracking of output voltages in the system. And when fault propagation is set to go from one group to another, the encoding is passed along un-changed.

7.7 Switching and Compensation

dPWER® dPOLs utilize the digital PWM controller. The controller enables users to program most of the PWM performance parameters, such as switching frequency, interleave, duty cycle, and feedback loop compensation.

7.7.1 Switching Frequency

The switching frequency of the DP7010 can be programmed to either 500KHz or 1MHz in the GUI PWM Controller window shown in Figure 37 or directly via the I2C bus by writing into the INT register shown in Figure 38.

Each dPOL is equipped with a PLL that locks to the 500 KHzSD signal which is generated by the DPM. This sets up for switching actions to be synchronous to the falling edge of SD by all dPOLs, which are thereby kept coordinated to each other.

Although synchronized to SD, switching frequency selection is independent for each dPOL, with the exception of shared load bus groups, where dPOLs attached to a shared load bus are forced to use the same frequency by the GUI.

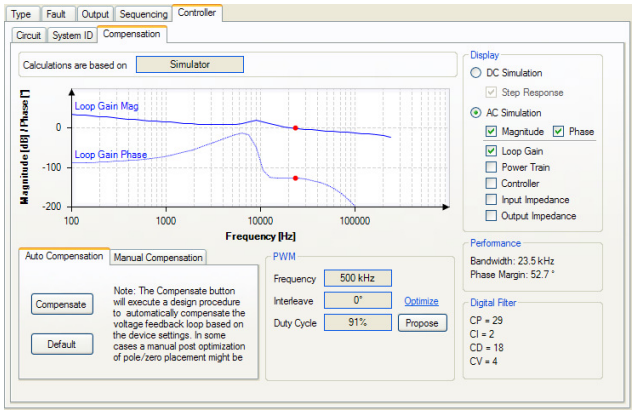


Figure 37. PWM Controller Window

In some applications, switching at higher frequencies is desirable even though efficiency is lower, because it allows for better transient response or lower application system noise.

7.7.2 Interleave Selection

Within the same PWM dialog is the switching Interleave control. Interleave is defined as a phase delay between the synchronizing slope of the master clock on the SD pin and the start of each dPOL PWM cycle. This parameter can be programmed in the dPOL Controller Configure Compensation window or directly via the I^2C bus by writing into the INT register in 22.5° steps.

INT: Interleave Configuration
Address: 0x04

R	R	R/W-0	U	R/W-0	R/W-0	R/W-0	R/W-0
PHS1	**PHS0**	**FRQ**	**---**	**INT3**	**INT2**	**INT1**	**INT0**
Bit 7							Bit 0

Bit 7:6 **PHS[1:0]**: Phase selection
0 = Single phase (PWM0)
1 = Dual phase (PWM0 and PWM2)
2 = Triple phase (PWM0, PWM1 and PWM2)
3 = Quad phase (PWM0, PWM1, PWM2 and PMW3)

Bit 5 **FRQ**: PWM frequency selection
0 = 500 kHz (default)
1 = 1000 kHz

Bit 4 **Unimplemented**: Read as '0'

Bit 3:0 **INT[3:0]**: PWM interleave phase with respect to SD line
0x00 = 0° phase lag
0x01 = 22.5° phase lag
0x02 = 45° phase lag
…
0x1F = 337.5° phase lag

Figure 38. Interleave Configuration Register INT

7.7.3 Interleave and Input Bus Noise

When a dPOL turns on its high side switch there is an inrush of current. If no interleave is programmed, inrush current spikes from all dPOLs in the system reflect back into the input source at the same time, adding together as shown in Figure 39.

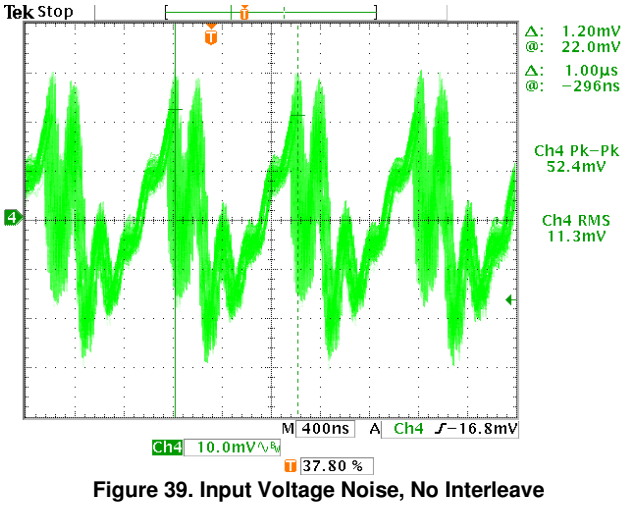


Figure 39. Input Voltage Noise, No Interleave

Figure 40 shows the input voltage noise of the three-output system with programmed interleave. Instead of all three dPOLs switching at the same time as in the previous example, the switching cycle of dPOLs V1, V2, and V3 start at 67.5°, 180°, and 303.75° of phase delay, respectively. Noise is spread evenly across the switching cycle resulting in more than 1.5 times reduction.

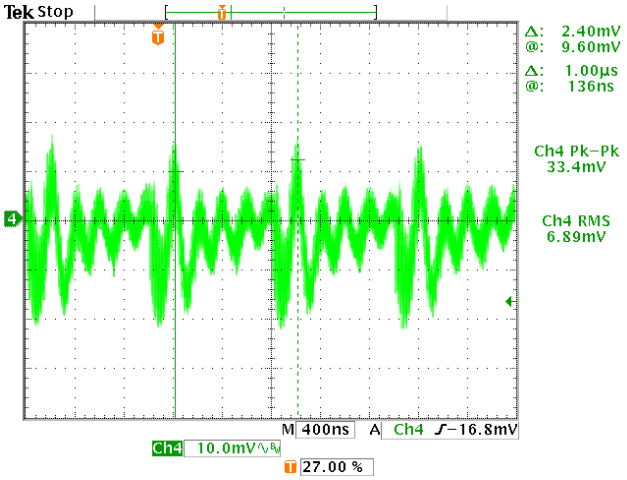


Figure 40. Input Voltage Noise with Interleave

7.7.4 Interleave and Current Sharing Noise

Similar noise reduction can be achieved on the output of dPOLs connected in parallel. Figure 41 and Figure 42 show the output noise of two dPOLs connected in parallel without and with a 180° interleave, respectively. Resulting noise reduction is more than 2 times and is equivalent to doubling switching frequency or adding extra capacitance on the output of the dPOLs.

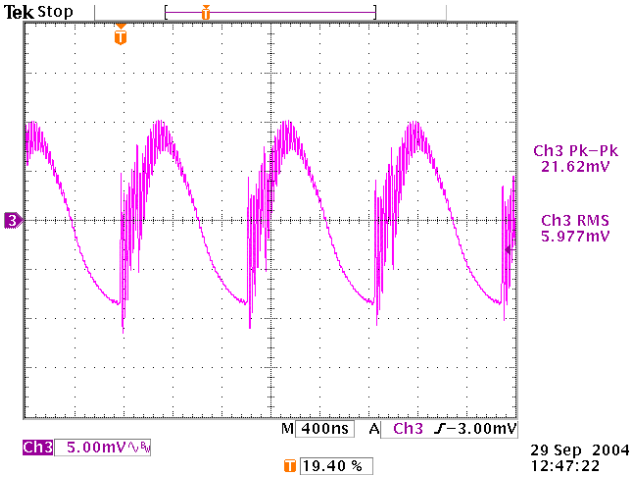


Figure 41. Output Voltage Noise, Full Load, No Interleave

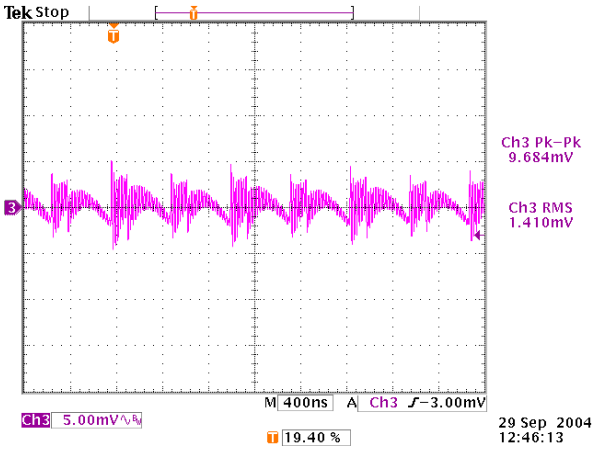


Figure 42. Output Voltage Noise, Full Load, 180° Interleave

7.7.5 Duty Cycle Limit

The DP7010 is a step-down converter therefore V_{OUT} is always less than V_{IN}. The relationship between the two parameters is characterized by the duty cycle and can be estimated from the following equation:

$$DC = \frac{V_{OUT}}{V_{IN.MIN}},$$

Where, DC is the duty cycle, V_{OUT} is the required maximum output voltage (including margining), $V_{IN.MIN}$ is the minimum input voltage.

The dPOL controller sets PWM duty cycle higher or lower than the above to compensate for drive train losses or to pull excess charge out of the output filter to keep the output voltage where it is supposed to be.

A side effect of PWM duty cycle is it also sets the rate of change of current into the output filter. A high limit helps deal with transients. However, if this is too high, an overcurrent alarm can be tripped. Thus DC limiting must be a compromise between supplying drive train losses and avoiding nuisance trips from transient load responses.

The duty cycle limit can be programmed in the GUI PWM Controller window Figure 37 or directly via the I2C bus by writing into the DCL register shown in Figure 43. The GUI will supply its own estimate of the best DC limit if the Propose button is clicked.

DCL: Duty Cycle Limitation
Address: 0x09

R/W-1	R/W-1	R/W-1	R/W-0	R/W-1	R/W-0	U	U
DCL5	**DCL4**	**DCL3**	**DCL2**	**DCL1**	**DCL0**	**---**	**---**
Bit 7							Bit 0

Bit 7:2 **DCL[5:0]**: Duty Cycle Limitation
0x00 = 0
0x01 = 1/64
0x02 = 2/64
…
0x1F = 63/64

Bit 1:0 **Unimplemented**: Read as '0'

Figure 43. Duty Cycle Limit Register

7.7.6 Feedback Loop Compensation

Programming feedback loop compensation allows optimizing dPOL performance for various application conditions. For example, increase in bandwidth can significantly improve dynamic response.

The dPOL implements a programmable PID (Proportional, Integral, and Derivative) digital controller to shape the open loop transfer function for desired bandwidth and phase/gain margin.

Feedback loop compensation can be programmed in the GUI PWM Controller window by setting Kr (Proportional), Ti (Integral), Td (Derivative), and Tv (Derivative roll-off) parameters or directly writing into the respective registers (CP, CI, CD, B1). Note that the coefficient Kr and the timing parameters (Ti, Td, Tv) displayed in the GUI do not map directly to the register values. It is therefore strongly recommended to use only the GUI to set the compensation values.

The GUI offers 3 ways to compensate the feedback loop:

Auto-Compensation: The GUI will calculate compensation settings from either information entered as to output capacitors in the application

circuit, or, if the SysID function has been run, the frequency response measured through the SysID function in the target dPOL. This method is usually sufficient, but is sensitive to accurate accounting of capacitor values and esr. The GUI displays the results of running Auto-Compensation as a set of graphs and compensation values.

Manual Compensation: The GUI supports manually adjusting feedback compensation parameters. As the parameters are changed the GUI recalculates expected frequency and phase performance.

System Identification (SysID) and Auto-Compensation: Hardware built into the dPOL controller that injects pseudo random bit sequence (PRBS) noise into PWM calculations and observes the response of the output voltage. The GUI collects this data and calculates actual system frequency response. Having frequency response data allows the Auto-Compensation function to have a better idea of actual output filter characteristics when it calculates feedback coefficients.

Using noise to plumb the output filter requires current values for compensation be good enough that injected signal can be extracted from system noise and the added noise does not trip a fault or error response. A moderately workable solution for compensation must be obtained by calculating from assumed system component values before invoking SysID.

7.8 Transient Response

The following figures show the deviation of the output voltage in response to alternating 25 / 75 % step loads applied at 2.5A/µs. The dPOL converter switching at 500KHz and had 10 x 22µF ceramic capacitors connected across the output pins. Bandwidth of the feedback loop was programmed for faster transient response.

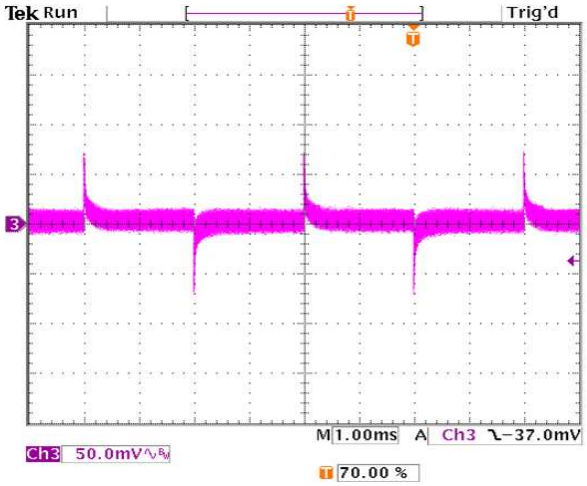


Figure 44. Transient Response with Regulation set to 0.0 mV/A.

As noted earlier, increasing the Load Regulation parameter provides a droop, or offset, in the output at the higher current load. This shows up in Figure 45.

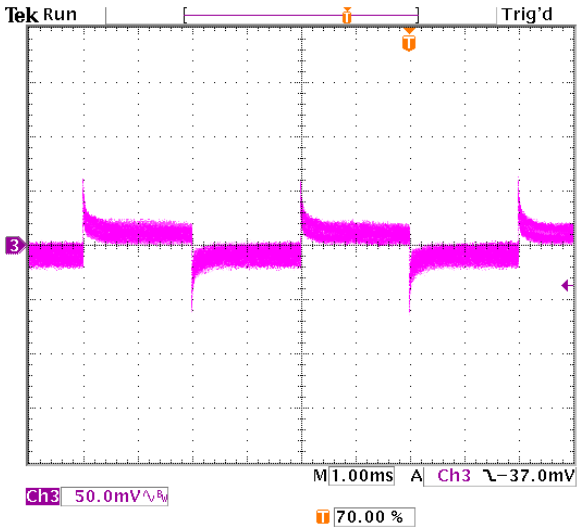


Figure 45. Transient Response with Regulation set to 1.48 mV/A

7.9 Load Current Sharing

The DP7010 is equipped with a patented active digital current share function. Setting up for current sharing requires both hardware and software configuration actions.

To set up for the current sharing, interconnect the CS pins of the dPOLs that are to share the load in parallel. This pulse width modulated digital signal drives the output currents of all dPOLs to